

# 2 ADC, 8 DAC, 96 kHz, 24-Bit $\Sigma$ - $\Delta$ Codec

AD1837

#### **FEATURES**

5 V Stereo Audio System with 3.3 V Tolerant **Digital Interface** 

Supports up to 96 kHz Sample Rates 192 kHz Sample Rate Available on One DAC Supports 16-/20-/24-Bit Word Lengths

Multibit  $\Sigma$ - $\Delta$  Modulators with Perfect Differential Linearity Restoration for

Reduced Idle Tones and Noise Floor Data Directed Scrambling DACs-Least

Sensitive to Jitter **Single-Ended Outputs** 

ADCs: -95 dB THD + N, 105 dB SNR and

**Dynamic Range** 

DACs: -92 dB THD + N, 108 dB SNR and

**Dynamic Range** 

On-Chip Volume Controls per Channel with 1024-Step Linear Scale

**DAC and ADC Software Controllable Clickless Mutes Digital De-emphasis Processing** 

Supports 256  $\times$   $f_S,\,512$   $\times$   $f_S,\,$  and 768  $\times$   $f_S$  Master Mode

Power-Down Mode Plus Soft Power-Down Mode Flexible Serial Data Port with Right-Justified, Left-Justified, I<sup>2</sup>S Compatible, and DSP Serial Port Modes TDM Interface Mode Supports 8 In/8 Out Using a Single SHARC® SPORT

52-Lead MQFP Plastic Package

**APPLICATIONS** 

**DVD Video and Audio Players Home Theater Systems Automotive Audio Systems Audio/Visual Receivers Digital Audio Effects Processors** 

### PRODUCT OVERVIEW

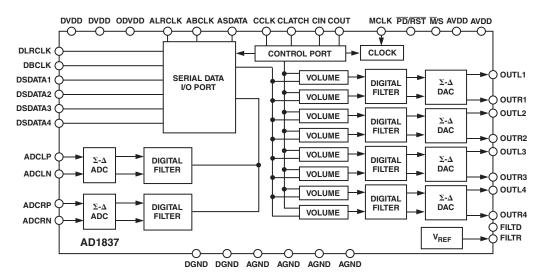
The AD1837 is a high performance single-chip codec featuring four stereo DACs and one stereo ADC. Each DAC comprises a high performance digital interpolation filter, a multibit  $\Sigma$ - $\Delta$ modulator featuring Analog Devices' patented technology, and a continuous-time voltage out analog section. Each DAC has independent volume control and clickless mute functions. The ADC comprises two 24-bit conversion channels with multibit  $\Sigma$ - $\Delta$  modulators and decimation filters.

The AD1837 also contains an on-chip reference with a nominal value of 2.25 V.

The AD1837 contains a flexible serial interface that allows for glueless connection to a variety of DSP chips, AES/EBU receivers, and sample rate converters. The AD1837 can be configured in left-justified, right-justified, I<sup>2</sup>S, or DSP compatible serial modes. Control of the AD1837 is achieved by means of an SPI compatible serial port. While the AD1837 can be operated from a single 5 V supply, it also features a separate supply pin for its digital interface that allows the device to be interfaced to other devices using 3.3 V power supplies.

The AD1837 is available in a 52-lead MQFP package and is specified for the industrial temperature range of -40°C to +85°C.

### FUNCTIONAL BLOCK DIAGRAM



REV. B

Information furnished by Analog Devices is believed to be accurate and reliable. However, no responsibility is assumed by Analog Devices for its use, nor for any infringements of patents or other rights of third parties that may result from its use. No license is granted by implication or otherwise under any patent or patent rights of Analog Devices. Trademarks and registered trademarks are the property of their respective companies.

One Technology Way, P.O. Box 9106, Norwood, MA 02062-9106, U.S.A. Tel: 781/329-4700 www.analog.com Fax: 781/326-8703 © 2003 Analog Devices, Inc. All rights reserved.

## AD1837-SPECIFICATIONS

## **TEST CONDITIONS**

 $\begin{array}{ll} \text{Supply Voltages (AV}_{DD},\,DV_{DD}) & 5.0\;V\\ \text{Ambient Temperature} & 25^{\circ}C \end{array}$ 

 $\begin{array}{lll} \text{Input Clock} & 12.288 \text{ MHz, } (256 \times f_S \text{ Mode}) \\ \text{ADC Input Signal} & 1.0078125 \text{ kHz, } -1 \text{ dBFS (Full Scale}) \\ \text{DAC Input Signal} & 1.0078125 \text{ kHz, } 0 \text{ dBFS (Full Scale}) \\ \end{array}$ 

Input Sample Rate (f<sub>S</sub>) 48 kHz

Measurement Bandwidth 20 Hz to 20 kHz

Word Width24 BitsLoad Capacitance100 pFLoad Impedance47 kΩ

Performance of all channels is identical (exclusive of the Interchannel Gain Mismatch and Interchannel Phase Deviation specifications).

Parameter	Min	Typ	Max	Unit
ANALOG-TO-DIGITAL CONVERTERS				
ADC Resolution		24		Bits
Dynamic Range (20 Hz to 20 kHz, -60 dB Input)				
No Filter		103		dB
A-Weighted	100	105		dB
Total Harmonic Distortion + Noise (THD+N)		-95	-88.5	dB
Interchannel Isolation		100		dB
Interchannel Gain Mismatch		0.025		dB
Analog Inputs				
Differential Input Range (± Full Scale)	-2.828		+2.828	V
Common-Mode Input Voltage		2.25		V
Input Impedance		4		kΩ
Input Capacitance		15		pF
$ m V_{REF}$		2.25		V
DC Accuracy				
Gain Error		±5		%
Gain Drift		35		ppm/°C
DIGITAL-TO-ANALOG CONVERTERS				
DAC Resolution				
Dynamic Range (20 Hz to 20 kHz, –60 dBFS Input)				
No Filter	103	105		dB
With A-Weighted Filter	105	108		dB
Total Harmonic Distortion + Noise	105	-92		dB
Interchannel Isolation		100		dB
DC Accuracy		100		ub
Gain Error		±4		%
Interchannel Gain Mismatch		0.025		dB
Gain Drift		200		ppm/°C
Interchannel Phase Deviation		±0.1		Degrees
Volume Control Step Size (1023 Linear Steps)		0.098		%
Volume Control Range (Max Attenuation)		60		dB
Mute Attenuation		-100		dB
De-emphasis Gain Error				dB
		±0.1		
Full-Scale Output Voltage at Each Pin (Single-Ended) Output Resistance at Each Pin		1.0 (2.8)		V rms (V p-p) Ω
Common-Mode Output Voltage		180 2.25		V
		2.25		V
ADC DECIMATION FILTER, 48 kHz*				
Pass Band		21.77		kHz
Pass-Band Ripple		$\pm 0.01$		dB
Stop Band		26.23		kHz
Stop-Band Attenuation		120		dB
Group Delay		910		μs

-2- REV. B

Parameter	Min	Тур	Max	Unit
ADC DECIMATION FILTER, 96 kHz*				
Pass Band		43.54		kHz
Pass-Band Ripple		$\pm 0.01$		dB
Stop Band		52.46		kHz
Stop-Band Attenuation		120		dB
Group Delay		460		μs
DAC INTERPOLATION FILTER, 48 kHz*				
Pass Band			21.77	kHz
Pass-Band Ripple		$\pm 0.06$		dB
Stop Band	28.0			kHz
Stop-Band Attenuation	55			dB
Group Delay		340		μs
DAC INTERPOLATION FILTER, 96 kHz*				
Pass Band			43.54	kHz
Pass-Band Ripple		$\pm 0.06$		dB
Stop Band	52.0			kHz
Stop-Band Attenuation	55			dB
Group Delay		160		μs
DAC INTERPOLATION FILTER, 192 kHz*				
Pass Band			81.2	kHz
Pass-Band Ripple		$\pm 0.06$		dB
Stop Band	97.0			kHz
Stop-Band Attenuation	80			dB
Group Delay		110		μs
DIGITAL I/O				
Input Voltage High	2.4			V
Input Voltage Low			0.8	V
Output Voltage High		$\mathrm{ODV}_{\mathrm{DD}}$ – 0.4		V
Output Voltage Low			0.4	V
Leakage Current			±10	μА
POWER SUPPLIES				
Supply Voltage (AV <sub>DD</sub> and DV <sub>DD</sub> )	4.5	5.0	5.5	V
Supply Voltage (OV <sub>DD</sub> )	3.0		$\mathrm{DV}_{\mathrm{DD}}$	V
Supply Current I <sub>ANALOG</sub>		84	95	mA
Supply Current I <sub>ANALOG</sub> , Power-Down		55	67	mA
Supply Current I <sub>DIGITAL</sub>		64	74	mA
Supply Current I <sub>DIGITAL</sub> , Power-Down		1	4.5	mA
Dissipation				
Operation, Both Supplies		740		mW
Operation, Analog Supply		420		mW
Operation, Digital Supply		320		mW
Power-Down, Both Supplies		280		mW
Power Supply Rejection Ratio				
1 kHz, 300 mV p-p Signal at Analog Supply Pins		-70		dB
20 kHz, 300 mV p-p Signal at Analog Supply Pins		-75		dB

<sup>\*</sup>Guaranteed by design.

REV. B -3-

Specifications subject to change without notice.

## TIMING SPECIFICATIONS

Parameter		Min	Max	Unit	Comments
MASTER CLO	OCK AND RESET				
t <sub>MH</sub>	MCLK High	15		ns	
t <sub>ML</sub>	MCLK Low	15		ns	
$t_{PDR}$	$\overline{\text{PD}}/\overline{\text{RST}}$ Low	20		ns	
SPI PORT					
t <sub>CCH</sub>	CCLK High	40		ns	
t <sub>CCL</sub>	CCLK Low	40		ns	
t <sub>CCP</sub>	CCLK Period	80		ns	
t <sub>CDS</sub>	CDATA Setup	10		ns	To CCLK Rising
t <sub>CDH</sub>	CDATA Hold	10		ns	From CCLK Rising
t <sub>CLS</sub>	CLATCH Setup	10		ns	To CCLK Rising
t <sub>CLH</sub>	CLATCH Hold	10		ns	From CCLK Rising
	COUT Enable	10	15	ns	From CLATCH Falling
toop	COUT Delay		20	ns	From CCLK Falling
t <sub>COD</sub>	COUT Three-State		25		From CLATCH Rising
t <sub>COTS</sub>			20	ns	From CLATCH Rising
DAC SERIAL					
Normal Mod	•				
$t_{ m DBH}$	DBCLK High	60		ns	
$t_{ m DBL}$	DBCLK Low	60		ns	
$ m f_{DB}$	DBCLK Frequency	$64  imes  ext{f}_{ ext{S}}$			
$t_{ m DLS}$	DLRCLK Setup	10		ns	To DBCLK Rising
$t_{ m DLH}$	DLRCLK Hold	10		ns	From DBCLK Rising
$t_{ m DDS}$	DSDATA Setup	10		ns	To DBCLK Rising
t <sub>DDH</sub> Packed 256	DSDATA Hold Modes (Slave)	10		ns	From DBCLK Rising
$t_{ m DBH}$	DBCLK High	15		ns	
$t_{ m DBL}$	DBCLK Low	15		ns	
$ m f_{DB}$	DBCLK Frequency	$256 \times f_{\rm S}$		===	
t <sub>DLS</sub>	DLRCLK Setup	10		ns	To DBCLK Rising
t <sub>DLH</sub>	DLRCLK Hold	5		ns	From DBCLK Rising
t <sub>DDS</sub>	DSDATA Setup	10		ns	To DBCLK Rising
	DSDATA Hold	10		ns	From DBCLK Rising
t <sub>DDH</sub>		10		113	Trom DBCLK Rising
ADC SERIAL					
Normal Mod	•		25		Enom MCI V Dising Edge
$t_{ m ABD}$	ABCLK Delay		25 5	ns	From MCLK Rising Edge
$t_{ALD}$	ALRCLK Delay Low		5	ns	From ABCLK Falling Edge
t <sub>ABDD</sub> Normal Mod	ASDATA Delay		10	ns	From ABCLK Falling Edge
	ABCLK High	60			
t <sub>ABH</sub>	- C	60		ns	
$t_{ m ABL}$	ABCLK Low			ns	
$f_{AB}$	ABCLK Frequency	$64 \times f_S$			To ADCI V Disire
$t_{ALS}$	ALRCLK Setup	5		ns	To ABCLK Rising
t <sub>ALH</sub>	ALRCLK Hold	15		ns	From ABCLK Rising
	Mode (Master)		20		Enter MOLK D E.
$t_{PABD}$	ABCLK Delay		20	ns	From MCLK Rising Edge
$t_{\mathrm{PALD}}$	LRCLK Delay		5	ns	From ABCLK Falling Edge
$t_{\mathrm{PABDD}}$	ASDATA Delay		10	ns	From ABCLK Falling Edge

-4- REV. B

Parameter		Min	Max	Unit	Comments
TDM256 MODE	(Master)				
$t_{ m TBD}$	BCLK Delay		20	ns	From MCLK Rising
$t_{FSD}$	FSTDM Delay		5	ns	From BCLK Rising
$t_{TABDD}$	ASDATA Delay		10	ns	From BCLK Rising
$t_{TDDS}$	DSDATA1 Setup	15		ns	To BCLK Falling
$t_{ m TDDH}$	DSDATA1 Hold	15		ns	From BCLK Falling
TDM256 MODE	(Slave)				
$ m f_{AB}$	BCLK Frequency	$256 \times f_S$			
$t_{TBCH}$	BCLK High	15		ns	
$t_{\mathrm{TBCL}}$	BCLK Low	15		ns	
$t_{TFS}$	FSTDM Setup	10		ns	To BCLK Falling
$t_{TFH}$	FSTDM Hold	10		ns	From BCLK Falling
$t_{\mathrm{TBDD}}$	ASDATA Delay		10	ns	From BCLK Rising
$t_{TDDS}$	DSDATA1 Setup	15		ns	To BCLK Falling
t <sub>TDDH</sub>	DSDATA1 Hold	15		ns	From BCLK Falling
AUXILIARY INT	TERFACE				
$t_{AXDS}$	AAUXDATA Setup	10		ns	To AUXBCLK Rising
$t_{AXDH}$	AAUXDATA Hold	10		ns	From AUXBCLK Rising
$\mathrm{f}_{\mathrm{ABP}}$	AUXBCLK Frequency	$64  imes  ext{f}_{ ext{S}}$			
Slave Mode					
$t_{AXBH}$	AUXBCLK High	15		ns	
$t_{\mathrm{AXBL}}$	AUXBCLK Low	15		ns	
$t_{AXLS}$	AUXLRCLK Setup	10		ns	To AUXBCLK Rising
$t_{AXLH}$	AUXLRCLK Hold	10		ns	From AUXBCLK Rising
Master Mode					
$t_{AUXBCLK}$	AUXLRCLK Delay	15		ns	From MCLK Rising
$t_{AUXLRCLK}$	AUXLRCLK Delay	15		ns	From AUXBCLK Falling

Specifications subject to change without notice.

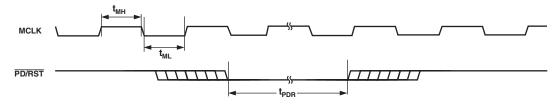


Figure 1. MCLK and PD/RST Timing

REV. B -5-

### **ABSOLUTE MAXIMUM RATINGS\***

 $(T_A = 25^{\circ}C, \text{ unless otherwise noted.})$ 

$AV_{DD}$ , $DV_{DD}$ , $ODV_{DD}$ to AGND, DGND $-0.3~V$ to $+6.0~V$
AGND to DGND0.3 V to +0.3 V
Digital I/O Voltage to DGND $\dots$ -0.3 V to ODV <sub>DD</sub> + 0.3 V
Analog I/O Voltage to AGND $-0.3 \text{ V}$ to AV <sub>DD</sub> + 0.3 V
Operating Temperature Range
Industrial (A Version)40°C to +85°C

<sup>\*</sup>Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those listed in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## TEMPERATURE RANGE

Parameter	Min	Typ	Max	Unit
Specifications		25		°C
Guaranteed				
Functionality	-40		+85	°C
Guaranteed				
Storage	-65		+150	°C

### **ORDERING GUIDE**

Model	Temperature	Package	Package
	Range	Description	Option
AD1837AS AD1837AS-REEL EVAL-AD1837EB	-40°C to +85°C -40°C to +85°C	52-Lead MQFP 52-Lead MQFP Evaluation Board	S-52A S-52A

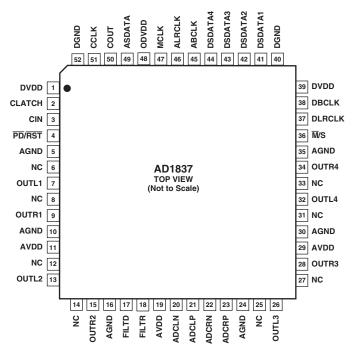
#### CAUTION \_

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although the AD1837 features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.



–6– REV. B

## PIN CONFIGURATION



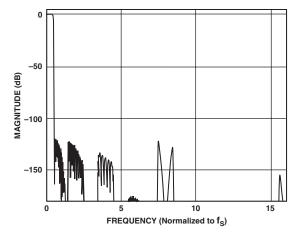
NC = NO CONNECT

## PIN FUNCTION DESCRIPTIONS

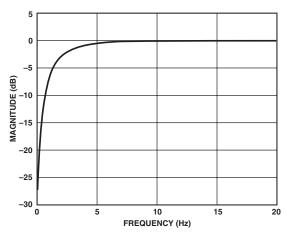
		Input/		
Pin No.	Mnemonic	Output	Description	
1, 39	DVDD		Digital Power Supply. Connect to digital 5 V supply.	
2	CLATCH	I	Latch Input for Control Data.	
3	CIN	I	Serial Control Input.	
4	$\overline{\text{PD}}/\overline{\text{RST}}$	I	Power-Down/Reset.	
5, 10, 16, 24, 30, 35	AGND		Analog Ground.	
6, 8, 12, 14, 25, 27, 31, 33	NC		Not Connected.	
7, 13, 26, 32	OUTLx	0	DACx Left Channel Output.	
9, 15, 28, 34	OUTRx	О	DACx Right Channel Output.	
11, 19, 29	AVDD		Analog Power Supply. Connect to analog 5 V supply.	
17	FILTD		Filter Capacitor Connection. Recommended 10 μF/100 nF.	
18	FILTR		Reference Filter Capacitor Connection. Recommended 10 μF/100 nF.	
20	ADCLN	I	ADC Left Channel Negative Input.	
21	ADCLP	I	ADC Left Channel Positive Input.	
22	ADCRN	I	ADC Right Channel Negative Input.	
23	ADCRP	I	ADC Right Channel Positive Input.	
36	$\overline{\mathrm{M}}/\mathrm{S}$	I	ADC Master/Slave Select.	
37	DLRCLK	I/O	DAC LR Clock.	
38	DBCLK	I/O	DAC Bit Clock.	
40, 52	DGND		Digital Ground.	
41–44	DSDATAx	I	DACx Input Data (Left and Right Channels).	
45	ABCLK	I/O	ADC Bit Clock.	
46	ALRCLK	I/O	ADC LR Clock.	
47	MCLK	I	Master Clock Input.	
48	ODVDD		Digital Output Driver Power Supply.	
49	ASDATA	0	ADC Serial Data Output.	
50	COUT	0	Output for Control Data.	
51	CCLK	I	Control Clock Input for Control Data.	

REV. B -7-

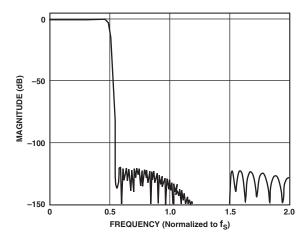
## **AD1837**—Typical Performance Characteristics



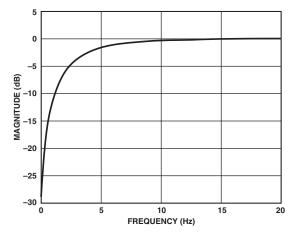
TPC 1. ADC Composite Filter Response



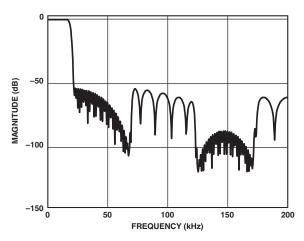
TPC 2. ADC High-Pass Filter Response,  $f_S = 48 \text{ kHz}$ 



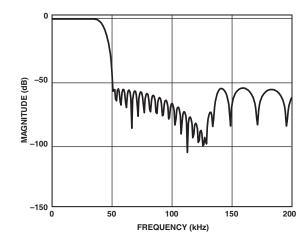
TPC 3. ADC Composite Filter Response (Pass-Band Section)



TPC 4. ADC High-Pass Filter Response,  $f_S = 96 \text{ kHz}$ 

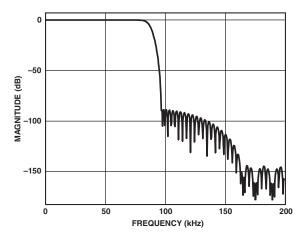


TPC 5. DAC Composite Filter Response,  $f_S = 48 \text{ kHz}$ 

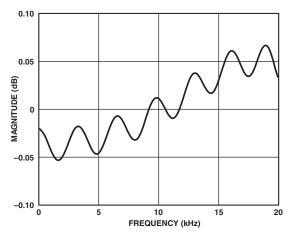


TPC 6. DAC Composite Filter Response,  $f_S = 96 \text{ kHz}$ 

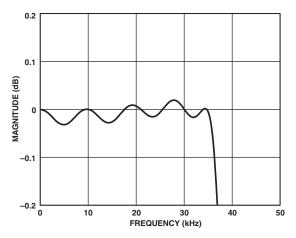
-8- REV. B



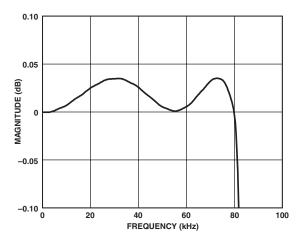
TPC 7. DAC Composite Filter Response,  $f_S = 192 \text{ kHz}$ 



TPC 8. DAC Composite Filter Response,  $f_S = 48 \text{ kHz}$  (Pass-Band Section)



TPC 9. DAC Composite Filter Response,  $f_S = 96 \text{ kHz}$  (Pass-Band Section)



TPC 10. DAC Composite Filter Response,  $f_S = 192 \text{ kHz}$  (Pass-Band Section)

REV. B -9-

#### **TERMINOLOGY**

## **Dynamic Range**

The ratio of a full-scale input signal to the integrated input noise in the pass band (20 Hz to 20 kHz), expressed in decibels (dB). Dynamic range is measured with a -60 dB input signal and is equal to (S/[THD + N]) + 60 dB. Note that spurious harmonics are below the noise with a -60 dB input, so the noise level establishes the dynamic range. The dynamic range is specified with and without an A-Weight filter applied.

## Signal-to-(Total Harmonic Distortion + Noise) [S/(THD + N)]

The ratio of the root-mean-square (rms) value of the fundamental input signal to the rms sum of all other spectral components in the pass band, expressed in decibels (dB).

#### **Pass Band**

The region of the frequency spectrum unaffected by the attenuation of the digital decimator's filter.

## Pass-Band Ripple

The peak-to-peak variation in amplitude response from equalamplitude input signal frequencies within the pass band, expressed in decibels.

## Stop Band

The region of the frequency spectrum attenuated by the digital decimator's filter to the degree specified by stop-band attenuation.

#### **Gain Error**

With a near full-scale input, the ratio of actual output to expected output, expressed as a percentage.

## **Interchannel Gain Mismatch**

With identical near full-scale inputs, the ratio of outputs of the two stereo channels, expressed in decibels.

### Gain Drift

Change in response to a near full-scale input with a change in temperature, expressed as parts-per-million (ppm) per °C.

#### Crosstalk (EIAJ Method)

Ratio of response on one channel with a grounded input to a full-scale 1 kHz sine wave input on the other channel, expressed in decibels.

### **Power Supply Rejection**

With no analog input, signal present at the output when a 300 mV p-p signal is applied to power supply pins, expressed in decibels of full scale.

### Group Delay

Intuitively, the time interval required for an input pulse to appear at the converter's output, expressed in milliseconds (ms). More precisely, the derivative of radian phase with respect to radian frequency at a given frequency.

### **Group Delay Variation**

The difference in group delays at different input frequencies. Specified as the difference between the largest and the smallest group delays in the pass band, expressed in microseconds (µs).

#### Acronyms

ADC-Analog-to-Digital Converter

DAC—Digital-to-Analog Converter

DSP—Digital Signal Processor

IMCLK—Internal Master Clock signal used to clock the ADC and DAC engines

MCLK-External Master Clock signal applied to the AD1837

–10– REV. B

## FUNCTIONAL OVERVIEW ADCs

There are two ADC channels in the AD1837, configured as a stereo pair. Each ADC has fully differential inputs. The ADC section can operate at a sample rate of up to 96 kHz. The ADCs include on-board digital decimation filters with 120 dB stop-band attenuation and linear phase response, operating at an oversampling ratio of 128 (for 48 kHz operation) or 64 (for 96 kHz operation).

ADC peak level information for each ADC may be read from the ADC Peak 0 and ADC Peak 1 registers. The data is supplied as a 6-bit word with a maximum range of 0 dB to -63 dB and a resolution of 1 dB. The registers will hold peak information until read; after reading, the registers are reset so that new peak information can be acquired. Refer to the register description for details on the format. The two ADC channels have a common serial bit clock and a left-right framing clock. The clock signals are all synchronous with the sample rate.

The ADC digital pins, ABCLK and ALRCLK, can be set to operate as inputs or outputs by connecting the  $\overline{M}/S$  pin to ODVDD or DGND, respectively. When the pins are set as outputs, the AD1837 will generate the timing signals. When the pins are set as inputs, the timing must be generated by the external audio controller.

#### **DACs**

The AD1837 has eight DAC channels arranged as four independent stereo pairs, with eight single-ended analog outputs for improved noise and distortion performance. Each channel has its own independently programmable attenuator, adjustable in 1024 linear steps. Digital inputs are supplied through four serial data input pins (one for each stereo pair) and a common frame (DLRCLK) and bit (DBLCK) clock. Alternatively, one of the "packed data" modes may be used to access all eight channels on a single TDM data pin. A stereo replicate feature is included where the DAC data sent to the first DAC pair is also sent to the other DACs in the part. The AD1837 can accept DAC data at a sample rate of 192 kHz on DAC 1 only. The stereo replicate feature can then be used to copy the audio data to the other DACs.

Each of the output pins sits at a dc level of  $V_{REF}$  and swings  $\pm 1.4~V$  for a 0 dB digital input signal. A single op amp third

order external low-pass filter is recommended to remove high frequency noise present on the output pins. Note that the use of op amps with low slew rate or low bandwidth may cause high frequency noise and tones to fold down into the audio band; care should be exercised in selecting these components.

The FILTD pin should be connected to an external grounded capacitor. This pin is used to reduce the noise of the internal DAC bias circuitry, thereby reducing the DAC output noise. In some cases, this capacitor may be eliminated with little effect on performance.

### DAC and ADC Coding

The DAC and ADC output data stream is in a twos complement encoded format. The word width can be selected from 16-bit, 20-bit, or 24-bit. The coding scheme is detailed in Table I.

Table I. Coding Scheme

Code	Level
011111111	+FS
000000000	0 (Ref Level)
100000000	-FS

## Clock Signals

The DAC and ADC engines in the AD1837 are designed to operate from a 24.576 MHz internal master clock (IMCLK). This clock is used to generate 48 kHz and 96 kHz sampling on the ADC and 48 kHz, 96 kHz, and 192 kHz on the DAC, although the 192 kHz option is available only on one DAC pair. The stereo replicate feature can be used to copy this DAC data to the other DACs if required.

To facilitate the use of different MCLK values, the AD1837 provides a clock scaling feature. The MCLK scaler can be programmed via the SPI port to scale the MCLK by a factor of 1 (pass through), 2 (doubling), or 2/3. The default setting of the MCLK scaler is 2, which will generate 48 kHz sampling from a 12.288 MHz MCLK. Additional sample rates can be achieved by changing the MCLK value. For example, the CD standard sampling frequency of 44.1 kHz can be achieved using an 11.2896 kHz MCLK. Figure 2 shows the internal configuration of the clock scaler and converter engines.

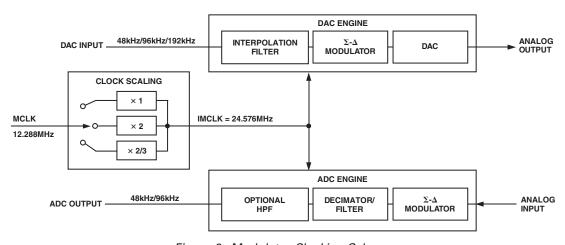


Figure 2. Modulator Clocking Scheme

REV. B –11–

To maintain the highest performance possible, it is recommended that the clock jitter of the master clock signal be limited to less than 300 ps rms, measured using the edge-to-edge technique. Even at these levels, extra noise or tones may appear in the DAC outputs if the jitter spectrum contains large spectral peaks. It is highly recommended that the master clock be generated by an independent crystal oscillator. In addition, it is especially important that the clock signal should not be passed through an FPGA or other large digital chip before being applied to the AD1837. In most cases, this will induce clock jitter due to the fact that the clock signal is sharing common power and ground connections with other unrelated digital output signals.

### **RESET and Power-Down**

PD/RST will power down the chip and set the control registers to their default settings. After PD/RST is deasserted, an initialization routine will run inside the AD1837 to clear all memories to zero. This initialization lasts for approximately 20 LRCLK intervals. During this time, it is recommended that no SPI writes occur.

## Power Supply and Voltage Reference

The AD1837 is designed for 5 V supplies. Separate power supply pins are provided for the analog and digital sections. These pins should be bypassed with 100 nF ceramic chip capacitors, as close to the pins as possible, to minimize noise pickup. A bulk aluminum electrolytic capacitor of at least 22  $\mu$ F should also be provided on the same PC board as the codec. For critical applications, improved performance will be obtained with separate supplies for the analog and digital sections. If this is not possible, it is recommended that the analog and digital supplies be isolated by means of two ferrite beads in series with the bypass capacitor of each supply. It is important that the analog supply be as clean as possible.

The internal voltage reference is brought out on the FILTR pin and should be bypassed as close as possible to the chip, with a parallel combination of 10  $\mu$ F and 100 nF. The reference voltage may be used to bias external op amps to the common-mode voltage of the analog input and output signal pins. The current drawn from the FILTR pin should be limited to less than 5  $\mu$ A.

#### **Serial Control Port**

The AD1837 has an SPI® compatible control port to permit programming the internal control registers for the ADCs and DACs and for reading the ADC signal levels from the internal peak detectors. The SPI control port is a 4-wire serial control port. The format is similar to the Motorola SPI format except the input data-word is 16 bits wide. The maximum serial bit clock frequency is 12.5 MHz and may be completely asynchronous to the sample rate of the ADCs and DACs. Figure 3 shows the format of the SPI signal.

### Serial Data Ports—Data Format

The ADC serial data output mode defaults to the popular I<sup>2</sup>S format, where the data is delayed by 1 BCLK interval from the edge of the LRCLK. By changing Bits 6 to 8 in ADC Control Register 2, the serial mode can be changed to right-justified (RJ), left-justified DSP (DSP), or left-justified (LJ). In the RJ mode, it is necessary to set Bits 4 and 5 to define the width of the data-word.

The DAC serial data input mode defaults to I<sup>2</sup>S. By changing Bits 5, 6, and 7 in DAC Control Register 1, the mode can be changed to RJ, DSP, LJ, Packed Mode 1, or Packed Mode 2. The word width defaults to 24 bits but can be changed by reprogramming Bits 3 and 4 in DAC Control Register 1.

#### **Packed Modes**

The AD1837 has a packed mode that allows a DSP or other controller to write to all DACs and read all ADCs using one input data pin and one output data pin. Packed Mode 256 refers to the number of BCLKs in each frame. The LRCLK is low while data from a left channel DAC or ADC is on the data pin and high while data from a right channel DAC or ADC is on the data pin. DAC data is applied on the DSDATA1 pin and ADC data is available on the ASDATA pin. Figures 7 to 10 show the timing for the packed mode. Packed mode is available only for 48 kHz (based on MCLK = 12.288 MHz) and when the ADC is set as a master  $(\overline{M}/S = 0)$ .

## Auxiliary (TDM) Mode

A special auxiliary mode is provided to allow three external stereo ADCs to be interfaced to the AD1837 to provide 8-in/8-out operation. In addition, this mode supports glueless interface to a single SHARC DSP serial port, allowing a SHARC DSP to access all eight channels of analog I/O. In this special mode, many pins are redefined; see Table II for a list of redefined pins.

The auxiliary and the TDM interfaces are independently configurable to operate as masters or slaves. When the auxiliary interface is set as a master, by programming the Aux Mode bit in ADC Control Register 2, AUXLRCLK and AUXBCLK are generated by the AD1837. When the auxiliary interface is set as a slave, the AUXLRCLK and AUXBCLK need to be generated by an external ADC as shown in Figure 13.

The TDM interface can be set to operate as a master or slave by connecting the  $\overline{\text{M}}/\text{S}$  pin to DGND or ODVDD, respectively. In master mode, the FSTDM and BCLK signals are outputs and are generated by the AD1837. In slave mode, the FSTDM and BCLK are inputs and should be generated by the SHARC. Slave mode operation is available for 48 kHz and 96 kHz operation (based on a 12.288 MHz or 24.576 MHz MCLK), and master mode operation is available for 48 kHz only.

–12– REV. B

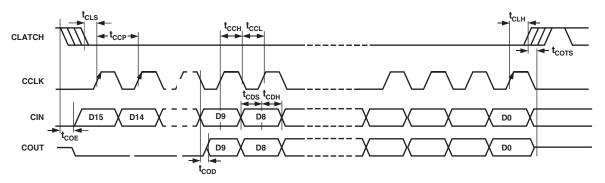
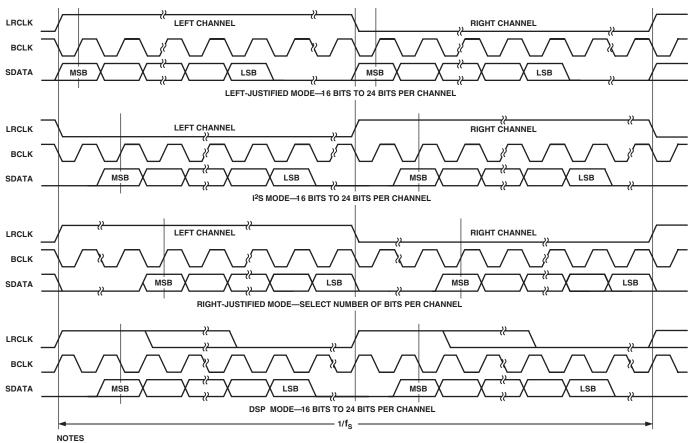


Figure 3. Format of SPI Timing



1. DSP MODE DOES NOT IDENTIFY CHANNEL. 2. LRCLK NORMALLY OPERATES AT  $f_S$  EXCEPT FOR DSP MODE WHICH IS 2  $\times$   $f_S$ . 3. BCLK FREQUENCY IS NORMALLY 64  $\times$  LRCLK BUT MAY BE OPERATED IN BURST MODE.

Figure 4. Stereo Serial Modes

REV. B -13-

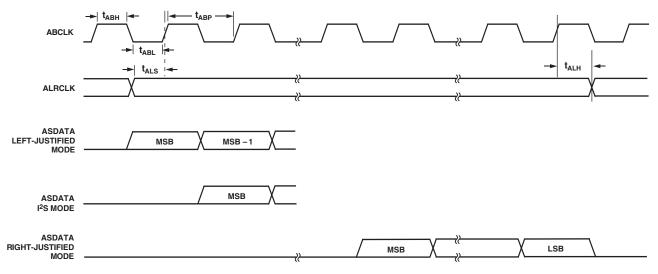


Figure 5. ADC Serial Mode Timing

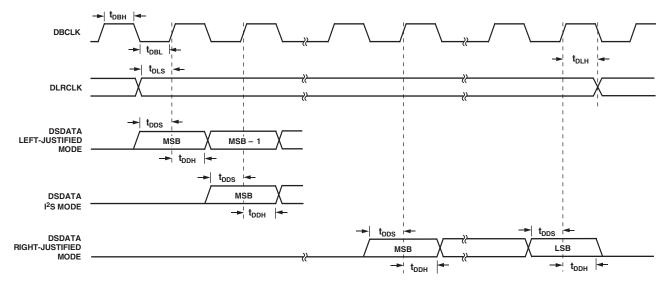


Figure 6. DAC Serial Mode Timing

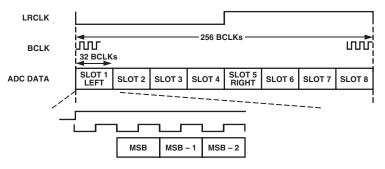


Figure 7. ADC Packed Mode 256

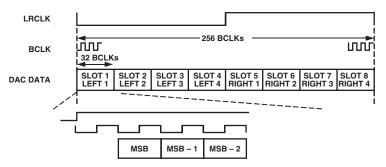


Figure 8. DAC Packed Mode 256

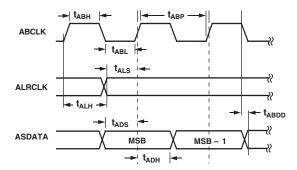


Figure 9. ADC Packed Mode Timing

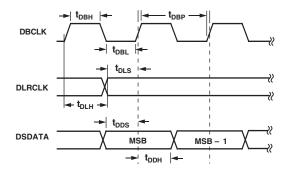
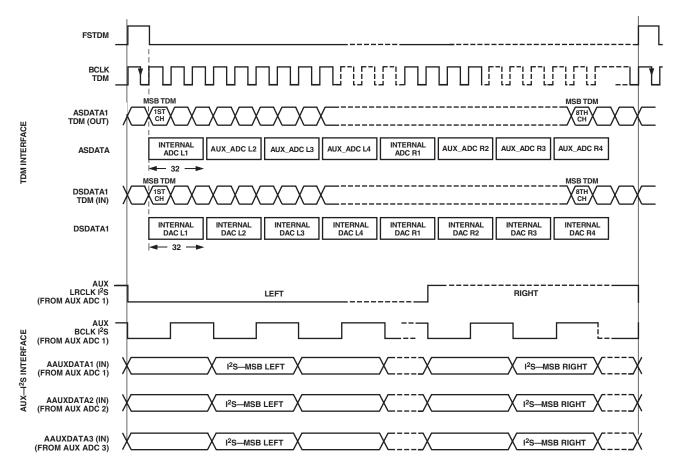


Figure 10. DAC Packed Mode Timing

REV. B –15–

Table II. Pin Function Changes in Auxiliary Mode

Pin Name	I <sup>2</sup> S Mode	Aux Mode
ASDATA (O)	I <sup>2</sup> S Data Out, Internal ADC	TDM Data Out to SHARC
DSDATA1 (I)	I <sup>2</sup> S Data In, Internal DAC1	TDM Data In from SHARC
DSDATA2 (I)/AAUXDATA1 (I)	I <sup>2</sup> S Data In, Internal DAC2	AUX-I <sup>2</sup> S Data In 1 (from Ext. ADC)
DSDATA3 (I)/AAUXDATA2 (I)	I <sup>2</sup> S Data In, Internal DAC3	AUX-I <sup>2</sup> S Data In 2 (from Ext. ADC)
DSDATA4 (I)/AAUXDATA3 (I)	I <sup>2</sup> S Data In, Internal DAC4	AUX-I <sup>2</sup> S Data In 3 (from Ext. ADC)
ALRCLK (O)	LRCLK for ADC	TDM Frame Sync Out to SHARC (FSTDM)
ABCLK (O)	BCLK for ADC	TDM BCLK Out to SHARC
DLRCLK (I)/AUXLRCLK(I/O)	LRCLK In/Out Internal DACs	AUX LRCLK In/Out. Driven by Ext. LRCLK from ADC
		in slave mode. In master mode, driven by MCLK/512.
DBCLK (I)/AUXBCLK	(I/O) BCLK In/Out Internal DACs	AUX BCLK In/Out. Driven by Ext. BCLK from ADC in
		slave mode. In master mode, driven by MCLK/8.



AUX BCLK FREQUENCY IS 64  $\times$  FRAME-RATE; TDM BCLK FREQUENCY IS 256  $\times$  FRAME-RATE.

Figure 11. Aux Mode Timing

-16- REV. B

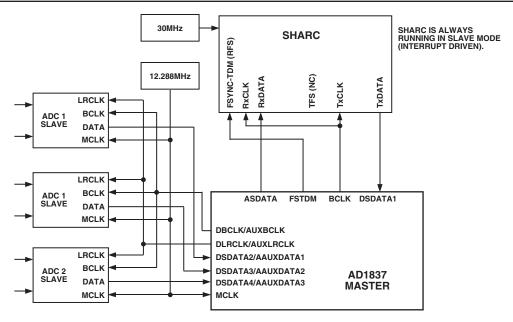


Figure 12. Aux Mode Connection to SHARC (Master Mode)

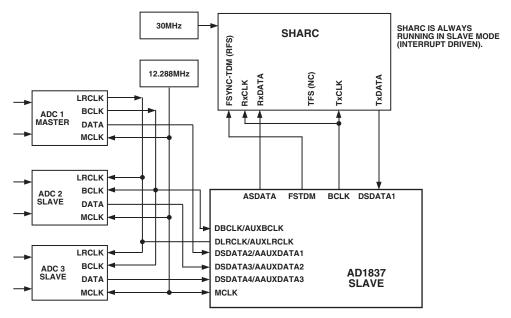


Figure 13. Aux Mode Connection to SHARC (Slave Mode)

REV. B –17–

#### **CONTROL/STATUS REGISTERS**

The AD1837 has 15 control registers, 13 of which are used to set the operating mode of the part. The other two registers, ADC Peak 0 and ADC Peak 1, are read-only and should not be programmed. Each of the registers is 10 bits wide with the exception of the ADC peak reading registers, which are 6 bits wide. Writing to a control register requires a 16-bit data frame to be transmitted. Bits 15 to 12 are the address bits of the required register. Bit 11 is a read/write bit. Bit 10 is reserved and should always be programmed to 0. Bits 9 to 0 contain the 10-bit value that is to be written to the register or, in the case of a read operation, the 10-bit register contents. Figure 3 shows the format of the SPI read and write operation.

## DAC CONTROL REGISTERS

The AD1837 register map has 10 registers that are used to control the functionality of the DAC section of the part. The function of the bits in these registers is discussed below.

### Sample Rate

These bits control the sample rate of the DACs. Based on a 24.576 MHz IMCLK, sample rates of 48 kHz, 96 kHz, and 192 kHz are available. The MCLK scaling bits in ADC Control 3 should be programmed appropriately, based on the master clock frequency.

#### Power-Down/Reset

This bit controls the power-down status of the DAC section. By default, normal mode is selected, but by setting this bit the digital section of the DAC stage can be put into a low power mode, thus reducing the digital current. The analog output section of the DAC stage is not powered down.

### **DAC Data-Word Width**

These two bits set the word width of the DAC data. Compact disc (CD) compatibility may require 16 bits, but many modern digital audio formats require 24-bit sample resolution.

## **DAC Data Format**

The AD1837 serial data interface can be configured to be compatible with a choice of popular interface formats including I<sup>2</sup>S, LJ, RJ, or DSP modes. Details of these interface modes are given in the Serial Data Port Section of this data sheet.

#### De-emphasis

The AD1837 provides built-in de-emphasis filtering for the three standard sample rates of 32.0 kHz, 44.1 kHz, and 48 kHz.

#### Mute DAC

Each of the eight DACs in the AD1837 has its own independent mute control. Setting the appropriate bit will mute the DAC output. The AD1837 uses a clickless mute function that attenuates the output to approximately  $-100~\mathrm{dB}$  over a number of cycles.

## Stereo Replicate

Setting this bit copies the digital data sent to the stereo pair DAC 1 to the three other stereo DACs in the system. This allows all four stereo DACs to be driven by one digital data stream. Note that in this mode, DAC data sent to the other DACs is ignored.

#### DAC Volume Control

Each DAC in the AD1837 has its own independent volume control. The volume of each DAC can be adjusted in 1024 linear steps by programming the appropriate register. The default value for this register is 1023, which provides no attenuation, i.e., full volume.

### ADC CONTROL REGISTERS

The AD1837 register map has five registers that are used to control the functionality and read the status of the ADCs. The function of the bits in each of these registers is discussed below.

#### ADC Peak Level

These two registers store the peak ADC result from each channel when the ADC peak readback function is enabled. The peak result is stored as a 6-bit number from 0 dB to -63 dB in 1 dB steps. The value contained in the register is reset once it has been read, allowing for continuous level adjustment as required. Note that the ADC peak level registers use the six most significant bits in the register to store the results.

#### Sample Rate

This bit controls the sample rate of the ADCs. Based on a 24.576 MHz IMCLK, sample rates of 48 kHz and 96 kHz are available. The MCLK scaling bits in ADC Control 3 should be programmed appropriately based on the master clock frequency.

### ADC Power-Down

This bit controls the power-down status of the ADC section and operates in a similar manner to the DAC power-down.

## **High-Pass Filter**

The ADC signal path has a digital high-pass filter. Enabling this filter will remove the effect of any dc offset in the analog input signal from the digital output codes.

#### Dither

Enabling the dither function will add a small amount of random charge to the sampling capacitors on the ADC inputs. This will eliminate the effect of any idle tones that could occur if there were no input signal present.

## **ADC Data-Word Width**

These two bits set the word width of the ADC data.

## ADC Data Format

The AD1837 serial data interface can be configured to be compatible with a choice of popular interface formats, including I<sup>2</sup>S, LJ, RJ, or DSP modes.

## Master/Slave Auxiliary Mode

When the AD1837 is operating in the auxiliary mode, the auxiliary ADC control pins, AUXBCLK and AUXLRCLK, that connect to the external ADCs, can be set to operate as a master or slave. If the pins are set in slave mode, one of the external ADCs should provide the LRCLK and BCLK signals.

#### ADC Peak Readback

Setting this bit enables ADC peak reading. See the ADC section for more information.

–18– REV. B

Table III. Control Register Map

Register Address	Register Name	Description	Type	Width	Reset Setting (Hex)
0000	DACCTRL1	DAC Control 1	$R/\overline{W}$	10	000
0001	DACCTRL2	DAC Control 2	$R/\overline{W}$	10	000
0010	DACVOL1	DAC Volume–Left 1	$R/\overline{W}$	10	3FF
0011	DACVOL2	DAC Volume–Right 1	$R/\overline{W}$	10	3FF
0100	DACVOL3	DAC Volume–Left 2	$R/\overline{W}$	10	3FF
0101	DACVOL4	DAC Volume–Right 2	$R/\overline{W}$	10	3FF
0110	DACVOL5	DAC Volume–Left 3	$R/\overline{W}$	10	3FF
0111	DACVOL6	DAC Volume–Right 3	$R/\overline{W}$	10	3FF
1000	DACVOL7	DAC Volume–Left 4	$R/\overline{W}$	10	3FF
1001	DACVOL8	DAC Volume–Right 4	$R/\overline{W}$	10	3FF
1010	ADCPeak0	ADC Left Peak	R	6	000
1011	ADCPeak1	ADC Right Peak	R	6	000
1100	ADCCTRL1	ADC Control 1	$R/\overline{W}$	10	000
1101	ADCCTRL2	ADC Control 2	$R/\overline{W}$	10	000
1110	ADCCTRL3	ADC Control 3	$R/\overline{W}$	10	000
1111	Reserved	Reserved	R/W	10	Reserved

Table IV. DAC Control 1

				Function						
Address	R/W	RES	De-emphasis	DAC Data Format	DAC Data- Word Width	Power-Down Reset	Sample Rate			
15, 14, 13, 12	11	10	9, 8	7, 6, 5	4, 3	2	1, 0			
0000	0	0	00 = None 01 = 44.1 kHz 10 = 32.0 kHz 11 = 48.0 kHz	000 = I <sup>2</sup> S 001 = RJ 010 = DSP 011 = LJ 100 = Pack Mode 256 101 = Reserved 110 = Reserved 111 = Reserved	00 = 24 Bits 01 = 20 Bits 10 = 16 Bits 11 = Reserved	0 = Normal 1 = Power-Down	00 = 48 kHz 01 = 96 kHz 10 = 192 kHz 11 = 48 kHz			

Table V. DAC Control 2

					Function							
				Stereo	MUTE DAC							
Address	R/W	RES	Reserved	Replicate	OUTR4	OUTL4	OUTR3	OUTL3	OUTR2	OUTL2	OUTR1	OUTL1
15, 14,												
13, 12	11	10	9	8	7	6	5	4	3	2	1	0
0001	0	0	0	0 = Off 1 = Replicate	0 = On 1 = Mute							

REV. B -19-

**Table VI. DAC Volume Control** 

Address	R/W	RES	Function DAC Volume
15, 14, 13, 12	11	10	9, 8, 7, 6, 5, 4, 3, 2, 1, 0
0010 = DACL1 0011 = DACR1 0100 = DACL2 0101 = DACR2 0110 = DACL3 0111 = DACR3 1000 = DACL4 1001 = DACR4	0	0	0000000000 = 1/1024 0000000001 = 2/1024 0000000010 = 3/1024 11111111110 = 1022/1024 11111111111 = 1023/1024

## Table VII. ADC Peak

			Function	ı
Address	R/W	RES	Six Data Bits	Four Fixed Bits
15, 14, 13, 12	11	10	9, 8, 7, 6, 5, 4	3, 2, 1, 0
0010 = Left ADC 1011 = Right ADC	1	0	000000 = 0.0 dBFS 000001 = -1.0 dBFS 000010 = -2.0 dBFS	

## Table VIII. ADC Control 1

				Function				
Address	R/W	RES	Dither	Filter	ADC Power-Down	Sample Rate	Reserved	
15, 14, 13, 12	11	10	9	8	7	6	5, 4, 3, 2, 1, 0	
1100	0	0	0 = Disabled 1 = Enabled	0 = All Pass 1 = High-Pass	0 = Normal 1 = Power-Down	0 = 48 kHz 1 = 96 kHz	0, 0, 0, 0, 0, 0 0, 0 0, 0, 0, 0, 0, 0,	

## Table IX. ADC Control 2

					Function				
	R/W		Master/Slave	ADC	ADC Data-		ADC Mute		
Address	RES	RES	Aux Mode	Data Format	Word Width	Reserved	Right	Left	
15, 14, 13, 12	11	10	9	8, 7, 6	5, 4	3, 2	1	0	
1101	0	0	0 = Slave 1 = Master	000 = I <sup>2</sup> S 001 = RJ 010 = DSP 011 = LJ 100 = Packed 256 101 = Reserved 110 = Auxiliary 256 111 = Reserved	00 = 24 Bits 01 = 20 Bits 10 = 16 Bits 11 = Reserved	0, 0	0 = On 1 = Mute	0 = On 1 = Mute	

## Table X. ADC Control 3

				Function				
	R/W			IMCLK	ADC	DAC	ADC	
Address	RES	RES	Reserved	Clocking Scaling	Peak Readback	Test Mode	Test Mode	
15, 14, 13, 12	11	10	9	8, 7, 6	5	4, 3, 2	1, 0	
1110	0	0	0, 0	00 = MCLK × 2 01 = MCLK 10 = MCLK × 2/3 11 = MCLK × 2	0 = Disabled Peak Readback 1 = Enabled Peak Readback			

-20- REV. B

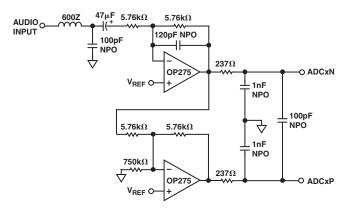


Figure 14. Typical ADC Input Filter Circuit

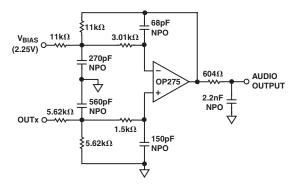


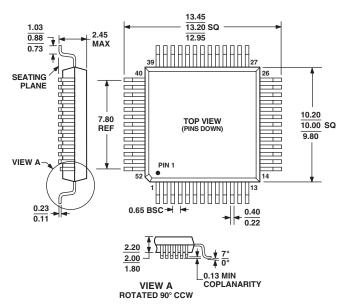
Figure 15. Typical DAC Output Filter Circuit

REV. B –21–

## **OUTLINE DIMENSIONS**

## 52-Lead Metric Quad Flat Package [MQFP] (S-52)

Dimensions shown in millimeters



COMPLIANT TO JEDEC STANDARDS MS-022-AC

-22- REV. B

# **Revision History**

Location	Page
7/03—Data Sheet changed from REV. A to REV. B.	
Changes to SPECIFICATIONS	
Changes to ORDERING GUIDE	
Changes to Figure 3	
Changes to Table IV	
Changes to Figure 15	
Updated OUTLINE DIMENSIONS	
10/02—Data Sheet changed from REV. B to REV. A.	
Changes to FEATURES	
Changes to SPECIFICATIONS	
Changes to PIN CONFIGURATION	
Changes to FUNCTIONAL OVERVIEW	
Changes to OUTLINE DIMENSIONS	

REV. B –23–



OOO «ЛайфЭлектроникс" "LifeElectronics" LLC

ИНН 7805602321 КПП 780501001 P/C 40702810122510004610 ФАКБ "АБСОЛЮТ БАНК" (ЗАО) в г.Санкт-Петербурге К/С 3010181090000000703 БИК 044030703

Компания «Life Electronics» занимается поставками электронных компонентов импортного и отечественного производства от производителей и со складов крупных дистрибьюторов Европы, Америки и Азии.

С конца 2013 года компания активно расширяет линейку поставок компонентов по направлению коаксиальный кабель, кварцевые генераторы и конденсаторы (керамические, пленочные, электролитические), за счёт заключения дистрибьюторских договоров

## Мы предлагаем:

- Конкурентоспособные цены и скидки постоянным клиентам.
- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
- Поставку компонентов, требующих военную и космическую приемку.
- Входной контроль качества.
- Наличие сертификата ISO.

В составе нашей компании организован Конструкторский отдел, призванный помогать разработчикам, и инженерам.

Конструкторский отдел помогает осуществить:

- Регистрацию проекта у производителя компонентов.
- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



Тел: +7 (812) 336 43 04 (многоканальный) Email: org@lifeelectronics.ru