

8-Channel, 12-Bit, Configurable ADC/DAC with On-Chip Reference, SPI Interface

Data Sheet **[AD5592R](https://www.analog.com/AD5592R?doc=AD5592R.pdf)**

FEATURES

8-channel, configurable ADC/DAC/GPIO Configurable as any combination of 8 × 12-bit DAC channels 8 × 12-bit ADC channels 8 × general-purpose digital input/output pins Integrated temperature sensor SPI interface Available in 16-ball, 2 mm × 2 mm WLCSP 16-lead, 3 mm × 3 mm LFCSP 16-lead TSSOP

APPLICATIONS

Control and monitoring General-purpose analog and digital inputs/outputs

GENERAL DESCRIPTION

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have eight I/Ox pins (I/O0 to I/O7) that can be independently configured as digital-to-analog converter (DAC) outputs, analog-to-digital converter (ADC) inputs, digital outputs, or digital inputs. When an I/Ox pin is configured as an analog output, it is driven by a 12-bit DAC. The output range of the DAC is 0 V to V_{REF} or 0 V to 2 \times V_{REF}.

When an I/Ox pin is configured as an analog input, it is connected to a 12-bit ADC via an analog multiplexer. The input range of the ADC is 0 V to VREF or 0 V to $2 \times V$ REF. The ADC has a total throughput rate of 400 kSPS. The I/Ox pins can also be configured as digital, general-purpose input or output (GPIO) pins. The state of the GPIO pins can be set or read back by accessing the GPIO write data register or the GPIO read configuration register, respectively, via a serial peripheral interface (SPI) write or read operation.

Th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have an integrated 2.5 V, 25 ppm/°C reference, which is turned off by default, and an integrated temperature indicator, which gives an indication of the die temperature. The temperature value is read back as part of an ADC read sequence.

The $AD5592R/AD5592R-1$ are available in 16-ball, 2 mm \times 2 mm WLCSP, 16-lead, 3 mm × 3 mm LFCSP, and 16-lead TSSOP. Th[e AD5592R/AD5592R-1 o](http://analog.com/AD5592R?doc=AD5592R.pdf)perate over a temperature range of −40°C to +105°C.

Table 1. Related Products

Figure 1[. AD5592R F](http://analog.com/AD5592R?doc=AD5592R.pdf)unctional Block Diagram

Rev. E [Document Feedback](https://form.analog.com/Form_Pages/feedback/documentfeedback.aspx?doc=AD5592R.pdf&product=AD5592R&rev=E)

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REVISION HISTORY

8/2017-Rev. C to Rev. D

2/2017-Rev. B to Rev. C

$2/2016$ –Rev. A to Rev. B

10/2014-Rev. 0 to Rev. A

8/2014-Revision 0: Initial Version

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FUNCTIONAL BLOCK DIAGRAM [\(AD5592R-1\)](https://www.analog.com/AD5592R?doc=AD5592R.pdf)

Figure 2[. AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) Functional Block Diagram

SPECIFICATIONS

 $V_{DD} = 2.7$ V to 5.5 V, $V_{REF} = 2.5$ V (external), $R_L = 2$ kΩ to GND, $C_L = 200$ pF to GND, $T_A = T_{MIN}$ to T_{MAX} , temperature range = -40°C to +105°C, unless otherwise noted.

Table 2.

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¹ All specifications expressed in decibels are referred to full-scale input (FSR) and tested with an input signal at 0.5 dB below full scale, unless otherwise noted. ² Guaranteed by design and characterization; not production tested.

³ DC specifications tested with the outputs unloaded, unless otherwise noted. Linearity calculated using a code range of 8 to 4095. There is an upper dead band of 10 mV when $V_{REF} = V_{DD}$.

⁴ When drawing a load current at either rail, the output voltage headroom with respect to that rail is limited by the 25 Ω typical channel resistance of the output devices. For example, when sinking 1 mA, the minimum output voltage = $25 \Omega \times 1$ mA = 25 mV (se[e Figure 33\).](#page-17-0)

TIMING CHARACTERISTICS

Guaranteed by design and characterization, not production tested; all input signals are specified with $t_R = t_F = 5$ ns (10% to 90% of V_{DD}) and timed from a voltage level of $(V_{IL} + V_{IH})/2$; $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.

Table 3[. AD5592R](http://analog.com/AD5592R?doc=AD5592R.pdf) Timing Characteristics

¹ When reading an ADC conversion.

Table +, AD3372R + Thinnig Characteristics					
Parameter	1.62 $V \leq V_{\text{LOGIC}} < 3 V$	$3 V \leq V_{LOGIC} \leq 5.5 V$	Unit	Test Conditions/Comments	
t1	33	20	ns min	SCLK cycle time, write operation	
	65	50	ns min	SCLK cycle time, read operation	
t ₂	16	10	ns min	SCLK high time	
t_3	16	10	ns min	SCLK low time	
t4	15	10	ns min	SYNC to SCLK falling edge setup time	
	2		us max	SYNC to SCLK falling edge setup time	
t ₅			ns min	Data setup time	
t6	5	5	ns min	Data hold time	
t ₇	15	10	ns min	SCLK falling edge to SYNC rising edge	
ts	30	30	ns min	Minimum SYNC high time for write operations	
	60	60	ns min	Minimum SYNC high time for register read operations	
t9	Ω	0	ns min	SYNC rising edge to next SCLK falling edge	
t_{10}	56	25	ns max	SCLK rising edge to SDO valid	

Table 4[. AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) Timing Characteristics

Figure 3. Load Circuit for Logic Output (SDO) Timing Specifications

Figure 4. Timing Diagram

ABSOLUTE MAXIMUM RATINGS

 $T_A = 25$ °C, unless otherwise noted. Transient currents of up to 100 mA do not cause SCR latch-up.

Table 5.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

 θ_{JA} is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

Table 6. Thermal Resistance

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

Figure 5[. AD5592R](http://analog.com/AD5592R?doc=AD5592R.pdf) 16-Ball WLCSP Pin Configuration

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Figure 9[. AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) 16-Ball WLCSP Pin Configuration

Table 11[. AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) 16-Lead WLCSP Pin Function Descriptions

TYPICAL PERFORMANCE CHARACTERISTICS

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2.58 2.56 2.54 2.52 VOUT (V) 2.50 2.48 2.46 2.44 2.42 12506-119 12506-119 **–10 –5 0 5 10 TIME (µs)**

Figure 22. DAC Settling Time (100 Code Change, Rising Edge)

Figure 25. DAC Settling Time, Output Range = 0 V to 2 \times V_{REF}

Bandwidth = 0 Hz to 20 kHz

Figure 36. Reference Line Regulation

Figure 35. Reference Noise Spectral Density (NSD)

TERMINOLOGY ADC TERMINOLOGY

Integral Nonlinearity (INL)

INL is the maximum deviation from a straight line passing through the endpoints of the ADC transfer function. The end-points of the transfer function are zero scale, a point that is 1 LSB below the first code transition, and full scale, a point that is 1 LSB above the last code transition.

Differential Nonlinearity (DNL)

DNL is the difference between the measured and the ideal 1 LSB change between any two adjacent codes in the ADC.

Offset Error

Offset error is the deviation of the first code transition (00 … 000) to $(00 \dots 001)$ from the ideal, that is, $AGND + 1$ LSB.

Offset Error Match

Offset error match is the difference in offset error between any two channels.

Gain Error

Gain error is the deviation of the last code transition (111 … 110) to (111 … 111) from the ideal (that is, V_{REF} – 1 LSB) after the offset error has been adjusted out.

Channel-to-Channel Isolation

Channel-to-channel isolation is a measure of the level of crosstalk between channels. It is measured by applying a full-scale, 5 kHz sine wave signal to all nonselected ADC input channels and determining how much that signal is attenuated in the selected channel. This specification is the worst case across all ADC channels for the [AD5592R/AD5592R-1.](http://analog.com/AD5592R?doc=AD5592R.pdf)

Track-and-Hold Acquisition Time

The track-and-hold amplifier enters hold mode on the falling edge of SYNC and returns to track mode when the conversion is complete. The track-and-hold acquisition time is the minimum time required for the track-and-hold amplifier to remain in track mode for its output to reach and settle to within ±1 LSB of the applied input signal, given a step change to the input signal.

Signal-to-Noise-and-Distortion (SINAD) Ratio

SINAD is the measured ratio of signal-to-noise-and-distortion at the output of the ADC. The signal is the rms amplitude of the fundamental. Noise is the sum of all nonfundamental signals up to half the sampling frequency $(f_s/2)$, excluding dc. The ratio is dependent on the number of quantization levels in the digitization process; the more levels, the smaller the quantization noise. The theoretical SINAD ratio for an ideal N-bit converter with a sine wave input is given by

 $SIMAD$ (dB) = $6.02N + 1.76$

Thus, for a 12-bit converter, SINAD is 74 dB.

Total Harmonic Distortion (THD)

THD is the ratio of the rms sum of harmonics to the fundamental. For th[e AD5592R/AD5592R-1,](http://analog.com/AD5592R?doc=AD5592R.pdf) it is defined as

$$
THD(dB) = 20 \times \log \frac{\sqrt{V_2^2 + V_3^2 + V_4^2 + V_5^2 + V_6^2}}{V_1}
$$

where:

V1 is the rms amplitude of the fundamental. V_2 , V_3 , V_4 , V_5 , and V_6 are the rms amplitudes of the second

through the sixth harmonics.

Peak Harmonic or Spurious Noise (SFDR)

Peak harmonic or spurious noise is defined as the ratio of the rms value of the next largest component in the ADC output spectrum (up to fs/2 and excluding dc) to the rms value of the fundamental. Normally, the value of this specification is determined by the largest harmonic in the spectrum, but for ADCs where the harmonics are buried in the noise floor, it is a noise peak.

DAC TERMINOLOGY

Relative Accuracy or Integral Nonlinearity (INL)

For the DAC, relative accuracy or integral nonlinearity is a measurement of the maximum deviation, in LSBs, from a straight line passing through the endpoints of the DAC transfer function. A typical INL vs. code plot is shown i[n Figure 17.](#page-15-0)

Differential Nonlinearity (DNL)

Differential nonlinearity is the difference between the measured change and the ideal 1 LSB change between any two adjacent codes. A specified differential nonlinearity of ±1 LSB maximum ensures monotonicity. This DAC is guaranteed monotonic by design. A typical DNL vs. code plot can be seen i[n Figure 18.](#page-15-1)

Zero Code Error

Zero code error is a measurement of the output error when zero code (0x000) is loaded to the DAC register. Ideally, the output is 0 V. The zero code error is always positive in th[e AD5592R/](http://analog.com/AD5592R?doc=AD5592R.pdf) [AD5592R-1 b](http://analog.com/AD5592R?doc=AD5592R.pdf)ecause the output of the DAC cannot go below 0 V due to a combination of the offset errors in the DAC and the output amplifier. Zero code error is expressed in mV.

Gain Error

Gain error is a measure of the span error of the DAC. It is the deviation in slope of the DAC transfer characteristic from the ideal expressed as % FSR.

Offset Error Drift

Offset error drift is a measurement of the change in offset error with a change in temperature. It is expressed in μV/°C.

Gain Temperature Coefficient

Gain temperature coefficient is a measurement of the change in gain error with changes in temperature. It is expressed in ppm of FSR/°C.

Offset Error

Offset error is a measurement of the difference between $\rm V_{OUT}$ (actual) and V_{OUT} (ideal), expressed in mV, in the linear region of the transfer function. Offset error can be negative or positive.

DC Power Supply Rejection Ratio (PSRR)

PSRR indicates how the output of the DAC is affected by changes in the supply voltage. PSRR is the ratio of the change in V_{OUT} to a change in V_{DD} for a full-scale output of the DAC. It is measured in mV/V. VREF is held at 2 V, and V_{DD} is varied by $\pm 10\%$.

Output Voltage Settling Time

Output voltage settling time is the amount of time it takes for the output of a DAC to settle to a specified level for a ¼ to ¾ full-scale input change and is measured from the rising edge of SYNC.

Digital-to-Analog Glitch Impulse

Digital-to-analog glitch impulse is the impulse injected into the analog output when the input code in the DAC register changes state. It is normally specified as the area of the glitch in nV-sec, and is measured when the digital input code is changed by 1 LSB at the major carry transition (0x7FF to 0x800).

Digital Feedthrough

Digital feedthrough is a measure of the impulse injected into the analog output of the DAC from the digital inputs of the DAC, but is measured when the DAC output is not updated. It is specified in nV-sec, and measured with a full-scale code change on the data bus, that is, from all 0s to all 1s and vice versa.

Reference Feedthrough

Reference feedthrough is the ratio of the amplitude of the signal at the DAC output to the reference input when the DAC output is not being updated. It is expressed in dB.

Noise Spectral Density

Noise spectral density is a measurement of the internally generated random noise. Random noise is characterized as a spectral density ($\frac{nV}{\text{Hz}}$). It is measured by loading the DAC to midscale and measuring noise at the output. It is measured in nV/\sqrt{Hz} .

DC Crosstalk

DC crosstalk is the dc change in the output level of one DAC in response to a change in the output of another DAC. It is measured with a full-scale output change on one DAC (or soft power-down and power-up) while monitoring another DAC maintained at midscale. It is expressed in μV.

DC crosstalk due to load current change is a measure of the impact that a change in load current on one DAC has to another DAC kept at midscale. It is expressed in μV/mA.

Digital Crosstalk

Digital crosstalk is the glitch impulse transferred to the output of one DAC at midscale in response to a full-scale code change (all 0s to all 1s and vice versa) in the input register of another DAC. It is measured in standalone mode and is expressed in nV-sec.

Analog Crosstalk

Analog crosstalk is the glitch impulse transferred to the output of one DAC due to a change in the output of another DAC. It is measured by loading one of the input registers with a full-scale code change (all 0s [to all 1s an](#page-38-2)d v[ice versa\),](#page-39-0) then executing a software LDAC (see Table 45 and Table 46), and monitoring the output of the DAC whose digital code was not changed. The area of the glitch is expressed in nV-sec.

DAC-to-DAC Crosstalk

DAC-to-DAC crosstalk is the glitch impulse transferred to the output of one DAC due to a digital code change and subsequent analog output change of another DAC. It is measured by loading the attack channel with a full-scale code change (all 0s to all 1s and vice versa), using the write to and update commands while monitoring the output of the victim channel that is at midscale. The energy of the glitch is expressed in nV-sec.

Multiplying Bandwidth

The amplifiers within the DAC have a finite bandwidth; the multiplying bandwidth is a measure of this. A sine wave on the reference (with full-scale code loaded to the DAC) appears on the output. The multiplying bandwidth is the frequency at which the output amplitude falls to 3 dB below the input.

Voltage Reference Temperature Coefficient (TC)

Voltage reference TC is a measure of the change in the reference output voltage with a change in temperature. The voltage reference TC is calculated using the box method, which defines the TC as the maximum change in the reference output over a given temperature range expressed in ppm/°C, as follows:

$$
TC = \left[\frac{V_{REF(MAX)} - V_{REF(MIN)}}{V_{REF(NOM)} \times Temp Range}\right] \times 10^6
$$

where:

VREF(MAX) is the maximum reference output measured over the total temperature range.

VREF(MIN) is the minimum reference output measured over the total temperature range.

VREF(NOM) is the nominal reference output voltage, 2.5 V. *Temp Range* is the specified temperature range of −40°C to $+105$ °C.

THEORY OF OPERATION

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) are 8-channel configurable analog and digital input/output ports. Th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have eight pins that can be independently configured as a 12-bit DAC output channel, a 12-bit ADC input channel, a digital input pin, or a digital output pin.

The function of each pin is determined by programming the ADC, DAC, or GPIO configuration registers as appropriate. See the [Configuring the AD5592R/AD5592R-1](#page-26-0) section and [Table 16](#page-26-2) for more information.

DAC SECTION

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) contain eight 12-bit DACs and implement a segmented string DAC architecture with an internal output buffer. [Figure 37](#page-22-2) shows the internal block diagram of the DAC architecture.

Figure 37. Internal Block Diagram of the DAC Architecture

The DAC channels have a shared gain bit that sets the output range as 0 V to V_{REF} or 0 V to 2 \times V_{REF}. Because the gain bit is shared by all channels, it is not possible to set different output ranges on a per channel basis. The input coding to the DAC is straight binary. The ideal output voltage is given by

$$
V_{OUT} = G \times V_{REF} \times \left(\frac{D}{2^N}\right)
$$

where:

D is the decimal equivalent of the binary code (0 to 4095) that is loaded to the DAC register.

 $G = 1$ for an output range of 0 V to V_{REF} , or $G = 2$ for an output range of 0 V to $2 \times V_{REF}$. $N = 12$.

Resistor String

The simplified segmented resistor string DAC structure is shown in [Figure 38.](#page-22-3) The code loaded to the DAC register determines the switch on the string that is connected to the output buffer.

Because each resistance in the string has the same value, R, the string DAC is guaranteed monotonic.

Figure 38. Simplified Resistor String Structure

Output Buffer

The output buffer is designed as an input/output rail-to-rail buffer. The output buffer can drive 2 nF capacitance with a 1 kΩ resistor in parallel. The slew rate is 1.25 V/µs with a ¼ to ¾ scale settling time of 6 µs. By default, the DAC outputs update directly after data has been written to the input register. The LDAC register is used to delay the updates until additional channels have been written to, if required. See the [Readback](#page-38-1) [and LDAC Mode Register](#page-38-1) section for more information.

ADC SECTION

The 12-bit, single-supply ADC is capable of throughput rates of 400 kSPS. The ADC is preceded by a multiplexer that switches selected I/Ox pins to the ADC. A sequencer is included to automatically switch the multiplexer to the next selected channel. Channels are selected for conversion by writing to the ADC sequence register. When the write to the ADC sequence register has completed, the first channel in the conversion sequence is put into track mode. Allow each channel to track the input signal for a minimum of 500 ns. The first SYNC falling edge following the write to the ADC sequence register begins the conversion of the first channel in the sequence. The next SYNC falling edge starts a conversion on the second channel in the sequence and also begins to clock the first ADC result onto the serial interface. ADC data is clocked out of the [AD5592R](http://analog.com/AD5592R?doc=AD5592R.pdf) in a 16-bit frame. D15 is 0 to indicate that the data contains ADC data, D14 to D12 is the binary representation of the ADC address, and D11 to D0 is the ADC result (see [Table](#page-23-1) 12).

Each conversion takes 2 µs, and the conversion must be completed before another conversion is initiated. Only write to the [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) when no conversion is taking place. I/O7 can be configured as a BUSY signal to indicate when a conversion is taking place. BUSY goes low while a conversion is in progress, and high when an ADC result is available. The ADC has an input range selection bit (Bit D5 in the generalpurpose control register), which sets the input range as 0 V to V_{REF} or 0 V to 2 \times V_{REF}. All input channels share the same range. The output coding of the ADC is straight binary. It is possible to set each I/Ox pin as both a DAC and an ADC. When an I/Ox pin is set as both a DAC and an ADC, the primary function is that of the DAC. If the pin is selected for inclusion in an ADC conversion sequence, the voltage on the pin is converted and made available via the serial interface, allowing the DAC voltage to be monitored.

Table 12. ADC Conversion Format

Calculating ADC Input Current

The current flowing into the I/Ox pins configured as ADC inputs vary with the sampling rate (f_s) , the voltage difference between successive channels (V_{DIFF}), and whether buffered or unbuffered mode is used[. Figure 39](#page-23-2) shows a simplified version of the ADC input structure. When a new channel is selected for conversion, the 5.8 pF capacitor must be charged or discharged of the voltage that was on the previously selected channel. The time required by the charge or discharge depends on the voltage difference between the two channels. This affects the input impedance of the multiplexer and therefore the input current flowing into the I/Ox pins. In buffered mode, Switch S1 is open and Switch S2 is closed, in which case the U1 buffer is directly driving the 23.1 pF capacitor, and its charging time is negligible. In unbuffered mode, Switch S1 is closed and Switch S2 is closed. In unbuffered mode, the 23.1 pF capacitor must be charged from the I/Ox pins, which contributes to the input current. For applications where the ADC input current is too high, an external input buffer may be required. The choice of buffer is a function of the particular application.

Calculate the input current for buffered mode as follows:

$$
f_s \times C \times V_{\text{DIFF}} + 1 \text{ nA}
$$

where:

fs is the ADC sample rate in Hertz.

C is the sampling capacitance in Farads.

VDIFF is the voltage change between successive channels.

1 nA is the dc leakage current associated with buffered mode.

Calculate the input current for unbuffered mode as follows:

$$
f_{\rm S}\times C\times V_{\rm DIFF}
$$

An example solution is as follows: for the ADC input current in buffered mode, where $I/O0 = 0.5$ V, $I/O1 = 2$ V, and $f_s = 10$ kHz,

 $(10,000 \times 5.8 \times 10^{-12} \times 1.5) + 1 \text{ nA} = 88 \text{ nA}$

Under the same conditions, the ADC input current in unbuffered mode is as follows:

$$
(10,000 \times 28.9 \times 10^{-12} \times 1.5) = 433.5 \text{ nA}
$$

Figure 39. ADC Input Structure

GPIO SECTION

Each of the eight I/Ox pins can be configured as a generalpurpose digital input pin by programming the GPIO read configuration register or as a digital output pin by programming the GPIO write configuration register. When an I/Ox pin is configured as an output, the pin can be set high or low by programming the GPIO write data register. Logic levels for general-purpose outputs are relative to V_{DD} and GND. When an I/Ox pin is configured as an input, its status can be determined by setting Bit D10 in the GPIO read configuration register (see [Table 37\)](#page-35-0). The next SPI operation clocks out the state of the GPIO pins. When an I/Ox pin is set as an output, it is possible to read its status by also setting it as an input pin. When reading the status of the I/Ox pins set as inputs, the status of an I/Ox pin set as both an input and output pin is also returned.

INTERNAL REFERENCE

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) contain an on-chip 2.5 V reference. The reference is powered down by default and is enabled by setting Bit D9 in the power-down register to 1 (se[e Table 43\)](#page-37-1). When the on-chip reference is powered up, the reference voltage appears on the V_{REF} pin and may be used as a reference source for other components. When the internal reference is used, it is recommended to decouple the internal reference to GND using a 100 nF capacitor. It is recommended that the internal reference be buffered before using it elsewhere in the system. When the reference is powered down, an external reference must be connected to the VREF pin. Suitable external reference sources for th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) include the [AD780,](https://www.analog.com/AD780?doc=AD5592R.pdf) [AD1582,](https://www.analog.com/AD1582?doc=AD5592R.pdf) [ADR431,](https://www.analog.com/ADR431?doc=AD5592R.pdf) [REF193,](https://www.analog.com/REF193?doc=AD5592R.pdf) an[d ADR391.](https://www.analog.com/ADR391?doc=AD5592R.pdf)

RESET FUNCTION

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have an asynchronous RESET pin. For normal operation, RESET is tied high. A falling edge on RESET resets all registers to their default values and reconfigures the I/Ox pins to their default values (85 kΩ pull-down to GND). The reset function takes 250 µs maximum; do not write new data to th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) during this time. The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have a software reset that performs the same function as the RESET pin. The reset function is activated by writing 0x5AC to the reset register (see [Table 44\)](#page-38-3).

TEMPERATURE INDICATOR

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) contain an integrated temperature indicator, which can be read to provide an estimation of the die temperature. The temperature reading can be used in fault detection where a sudden rise in die temperature may indicate a fault condition such as a shorted output. Temperature readback is enabled by setting Bit D8 in the ADC sequence register to 1 (see [Table 28\)](#page-31-0). The temperature result is then added to the ADC sequence. The temperature result has an address of 0b1000; take care that this result is not confused with the readback from DAC0. The temperature conversion takes 5 us with the ADC buffer enabled and 20 µs when the buffer is disabled. Calculate the temperature by using the following formulae:

For ADC gain =
$$
1
$$
,

Temperature (°C) =
25 +
$$
\frac{(ADC Code - (0.5/V_{REF}) \times 4095)}{(2.654 \times (2.5/V_{REF}))}
$$

For ADC gain $= 2$,

Temperature (°C) =
25 +
$$
\frac{(ADC Code - (0.5/(2 \times V_{REF})) \times 4095)}{(1.327 \times (2.5/V_{REF}))}
$$

The range of codes returned by the ADC when reading from the temperature indicator is approximately 645 to 1035, corresponding to a temperature between −40°C to +105°C. The accuracy of the temperature indicator, averaged over five samples, is typically 3°C.

SERIAL INTERFACE

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have a serial interface (SYNC, SCLK, SDI, and SDO), which is compatible with SPI standards, as well as with most DSPs. The input shift register is 16 bits wide (see [Table](#page-25-4) 13). The MSB (D15) determines what type of write function is required. When D15 is 0, a write to the control register is selected. The control register address is selected by D14 to D11. D10 and D9 are reserved and are 0s. D8 to D0 set the data that is written to the selected control register. When D15 is 1, data is written to a DAC channel (assuming that channel has been set to be a DAC). D14 to D12 select which DAC is addressed. D11 to D0 is the 12-bit data loaded to the selected DAC, with D11 being the MSB of the DAC data. [Table](#page-25-5) 14 shows the control register map for th[e AD5592R/AD5592R-1.](http://analog.com/AD5592R?doc=AD5592R.pdf) The register map allows the operation of each of the I/Ox pins to be configured. ADCs can be selected for inclusion in sampling sequences. DACs can be updated individually or simultaneously (see th[e LDAC Mode Operation](#page-28-1) section). GPIO settings are also controlled via the register map.

POWER-UP TIME

When power is applied to the [AD5592R/AD5592R-1,](http://analog.com/AD5592R?doc=AD5592R.pdf) the power-on reset block begins to configure the device and to load the registers with their default values. The configuration process takes 250 µs; do not write to any of the registers during this time.

WRITE MODE

[Figure 4](#page-7-0) shows the read and write timing for th[e AD5592R/](http://analog.com/AD5592R?doc=AD5592R.pdf) [AD5592R-1.](http://analog.com/AD5592R?doc=AD5592R.pdf) A write sequence begins by bringing the SYNC line low. Data on SDI is clocked into the 16-bit shift register on the falling edge of SCLK. After the 16th falling clock edge, the last data bit is clocked in. SYNC is brought high, and the programmed function is executed (that is, a change in a DAC input register or a change in a control register). SYNC must be brought high for a minimum of 20 ns before the next write. All interface pins must be operated close to the V_{DD} or V_{LOGIC} rails to minimize power consumption in the digital input buffers.

READ MODE

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) allow data readback from the ADCs and control registers via the serial interface. ADC conversions are automatically clocked out on the serial interface as part of a sequence or as a single ADC conversion. Reading from a register first requires a write to the readback and LDAC mode register to select the register to read back. The contents of the selected register are clocked out on the next 16 SCLKs following a falling edge of SYNC. Note that due to timing requirements of t_{10} (25 ns), the maximum speed of the SPI interface during a read operation must not exceed 20 MHz.

Table 13. Input Shift Register Format

¹ This register is also used to set I/O7 as a BUSY output.

² D14 to D11 is the DAC register address (se[e Table](#page-25-4) 13).

CONFIGURING TH[E AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf)

Th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) I/Ox pins are configured by writing to a series of configuration registers. The control registers are accessed when the MSB of a serial write is 0, as shown in [Table](#page-25-4) 13. The control register map for th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) is shown in [Table](#page-25-5) 14. At power-up, the I/Ox pins are configured as 85 kΩ pulldown resistors connected to GND.

The input/output channels of th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) can be configured to operate as DAC outputs, ADC inputs, digital outputs, digital inputs, three-state, or connected to GND with 85 kΩ pulldown resistors. When configured as digital outputs, the I/Ox pins have the additional option of being configured as push/pull or open-drain. The input/output channels are configured by writing to the appropriate configuration registers, as shown in [Table 15](#page-26-1) and [Table 16.](#page-26-2) To assign a particular function to an input/output channel, the user writes to the appropriate register and sets the corresponding bit to 1. For example, setting Bit D0 in the DAC configuration register to 1 configures I/O0 as a DAC (se[e Table 20\)](#page-28-2).

In the event that the bit for an input/output channel is set in multiple configuration registers, the input/output channel takes the function dictated by the last write operation. The exceptions to this rule are that an I/Ox pin can be set as both a DAC and an ADC or

Table 15. I/Ox Pin Configuration Registers

as a digital input and output. When an I/Ox pin is configured as a DAC and ADC, its primary function is as a DAC, and the ADC can measure the voltage being provided by the DAC. This feature can monitor the output voltage to detect short circuits or overload conditions.

When a pin is configured as both a general-purpose input and output, the primary function is as an output pin. This configuration allows the status of the output pin to be determined by reading the GPIO register. [Figure 40](#page-26-3) shows a typical configuration example where I/O0 and I/O1 are configured as ADCs, I/O2 and I/O3 are configured as DACs, I/O4 is a general-purpose output pin, I/O5 is a general-purpose input pin, and I/O6 and I/O7 are three-state.

The general-purpose control register also contains other functions associated with the DAC and ADC, such as the lock configuration bit. When the lock configuration bit is set to 1, any writes to the pin configuration registers are ignored, thus preventing the function of the I/Ox pins from being changed.

The I/Ox pins can be reconfigured at any time when th[e AD5592R/](http://analog.com/AD5592R?doc=AD5592R.pdf) [AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) is in an idle state, that is, no ADC conversions are taking place and no registers are being read back. The lock configuration bit must also be 0.

Table 16. Bit Descriptions for the I/Ox Pin Configuration Registers

GENERAL-PURPOSE CONTROL REGISTER

The general-purpose control register enables or disables certain functions associated with the DAC, ADC, and I/Ox pin configuration (see [Table 17](#page-27-1) and [Table 18\)](#page-27-2). The general-purpose control register sets the gain of the DAC and ADC. Bit D5 sets the input range for the ADC, and Bit D4 sets the output range of the DAC.

The general-purpose control register also enables/disables the ADC buffer and precharge function (see th[e ADC Section](#page-23-0) for more details). The register can also be used to lock the I/Ox pin configuration to prevent accidental change. When Bit D7 is set to 1, writes to the configuration registers are ignored.

Table 17. General-Purpose Control Register

Table 18. Bit Descriptions for the General-Purpose Control Register

DAC WRITE OPERATION

To set a pin as a DAC, set the appropriate bit in the DAC pin configuration register to 1 (see [Table 19](#page-28-3) and [Table 20\)](#page-28-2). For example, setting Bit 0 to Bit 1 enables I/O0 as a DAC output. Data is written to a DAC when the MSB (D15) of the serial write is 1. D14, D13, and D12 determine which DAC is addressed, and D11 to D0 contain the 12-bit data to be written to the DAC, as shown i[n Table 21](#page-28-4) and [Table 22.](#page-28-5) Data is written to the selected DAC input register. Data written to the input register can be automatically copied to the DAC register, if required. Data is transferred to the DAC register based on the setting of the LDAC mode register (se[e Table 45](#page-38-2) and [Table 46\)](#page-39-0).

Table 19. DAC Pin Configuration Register

LDAC Mode Operation When the LDAC mode bits (D1 and D0) are 00 respectively,

new data is automatically transferred from the input register to the DAC register, and the analog output updates. When the LDAC mode bits are 01, data remains in the input register. This LDAC mode allows writes to input registers without affecting the analog outputs. When the input registers have been loaded with the desired values, setting the LDAC mode bits to 10 transfers the values in the input registers to the DAC registers, and the analog outputs update simultaneously. The LDAC mode bits then revert back to 01, assuming their previous setting was 01. See [Table 45](#page-38-2) and [Table 46.](#page-39-0)

Table 20. Bit Descriptions for the DAC Pin Configuration Register

Table 21. DAC Write Register

Table 22. Bit Descriptions for the DAC Data Register

DAC READBACK

The input register of each DAC can be read back via the SPI interface. Reading back the DAC register value can be used to confirm that the data was received correctly before writing to the LDAC register, or to check what value was last loaded to a DAC. Data can only be read back from a DAC when there is no ADC conversion sequence taking place.

To read back a DAC input register, it is first necessary to enable the readback function and select which DAC register is required.

This is achieved by writing to the DAC read back register (shown in [Table 23](#page-29-1) and [Table 24\)](#page-29-2). Set the D4 and D3 bits to 1 to enable the readback function. The D2 to D0 bits select which DAC data is required. The DAC data is clocked out of the [AD5592R/](http://analog.com/AD5592R?doc=AD5592R.pdf) [AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) on the subsequent SPI operation. [Figure 41](#page-29-3) shows an example of setting I/O3, configured as a DAC, to midscale. The input data is then read back. D14 to D12 contain the address of the DAC register being read back, and D15 is 1.

Table 23. DAC Readback Register

Table 24. Bit Descriptions for the DAC Readback Register

Figure 41. DAC Readback Operation

ADC OPERATION

To set a pin as an ADC, set the appropriate bit in the ADC pin configuration register to 1 (se[e Table 25](#page-30-1) and [Table 26\)](#page-30-2). For example, setting Bit 0 to Bit 1 enables I/O0 as an ADC input. The ADC channels of th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) operate as a traditional multichannel ADC, where each serial transfer selects the next channel for conversion. Writing to the ADC sequence register (see [Table 27](#page-31-1) and [Table 28\)](#page-31-0) selects the ADC channels to be included in the sequence, and the REP bit determines if the sequence is repeated. The $\overline{\text{SYNC}}$ signal is used to frame the write to the converter on the SDI pin. The data that appears on the SDO pin during the initial write to the ADC sequence register is invalid. When the sequence register is written to, the ADC begins to track the first channel in the sequence. Tracking takes 500 ns; do not initiate a conversion until this time has passed. The next SYNC falling edge initiates a conversion on the selected channel. The subsequent SYNC falling edge begins clocking out the ADC result and also initiates the next conversion. The ADC operates with one cycle latency, thus the conversion result corresponding to each conversion is available one serial read cycle after the cycle in which the conversion was initiated.

If more than one channel is selected in the ADC sequence register, the ADC converts all selected channels sequentially in ascending

Table 25. ADC Pin Configuration Register

order on successive SYNC falling edges. Once all the selected channels in the control register are converted, the ADC repeats the sequence if the REP bit is set. If the REP bit is clear, the ADC goes three-state. [Figure 42](#page-31-2) t[o Figure 45](#page-32-0) show typical ADC modes of operation. I/O7 can be configured as a BUSY output pin to indicate when a conversion result is available. \overline{BUSY} goes low while a conversion takes place and goes high when the conversion result is available. The conversion result is clocked out on the SDO pin on the following read/write operation. For an ADC conversion, D15 is 0, D14 to D12 contain the ADC address, and D11 to D0 contain the 12-bit conversion result, as shown in [Table 29.](#page-31-3)

Changing an ADC Sequence

The channels included in an ADC sequence can be changed by first stopping an existing conversion sequence (see [Figure 46\)](#page-33-0). The ADC conversion sequence is stopped by clearing the REP, TEMP, and ADC7 to ADC0 bits in the ADC sequence register to 0.

As the command to stop the sequence is written, an ADC conversion is also taking place. This conversion must finish before a new sequence can be written to the ADC sequence register. Allow a minimum of 2 µs between starting the write to end the current sequence and starting the write to select a new sequence. After selecting the new sequence, allow an ADC track time of 500 ns before initiating the next conversion.

Bit(s)	Bit Name	Description		
D ₁₅	MSB	Set this bit to 0.		
D ₁₄ to D ₁₁	Register address	Set these bits to 0b0100.		
D ₁₀ to D ₈	Reserved	Reserved. Set these bits to 0.		
$D7$ to $D0$	ADC7 to ADC0	Select I/Ox pins as ADC inputs.		
		1: I/Ox is an ADC input.		
		0: I/Ox function is determined by the pin configuration registers (default).		

Table 26. Bit Descriptions for the ADC Pin Configuration Register

Table 27. ADC Sequence Register

Table 28. Bit Descriptions for the ADC Sequence Register

Table 29. ADC Conversion Result

¹ The ADC addresses are as follows: $000 =$ ADC0 ... $111 =$ ADC7.

Figure 42. Single-Channel ADC Conversion Sequence, No Repeat

Figure 43. Single-Channel, Repeating, ADC Conversion Sequence

Data Sheet **[AD5592R](https://www.analog.com/AD5592R?doc=AD5592R.pdf)**

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Figure 45. Multichannel, Repeating, ADC Conversion Sequence

[AD5592R](https://www.analog.com/AD5592R?doc=AD5592R.pdf) Data Sheet

GPIO OPERATION

Each of the I/Ox pins of th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) can operate as a general-purpose, digital input or output pin. The function of the pins is determined by writing to the appropriate bit in the GPIO read configuration and GPIO write configuration registers.

Setting Pins as Outputs

To set a pin as a general-purpose output, set the appropriate bit in the GPIO write configuration register to 1 (se[e Table 30](#page-34-1) and [Table 31\)](#page-34-2). For example, setting Bit 0 to Bit 1 enables I/O0 as a general-purpose output. The state of the output pin is controlled by setting or clearing the bits in the GPIO write data register (see [Table 34\)](#page-35-1). A data bit is ignored if it is written to a location that is not configured as an output.

The outputs can be independently configured as push/pull or open-drain outputs. When in a push/pull configuration, the output is driven to V_{DD} or GND, as determined by the data in the GPIO write data register. To set a pin as an open-drain output, set the appropriate bit in the GPIO open-drain configuration register to 1 (see [Table 32](#page-35-2) and [Table 33\)](#page-35-3). When in an open-drain configuration, the output is driven to GND when a data bit in the GPIO write data register sets the pin low. When the pin is set high, the output is not driven and must be pulled high by an external resistor. Open-drain configuration allows for multiple output pins to be tied together. If all the pins are normally high,

Table 30. GPIO Write Configuration Register

the open-drain configuration allows for one pin to pull down the others pins. This method is commonly used where multiple pins are used to trigger an alarm or an interrupt pin.

To change the state of the I/Ox pins, a write to the GPIO write data register is required. Setting a bit to 1 gives a Logic 1 on the selected output. Clearing a bit to 0 gives a Logic 0 on the selected output.

Setting Pins as Inputs

To set a pin as a general-purpose input, set the appropriate bit in the GPIO read configuration register to 1 (se[e Table 36](#page-35-4) and [Table 37\)](#page-35-0). For example, setting Bit 0 to Bit 1 enables I/O0 as a general-purpose input. To read the state of the general-purpose inputs, write to the GPIO read and configuration register to set Bit D10 to 1 and also any of Bit D7 to Bit D0 that correspond to a general-purpose input pin. The following SPI operation clocks out the state of any pins set as general-purpose inputs[. Figure 47](#page-33-1) shows an example where I/O4 to I/O7 are set as general-purpose inputs. I/O3 is assumed to be a DAC. To read the status of I/O7 to I/O4, Bit D10 and Bit D7 to Bit D4 are set to 1. To read the status of I/O5 and I/O4, only Bit D10, Bit D5, and Bit D4 need to be set to 1. The status of I/O7 and I/O6 are not read, and Bit D7 and Bit D6 are read as 0[. Figure 47](#page-33-1) also has a write to a DAC to show that other operations can be included when reading the status of the general-purpose pins.

Table 31. Bit Descriptions for the GPIO Write Configuration Register

Table 32.GPIO Open-Drain Configuration Register

Table 33. Bit Descriptions for the GPIO Open-Drain Configuration Register

Table 34. GPIO Write Data Register

Table 35. Bit Descriptions for the GPIO Write Data Register

Table 36. GPIO Read Configuration Register

Table 37. Bit Descriptions for the GPIO Read Configuration Register

THREE-STATE PINS

The I/Ox pins can be set to three-state by writing to the threestate configuration register, as shown in [Table 38](#page-36-2) and [Table 39.](#page-36-3)

85 kΩ PULL-DOWN RESISTOR PINS

The I/Ox pins can be connected to GND via a pull-down resistor (85 kΩ) by setting the appropriate bits in the pull-down configuration register, as shown in [Table 40](#page-36-4) and [Table 41.](#page-36-5)

Table 38. Three-State Configuration Register

Table 39. Bit Descriptions for the Three-State Configuration Register

Table 40. Pull-Down Configuration Register

Table 41. Bit Descriptions for the Pull-Down Configuration Register

POWER-DOWN MODE

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) have a power configuration register to reduce the power consumption when certain functions are not needed. The power-down register allows any channels set as DACs to be individually placed in a power-down state. When in a power-down state, the DAC outputs are three-state. When a DAC channel is put back into normal mode, the DAC output returns to its previous value. The internal reference and its

buffer are powered down by default and are enabled by setting the EN_REF bit in the power-down register. The internal reference voltage then appears at the VREF pin.

There is no dedicated power-down function for the ADC, but the ADC is automatically powered down if none of the I/Ox pins are selected as ADCs. The PD_ALL bit powers down all the DACs, the reference and its buffer, and the ADC simultaneously. [Table 42](#page-37-2) and [Table 43](#page-37-1) show the power-down register.

Table 43. Bit Descriptions for the Power-Down/Reference Control Register

RESET FUNCTION

Th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) can be reset to their default conditions by writing to the reset register, as shown in [Table 44.](#page-38-3) This write resets all registers to their default values and reconfigures the I/Ox pins to their default values (85 kΩ pull-down resistor to GND). The reset function takes 250 µs maximum; do not write new data to the [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) during this time. Th[e AD5592R](http://analog.com/AD5592R?doc=AD5592R.pdf) has a RESET pin that performs the same function. For normal operation, RESET is tied high. A falling edge on RESET triggers the reset function.

READBACK AND LDAC MODE REGISTER

The values contained in th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) registers can be read back to ensure that the registers are correctly set up. The register readback is initiated by writing to the readback and LDAC mode register with Bit D6 set to 1. Bit D5 to Bit D2 select which register is to be read back. The register data is clocked out of the [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) on the next SPI transfer.

Bit D1 to Bit D0 of the readback and LDAC mode register select the LDAC mode. The LDAC mode determines if data written to a DAC input register is also transferred to the DAC register. See the [LDAC Mode Operation](#page-28-1) section for details of the LDAC mode function.

Table 44. Software Reset

Table 45. Readback and LDAC Mode Register

Table 46. Bit Descriptions for the Readback and LDAC Mode Register

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APPLICATIONS INFORMATION

MICROPROCESSOR INTERFACING

Microprocessor interfacing to th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) is via a serial bus that uses a standard protocol compatible with DSPs and microcontrollers. The communications channel requires a 4-wire interface consisting of a clock signal, a data input signal, a data output signal, and a synchronization signal. The devices require a 16-bit data-word with data valid on the falling edge of SCLK.

[AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) TO SPI INTERFACE

The SPI interface of th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) is designed to be easily connected to industry-standard DSPs and microcontrollers. [Figure 48](#page-40-5) shows th[e AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) connected to the Analog Devices, Inc., [ADSP-BF531](https://www.analog.com/ADSP-BF531?doc=AD5592R.pdf) Blackfin® DSP. The Blackfin has an integrated SPI port that can be connected directly to the SPI pins of the [AD5592R/AD5592R-1.](http://analog.com/AD5592R?doc=AD5592R.pdf)

Figure 48[. ADSP-BF531](https://www.analog.com/ADSP-BF531?doc=AD5592R.pdf) SPI Interface

[AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) TO SPORT INTERFACE

The Analog Device[s ADSP-BF527](https://www.analog.com/ADSP-BF527?doc=AD5592R.pdf) has two serial ports (SPORT). [Figure 49](#page-40-6) shows how a SPORT interface can be used to control the [AD5592R/AD5592R-1.](http://analog.com/AD5592R?doc=AD5592R.pdf) Th[e ADSP-BF527](https://www.analog.com/ADSP-BF527?doc=AD5592R.pdf) has an SPI port that can also be used. This method is the same as when using the [ADSP-BF531.](https://www.analog.com/ADSP-BF531?doc=AD5592R.pdf)

Figure 49[. ADSP-BF527](https://www.analog.com/ADSP-BF527?doc=AD5592R.pdf) SPORT Interface

LAYOUT GUIDELINES

In any circuit where accuracy is important, careful consideration of the power supply and ground return layout helps to ensure the rated performance. The printed circuit board (PCB) on which th[e AD5592R](http://analog.com/AD5592R?doc=AD5592R.pdf) or the [AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) is mounted must be designed so that the [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) lie on the analog plane.

The [AD5592R/AD5592R-1](http://analog.com/AD5592R?doc=AD5592R.pdf) must have ample supply bypassing of 10 μ F in parallel with 0.1 μ F on each supply, located as close to the package as possible, ideally right up against the device. The 10 µF capacitors are the tantalum bead type. The 0.1 µF capacitor must have low effective series resistance (ESR) and low effective series inductance (ESI). Ceramic capacitors, for example, provide a low impedance path to ground at high frequencies to handle transient currents due to internal logic switching.

OUTLINE DIMENSIONS

Figure 50. 16-Lead Thin Shrink Small Outline Package [TSSOP] (RU-16) Dimensions shown in millimeters

Dimensions shown in millimeters

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Figure 52. 16-Ball Wafer Level Chip Scale Package [WLCSP] (CB-16-3) Dimensions shown in millimeters

ORDERING GUIDE

¹ Z = RoHS Compliant Part.

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