

General Description

The MAX690 family of supervisory circuits reduces the complexity and number of components required for power supply monitoring and battery control functions in microprocessor systems. These include μP reset and backup-battery switchover, watchdog timer, CMOS RAM write protection, and power-failure warning. The MAX690 family significantly improves system reliability and accuracy compared to that obtainable with separate ICs or discrete components.

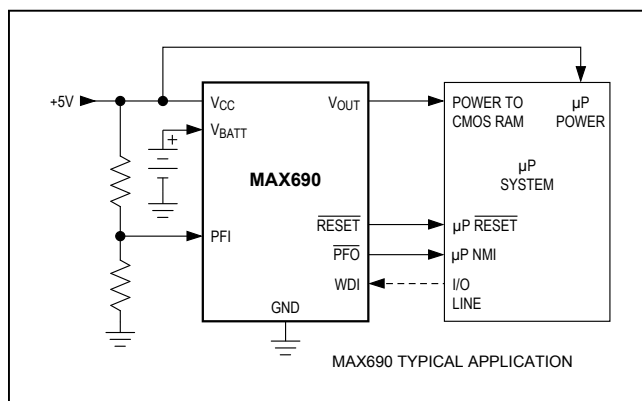
The MAX690, MAX692, and MAX694 are supplied in 8-pin packages and provide four functions:

- A reset output during power-up, power-down, and brownout conditions.
- Battery backup switching for CMOS RAM, CMOS microprocessor or other low power logic.
- A Reset pulse if the optional watchdog timer has not been toggled within a specified time.
- A 1.3V threshold detector for power fail warning, low battery detection, or to monitor a power supply other than +5V.

The MAX691, MAX693, and MAX695 are supplied in 16-pin packages and perform all MAX690, MAX692, MAX694 functions, plus:

- Write protection of CMOS RAM or EEPROM.
- Adjustable reset and watchdog timeout periods.
- Separate outputs for indicating a watchdog timeout, backup-battery switchover, and low V_{CC} .

Typical Operating Circuit



Benefits and Features

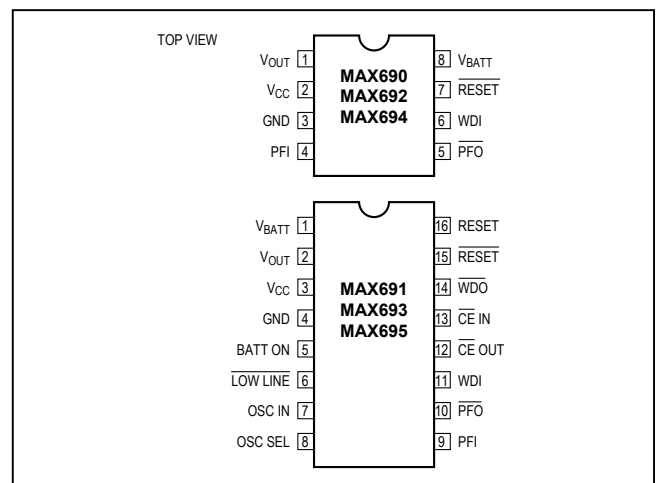
- Supervisory Function Integration Saves Board Space while Fully Protecting Microprocessor-Based Systems
 - Precision Voltage Monitor
 - 4.65V (MAX690, MAX691, MAX694, MAX695)
 - 4.40V (MAX692, MAX693)
 - Power OK/Reset Time Delay
 - 50ms, 200ms, or Adjustable
 - Watchdog Timer
 - 100ms, 1.6s, or Adjustable
 - Battery Backup Power Switching
 - Voltage Monitor for Power Fail or Low Battery Warning
 - Minimum External Component Count
- Low Power Consumption in Battery Backup Mode Extends Battery Life
 - 1 μA Standby Current
- Onboard Gating of Chip Enable Signals Protects Against Erroneous Data Written to RAM During Low V_{CC} Events

Applications

- Computers
- Controllers
- Intelligent Instruments
- Automotive Systems
- Critical μP Power Monitoring

Ordering information appears at end of data sheet.

Pin Configurations



Absolute Maximum Ratings

| | | |
|---|------------------------------------|--|
| Terminal Voltage (with respect to GND) | | Power Dissipation |
| V _{CC} | -0.3V to +6.0V | 8-Pin Plastic DIP |
| V _{BATT} | -0.3V to +6.0V | (derate 5mW/°C above +70°C)..... |
| All Other Inputs (Note 1)..... | -0.3V to (V _{OUT} + 0.5V) | 8-Pin CERDIP |
| Input Current | | (derate 8mW/°C above +85°C)..... |
| V _{CC} | 200mA | 16-Pin Plastic DIP |
| V _{BATT} | 50mA | (derate 7mW/°C above +70°C)..... |
| GND..... | 20mA | 16-Pin Small Outline |
| Output Current | | (derate 7mW/°C above +70°C)..... |
| V _{OUT} | Short circuit protected | 16-Pin CERDIP |
| All Other Outputs..... | 20mA | (derate 10mW/°C above +85°C)..... |
| Rate-of-Rise, V _{BATT} , V _{CC} | 100V/μs | Storage Temperature Range..... |
| Operating Temperature Range | | Lead Temperature (Soldering, 10s)..... |
| C suffix..... | 0°C to +70°C | |
| E suffix..... | -40°C to +85°C | |
| M suffix..... | -55°C to +125°C | |

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Electrical Characteristics

V_{CC} = full operating range, V_{BATT} = 2.8V, T_A = +25°C, unless otherwise noted.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|--|---------------------------------------|--------------------------|-------|-------|
| BATTERY BACKUP SWITCHING | | | | | |
| Operating Voltage Range (MAX690, MAX691, MAX694, MAX695 V _{CC}) | | 4.75 | | 5.5 | V |
| Operating Voltage Range (MAX690, MAX691, MAX694, MAX695 V _{BATT}) | | 2.0 | | 4.25 | |
| Operating Voltage Range (MAX692, MAX693 V _{CC}) | | 4.5 | | 5.5 | |
| Operating Voltage Range (MAX692, MAX693 V _{BATT}) | | 2.0 | | 4.0 | |
| V _{OUT} Output Voltage | I _{OUT} = 1mA | V _{CC} - 0.3 | V _{CC} - 0.1 | | V |
| | I _{OUT} = 50mA | V _{CC} - 0.5 | V _{CC} - 0.25 | | |
| V _{OUT} in Battery Backup Mode | I _{OUT} = 250μA, V _{CC} < V _{BATT} - 0.2V | V _{BATT} - 0.1 | V _{BATT} - 0.02 | | V |
| Supply Current (Excluded I _{OUT}) | I _{OUT} = 1mA | | 2 | 5 | mA |
| | I _{OUT} = 50mA | | 3.5 | 10 | |
| Supply Current in Battery Backup Mode | V _{CC} = 0V, V _{BATT} = 2.8V | | 0.6 | 1 | μA |
| Battery Standby Current (+ = Discharge, - = Charge) | 5.5V > V _{CC} > V _{BATT} + 1V | T _A = +25°C | -0.1 | +0.02 | μA |
| | | T _A = full operating range | -1.0 | +0.02 | |
| Battery Switchover Threshold (V _{CC} - V _{BATT}) | Power-up | | 70 | | mV |
| | Power-down | | 50 | | |

Electrical Characteristics (continued)

V_{CC} = full operating range, $V_{BATT} = 2.8V$, $T_A = +25^\circ C$, unless otherwise noted.)

| PARAMETER | CONDITIONS | | MIN | TYP | MAX | UNITS |
|---|--|--------------------------------|------|------------|----------|--------------|
| Battery Switchover Hysteresis | | | | 20 | | mV |
| BATT ON Output Voltage | $I_{SINK} = 3.2mA$ | | | | 0.4 | V |
| BATT ON Output Short-Circuit Current | BATT ON = $V_{OUT} = 4.5V$ sink current | | | 25 | | mA |
| | BATT ON = 0V source current | | 0.5 | 1 | 25 | μA |
| RESET AND WATCHDOG TIMER | | | | | | |
| Reset Voltage Threshold | $T_A =$ full operating range | MAX690, MAX691, MAX694, MAX695 | 4.5 | 4.65 | 4.75 | V |
| | | MAX692, MAX693 | 4.25 | 4.4 | 4.5 | |
| Reset Threshold Hysteresis | | | | 40 | | mV |
| Reset Timeout Delay (MAX690/MAX691/MAX692/MAX693) | Figure 6, OSC SEL HIGH, $V_{CC} = 5V$ | | 35 | 50 | 70 | ms |
| Reset Timeout Delay (MAX694/MAX695) | Figure 6, OSC SEL HIGH, $V_{CC} = 5V$ | | 140 | 200 | 280 | ms |
| Watchdog Timeout Period, Internal Oscillator | Long period, $V_{CC} = 5V$ | | 1.0 | 1.6 | 2.25 | s |
| | Short period, $V_{CC} = 5V$ | | 70 | 100 | 140 | ms |
| Watchdog Timeout Period, External Clock | Long period | | 3840 | | 4097 | Clock Cycles |
| | Short period | | 768 | | 1025 | |
| Minimum WDI Input Pulse Width | $V_{IL} = 0.4, V_{IH} = 0.8V_{CC}$ | | 200 | | | ns |
| \overline{RESET} and $\overline{LOW LINE}$ Output Voltage | $I_{SINK} = 1.6mA, V_{CC} = 4.25V$ | | | | 0.4 | V |
| | $I_{SOURCE} = 1\mu A, V_{CC} = 5V$ | | 3.5 | | | |
| RESET and \overline{WDO} Output Voltage | $I_{SINK} = 1.6mA$ | | | | 0.4 | V |
| | $I_{SOURCE} = 1\mu A, V_{CC} = 5V$ | | 3.5 | | | |
| Output Short-Circuit Current | RESET, \overline{RESET} , \overline{WDO} , $\overline{LOW LINE}$ | | 1 | 3 | 25 | μA |
| WDI Input Threshold Logic-Low | $V_{CC} = 5V$ (Note 2) | | | | 0.8 | V |
| WDI Input Threshold Logic-High | $V_{CC} = 5V$ (Note 2) | | 3.5 | | | |
| WDI Input Current | WDI = V_{OUT} | | | 20 | 50 | μA |
| | WDI = 0V | | -50 | -15 | | |
| POWER-FAIL DETECTOR | | | | | | |
| PFI Threshold | $V_{CC} = 5V, T_A =$ full | | 1.2 | 1.3 | 1.4 | V |
| PFI Current | | | | ± 0.01 | ± 25 | nA |
| \overline{PFO} Output Voltage | $I_{SINK} = 3.2mA$ | | | | 0.4 | V |
| | $I_{SOURCE} = 1\mu A$ | | 3.5 | | | V |
| \overline{PFO} Short Circuit Source Current | PFI = $V_{IH}, \overline{PFO} = 0V$ | | 1 | 3 | 25 | μA |

Electrical Characteristics (continued)

V_{CC} = full operating range, V_{BATT} = 2.8V, T_A = +25°C, unless otherwise noted.)

| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|--|--|------------------|---------|-----|---------|
| CHIP ENABLE GATING | | | | | |
| \overline{CE} IN Thresholds | V_{IL} | | | 0.8 | V |
| | V_{IH} | 3.0 | | | |
| \overline{CE} IN Pullup Current | | | 3 | | μ A |
| \overline{CE} OUT Output Voltage | $I_{SINK} = 3.2\text{mA}$ | | | 0.4 | V |
| | $I_{SOURCE} = 3.0\text{mA}$ | $V_{OUT} - 1.5$ | | | |
| | $I_{SOURCE} = 1\mu\text{A}$, $V_{CC} = 0\text{V}$ | $V_{OUT} - 0.05$ | | | |
| \overline{CE} Propagation Delay | $V_{CC} = 5\text{V}$ | | 50 | 200 | ns |
| OSCILLATOR | | | | | |
| OSC IN Input Current | | | ± 2 | | μ A |
| OSC SEL Input Pullup Current | | | 5 | | μ A |
| OSC IN Frequency Range | OSC SEL = 0V | 0 | | 250 | kHz |
| OSC IN Frequency with External Capacitor | OSC SEL = 0V $C_{OSC} = 47\text{pF}$ | | 4 | | kHz |

Note 1: The input voltage limits on PFI and WDI may be exceeded provided the input current is limited to less than 10mA.

Note 2: WDI is guaranteed to be in the mid-level (inactive) state if WDI is floating and V_{CC} is in the operating voltage range. WDI is internally biased to 38% of V_{CC} with an impedance of approximately 125k Ω .

Pin Description

| PIN | | NAME | FUNCTION |
|------------------------------|------------------------------|--------------------|---|
| MAX690/ MAX692/ MAX694 | MAX691/ MAX693/ MAX695 | | |
| 2 | 3 | V_{CC} | The +5V Input |
| 8 | 1 | V_{BATT} | Backup Battery Input. Connect to Ground if a backup battery is not used. |
| 1 | 2 | V_{OUT} | The higher of V_{CC} or V_{BATT} is internally switched to V_{OUT} . Connect V_{OUT} to V_{CC} if V_{OUT} and V_{BATT} are not used. Connect a 0.1 μ F or larger bypass capacitor to V_{OUT} . |
| 3 | 4 | GND | 0V Ground Reference for All Signals |
| 7 | 15 | \overline{RESET} | \overline{RESET} goes low whenever V_{CC} falls below either the reset voltage threshold or the V_{BATT} input voltage. The reset threshold is typically 4.65V for the MAX690/691/694/695, and 4.4V for the MAX692 and MAX693. \overline{RESET} remains low for 50ms after V_{CC} returns to 5V, (except 200ms in MAX694/695). \overline{RESET} also goes low for 50ms if the Watchdog Timer is enabled but not serviced within its timeout period. The \overline{RESET} pulse width can be adjusted as shown in Table 1. |
| 6 | 11 | WDI | Watchdog Input (WDI). WDI is a three level input. If WDI remains either high or low for longer than the watchdog timeout period, \overline{RESET} pulses low and \overline{WDO} goes low. The Watchdog Timer is disabled when WDI is left floating or is driven to mid-supply. The timer resets with each transition at the Watchdog Timer input. |
| 4 | 9 | PFI | Noninverting Input to the Power-Fail Comparator. When PFI is less than 1.3V, \overline{PFO} goes low. Connect PFI to GND or V_{OUT} when not used. See Figure 1. |

Pin Description (continued)

| PIN | | NAME | FUNCTION |
|------------------------------|------------------------------|------------------------------|--|
| MAX690/ MAX692/ MAX694 | MAX691/ MAX693/ MAX695 | | |
| 5 | 10 | $\overline{\text{PFO}}$ | Output of the Power-Fail Comparator. It goes low when PFI is less than 1.3V. The comparator is turned off and $\overline{\text{PFO}}$ goes low when V_{CC} is below V_{BATT} . |
| — | 13 | $\overline{\text{CE IN}}$ | $\overline{\text{CE}}$ Gating Circuit Input. Connect to GND or V_{OUT} if not used. |
| — | 12 | $\overline{\text{CE OUT}}$ | $\overline{\text{CE OUT}}$ goes low only when $\overline{\text{CE IN}}$ is low and V_{CC} is above the reset threshold (4.65V for MAX691 and MAX695, 4.4V for MAX693). See Figure 6. |
| — | 5 | BATT ON | BATT ON goes high when V_{OUT} is internally switched to the V_{BATT} input. It goes low when V_{OUT} is internally switched to V_{CC} . The output typically sinks 25mA and can directly drive the base of an external PNP transistor to increase the output current above the 50mA rating of V_{OUT} . |
| — | 6 | $\overline{\text{LOW LINE}}$ | $\overline{\text{LOW LINE}}$ goes low when V_{CC} falls below the reset threshold. It returns high as soon as V_{CC} rises above the reset threshold. See Figure 6, Reset Timing. |
| — | 16 | RESET | Active-High Output. It is the inverse of $\overline{\text{RESET}}$. |
| — | 8 | OSC SEL | When OSC SEL is unconnected or driven high, the internal oscillator sets the reset time delay and watchdog timeout period. When OSC SEL is low, the external oscillator input, OSC IN, is enabled. OSC SEL has a 3 μ A internal pullup. See Table 1. |
| — | 7 | OSC IN | When OSC SEL is low, OSC IN can be driven by an external clock to adjust both the reset delay and the watchdog timeout period. The timing can also be adjusted by connecting an external oscillator to this pin. See Figure 8. When OSC SEL is high or floating, OSC IN selects between fast and slow Watchdog timeout periods. |
| — | 14 | $\overline{\text{WDO}}$ | The Watchdog Output ($\overline{\text{WDO}}$). $\overline{\text{WDO}}$ goes low if WDI remains either high or low for longer than the watchdog timeout period. $\overline{\text{WDO}}$ is set high by the next transition at WDI. If WDI is unconnected or at mid-supply, $\overline{\text{WDO}}$ remains high. $\overline{\text{WDO}}$ also goes high when $\overline{\text{LOW LINE}}$ goes low. |

Typical Applications

MAX691, MAX693, and MAX695

A typical connection for the MAX691/693/695 is shown in Figure 1. CMOS RAM is powered from V_{OUT}. V_{OUT} is internally connected to V_{CC} when 5V power is present, or to V_{BATT} when V_{CC} is less than the battery voltage. V_{OUT} can supply 50mA from V_{CC}, but if more current is required, an external PNP transistor can be added. When V_{CC} is higher than V_{BATT}, the BATT ON output goes low, providing 25mA of base drive for the external transistor. When V_{CC} is lower than V_{BATT}, an internal 200Ω MOSFET connects the backup battery to V_{OUT}. The quiescent current in the battery backup mode is 1μA maximum when V_{CC} is between 0V and V_{BATT}–700mV.

Reset Output

A voltage detector monitors V_{CC} and generates a $\overline{\text{RESET}}$ output to hold the microprocessor's Reset line low when V_{CC} is below 4.65V (4.4V for MAX693). An internal monostable holds $\overline{\text{RESET}}$ low for 50ms* after V_{CC} rises above 4.65V (4.4V for MAX693). This prevents repeated toggling of $\overline{\text{RESET}}$ even if the 5V power drops out and recovers with each power line cycle.

The crystal oscillator normally used to generate the clock for microprocessors takes several milliseconds to start. Since most microprocessors need several clock cycles to reset, $\overline{\text{RESET}}$ must be held low until the microprocessor clock oscillator has started. The MAX690 family

power-up $\overline{\text{RESET}}$ pulse lasts 50ms* to allow for this oscillator start-up time. The manual reset switch and the 0.1μF capacitor connected to the reset bus can be omitted if manual reset is not needed. An inverted, active high, $\overline{\text{RESET}}$ output is also supplied.

Power-Fail Detector

The MAX691/93/95 issues a nonmaskable interrupt (NMI) to the microprocessor when a power failure occurs. The +5V power line is monitored via two external resistors connected to the power-fail input (PFI). When the voltage at PFI falls below 1.3V, the power-fail output ($\overline{\text{PFO}}$) drives the processor's NMI input low. If a power-fail threshold of 4.8V is chosen, the microprocessor will have the time when V_{CC} fails from 4.8V to 4.65V to save data into RAM. An earlier power-fail warning can be generated if the unregulated DC input of the 5V regulator is available for monitoring.

RAM Write Protection

The MAX691/MAX693/MAX695 $\overline{\text{CE}}$ OUT line drives the $\overline{\text{Chip Select}}$ inputs of the CMOS RAM. $\overline{\text{CE}}$ OUT follows $\overline{\text{CE}}$ IN as long as V_{CC} is above the 4.65V (4.4V for MAX693) reset threshold. If V_{CC} falls below the reset threshold, $\overline{\text{CE}}$ OUT goes high, independent of the logic level at $\overline{\text{CE}}$ IN. This prevents the microprocessor from writing erroneous data into RAM during power-up, power-down, brownouts, and momentary power interruptions. The $\overline{\text{LOW LINE}}$ output goes low when V_{CC} falls below 4.65V (4.4V for MAX693).

*200ms for MAX695

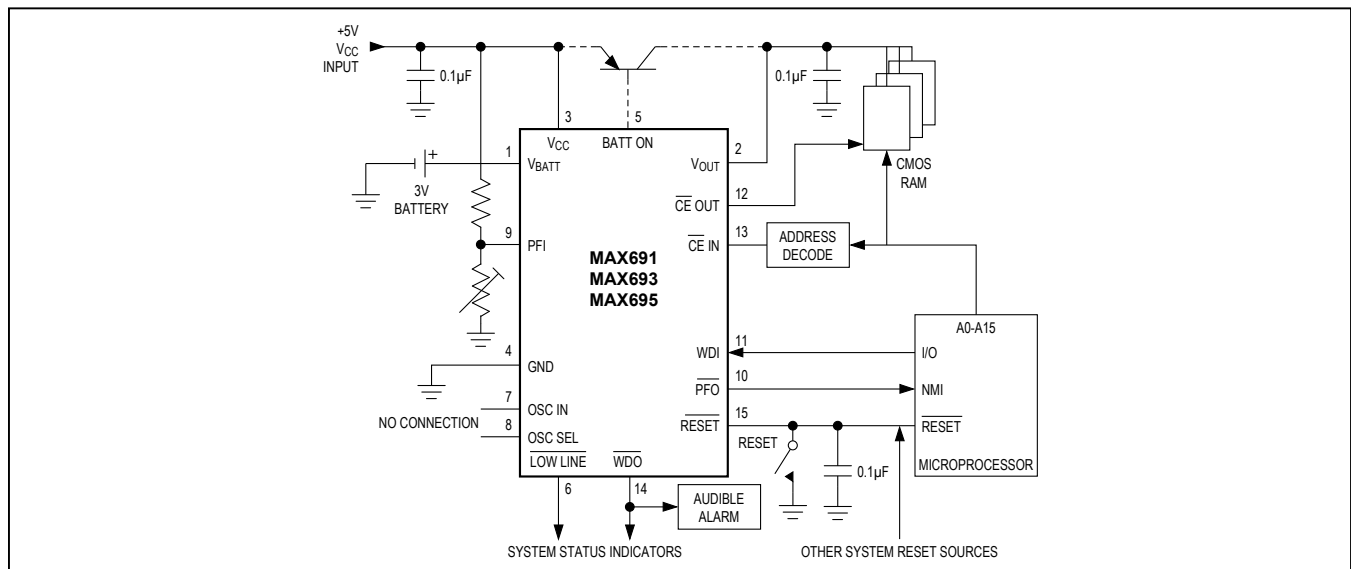


Figure 1. MAX691/693/695 Typical Application

Watchdog Timer

The microprocessor drives the WATCHDOG INPUT (WDI) with an I/O line. When OSC IN and OSC SEL are unconnected, the microprocessor must toggle the WDI pin once every 1.6s to verify proper software execution. If a hardware or software failure occurs such that WDI not toggled the MAX691/MAX693 will issue a 50ms* $\overline{\text{RESET}}$ pulse after 1.6s. This typically restarts the microprocessor’s power-up routine. A new $\overline{\text{RESET}}$ pulse is issued every 1.6s until the WDI is again strobed.

The WATCHDOG OUTPUT ($\overline{\text{WDO}}$) goes low if the watchdog timer is not serviced within its timeout period. Once $\overline{\text{WDO}}$ goes low it remains low until a transition occurs at WDI. The watchdog timer feature can be disabled by leaving WDI unconnected. OSC IN and OSC SEL also allow other watchdog timing options, as shown in Table 1 and Figure 8.

MAX690, MAX692, and MAX694

The 8 pin MAX690, MAX692, and MAX694 have most of the features of the MAX691, MAX693, and MAX695. Figure 2 shows the MAX690/MAX692/MAX694 in a typical

application. Operation is much the same as with the MAX691/MAX693/MAX695 (Figure 1), but in this case, the power-fail input (PFI) monitors the unregulated input to the 7805 regulator. The MAX690/MAX694 $\overline{\text{RESET}}$ output goes low when V_{CC} falls below 4.65V. The $\overline{\text{RESET}}$ output of the MAX692 goes low when V_{CC} drops below 4.4V.

The current consumption of the battery-backed-up power bus must be less than 50mA. The MAX690/MAX692/MAX694 does not have a BATT ON output to drive an external transistor. The MAX690/MAX692/MAX694 also does not include chip enable gating circuitry that is available on the MAX690/MAX692/MAX694. In many systems though, CE gating is not needed since a low input to the microprocessor $\overline{\text{RESET}}$ line prevents the processor from writing to RAM during power-up and power-down transients.

The MAX690/MAX692/MAX694 watchdog timer has a fixed 1.6s timeout period. If WDI remains either low or high for more than 1.6s, a $\overline{\text{RESET}}$ pulse is sent to the microprocessor. The watchdog timer is disabled if WDI is left unconnected.

*200ms for MAX695

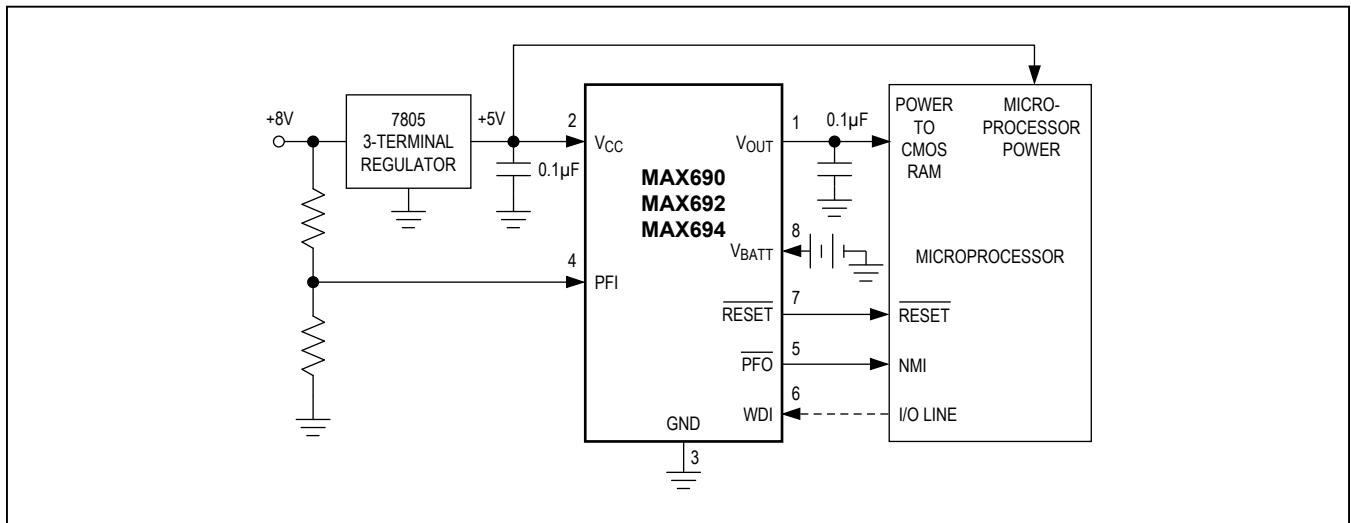


Figure 2. MAX690/692/694 Typical Application

Detailed Description

Battery-Switchover and V_{OUT}

The battery switchover circuit compares V_{CC} to the V_{BATT} input, and connects V_{OUT} to whichever is higher. Switchover occurs when V_{CC} is 50mV greater than V_{BATT} as V_{CC} falls, and V_{CC} is 70mV more than V_{BATT} as V_{CC} rises (see Figure 4). The switchover comparator has 20mV of hysteresis to prevent repeated, rapid switching if V_{CC} falls very slowly or remains nearly equal to the battery voltage.

When V_{CC} is higher than V_{BATT} , V_{CC} is internally switched to V_{OUT} via a low saturation PNP transistor. V_{OUT} has 50mA output current capability. Use an external PNP pass transistor in parallel with internal transistor if the output current requirement at V_{OUT} exceeds 50mA or if a lower $V_{CC}-V_{OUT}$ voltage differential is desired. The BATT ON output (MAX691/MAX693/MAX695 only) can directly drive the base of the external transistor.

It should be noted that the MAX690–MAX695 need only supply the average current drawn by the CMOS RAM if there is adequate filtering. Many RAM data sheets specify a 75mA maximum supply current, but this peak current spike lasts only 100ns. A 0.1 μ F bypass capacitor at V_{OUT} supplies the high instantaneous current, while V_{OUT} need only supply the average load current, which is much less. A capacitance of 0.1 μ F or greater must be connected to the V_{OUT} terminal to ensure stability.

A 200 Ω MOSFET connects V_{BATT} input to V_{OUT} during battery backup. This MOSFET has very low input-to-output differential (dropout voltage) at the low current

levels required for battery backup of CMOS RAM or other low power CMOS circuitry. When V_{CC} equals V_{BATT} the supply current is typically 12 μ A. When V_{CC} is between 0V and ($V_{BATT} - 700$ mV) the typical supply current is only 600nA typical, 1 μ A maximum.

The MAX690/MAX691/MAX694/MAX695 operate with battery voltages from 2.0V to 4.25V while MAX692/MAX693 operate with battery voltages from 2.0V to 4.0V. High value capacitors can also be used for short-term memory backup. External circuitry is required to ensure that the capacitor voltage does not rise above the reset threshold, and that the charging resistor does not discharge the capacitor when in backup mode. The MAX691A and the MAX791 provide solutions requiring fewer external components.

A small charging current of typically 10nA (0.1 μ A max) flows out of the V_{BATT} terminal. This current varies with the amount of current that is drawn from V_{OUT} but its polarity is such that the backup battery is always slightly charged, and is never discharged while V_{CC} is in its operating voltage range. This extends the shelf life of the backup battery by compensating for its self-discharge current. Also note that this current poses no problem when lithium batteries are used for backup since the maximum charging current (0.1 μ A) is safe for even the smallest lithium cells.

If the battery-switchover section is not used, connect V_{BATT} to GND and connect V_{OUT} to V_{CC} . Table 2 shows the state of the inputs and output in the low power battery backup mode.

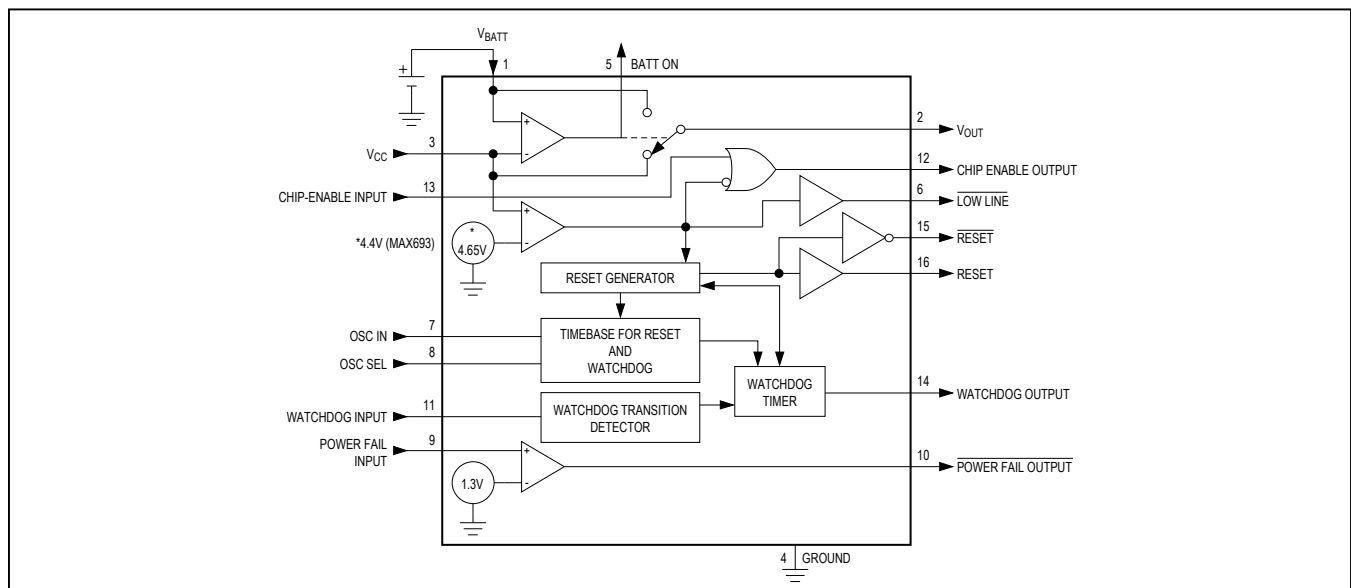


Figure 3. MAX691/MAX693/MAX695 Block Diagram

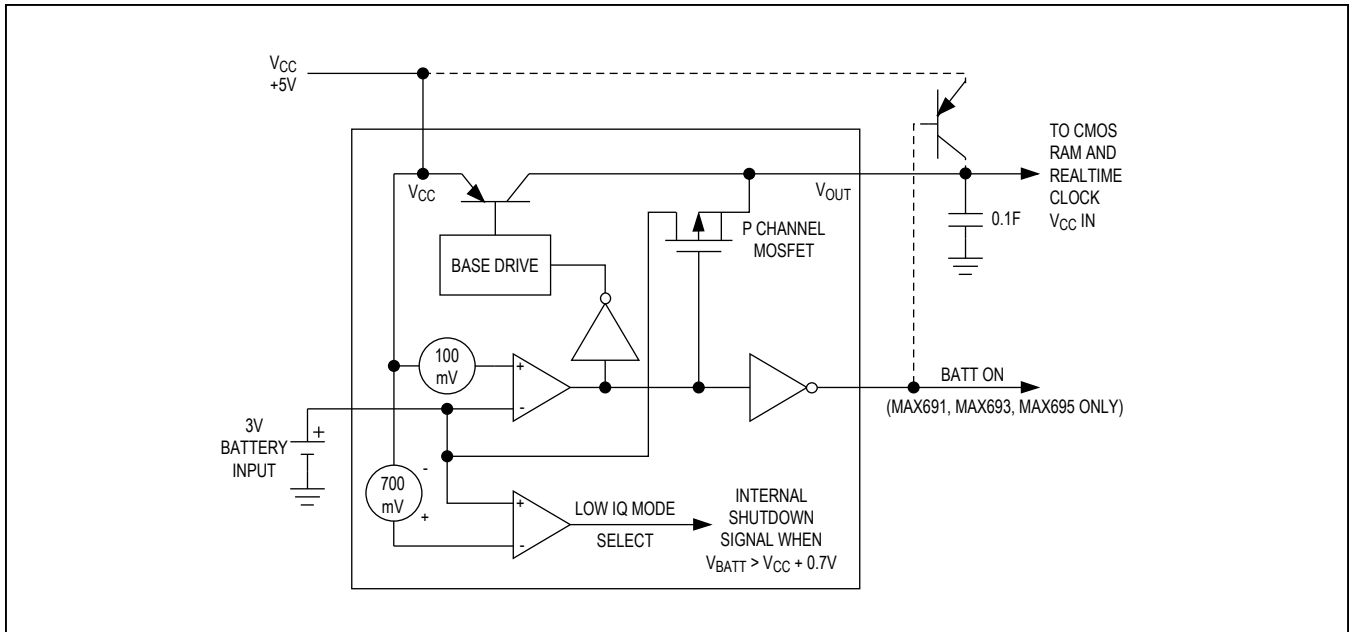


Figure 4. Battery-Switchover Block Diagram

Reset Output

$\overline{\text{RESET}}$ is an active-low output which goes low whenever V_{CC} falls below 4.5V (MAX690/MAX691/MAX694/MAX695) or 4.25V (MAX692/MAX693). It will remain low until V_{CC} rises above 4.75V (MAX690/691/694/695) or 4.5V (MAX692/MAX693) for milliseconds*. See Figures 5 and 6.

The guaranteed minimum and maximum thresholds of MAX690/MAX691/MAX694/MAX695 are 4.5V and 4.75V, while the guaranteed thresholds of the MAX692/MAX693 are 4.25V and 4.5V. The MAX690/MAX691/MAX694/MAX695 is compatible with 5V supplies with a +10%, -5% tolerance while the MAX692/MAX693 is compatible with 5V $\pm 10\%$ supplies. The reset threshold comparator has approximately 50mV of hysteresis, with a nominal threshold of 4.65V in the MAX690/MAX691/MAX694/MAX695, and 4.4V in the MAX692/MAX693.

The response time of the reset voltage comparator is about 100 μ s. V_{CC} should be bypassed to ensure that glitches do not activate the $\overline{\text{RESET}}$ output.

$\overline{\text{RESET}}$ also goes low if the watchdog timer is enabled and WDI remains either high or low longer than the watchdog timeout period. $\overline{\text{RESET}}$ has an internal 3 μ A pullup, and can either connect to and open collector Reset bus or directly drive a CMOS gate without an external pullup resistor.

*200ms for MAX694 and MAX695

$\overline{\text{CE}}$ Gating and RAM Write Protection

The MAX691/MAX693/MAX695 use two pins to control the $\overline{\text{Chip Enable}}$ or $\overline{\text{Write}}$ inputs of CMOS RAMs. When V_{CC} is +5V, $\overline{\text{CE OUT}}$ is a buffered replica of $\overline{\text{CE IN}}$, with a 50ns propagation delay. If V_{CC} input falls below 4.65V (4.5V min, 4.75V max) an internal gate forces $\overline{\text{CE OUT}}$ high, independent of $\overline{\text{CE IN}}$. The MAX693 $\overline{\text{CE OUT}}$ goes high whenever V_{CC} is below 4.4V (4.25V min, 4.5V max). The $\overline{\text{CE}}$ output of both devices is also forced high when V_{CC} is less than V_{BATT} . (See Figure 5.)

$\overline{\text{CE OUT}}$ typically drives the $\overline{\text{CE}}$, $\overline{\text{CS}}$, or $\overline{\text{Write}}$ input of battery backed up CMOS RAM. This ensures the integrity of the data in memory by preventing write operations when V_{CC} is at an invalid level. Similar protection of EEPROMs can be achieved by using the $\overline{\text{CE OUT}}$ to drive the $\overline{\text{Store}}$ or $\overline{\text{Write}}$ inputs of an EEPROM, EAROM, or NOVROM.

If the 50ns typical propagation delay of $\overline{\text{CE OUT}}$ is too long, connect $\overline{\text{CE IN}}$ to GND and use the resulting $\overline{\text{CE OUT}}$ to control a high speed external logic gate. A second alternative is to AND the LOW LINE output with the $\overline{\text{CE}}$ or $\overline{\text{WR}}$ signal. An external logic gate and the $\overline{\text{RESET}}$ output of the MAX690/MAX692/MAX694 can also be used for CMOS RAM write protection.

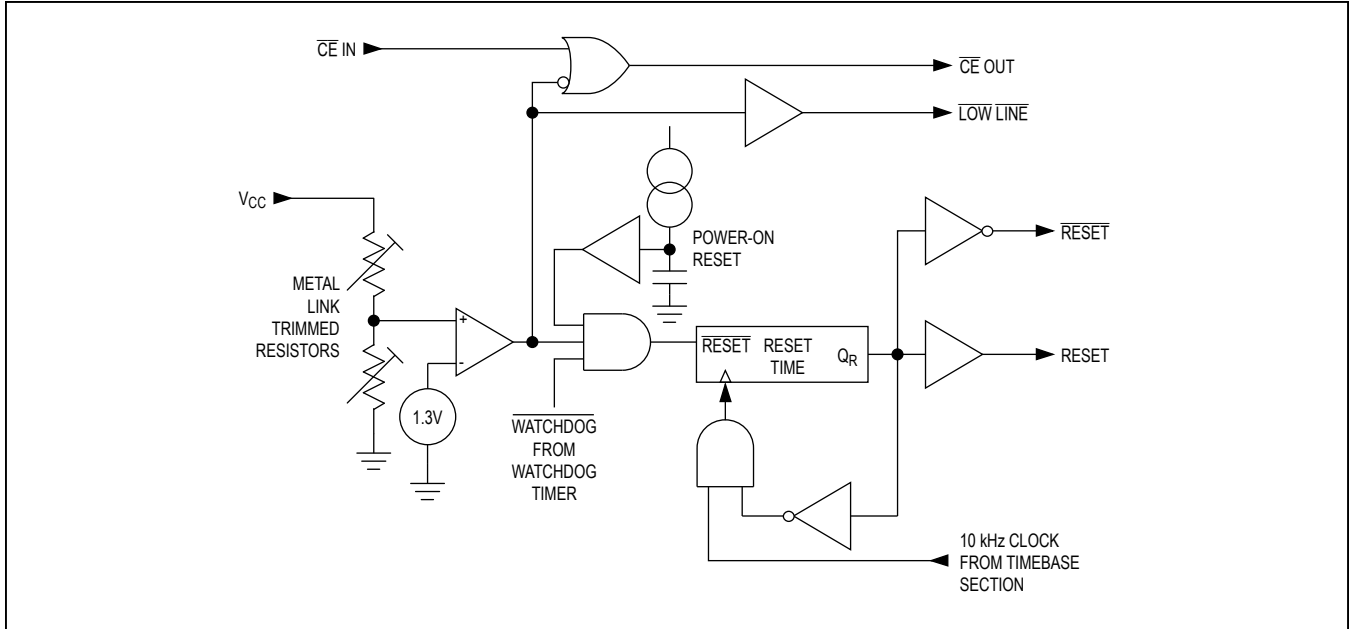
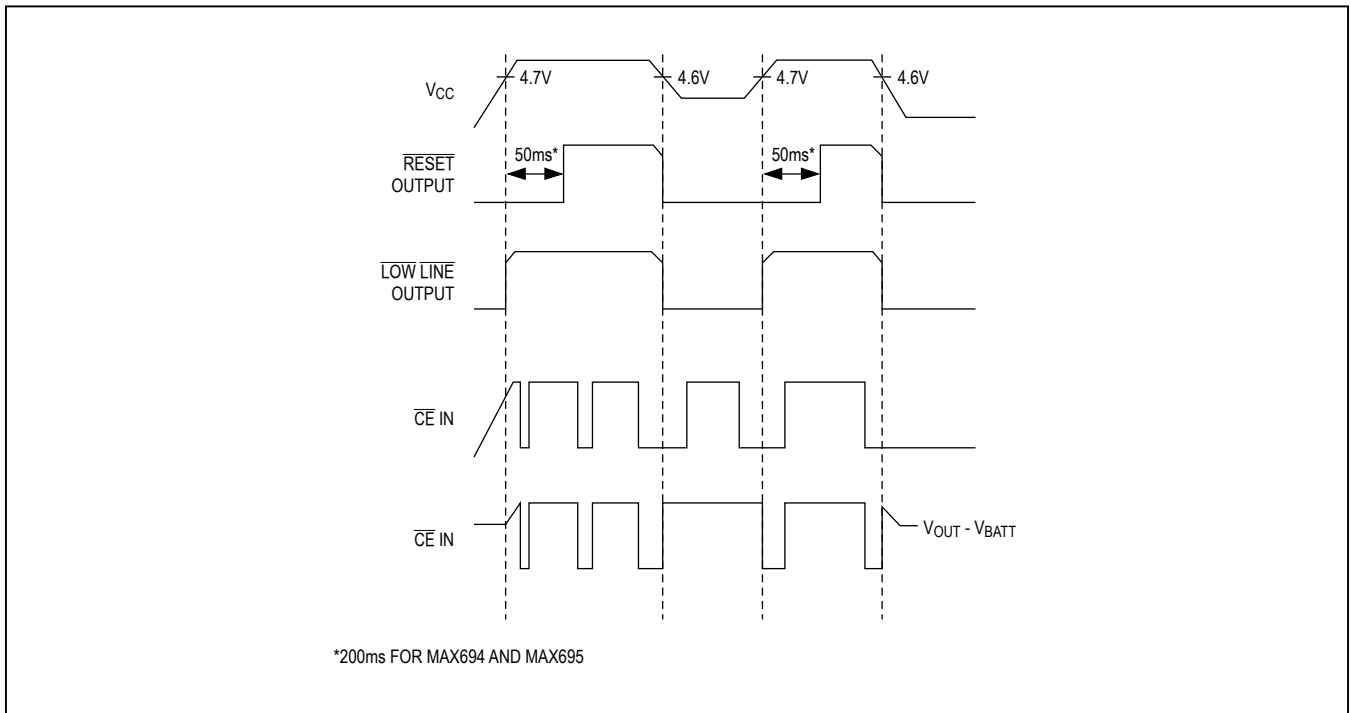


Figure 5. Reset Block Diagram



*200ms FOR MAX694 AND MAX695

Figure 6. Reset Timing

1.3V Comparator and Power-Fail Warning

The power-fail input (PFI) is compared to an internal 1.3V reference. The power-fail output ($\overline{\text{PFO}}$) goes low when the voltage at PFI is less than 1.3V. Typically, PFI is driven by an external voltage divider which senses either the unregulated DC input to the system's 5V regulator or the regulated 5V output. The voltage divider ratio can be chosen such that the voltage at PFI falls below 1.3V several milliseconds before the +5V supply falls below 4.75V. $\overline{\text{PFO}}$ is normally used to interrupt the microprocessor so that data can be stored in RAM before V_{CC} falls below 4.75V and the RESET output goes low (4.5V for MAX692/MAX693).

The power-fail detector can also monitor the backup battery to warn of a low battery condition. To conserve battery power, the power-fail detector comparator is turned off and $\overline{\text{PFO}}$ is forced when V_{CC} is lower than V_{BATT} input voltage.

Watchdog Timer and Oscillator

The watchdog circuit monitors the activity of the microprocessor. If the microprocessor does not toggle the Watchdog Input (WDI) within the selected timeout period, a 50ms* RESET pulse is generated. Since many systems cannot service the watchdog timer immediately after a reset, the MAX691/MAX693/MAX695 has a longer timeout period after reset is issued. The normal timeout period becomes effective following the first transition of WDI after RESET has gone high. The watchdog timer is restarted

at the end of reset, whether the reset was caused by lack of activity on WDI or by V_{CC} falling below the reset threshold. If WDI remains either high or low, reset pulses will be issued every 1.6s. The watchdog monitor can be deactivated by floating the watchdog input (WDI).

The watchdog output ($\overline{\text{WDO}}$, MAX691/MAX693/MAX695 only) goes low if the watchdog timer times out and remains low until set high by the next transition on the watchdog input. $\overline{\text{WDO}}$ is also set high when V_{CC} goes below the reset threshold.

The watchdog timeout period is fixed at 1.6s and the reset pulse width is fixed at 50ms* on the 8-pin MAX690/MAX692/MAX694. The MAX691/MAX693/MAX695 allow these times to be adjusted per Table 1. Figures 8 shows various oscillator configurations.

The internal oscillator is enabled when OSC SEL is floating. In this mode, OSC IN selects between the 1.6s and 100ms watchdog timeout periods. In either case, immediately after a reset the timeout period 1.6s. This gives the microprocessors time to reinitialize the system. If OSC IN is low, then the 100ms watchdog period becomes effective after the first transition of WDI. The software should be written such that the I/O port driving WDI is left in its power-up reset state until the initialization routines are completed and the microprocessor is able to toggle WDI at the minimum watchdog timeout period of 70ms.

*200ms for MAX694

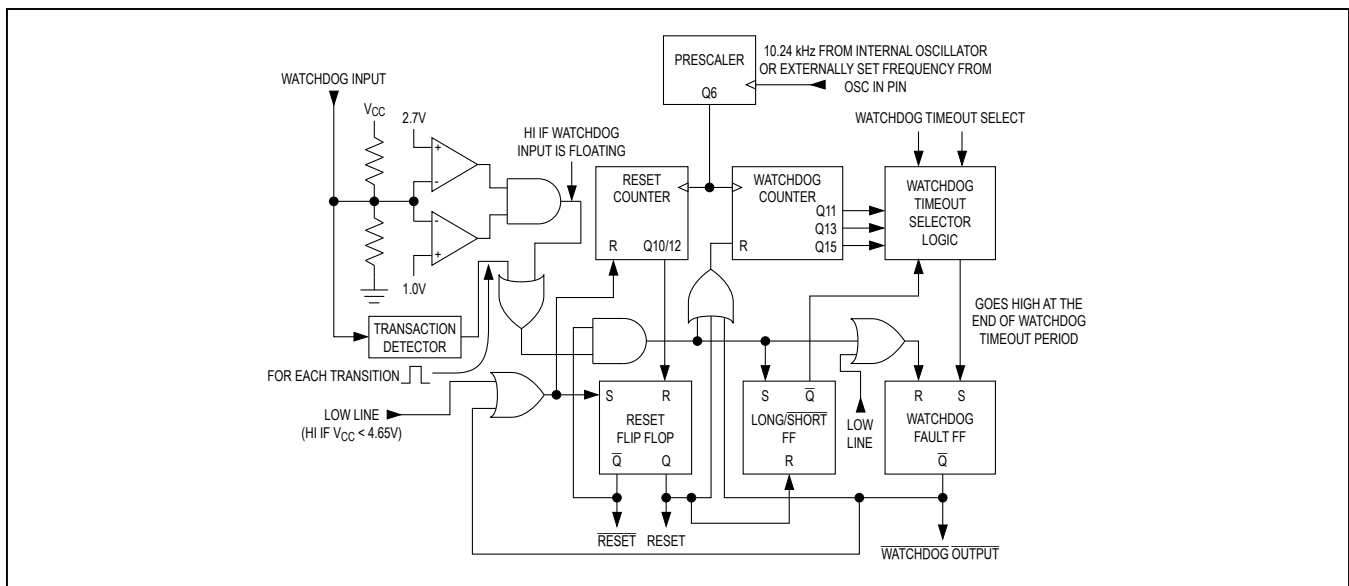


Figure 7. Watchdog Timer Block Diagram

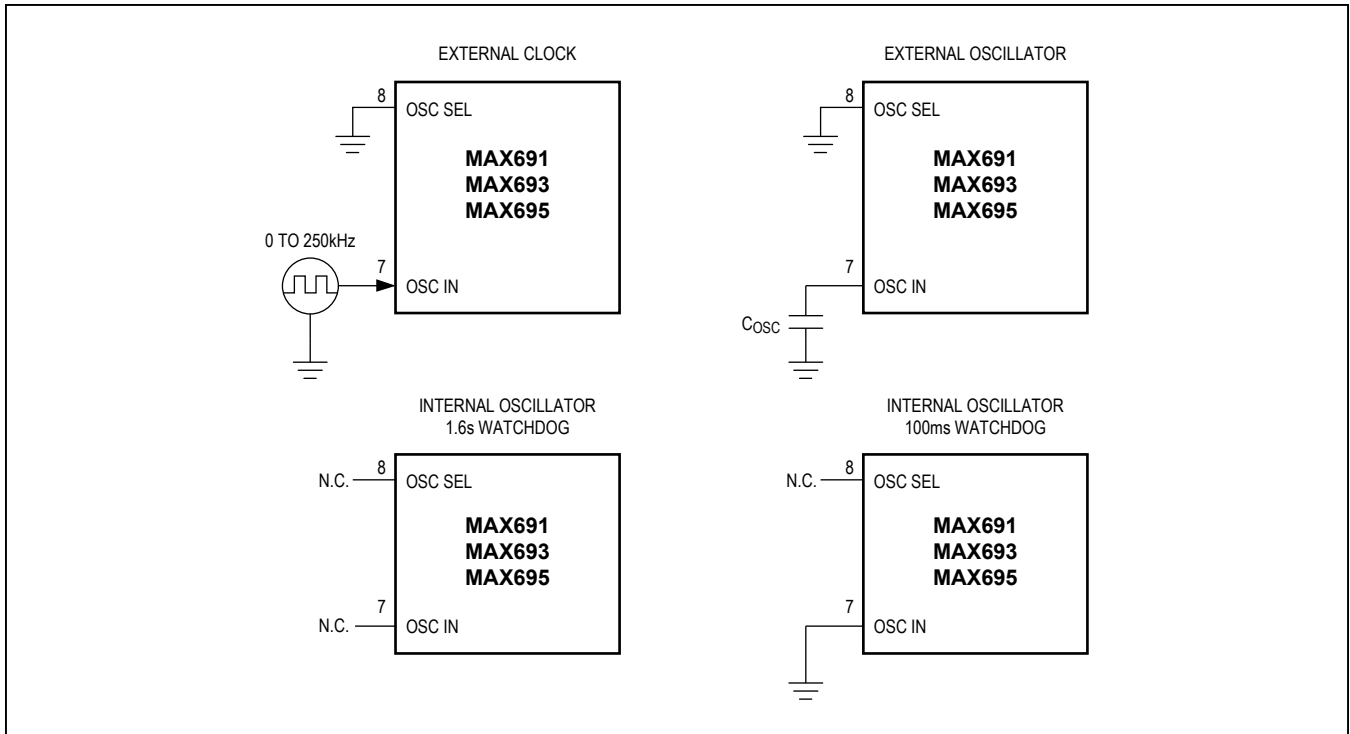


Figure 8. Oscillator Circuits

Table 1. MAX691/MAX693/MAX695 Reset Pulse Width and Watchdog Timeout Selections

| OSC SEL | OSC IN | WATCHDOG TIMEOUT PERIOD | | RESET TIMEOUT PERIOD | |
|----------|----------------------|-------------------------|------------------------|----------------------|----------------|
| | | NORMAL | IMMEDIATELY AFTER REST | MAX691/MAX693 | MAX695 |
| Low | External Clock Input | 1024 clks | 4096 clks | 512 clks | 2048 clks |
| Low | External Capacitor | 400ms/47pF x C | 1.6s/47pF x C | 200ms/47pF x C | 800ms/47pF x C |
| Floating | Low | 100ms | 1.6s | 50ms | 200ms |
| Floating | Floating | 1.6s | 1.6s | 50ms | 200ms |

Note 1: The MAX690/MAX692/MAX694 watchdog timeout period is fixed at 1.6s nominal, the MAX690/692 reset pulse width is fixed at 50ms nominal and the MAX694 is 200ms nominal.

Note 2: When the MAX691 OSC SEL pin is low, OSC IN can be driven by an external clock signal or an external capacitor can be connected between OSC IN and GND. The nominal internal oscillator frequency is 6.55kHz. The nominal oscillator frequency with capacitor is:

$$f_{OSC}(Hz) = \frac{120,000}{C(pF)}$$

Note 3: See *Electrical Characteristics* table for minimum and maximum timing values.

Application Hints

Other Uses of the Power-Fail Detector

In Figure 9, the power-fail detector is used to initiate a system reset when V_{CC} falls to 4.85V. Since the threshold of the power-fail detector is not as accurate as the onboard reset-voltage detector, a trimpot must be used to adjust the voltage detection threshold. Both the \overline{PFO} and \overline{RESET} outputs have high sink current capability and only 10 μ A of source current drive. This allows the two outputs to be connected directly to each other in a wired OR fashion.

The overvoltage detector circuit in Figure 10 resets the microprocessor whenever the nominal 5V V_{CC} is above 5.5V. The battery monitor circuit (Figure 11) shows the status of the memory backup battery. If desired, the \overline{CE} OUT can be used to apply a test load to the battery. Since \overline{CE} OUT is forced high during the battery backup mode, the test load will not be applied to the battery while it is in use even if the microprocessor is not powered.

Adding Hysteresis to the Power Fail Comparator

Since the power fail comparator circuit is noninverting, hysteresis can be added by connecting a resistor between the \overline{PFO} output and the PFI input as shown in Figure 12. When \overline{PFO} is low, resistor R3 sinks current from the summing junction at the PFI pin. When \overline{PFO} is high, the series combination of R3 and R4 source current into the PFI summing junction.

Alternate Watchdog Input Drive Circuits

The Watchdog feature can be enabled and disabled under program control by driving WDI with a three-state buffer (Figure 13). The drawback to this circuit is that a software fault may be erroneously three-state the buffer, thereby preventing the MAX690 from detecting that the microprocessor is no longer working. In most cases a better method is to extend the watchdog period rather than disabling the watchdog. See Figure 14. When the control input is high, the OSC SEL pin is low and the watchdog timeout is set by the external capacitor. A 0.01 μ F capacitor sets a watchdog timeout delay of 100s. When the control input is low the OSC SEL pin is high, selecting the internal oscillator. The 100ms or the 1.6s period is chosen, depending on which diode in Figure 14 is used.

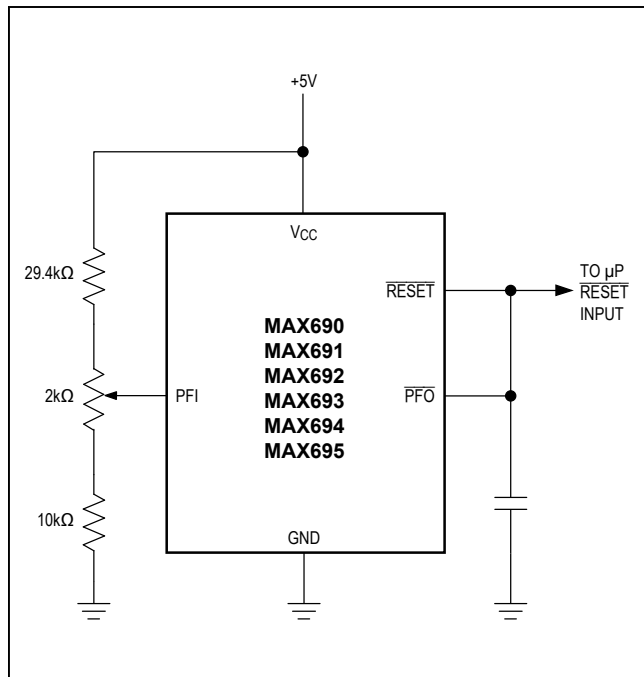


Figure 9. Externally Adjustable V_{CC} Reset Threshold

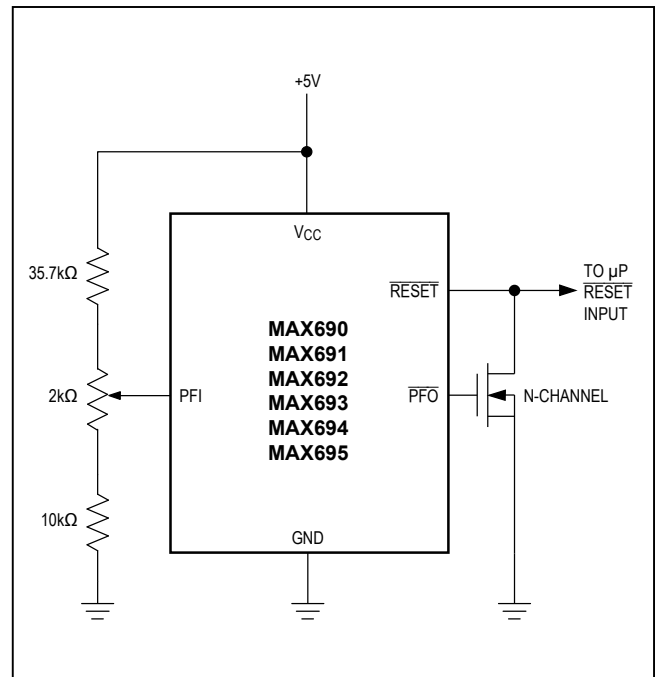


Figure 10. Reset on Overvoltage or Undervoltage

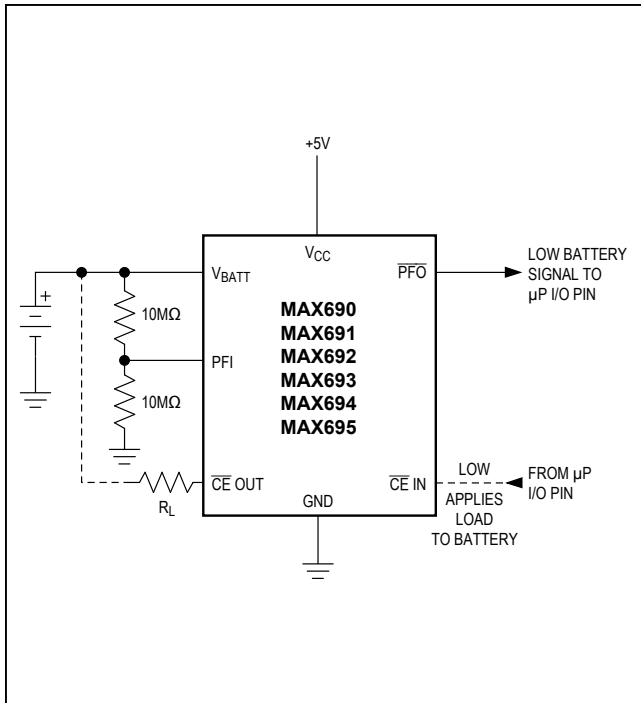


Figure 11. Backup VBattery Monitor with Optional Test Load

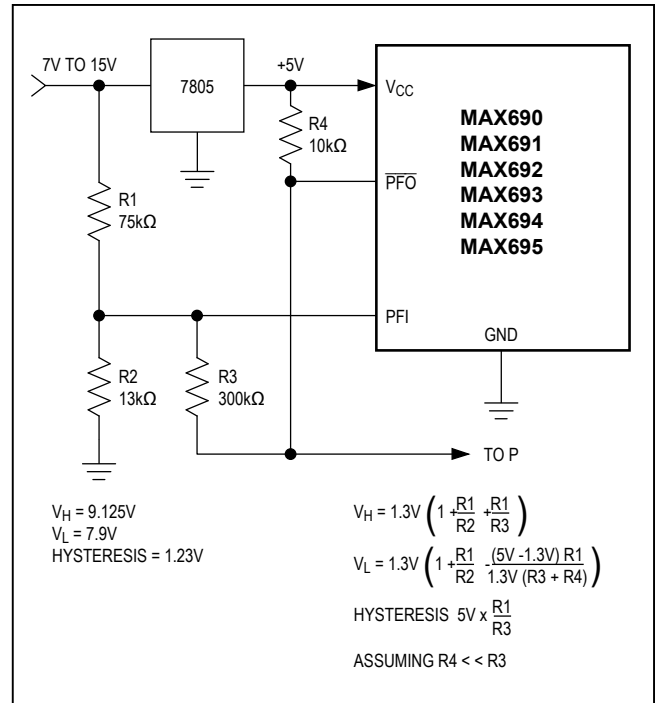


Figure 12. Adding Hysteresis to the Power-Fail Voltage Comparator

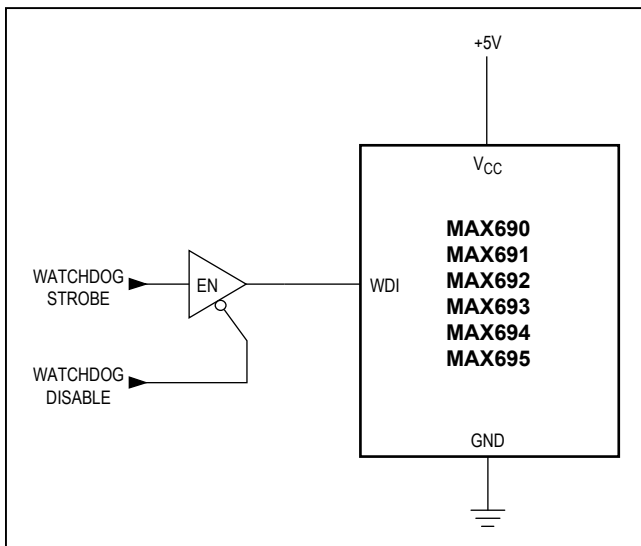


Figure 13. Disabling the Watchdog Under Program Control

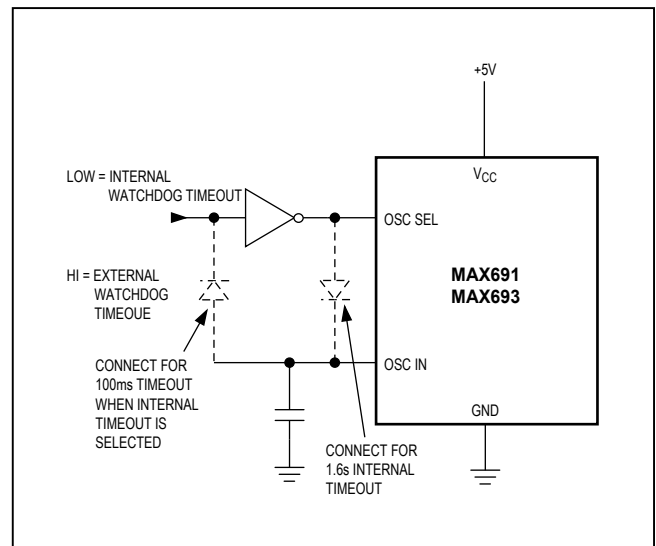


Figure 14. Selecting Internal or External Watchdog Timeout

Table 2. Input and Output Status In Battery Backup Mode

| | |
|------------------------|---|
| V_{BATT} , V_{OUT} | V_{BATT} is connected to V_{OUT} via internal MOSFET. |
| \overline{RESET} | Logic-low |
| RESET | Logic-high. The open circuit output voltage is equal to V_{OUT} . |
| $\overline{LOW\ LINE}$ | Logic-low |
| BATT ON | Logic-high |
| WDI | WDI is internally disconnected from its internal pullup and does not source or sink current as long as its input voltage is between GND and V_{OUT} . The input voltage does not affect the source current. |
| \overline{WDO} | Logic-high |
| PFI | The power-fail comparator is turned off and the power-fail input voltage has no effect on the power-fail output. |
| \overline{PFO} | Logic-low |
| $\overline{CE\ IN}$ | $\overline{CE\ IN}$ is internally disconnected from its internal pullup and does not source or sink current as long as its input voltage is between GND and V_{OUT} . The input voltage does not affect the source current. |
| $\overline{CE\ OUT}$ | Logic-high |
| OSC IN | OSC IN is ignored. |
| OSC SEL | OSC SEL is ignored. |
| V_{CC} | Approximately 12 μ A is drawn from the V_{BATT} input when V_{CC} is between $V_{BATT} + 100\text{mV}$ and $V_{BATT} - 700\text{mV}$. The supply current is 1 μ A maximum when V_{CC} is less than $V_{BATT} - 700\text{mV}$. |

Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
|-------------------|-----------------|---------------------|
| MAX690 CPA | 0°C to +70°C | 8 Lead Plastic DIP |
| MAX690C/D | 0°C to +70°C | Dice* |
| MAX690EPA | -40°C to +85°C | 8 Lead Plastic DIP |
| MAX690EJA | -40°C to +85°C | 8 Lead CERDIP |
| MAX690MJA | -55°C to +125°C | 8 Lead CERDIP |
| MAX691 CPE | 0°C to +70°C | 16 Lead Plastic DIP |
| MAX691CWE | 0°C to +70°C | 16 Lead Wide SO |
| MAX691C/D | 0°C to +70°C | Dice* |
| MAX691EPE | -40°C to +85°C | 16 Lead Plastic DIP |
| MAX691EWE | -40°C to +85°C | 16 Lead Wide SO |
| MAX691EJE | -40°C to +85°C | 16 Lead CERDIP |
| MAX691MJE | -55°C to +125°C | 16 Lead CERDIP |
| MAX692 C/D | 0°C to +70°C | Dice |
| MAX692CPA | 0°C to +70°C | 8 Lead Plastic DIP |
| MAX692EPA | -40°C to +85°C | 8 Lead Plastic DIP |
| MAX692EJA | -40°C to +85°C | 8 Lead CERDIP |
| MAX692MJA | -55°C to +125°C | 8 Lead CERDIP |
| MAX693 C/D | 0°C to +70°C | Dice |

| PART | TEMP RANGE | PIN-PACKAGE |
|-------------------|-----------------|---------------------|
| MAX693CPE | 0°C to +70°C | 16 Lead Plastic DIP |
| MAX693CWE | 0°C to +70°C | 16 Lead Wide SO |
| MAX693EPE | -40°C to +85°C | 16 Lead Plastic DIP |
| MAX693EJE | -40°C to +85°C | 16 Lead CERDIP |
| MAX693EWE | -40°C to +85°C | 16 Lead Wide SO |
| MAX693MJE | -55°C to +125°C | 16 Lead CERDIP |
| MAX694 C/D | 0°C to +70°C | Dice |
| MAX694CPA | 0°C to +70°C | 8 Lead Plastic DIP |
| MAX694EPA | -40°C to +85°C | 8 Lead Plastic DIP |
| MAX694EJA | -40°C to +85°C | 8 Lead CERDIP |
| MAX694MJA | -55°C to +125°C | 8 Lead CERDIP |
| MAX695 C/D | 0°C to +70°C | Dice |
| MAX695CPE | 0°C to +70°C | 16 Lead Plastic DIP |
| MAX695CWE | 0°C to +70°C | 16 Lead Wide SO |
| MAX695EPE | -40°C to +85°C | 16 Lead Plastic DIP |
| MAX695EJE | -40°C to +85°C | 16 Lead CERDIP |
| MAX695EWE | -40°C to +85°C | 16 Lead Wide SO |
| MAX695MJE | -55°C to +125°C | 16 Lead CERDIP |

*Contact factory for dice specifications.

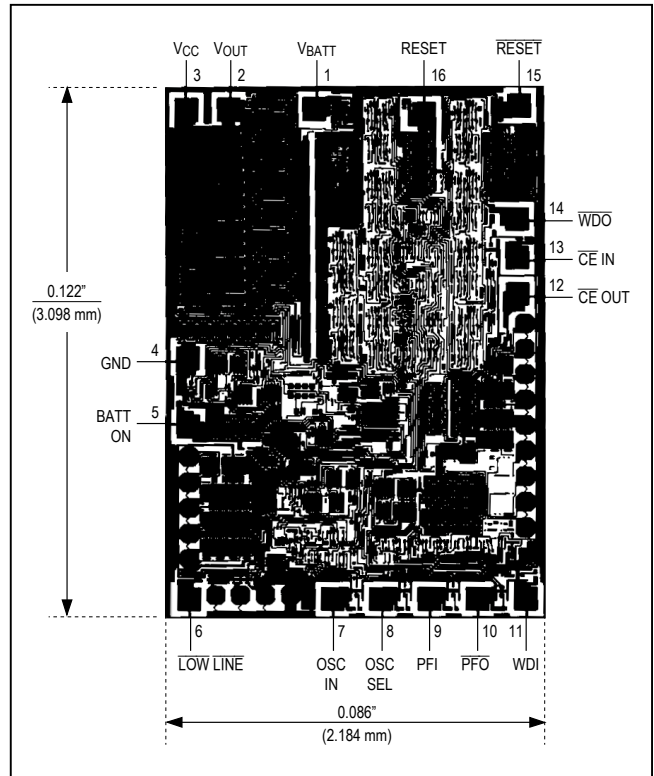
Devices in PDIP and SO packages are available in both leaded and lead-free packaging. Specify lead free by adding the + symbol at the end of the part number when ordering. Lead free not available for CERDIP package.

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a “+”, “#”, or “-” in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | OUTLINE NO. | LAND PATTERN NO. |
|--------------|--------------|-------------------------|------------------|
| 8 PDIP | P8-2 | 21-0043 | — |
| 8 CEDIP | J8-2 | 21-0045 | — |
| 16 PDIP | P16-1 | 21-0043 | — |
| 16 Wide SO | W16-1 | 21-0042 | — |
| 16 CERDIP | P16-1 | 21-0043 | — |

Chip Topography



Revision History

| REVISION NUMBER | REVISION DATE | DESCRIPTION | PAGES CHANGED |
|-----------------|---------------|--|---------------|
| 5 | 4/15 | Revised <i>Benefits and Features</i> section | 1 |

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim Integrated's website at www.maximintegrated.com.

Maxim Integrated cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim Integrated product. No circuit patent licenses are implied. Maxim Integrated reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.

Компания «Life Electronics» занимается поставками электронных компонентов импортного и отечественного производства от производителей и со складов крупных дистрибьюторов Европы, Америки и Азии.

С конца 2013 года компания активно расширяет линейку поставок компонентов по направлению коаксиальный кабель, кварцевые генераторы и конденсаторы (керамические, пленочные, электролитические), за счёт заключения дистрибьюторских договоров

Мы предлагаем:

- Конкурентоспособные цены и скидки постоянным клиентам.
- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
- Поставку компонентов, требующих военную и космическую приемку.
- Входной контроль качества.
- Наличие сертификата ISO.

В составе нашей компании организован Конструкторский отдел, призванный помогать разработчикам, и инженерам.

Конструкторский отдел помогает осуществить:

- Регистрацию проекта у производителя компонентов.
- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



Тел: +7 (812) 336 43 04 (многоканальный)

Email: org@lifeelectronics.ru