

**FEATURES**

- I<sub>DD</sub> total: 7.5 mA**
- Bandwidth RF/IF: 2.4 GHz/1.0 GHz**
- 2.7 V to 3.3 V power supply**
- Separate V<sub>P</sub> allows extended tuning voltage**
- Programmable dual modulus prescaler**  
RF and IF: 8/9, 16/17, 32/33, 64/65
- Programmable charge pump currents**
- 3-wire serial interface**
- Analog and digital lock detect**
- Fastlock mode**
- Power-down mode**
- 20-lead TSSOP and 20-lead LFCSP packages**

**APPLICATIONS**

- Wireless handsets (GSM, PCS, DCS, DSC1800, CDMA, WCDMA)**
- Base stations for wireless Radio (GSM, PCS, DCS, CDMA, WCDMA)**
- Wireless LANS**
- Cable TV tuners (CATV)**
- Communications test equipment**

**GENERAL DESCRIPTION**

The ADF4212L is a dual frequency synthesizer that can be used to implement local oscillators (LO) in the up-conversion and down-conversion sections of wireless receivers and transmitters. It can provide the LO for both the RF and IF sections. It consists of a low noise digital phase frequency detector (PFD), a precision charge pump, a programmable reference divider, programmable A and B counters, and a dual modulus prescaler (P/P + 1). The A (6-bit) and B (12-bit) counters, in conjunction with the dual modulus prescaler (P/P + 1), implement an N divider (N = BP + A). In addition, the 15-bit reference counter (R counter) allows selectable REF<sub>IN</sub> frequencies at the PFD input. A complete phase-locked loop (PLL) can be implemented if the synthesizer is used with external loop filters and voltage controlled oscillators (VCOs).

Control of all the on-chip registers is via a simple 3-wire interface with 1.8 V compatibility. The devices operate with a power supply ranging from 2.7 V to 3.3 V and can be powered down when not in use.

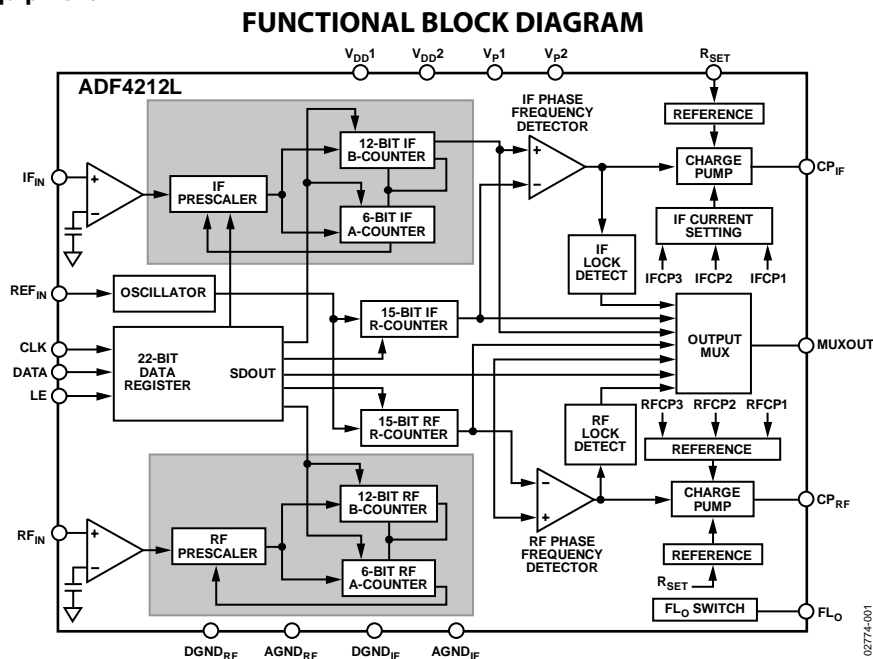


Figure 1.

Rev. E

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## REVISION HISTORY

### 1/14—Rev. D to Rev. E

Changes to Table 10..... 18

### 8/12—Rev. C to Rev D

Changed CP-20-1 to CP-20-6 .....Universal

Updated Outline Dimensions .....
 25 |

Changes to Ordering Guide .....
 26 |

### 11/10—Rev. B to Rev C

Changes to  $V_{P1}$ ,  $V_{P2}$  to GND Parameter, Table 4 and  $V_{P1}$ ,  $V_{P2}$   
to  $V_{DD1}$ ,  $V_{DD2}$  Parameter, Table 4 .....
 6 |

Changes to Ordering Guide .....
 26 |

### 9/08—Rev. A to Rev B

Updated Format..... Universal

Changes to Figure 1 and General Description Section .....
 1 |

Changes to Prescaler Output Frequency Parameter and RF

Input Frequency ( $RF_{IN}$ ) Parameter.....
 3 |

Changes to Table 3 and Figure 2.....
 5 |

Changes to Figure 4.....
 7 |

Changes to Figure 27, RF/IF A and B Counters Section, Pulse  
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 13 |

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Changes to Figure 32.....
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 24 |

Added PCB Design Guidelines for Lead Frame Chip Scale  
Package Section.....
 24 |

Updated Outline Dimensions.....
 25 |

Changes to Ordering Guide .....
 25 |

### 3/03—Data Sheet changed from REV. 0 to REV. A

Changes to General Description .....
 1 |

Changes to Specifications.....
 3 |

Changes to Table 9.....
 18 |

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 20 |

Changes to Figure 31.....
 23 |

### 11/02—Revision 0: Initial Version

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## SPECIFICATIONS

$V_{DD1} = V_{DD2} = 2.7\text{ V to }3.3\text{ V}$ ;  $V_{P1}, V_{P2} = V_{DD}$  to  $5.5\text{ V}$ ;  $AGND_{RF} = DGND_{RF} = AGND_{IF} = DGND_{IF} = 0\text{ V}$ ;  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted; dBm referred to  $50\ \Omega$ .

**Table 1.**

Parameter <sup>1</sup>	B Version	B Chips <sup>2</sup>	Unit	Test Conditions/Comments
<b>RF/IF CHARACTERISTICS</b>				
RF Input Frequency (RF <sub>IN</sub> )	0.2/2.4	0.2/2.4	GHz min/max	For lower frequencies, ensure that slew rate (SR) > 140 V/μs
RF Input Sensitivity	-10/0	-10/0	dBm min/max	$V_{DD} = 3\text{ V}$
IF Input Frequency (IF <sub>IN</sub> )	100/1000	100/1000	MHz min/max	
IF Input Sensitivity	-10/0	-10/0	dBm min/max	$V_{DD} = 3\text{ V}$
<b>MAXIMUM ALLOWABLE</b>				
Prescaler Output Frequency <sup>3</sup>	188	188	MHz max	
<b>REF<sub>IN</sub> CHARACTERISTICS</b>				
REF <sub>IN</sub> Input Frequency	10/150	10/150	MHz min/max	See Figure 26 for input circuit
REF <sub>IN</sub> Input Sensitivity	500 mV/V <sub>DD</sub>	500 mV/V <sub>DD</sub>	V p-p min/max	AC-coupled; when dc-coupled, 0 V to V <sub>DD</sub> maximum (CMOS compatible)
REF <sub>IN</sub> Input Capacitance	10	10	pF max	
REF <sub>IN</sub> Input Current	±100	±100	μA max	
<b>PHASE DETECTOR</b>				
Phase Detector Frequency <sup>4</sup>	75	75	MHz max	
<b>CHARGE PUMP</b>				
I <sub>CP</sub> Sink/Source				Programmable, see Table 10
High Value	5	5	mA typ	With R <sub>SET</sub> = 2.7 kΩ
Low Value	625	625	μA typ	
Absolute Accuracy	2	2	% typ	With R <sub>SET</sub> = 2.7 kΩ
R <sub>SET</sub> Range	1.5/5.6	1.5/5.6	kΩ min/max	
I <sub>CP</sub> Three-State Leakage Current	1	1	nA max	
Sink and Source Current Matching	6	6	% typ	0.5 V < V <sub>CP</sub> < V <sub>P</sub> - 0.5 V
I <sub>CP</sub> vs. V <sub>CP</sub>	2	2	% typ	0.5 V < V <sub>CP</sub> < V <sub>P</sub> - 0.5 V
I <sub>CP</sub> vs. Temperature	2	2	% typ	V <sub>CP</sub> = V <sub>P</sub> /2
<b>LOGIC INPUTS</b>				
V <sub>INH</sub> , Input High Voltage	1.4	1.4	V min	
V <sub>INL</sub> , Input Low Voltage	0.6	0.6	V max	
I <sub>INH</sub> /I <sub>INL</sub> , Input Current	±1	±1	μA max	
C <sub>IN</sub> , Input Capacitance	10	10	pF max	
<b>LOGIC OUTPUTS</b>				
V <sub>OH</sub> , Output High Voltage	1.4	1.4	V min	Open-drain 1 kΩ pull-up to 1.8 V
V <sub>OL</sub> , Output Low Voltage	0.4	0.4	V max	I <sub>OL</sub> = 500 μA
<b>POWER SUPPLIES</b>				
V <sub>DD1</sub>	2.7/3.3	2.7/3.3	V min/max	
V <sub>DD2</sub>	V <sub>DD1</sub>	V <sub>DD1</sub>	V min/max	
V <sub>P1</sub> , V <sub>P2</sub>	V <sub>DD1</sub> /5.5	V <sub>DD1</sub> /5.5	V min/max	
<b>I<sub>DD</sub> (RF and IF)<sup>5</sup></b>				
RF Only	5.0/6	5.0/6	mA typ/max	
IF Only	2.5/4	2.5/4	mA typ/max	
I <sub>P</sub> (I <sub>P1</sub> + I <sub>P2</sub> )	0.6	0.6	mA typ	
Low Power Sleep Mode	1	1	μA typ	

<sup>1</sup> Operating temperature range is as follows: B version: -40°C to +85°C.

<sup>2</sup> The B chip specifications are given as typical values.

<sup>3</sup> This is the maximum operating frequency of the CMOS counters. The prescaler value should be chosen to ensure that the RF input is divided down to a frequency less than this value.

<sup>4</sup> Guaranteed by design. Sample tested to ensure compliance.

<sup>5</sup> T<sub>A</sub> = 25°C. RF = 1 GHz; prescaler = 32/33. IF = 500 MHz; prescaler = 16/17.

$V_{DD1} = V_{DD2} = 2.7 \text{ V to } 3.3 \text{ V}$ ;  $V_{P1}, V_{P2} = V_{DD}$  to  $5.5 \text{ V}$ ;  $AGND_{RF} = DGND_{RF} = AGND_{IF} = DGND_{IF} = 0 \text{ V}$ ;  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted; dBm referred to  $50 \text{ V}$ .

**Table 2.**

Parameter <sup>1</sup>	B Version	B Chips <sup>2</sup>	Unit	Test Conditions/Comments
<b>NOISE CHARACTERISTICS</b>				
RF Phase Noise Floor <sup>3</sup>	-170	-170	dBc/Hz typ	25 kHz PFD frequency
	-162	-162	dBc/Hz typ	200 kHz PFD frequency
Phase Noise Performance <sup>4</sup>				VCO output
IF: 540 MHz Output <sup>5</sup>	-89	-89	dBc/Hz typ	1 kHz offset and 200 kHz PFD frequency
IF: 900 MHz Output <sup>6</sup>	-87	-87	dBc/Hz typ	1 kHz offset and 200 kHz PFD frequency
RF: 900 MHz Output <sup>6</sup>	-89	-89	dBc/Hz typ	1 kHz offset and 200 kHz PFD frequency
RF: 1750 MHz Output <sup>7</sup>	-84	-84	dBc/Hz typ	1 kHz offset and 200 kHz PFD frequency
RF: 2400 MHz Output <sup>8</sup>	-87	-87	dBc/Hz typ	1 kHz Offset and 1 MHz PFD frequency
<b>Spurious Signals</b>				
IF: 540 MHz Output <sup>5</sup>	-88/-90	-88/-90	dB typ	200 kHz/400 kHz and 200 kHz PFD frequency
IF: 900 MHz Output <sup>6</sup>	-90/-94	-90/-94	dB typ	200 kHz/400 kHz and 200 kHz PFD frequency
RF: 900 MHz Output <sup>6</sup>	-90/-94	-90/-94	dB typ	200 kHz/400 kHz and 200 kHz PFD frequency
RF: 1750 MHz Output <sup>7</sup>	-80/-82	-80/-82	dB typ	200 kHz/400 kHz and 200 kHz PFD frequency
RF: 2400 MHz Output <sup>8</sup>	-80/-82	-80/-82	dB typ	200 kHz/400 kHz and 200 kHz PFD frequency

<sup>1</sup> Operating temperature range is as follows: B version:  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ .

<sup>2</sup> The B Chip specifications are given as typical values.

<sup>3</sup> The synthesizer phase noise floor is estimated by measuring the in-band phase noise at the output of the VCO and subtracting  $20\log N$  (where N is the N divider value). See Figure 9.

<sup>4</sup> The phase noise is measured with the EVAL-ADF4212EB and the HP8562E spectrum analyzer. The spectrum analyzer provides the REFIN for the synthesizer ( $f_{REFOUT} = 10 \text{ MHz}$  at  $0 \text{ dBm}$ ).

<sup>5</sup>  $f_{REFIN} = 10 \text{ MHz}$ ;  $f_{PFD} = 200 \text{ kHz}$ ; offset frequency =  $1 \text{ kHz}$ ;  $f_{IF} = 540 \text{ MHz}$ ;  $N = 2700$ ; loop B/W =  $20 \text{ kHz}$

<sup>6</sup>  $f_{REFIN} = 10 \text{ MHz}$ ;  $f_{PFD} = 200 \text{ kHz}$ ; offset frequency =  $1 \text{ kHz}$ ;  $f_{RF} = 900 \text{ MHz}$ ;  $N = 4500$ ; loop B/W =  $20 \text{ kHz}$

<sup>7</sup>  $f_{REFIN} = 10 \text{ MHz}$ ;  $f_{PFD} = 200 \text{ kHz}$ ; offset frequency =  $1 \text{ kHz}$ ;  $f_{RF} = 1750 \text{ MHz}$ ;  $N = 8750$ ; loop B/W =  $20 \text{ kHz}$

<sup>8</sup>  $f_{REFIN} = 10 \text{ MHz}$ ;  $f_{PFD} = 1 \text{ MHz}$ ; offset frequency =  $1 \text{ kHz}$ ;  $f_{RF} = 2400 \text{ MHz}$ ;  $N = 9800$ ; loop B/W =  $20 \text{ kHz}$

**TIMING CHARACTERISTICS**

$V_{DD1} = V_{DD2} = 2.6 \text{ V to } 3.3 \text{ V}$ ;  $V_{P1}, V_{P2} = V_{DD}$  to  $5.5 \text{ V}$ ;  $AGND_{RF} = DGND_{RF} = AGND_{IF} = DGND_{IF} = 0 \text{ V}$ ;  $T_A = T_{MIN}$  to  $T_{MAX}$ , unless otherwise noted; dBm referred to  $50 \Omega$ .

Table 3.

Parameter <sup>1</sup>	Limit at $T_{MIN}$ to $T_{MAX}$ (B Version)	Unit	Test Conditions/Comments
$t_1$	20	ns min	LE setup time
$t_2$	10	ns min	Data to clock setup time
$t_3$	10	ns min	Data to clock hold time
$t_4$	25	ns min	Clock high duration
$t_5$	25	ns min	Clock low duration
$t_6$	10	ns min	Clock to LE setup time
$t_7$	20	ns min	LE pulse width

<sup>1</sup> Guaranteed by design but not production tested.

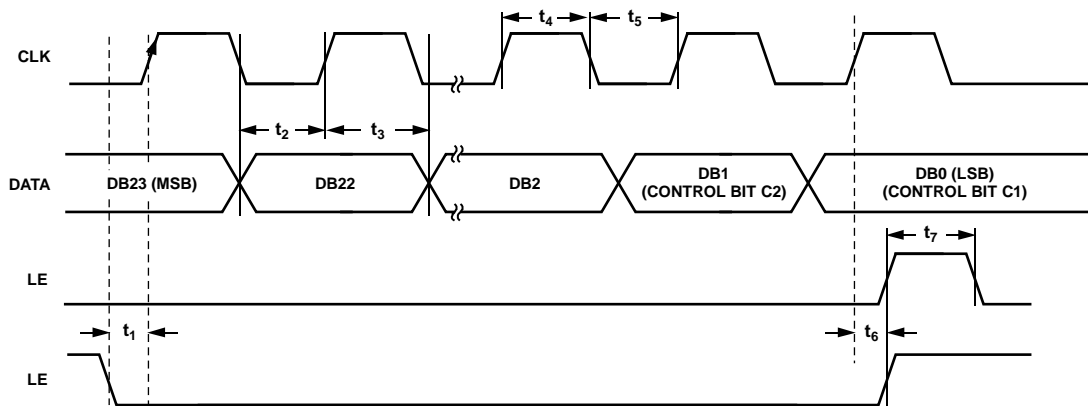


Figure 2. Timing Diagram

02774-002

## ABSOLUTE MAXIMUM RATINGS

$T_A = 25^\circ\text{C}$ , unless otherwise noted.

Table 4.

Parameter <sup>1,2</sup>	Rating
$V_{DD1}$ to GND	-0.3 V to +3.6 V
$V_{DD1}$ to $V_{DD2}$	-0.3 V to +0.3 V
$V_{P1}$ , $V_{P2}$ to GND	-0.3 V to +5.8 V
$V_{P1}$ , $V_{P2}$ to $V_{DD1}$ , $V_{DD2}$	-0.3 V to +5.8 V
Digital I/O Voltage to GND	-0.3 V to $DV_{DD} + 0.3$ V
Analog I/O Voltage to GND	-0.3 V to $V_{DD} + 0.3$ V
$REF_{IN}$ , $RF_{IN}$ , $IF_{IN}$ to GND	-0.3 V to $V_{DD} + 0.3$ V
Operating Temperature Range	
Industrial (B Version)	-40°C to +85°C
Storage Temperature Range	-65°C to +150°C
Maximum Junction Temperature	150°C
TSSOP $\theta_{JA}$ Thermal Impedance	150.4°C/W
LFCSP $\theta_{JA}$ Thermal Impedance (Paddle Soldered)	122°C/W
LFCSP $\theta_{JA}$ Thermal Impedance (Paddle Not Soldered)	216°C/W
Lead Temperature, Soldering	
Vapor Phase (60 sec)	215°C
Infrared (15 sec)	220°C

<sup>1</sup> This device is a high performance RF integrated circuit with an ESD rating of <2 kV, and is ESD sensitive. Proper precautions should be taken for handling and assembly.

<sup>2</sup> GND = AGND = DGND = 0 V.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### ESD CAUTION



#### ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

# PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

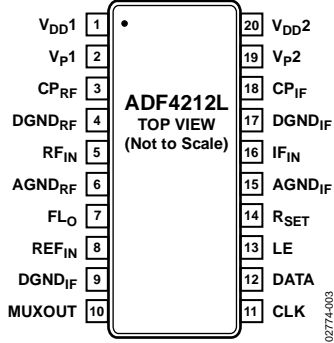


Figure 3. TSSOP Pin Configuration



**NOTES**

1. IT IS RECOMMENDED THAT THE EXPOSED PAD BE THERMALLY CONNECTED TO A COPPER PLANE FOR ENHANCED THERMAL PERFORMANCE. THE PAD SHOULD BE GROUNDING AS WELL.

Figure 4. LFCSP Pin Configuration

Table 5. Pin Function Descriptions

Mnemonic	Pin No.		Description
	TSSOP	LFCSP	
CP <sub>RF</sub>	3	1	RF Charge Pump Output. When enabled, this provides ±I <sub>CP</sub> to the external RF loop filter, which in turn drives the external RF VCO.
DGND <sub>RF</sub>	4	2	Digital Ground Pin for the RF Digital Circuitry.
RF <sub>IN</sub>	5	3	Input to the RF Prescaler. This small signal input is normally ac-coupled from the RF VCO.
AGND <sub>RF</sub>	6	4	Ground Pin for the RF Analog Circuitry.
FLO	7	5	Multiplexed Output of RF/IF Programmable or Reference Dividers, RF/IF Fastlock Mode. CMOS output.
REF <sub>IN</sub>	8	6	Reference Input. This is a CMOS input with a nominal threshold of V <sub>DD</sub> /2 and an equivalent input resistance of 100 kΩ. See Figure 26. This input can be driven from a TTL or CMOS crystal oscillator, or can be ac-coupled.
DGND <sub>IF</sub>	9, 17	7, 15	Digital Ground Pin for the IF Digital, Interface, and Control Circuitry.
MUXOUT	10	8	This multiplexer output allows either the IF/RF lock detect, the scaled RF, the scaled IF, or the scaled reference frequency to be accessed externally.
CLK	11	9	Serial Clock Input. This serial clock is used to clock in the serial data to the registers. The data is latched into the 24-bit shift register on the CLK rising edge. This input is a high impedance CMOS input.
DATA	12	10	Serial Data Input. The serial data is loaded MSB first with the two LSBs being the control bits. This input is a high impedance CMOS input.
LE	13	11	Load Enable, CMOS Input. When LE goes high, the data stored in the shift registers is loaded into one of the four latches, with the latch selected using the control bits.
RSET	14	12	Connecting a resistor between this pin and ground sets the maximum RF and IF charge pump output current. The nominal voltage potential at the R <sub>SET</sub> pin is 0.66 V. The relationship between I <sub>CP</sub> and R <sub>SET</sub> is, therefore, $I_{CP\ MAX} = \frac{13.5}{R_{SET}}$ where R <sub>SET</sub> = 2.7 kΩ and I <sub>CP MAX</sub> = 5 mA for both the RF and IF charge pumps.
AGND <sub>IF</sub>	15	13	Ground Pin for the IF Analog Circuitry.
IF <sub>IN</sub>	16	14	Input to the IF Prescaler. This small signal input is normally ac-coupled from the IF VCO.
CP <sub>IF</sub>	18	16	Output from the IF Charge Pump. This is normally connected to a loop filter that drives the input to an external VCO.
V <sub>p2</sub>	19	17	Power Supply for the IF Charge Pump. This should be greater than or equal to V <sub>DD2</sub> . In systems where V <sub>DD2</sub> is 3 V, it can be set to 5.5 V and used to drive a VCO with a tuning range up to 5.5 V.

Mnemonic	Pin No.		Description
	TSSOP	LFCSP	
V <sub>DD2</sub>	20	18	Power Supply for the IF, Digital, and Interface Section. Decoupling capacitors to the ground plane should be placed as close as possible to this pin. V <sub>DD2</sub> should have a value of between 2.6 V and 3.3 V. V <sub>DD2</sub> must have the same potential as V <sub>DD1</sub> .
V <sub>DD1</sub>	1	19	Power Supply for the RF Section. Decoupling capacitors to the ground plane should be placed as close as possible to this pin. V <sub>DD1</sub> should have a value of between 2.6 V and 3.3 V. V <sub>DD1</sub> must have the same potential as V <sub>DD2</sub> .
V <sub>p1</sub>	2	20	Power Supply for the RF Charge Pump. This should be greater than or equal to V <sub>DD1</sub> . In systems where V <sub>DD1</sub> is 3 V, it can be set to 5.5 V and used to drive a VCO with a tuning range up to 5.5 V.
EP		Exposed Pad	It is recommended that the exposed pad be thermally connected to a copper plane for enhanced thermal performance. The pad should be grounded as well.



TYPICAL PERFORMANCE CHARACTERISTICS

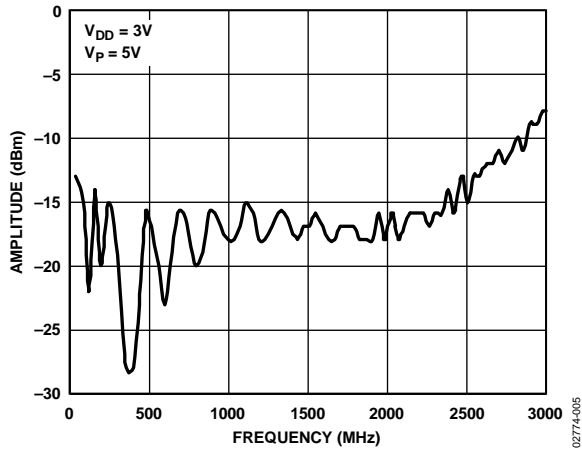


Figure 5. Input Sensitivity (RF Input)

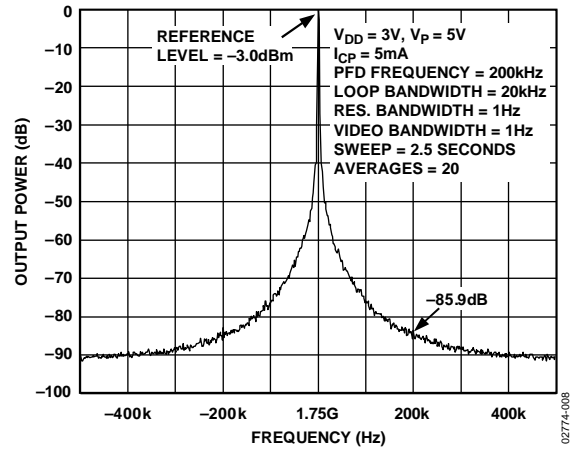


Figure 8. Reference Spurs, RF Side (1750 MHz, 200 kHz, 20 kHz)

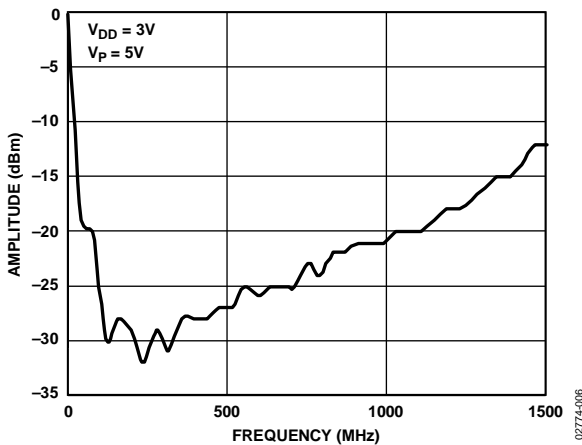


Figure 6. Input Sensitivity (IF Input)

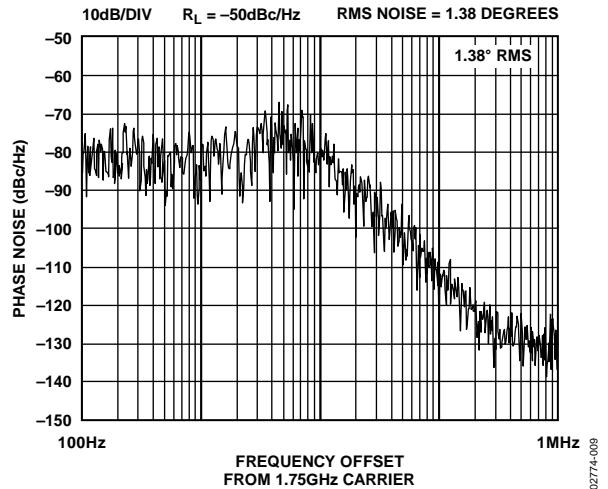


Figure 9. Integrated Phase Noise (1750 MHz, 200 kHz/20 kHz)

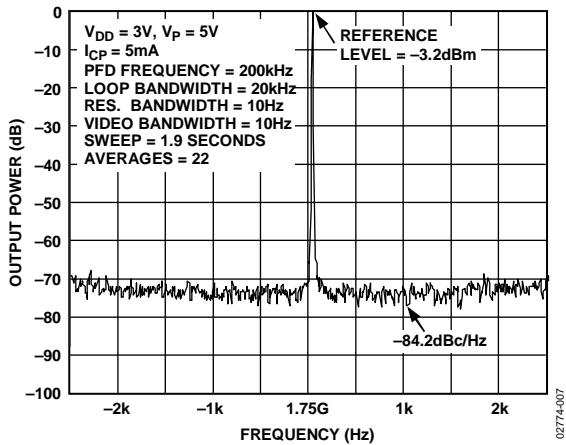


Figure 7. Phase Noise, RF Side (1750 MHz, 200 kHz, 20 kHz)

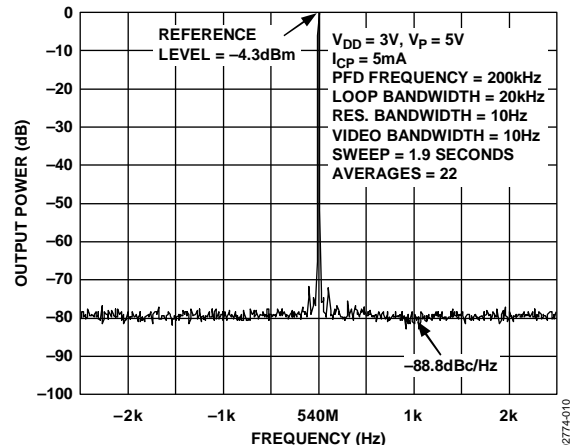


Figure 10. Phase Noise, IF Side (540 MHz, 200 kHz/20 kHz)



Figure 11. Reference Spurs, IF Side (540 MHz, 200 kHz, 20 kHz)

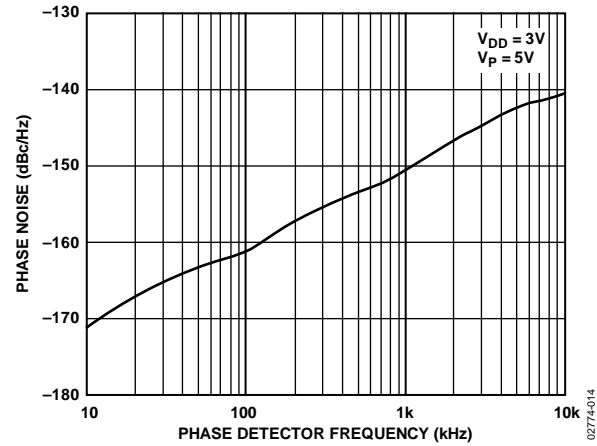


Figure 14. Phase Noise Referred to CP Output vs. PFD Frequency, IF Side

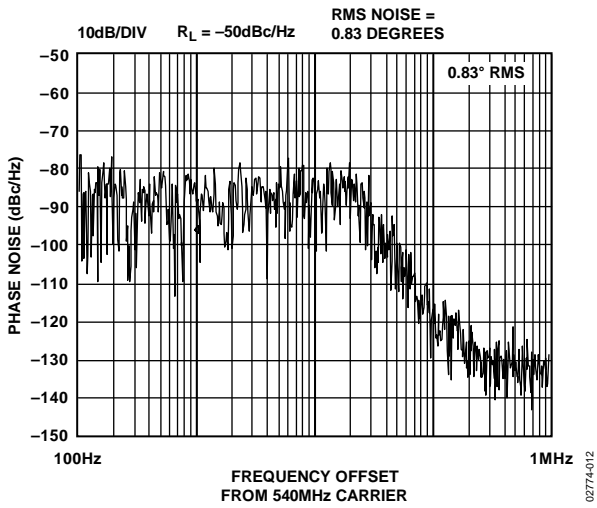


Figure 12. Integrated Phase Noise (540 MHz, 200 kHz/20 kHz)



Figure 15. RF Charge Pump Output Characteristics

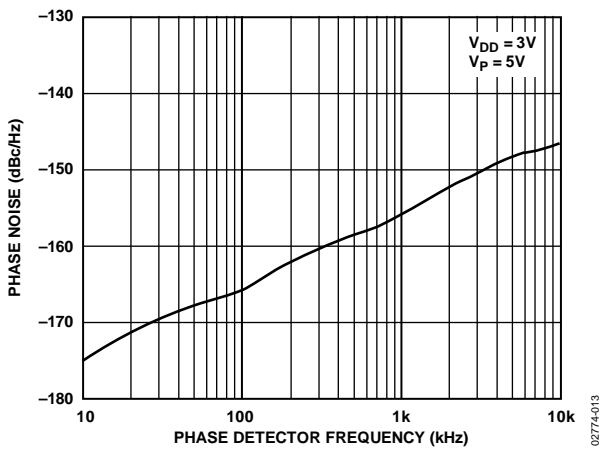


Figure 13. Phase Noise Referred to CP Output vs. PFD Frequency, RF Side



Figure 16. IF Charge Pump Output Characteristics

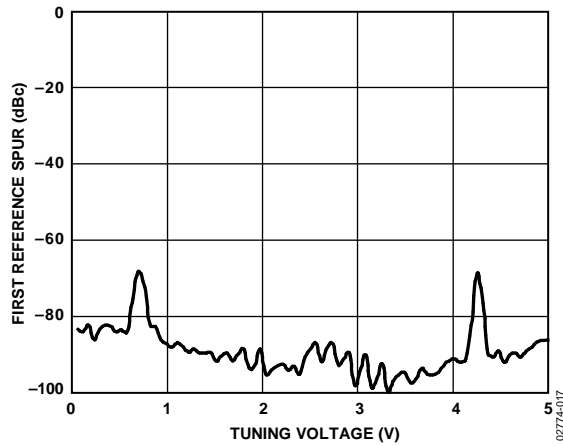


Figure 17. RF Reference Spurs (200 kHz) vs.  $V_{TUNE}$  (1750 MHz, 200 kHz, 20 kHz)



Figure 20. IF Phase Noise vs. Temperature (540 MHz, 200 kHz, 20 kHz)



Figure 18. IF Reference Spurs (200 kHz) vs.  $V_{TUNE}$  (1750 MHz, 200 kHz, 20 kHz)



Figure 21. RF Noise vs.  $V_{TUNE}$



Figure 19. RF Phase Noise vs. Temperature (1750 MHz, 200 kHz, 20 kHz)



Figure 22. IF Noise vs.  $V_{TUNE}$



Figure 23. RF Spurs vs. Temperature

FREQUENCY (MHz)	s11.REAL	s11.IMAG	FREQUENCY (MHz)	s11.REAL	s11.IMAG
50	0.97692	-0.021077	1550	0.561872	-0.646879
150	0.942115	-0.110459	1650	0.529742	-0.668172
250	0.961217	-0.085802	1750	0.514244	-0.702192
350	0.920667	-0.185830	1850	0.405754	-0.714541
450	0.897441	-0.245482	1950	0.379354	-0.703593
550	0.888164	-0.282399	2050	0.312959	-0.802878
650	0.850012	-0.305457	2150	0.322646	-0.803970
750	0.760189	-0.358884	2250	0.288881	-0.807055
850	0.767363	-0.541032	2350	0.199294	-0.758619
950	0.779511	-0.585687	2450	0.206914	-0.725029
1050	0.761034	-0.482539	2550	0.168344	-0.770837
1150	0.624825	-0.530106	2650	0.092764	-0.776619
1250	0.635364	-0.590526	2750	0.036125	-0.706197
1350	0.630242	-0.592498	2850	0.037007	-0.716939
1450	0.634506	-0.655932	2950	-0.053842	-0.736527

02774-025

Figure 25. S Parameter Data for the RF Input

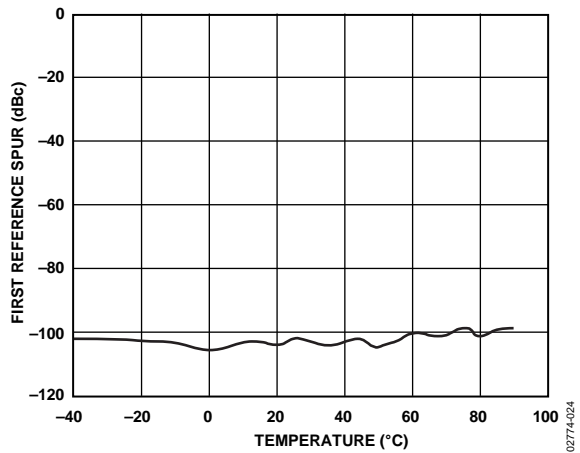


Figure 24. IF Spurs vs. Temperature

## CIRCUIT DESCRIPTION

### REFERENCE INPUT SECTION

The reference input stage is shown in Figure 26. SW1 and SW2 are normally closed switches. SW3 is normally open. When power-down is initiated, SW3 is closed and SW1 and SW2 are opened. This ensures that there is no loading of the REF<sub>IN</sub> pin on power-down.

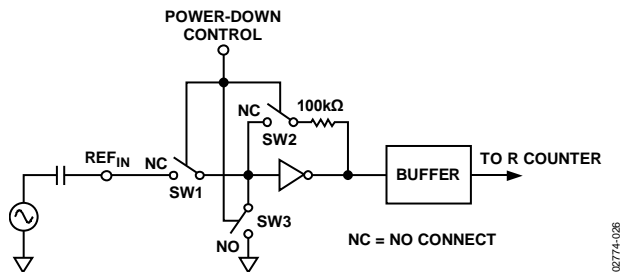


Figure 26. Reference Input Stage

### RF/IF INPUT STAGE

The RF/IF input stage is shown in Figure 27. It is followed by a two-stage limiting amplifier to generate the current mode logic (CML) clock levels needed for the prescaler.

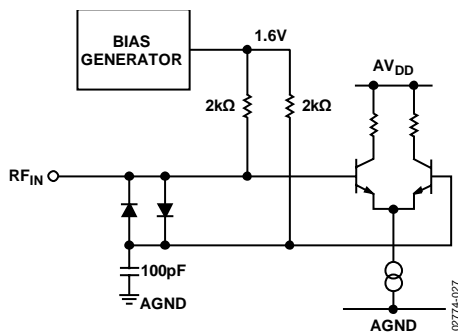


Figure 27. RF/IF Input Stage

### PRESCALER (P/P + 1)

The dual-modulus prescaler (P/P + 1), along with the A and B counters, enables the large division ratio, N, to be realized ( $N = PB + A$ ). The dual modulus prescaler, operating at CML levels, takes the clock from the RF/IF input stage and divides it down to a manageable frequency for the A and B CMOS counters in the RF and IF sections. The prescaler in both sections is programmable. It can be set in software to 8/9, 16/17, 32/33, or 64/65 (see Table 9 and Table 10). It is based on a synchronous 4/5 core.

### RF/IF A AND B COUNTERS

The A and B CMOS counters combine with the dual modulus prescaler to allow a wide ranging division ratio in the PLL feedback counter. The counters are specified to work when the prescaler output is 188 MHz or less. Thus, with an RF input frequency of 2.5 GHz, a prescaler value of 16/17 is valid, but a value of 8/9 is not valid.

### PULSE SWALLOW FUNCTION

The A and B CMOS counters, in conjunction with the dual modulus prescaler, make it possible to generate output frequencies that are spaced only by the reference frequency divided by R. The equation for the VCO frequency is as follows:

$$f_{VCO} = [(P \times B) + A] \times f_{REFIN}/R$$

where:

$f_{VCO}$  is the output frequency of external voltage controlled oscillator (VCO).

$P$  is the preset modulus of the dual modulus prescaler (8/9, 16/17, and so on).

$B$  is the preset divide ratio of the binary 12-bit counter (3 to 4095).

$A$  is the preset divide ratio of the binary 6-bit swallow counter (0 to 63).

$f_{REFIN}$  is the external reference oscillator frequency.

$R$  is the preset divide ratio of the binary 15-bit programmable reference counter (1 to 32,767).

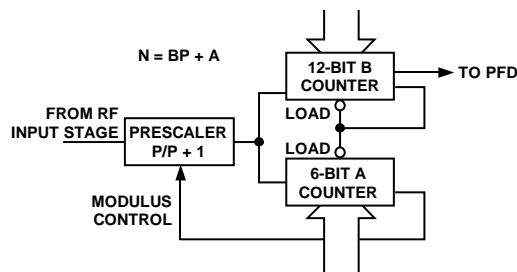


Figure 28. RF/IF A and B Counters

### RF/IF R COUNTER

The 15-bit RF/IF R counter allows the input reference frequency to be divided down to produce the input clock to the phase frequency detector (PFD). Division ratios from 1 to 16,383 are allowed.

**PHASE FREQUENCY DETECTOR (PFD) AND CHARGE PUMP**

The PFD takes inputs from the R counter and N counter and produces an output proportional to the phase and frequency difference between them. Figure 29 is a simplified schematic. The PFD includes a fixed delay element that sets the width of the antibacklash pulse. This is typically 3 ns. This pulse ensures that there is no dead zone in the PFD transfer function and gives a consistent reference spur level.

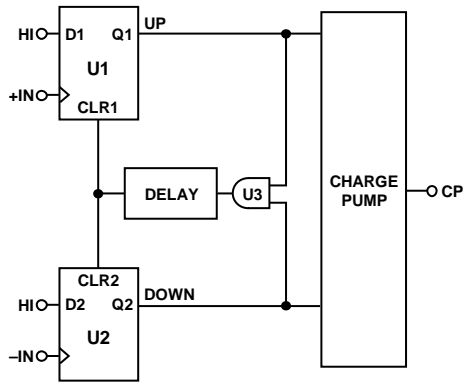


Figure 29. RF/IF PFD Simplified Schematic

**MUXOUT AND LOCK DETECT**

The output multiplexer on the ADF4212L allows the user to access various internal points on the chip. The state of MUXOUT is controlled by P3, P4, P11, and P12 (see Table 8 and Table 10). Figure 30 shows the MUXOUT section in block diagram form.

**LOCK DETECT**

MUXOUT can be programmed for two types of lock detect: digital lock detect and analog lock detect. Digital lock detect is active high. It is set high when the phase error on three consecutive phase detector cycles is less than 15 ns. It stays set high until a phase error of greater than 25 ns is detected on any subsequent PD cycle.

The N-channel open-drain analog lock detect should be operated with an external pull-up resistor of 10 kΩ nominal. When lock has been detected, it is high with narrow, low-going pulses.



Figure 30. MUXOUT Schematic

**RF/IF INPUT SHIFT REGISTER**

The ADF4212L digital section includes a 24-bit input shift register, a 15-bit IF R counter, and an 18-bit IF N counter (comprising a 6-bit IF A counter and a 12-bit IF B counter). Also present is a 15-bit RF R counter and an 18-bit RF N counter (comprising a 6-bit RF A counter and a 12-bit RF B counter). Data is clocked into the 24-bit shift register on each rising edge of CLK. The data is clocked in MSB first. Data is transferred from the shift register to one of four latches on the rising edge of LE. The destination latch is determined by the state of the two control bits (C2, C1) in the shift register. These are the two LSBs, DB1 and DB0, as shown in the timing diagram of Figure 2. The truth table for these bits is shown in Table 6.

Table 7 shows a summary of how the latches are programmed.

Table 6. C2, C1 Truth Table

Control Bits		Data Latch
C2	C1	
0	0	IF R counter
0	1	IF N counter (A and B)
1	0	RF R counter
1	1	RF N counter (A and B)

Table 7. Latch Summary

IF R COUNTER LATCH

IF CP CURRENT SETTING			IF F <sub>0</sub>	LOCK DETECT PRECISION	THREE-STATE CP	IF PD POLARITY	15-BIT REFERENCE COUNTER															CONTROL BITS	
DB23	DB22	DB21	DB20	DB19	DB18	DB17	DB16	DB15	DB14	DB13	DB12	DB11	DB10	DB9	DB8	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
IFCP2	IFCP1	IFCP0	P4	P3	P2	P1	R15	R14	R13	R12	R11	R10	R9	R8	R7	R6	R5	R4	R3	R2	R1	C2 (0)	C1 (0)

IF N COUNTER LATCH

IF CP GAIN	IF POWER-DOWN	IF PRESCALER			12-BIT B COUNTER												6-BIT A COUNTER						CONTROL BITS	
DB23	DB22	DB21	DB20	DB19	DB18	DB17	DB16	DB15	DB14	DB13	DB12	DB11	DB10	DB9	DB8	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0	
P8	P7	P6	P5	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	A6	A5	A4	A3	A2	A1	C2 (0)	C1 (1)	

RF R COUNTER LATCH

RF CP CURRENT SETTING			RF F <sub>0</sub>	RF LOCK DETECT	THREE-STATE CP	RF PD POLARITY	15-BIT RF REFERENCE COUNTER															CONTROL BITS	
DB23	DB22	DB21	DB20	DB19	DB18	DB17	DB16	DB15	DB14	DB13	DB12	DB11	DB10	DB9	DB8	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
RFCP2	RFCP1	RFCP0	P12	P11	P10	P9	R15	R14	R13	R12	R11	R10	R9	R8	R7	R6	R5	R4	R3	R2	R1	C2 (1)	C1 (0)

RF N COUNTER LATCH

RF C GAIN	RF POWER-DOWN	RF PRESCALER			12-BIT B COUNTER												6-BIT A COUNTER						CONTROL BITS	
DB23	DB22	DB21	DB20	DB19	DB18	DB17	DB16	DB15	DB14	DB13	DB12	DB11	DB10	DB9	DB8	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0	
P17	P16	P15	P14	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	A6	A5	A4	A3	A2	A1	C2 (1)	C1 (1)	

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IF R COUNTER LATCH

Table 8. IF R Counter Latch Map

IF CP CURRENT SETTING			IF Fo	LOCK DETECT PRECISION	THREE-STATE CP	IF PD POLARITY	15-BIT IF REFERENCE COUNTER															CONTROL BITS	
DB23	DB22	DB21	DB20	DB19	DB18	DB17	DB16	DB15	DB14	DB13	DB12	DB11	DB10	DB9	DB8	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
IFCP2	IFCP1	IFCP0	P4	P3	P2	P1	R15	R14	R13	R12	R11	R10	R9	R8	R7	R6	R5	R4	R3	R2	R1	C2 (0)	C1 (0)

R15	R14	R13	.....	R3	R2	R1	DIVIDE RATIO
0	0	0	.....	0	0	1	1
0	0	0	.....	0	1	0	2
0	0	0	.....	0	1	1	3
0	0	0	.....	1	0	0	4
.	.	.	.....	.	.	.	.
.	.	.	.....	.	.	.	.
1	1	1	.....	1	0	0	32764
1	1	1	.....	1	0	1	32765
1	1	1	.....	1	1	0	32766
1	1	1	.....	1	1	1	32767

P1	IF PD POLARITY
0	NEGATIVE
1	POSITIVE

P2	CHARGE PUMP OUTPUT
0	NORMAL
1	THREE-STATE

P12	P11	P4	P3	MUXOUT
0	0	0	0	LOGIC LOW STATE
0	0	0	1	IF ANALOG LOCK DETECT
0	0	1	0	IF REFERENCE DIVIDER OUTPUT
0	0	1	1	IF N DIVIDER OUTPUT
0	1	0	0	RF ANALOG LOCK DETECT
0	1	0	1	RF/IF ANALOG LOCK DETECT
0	1	1	0	IF DIGITAL LOCK DETECT
0	1	1	1	LOGIC HIGH STATE
1	0	0	0	RF REFERENCE DIVIDER OUTPUT
1	0	0	1	RF N DIVIDER OUTPUT
1	0	1	0	THREE-STATE OUTPUT
1	0	1	1	IF COUNTER RESET
1	1	0	0	RF DIGITAL LOCK DETECT
1	1	0	1	RF/IF DIGITAL LOCK DETECT
1	1	1	0	RF COUNTER RESET
1	1	1	1	IF AND RF COUNTER RESET

IFCP2	IFCP1	IFCP0	I <sub>CP</sub> (mA)		
			1.5kΩ	2.7kΩ	5.6kΩ
0	0	0	1.1250	0.625	0.301
0	0	1	2.2500	1.250	0.602
0	1	0	3.3750	1.875	0.904
0	1	1	4.5000	2.500	1.205
1	0	0	5.6250	3.125	1.506
1	0	1	6.7500	3.750	1.808
1	1	0	7.7875	4.375	2.109
1	1	1	9.0000	5.000	2.411



IF N COUNTER LATCH

Table 9. IF N Counter Latch Map



RF R COUNTER LATCH

Table 10. RF R Counter Latch Map

RF CP CURRENT SETTING			RF F <sub>0</sub>	RF LOCK DETECT	THREE-STATE CP	RF PD POLARITY	15-BIT RF REFERENCE COUNTER														CONTROL BITS		
DB23	DB22	DB21	DB20	DB19	DB18	DB17	DB16	DB15	DB14	DB13	DB12	DB11	DB10	DB9	DB8	DB7	DB6	DB5	DB4	DB3	DB2	DB1	DB0
RFCP2	RFCP1	RFCP0	P12	P11	P10	P9	R15	R14	R13	R12	R11	R10	R9	R8	R7	R6	R5	R4	R3	R2	R1	C2 (1)	C1 (0)

R15	R14	R13	.....	R3	R2	R1	DIVIDE RATIO
0	0	0	.....	0	0	1	1
0	0	0	.....	0	1	0	2
0	0	0	.....	0	1	1	3
0	0	0	.....	1	0	0	4
.	.	.	.....	.	.	.	.
.	.	.	.....	.	.	.	.
1	1	1	.....	1	0	0	32764
1	1	1	.....	1	0	1	32765
1	1	1	.....	1	1	0	32766
1	1	1	.....	1	1	1	32767

P9	RF PD POLARITY
0	NEGATIVE
1	POSITIVE

P10	CHARGE PUMP OUTPUT
0	NORMAL
1	THREE-STATE

P12	P11	P4	P3	MUXOUT
FROM IF R LATCH				
0	0	0	0	LOGIC LOW STATE
0	0	0	1	IF ANALOG LOCK DETECT
0	0	1	0	IF REFERENCE DIVIDER OUTPUT
0	0	1	1	IF N DIVIDER OUTPUT
0	1	0	0	RF ANALOG LOCK DETECT
0	1	0	1	RF/IF ANALOG LOCK DETECT
0	1	1	0	IF DIGITAL LOCK DETECT
0	1	1	1	LOGIC HIGH STATE
1	0	0	0	RF REFERENCE DIVIDER OUTPUT
1	0	0	1	RF N DIVIDER OUTPUT
1	0	1	0	THREE-STATE OUTPUT
1	0	1	1	IF COUNTER RESET
1	1	0	0	RF DIGITAL LOCK DETECT
1	1	0	1	RF/IF DIGITAL LOCK DETECT
1	1	1	0	RF COUNTER RESET
1	1	1	1	IF AND RF COUNTER RESET

RFCP2	RFCP1	RFCP0	I <sub>CP</sub> (mA)		
			1.5kΩ	2.7kΩ	5.6kΩ
0	0	0	1.1250	0.625	0.301
0	0	1	2.2500	1.250	0.602
0	1	0	3.3750	1.875	0.904
0	1	1	4.5000	2.500	1.205
1	0	0	5.6250	3.125	1.506
1	0	1	6.7500	3.750	1.808
1	1	0	7.7875	4.375	2.109
1	1	1	9.0000	5.000	2.411

RF N COUNTER LATCH

Table 11. RF N Counter Latch Map



## PROGRAM MODES

Table 8 and Table 10 show how to set up the program modes in the [ADF4212L](#). The following should be noted:

- IF and RF analog lock detect indicate when the PLL is in lock. When the loop is locked and either IF or RF analog lock detect is selected, the MUXOUT pin shows a logic high with narrow, low-going pulses. When the IF/RF analog lock detect is chosen, the locked condition is indicated only when both IF and RF loops are locked.
- The IF counter reset mode resets the R, A, and B counters in the IF section and puts the IF charge pump into three-state mode. The RF counter reset mode resets the R, A, and B counters in the RF section and puts the RF charge pump into three-state. The IF and RF counter reset mode does both of the above. Upon removal of the reset bits, the A and B counters resume counting in close alignment with the R counter. (Maximum error is one prescaler output cycle.)
- The fastlock mode uses MUXOUT to switch a second loop filter damping resistor to ground during fastlock operation. Activation of fastlock occurs whenever RF CP gain in the RF reference counter is set to 1.

### IF AND RF POWER-DOWN

It is possible to program the [ADF4210](#) family for either synchronous or asynchronous power-down on either the IF or RF side.

#### **Synchronous IF Power-Down**

Programming a 1 to P7 of the [ADF4212L](#) initiates a power-down. If P2 of the [ADF4212L](#) has been set to 0 (normal operation), a synchronous power-down is conducted. The device automatically puts the charge pump into three-state mode and completes the power-down.

#### **Asynchronous IF Power-Down**

If P2 of the [ADF4212L](#) has been set to 1 (the IF charge pump in three-state mode) and P7 is subsequently set to 1, an asynchronous power-down is conducted. The device goes into power-down on the rising edge of LE, which latches the 1 to the IF power-down bit (P7).

#### **Synchronous RF Power-Down**

Programming a 1 to P16 of the [ADF4212L](#) initiates a power-down. If P10 of the [ADF4212L](#) has been set to 0 (normal operation), a synchronous power-down is conducted. The device automatically puts the charge pump into three-state mode and then completes the power-down.

#### **Asynchronous RF Power-Down**

If P10 of the [ADF4212L](#) has been set to 1 (the RF charge pump in three-state mode) and P16 is subsequently set to 1, an asynchronous power-down is conducted. The device goes into power-down on the rising edge of LE, which latches the 1 to the RF power-down bit (P16).

Activation of either synchronous or asynchronous power-down forces the IF/RF loop's R and A/B dividers to their load state conditions, and the IF/RF input section is debiased to a high impedance state.

The REF<sub>IN</sub> oscillator circuit is disabled only if both the IF and RF power-downs are set.

The input register and latches remain active and are capable of loading and latching data during all power-down modes.

The IF/RF section of the device returns to normal powered-up operation immediately upon LE latching a 0 to the appropriate power-down bit.

### IF SECTION

#### **Programmable IF Reference (R) Counter**

If Control Bits[C2:C1] = 00, the data is transferred from the input shift register to the 15-bit IF R counter. Table 8 shows the input shift register data format for the IF R counter and the divide ratios that are possible.

#### **IF Phase Detector Polarity**

P1 sets the IF phase detector polarity. When the IF VCO characteristics are positive, P1 should be set to 1. When the IF VCO characteristics are negative, it should be set to 0. See Table 8.

#### **IF Charge Pump Three-State**

P2 puts the IF charge pump into three-state mode when programmed to a 1. It should be set to 0 for normal operation. See Table 8.

#### **IF Program Modes**

Table 8 shows how to set up the program modes in the [ADF4212L](#).

#### **IF Charge Pump Currents**

IFCP2, IFCP1, and IFCP0 program the current setting for the IF charge pump. See Table 8.

#### **Programmable IF N Counter**

If Control Bits[C2:C1] = 01, the data in the input register is used to program the IF N (A + B) counter. The N counter consists of a 6-bit swallow counter (A counter) and 12-bit programmable counter (B counter). Table 9 shows the input register data format for programming the IF A and B counters and the divide ratios possible.

#### **IF Prescaler Value**

P5 and P6 in the IF N counter latch set the IF prescaler values. See Table 9.

**IF Power-Down**

Table 9 shows the power-down bits in the [ADF4212L](#).

**IF Fastlock**

The IF CP gain bit (P8) of the IF N counter latch register in the [ADF4212L](#) is the fastlock enable bit. Only when P8 is set to 1 is IF fastlock enabled. When fastlock is enabled, the IF CP current is set to the maximum value. Also, an extra loop filter damping resistor to ground is switched in using the FL<sub>o</sub> pin, thus compensating for the change in loop characteristics while in fastlock. Because the IF CP gain bit is contained in the IF N counter, only one write is needed to both program a new output frequency and initiate fastlock. To come out of fastlock, the IF CP gain bit on the IF N counter latch register must be set to 0 (see Table 9).

**RF SECTION****Programmable RF Reference (R) Counter**

If Control Bits[C2: C1] = 10, the data is transferred from the input shift register to the 15-bit RF R counter. Table 10 shows the input shift register data format for the RF R counter and the divide ratios possible.

**RF Phase Detector Polarity**

P9 sets the IF phase detector polarity. When the RF VCO characteristics are positive, P9 should be set to 1. When they are negative, it should be set to 0 (see Table 10).

**RF Charge Pump Three-State**

P10 puts the RF charge pump into three-state mode when programmed to a 1. It should be set to 0 for normal operation (see Table 10).

**RF Program Modes**

Table 10 shows how to set up the program modes in the [ADF4212L](#).

**RF Charge Pump Currents**

RFCP2, RFCP1, and RFCP0 program the current setting for the RF charge pump. See Table 10.

**Programmable RF N Counter**

If Control Bits[C2:C1] = 11, the data in the input register is used to program the RF N (A + B) counter. The N counter consists of a 6-bit swallow counter (A counter) and a 12-bit programmable counter (B counter). Table 11 shows the input register data format for programming the RF N counter and the divide ratios that are possible.

**RF Prescaler Value**

P14 and P15 in the RF N counter latch set the RF prescaler values. See Table 11.

**RF Power-Down**

Table 11 shows the power-down bits in the [ADF4212L](#).

**RF Fastlock**

The RF CP gain bit (P17) of the RF N counter latch register in the [ADF4212L](#) is the fastlock enable bit. Only when P17 is set to 1 is IF fastlock enabled. When fastlock is enabled, the RF CP current is set to the maximum value. Also, an extra loop filter damping resistor to ground is switched in using the FL<sub>o</sub> pin, thus compensating for the change in loop characteristics while in fastlock. Because the RF CP gain bit is contained in the RF N counter, only one write is needed to both program a new output frequency and initiate fastlock. To come out of fastlock, the RF CP gain bit on the RF N counter latch register must be set to 0. See Table 11.

## APPLICATIONS INFORMATION

### LOCAL OSCILLATOR FOR GSM HANDSET RECEIVER

Figure 31 shows the [ADF4212L](#) being used with a VCO to produce the required LOs for a GSM base station transmitter or receiver. The reference input signal is applied to the circuit at  $F_{REF_{IN}}$  and, in this case, is terminated in  $50\ \Omega$ . Typical GSM systems have a 13 MHz TCXO driving the reference input without any  $50\ \Omega$  termination. To have a channel spacing of 200 kHz (the GSM standard), the reference input must be divided by 65, using the on-chip reference.

The RF output frequency range is 880 MHz to 915 MHz. The loop filter is designed to give a 20 kHz loop bandwidth. The filter is set up for a 5 mA charge pump current, and the VCO sensitivity is 12 MHz/V. The IF output is fixed at 540 MHz. The filter is again designed to have a bandwidth of 20 kHz, and the system is programmed to give channel steps of 200 kHz.



DECOUPLING CAPACITORS (22 $\mu$ F/10pF) ON  $V_{DD}$ ,  $V_P$  OF THE ADF4212L AND ON  $V_{CC}$  OF THE VCOs HAVE BEEN OMITTED FROM THE DIAGRAM TO AID CLARITY.

Figure 31. GSM Handset Receiver Local Oscillator Using the [ADF4212L](#)

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**WIDEBAND PLL**

Many of the wireless applications for synthesizers and VCOs in PLLs are narrow band in nature. These applications include the various wireless standards such as GSM, DSC1800, CDMA, or WCDMA. In each of these cases, the total tuning range for the LO is less than 100 MHz. However, there are also wideband applications where the LO can have up to an octave tuning range. For example, cable television tuners have a total range of about 400 MHz. Figure 32 shows an application where the ADF4212L is used to control and program the Micronetics M3500-1324. The loop filter was designed for an RF output of 2100 MHz, a loop bandwidth of 40 kHz, a PFD frequency of 1 MHz,  $I_{CP}$  of 10 mA (2.5 mA synthesizer  $I_{CP}$  multiplied by the gain factor of 4), VCO  $K_D$  of 80 MHz/V (sensitivity of the M3500-1324 at an output of 2100 MHz), and a phase margin of 45 degrees.

In narrow-band applications, there is generally a small variation in output frequency (generally less than 10%) and a small variation in VCO sensitivity over the range (typically <10%). However, in wideband applications, both of these parameters have a much greater variation, which changes the loop bandwidth. This, in turn, can affect stability and lock time. By changing the programmable  $I_{CP}$ , it is possible to obtain compensation for these varying loop conditions and to ensure that the loop is always operating close to optimal conditions.

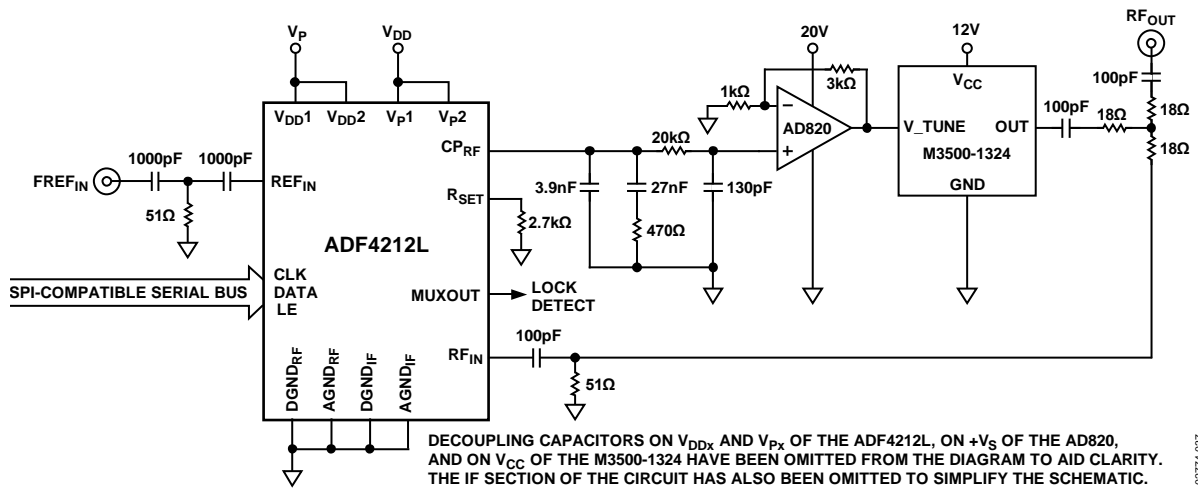


Figure 32. Wideband PLL Circuit

02774-037

**INTERFACING**

The ADF4212L has a simple SPI-compatible interface for writing to the device. CLK, DATA, and LE control the data transfer. When latch enable (LE) goes high, the 22 bits that have been clocked into the input register on each rising edge of CLK are transferred to the appropriate latch. See Figure 2 for the timing diagram and Table 6 for the latch truth table.

The maximum allowable serial clock rate is 20 MHz. This means that the maximum update rate possible for the device is 909 kHz or one update every 1.1  $\mu$ s, which is more than adequate for systems that have typical lock times in hundreds of microseconds.

**ADuC812 Interface**

Figure 33 shows the interface between the ADF4212L and the ADuC812 MicroConverter<sup>®</sup>. Because the ADuC812 is based on an 8051 core, this interface can be used with any 8051-based microcontroller. The microconverter is set up for SPI (serial port interface) master mode with CPHA = 0. To initiate the operation, the I/O port driving LE is brought low. Each latch of the ADF4212L needs a 24-bit word. This is accomplished by writing three 8-bit bytes from the microconverter to the device. When the third byte has been written, the LE input should be brought high to complete the transfer.

When first applying power to the ADF4212L, four writes (one each to the R counter latch and the N counter latch for both the IF and RF sides) are required for the output to become active.

When operating in the mode described, the maximum SCLOCK rate of the ADuC812 is 4 MHz. This means that the maximum rate at which the output frequency can be changed is 180 kHz.

**ADSP-2181 Interface**

Figure 34 shows the interface between the ADF4212L and the ADSP-21xx digital signal processor. As previously described, the ADF4212L needs a 24-bit serial word for each latch write. The easiest way to accomplish this with the ADSP-21xx family is to use the autobuffered transmit mode of operation with alternate framing. This provides a means for transmitting an entire block of serial data before an interrupt is generated. Set up the word length for eight bits and use three memory locations for each 24-bit word. To program each 24-bit latch, store the three 8-bit bytes, enable the autobuffered mode, and then write to the transmit register of the DSP. This last operation initiates the autobuffer transfer.

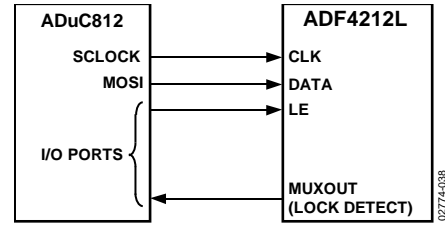


Figure 33. ADuC812 to ADF4212L Interface

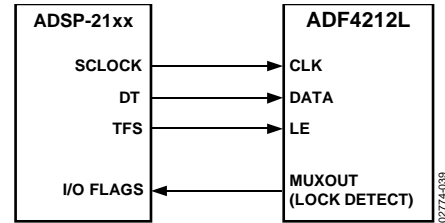


Figure 34. ADSP-21xx to ADF4212L Interface

**PCB DESIGN GUIDELINES FOR LEAD FRAME CHIP SCALE PACKAGE**

The lands on the LFCSP (CP-20-6) are rectangular. The printed circuit board (PCB) pad for these should be 0.1 mm longer than the package land length and 0.05 mm wider than the package land width. The land should be centered on the pad. This ensures that the solder joint size is maximized. The bottom of the LFCSP has a central thermal pad.

The thermal pad on the PCB should be at least as large as the exposed pad. On the PCB, there should be a clearance of at least 0.25 mm between the thermal pad and the inner edges of the pad pattern. This ensures that shorting is avoided.

Thermal vias can be used on the PCB thermal pad to improve thermal performance of the package. If vias are used, they should be incorporated in the thermal pad at 1.2 mm pitch grid. The via diameter should be between 0.3 mm and 0.33 mm, and the via barrel should be plated with 1 oz copper to plug the via.

The user should connect the PCB thermal pad to PCB ground.

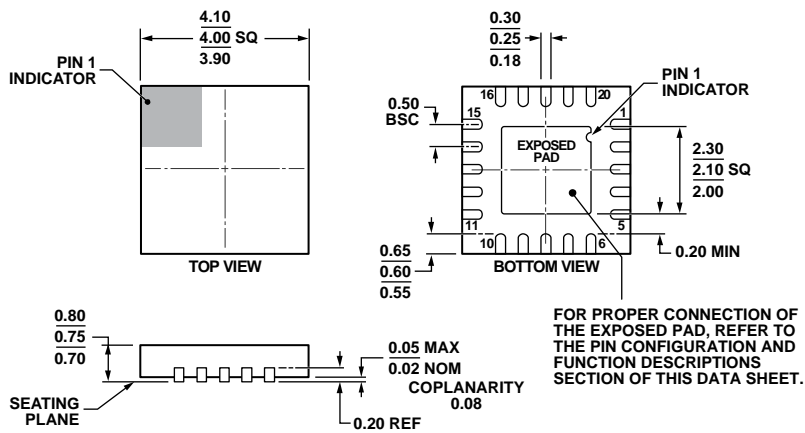


# OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-153-AC

Figure 35. 20-Lead Thin Shrink Small Outline Package [TSSOP] (RU-20)  
Dimensions shown in millimeters



COMPLIANT TO JEDEC STANDARDS MO-220-WGGD-1.

Figure 36. 20-Lead Lead Frame Chip Scale Package [LF CSP\_WQ] 4 mm x 4 mm Body, Very Very Thin Quad (CP-20-6)  
Dimensions shown in millimeters

08-16-2010-B

**ORDERING GUIDE**

<b>Model<sup>1</sup></b>	<b>Temperature Range</b>	<b>Package Description</b>	<b>Package Option<sup>2</sup></b>
ADF4212LBRUZ	-40°C to +85°C	20-Lead TSSOP	RU-20
ADF4212LBRUZ-RL	-40°C to +85°C	20-Lead TSSOP	RU-20
ADF4212LBRUZ-RL7	-40°C to +85°C	20-Lead TSSOP	RU-20
ADF4212LBCPZ	-40°C to +85°C	20-Lead LFCSP_WQ	CP-20-6
ADF4212LBCPZ-RL	-40°C to +85°C	20-Lead LFCSP_WQ	CP-20-6
ADF4212LBCPZ-RL7	-40°C to +85°C	20-Lead LFCSP_WQ	CP-20-6

<sup>1</sup> Z = RoHS Compliant Part.

<sup>2</sup> CP-20-6 package was formerly CP-20-1 package.

**NOTES**

**NOTES**

Компания «Life Electronics» занимается поставками электронных компонентов импортного и отечественного производства от производителей и со складов крупных дистрибьюторов Европы, Америки и Азии.

С конца 2013 года компания активно расширяет линейку поставок компонентов по направлению коаксиальный кабель, кварцевые генераторы и конденсаторы (керамические, пленочные, электролитические), за счёт заключения дистрибьюторских договоров

Мы предлагаем:

- Конкурентоспособные цены и скидки постоянным клиентам.
- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
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- Входной контроль качества.
- Наличие сертификата ISO.

В составе нашей компании организован Конструкторский отдел, призванный помогать разработчикам, и инженерам.

Конструкторский отдел помогает осуществить:

- Регистрацию проекта у производителя компонентов.
- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



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