

7.0V to 9.5V

38mA(Typ)

0.002%(Typ)

2.3Vrms(Typ)

-100dB(Typ)

3.8µVrms(Typ)

+15dB to -79dB

Sound Processor with Built-in 3-band Equalizer BD37532FV

General Description

BD37532FV is a sound processor with built-in 3-band equalizer for car audio. A stereo input selector is available that functions to switch single end input and ground isolation input, input-gain control, main volume, loudness, 5ch fader volume and LPF for subwoofer. Moreover, "Advanced switch circuit", which is an original ROHM technology, can reduce various switching noise (ex. No-signal, low frequency like 20Hz & large signal inputs). Also, "Advanced switch" makes control of microcomputer easier, and constructs a high quality car audio system.

Features

- Reduced switching noise of input gain control, mute, main volume, fader volume, bass, middle, treble, loudness by using advanced switch circuit
- Built-in differential input selector that can make various combination of single-ended / differential input.
- Built-in ground isolation amplifier inputs, which is ideal for external stereo input.
- Built-in input gain controller reduces volume switching noise for portable audio input.
- Decreased number of external components due to built-in 3-band equalizer filter, LPF for subwoofer and loudness filter. It is possible to freely control Q, Gv, fo of 3-band equalizer and fc of LPF, Gv of loudness by I²C BUS control.
- A gain adjustment quantity of ±20dB with a 1 dB step gain adjustment is possible for bass, middle and treble.
- Equipped with terminals for subwoofer outputs. Also, the audio signal outputs of the front, rear and subwoofer can be chosen using the I²C BUS control.
- Energy-saving design resulting in low current consumption is achieved utilizing the BiCMOS process. It has the advantage in quality over scaling down the power heat control of the internal regulators.
- Input pins and output pins are organized and separately laid out in such a way that it simplifies the pattern layout of the PCB and decreases the board dimensions.
- It is possible to control I²C BUS with 3.3V / 5V.

Applications

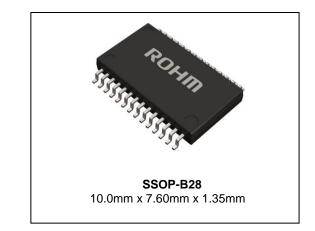
It is optimal for car audio systems. It can also be used for audio equipments like mini Compo, micro Compo, TV etc.

Key Specifications

- Power Supply Voltage Range:
- Circuit Current (No Signal):
- Total Harmonic Distortion 1: (FRONT,REAR) 0.001%(Typ)
 Total Harmonic Distortion 2:
- Total Harmonic Distortion 2: (SUBWOOFER)
- Maximum Input Voltage:
- Crosstalk Between Selectors:
- Volume Control Range:
- Output Noise Voltage 1: (FRONT,REAR)
- Output Noise Voltage 2: (SUBWOOFER)
- (SUBWOOFER)4.8µVrms(Typ)■ Residual Output Noise Voltage:1.8µVrms(Typ)
- Operating Temperature Range: -40°C to +85°C

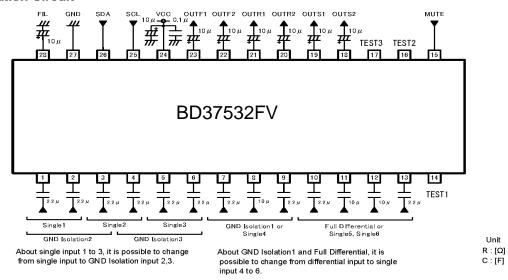
Package

W(Typ) x D(Typ) x H(Max)

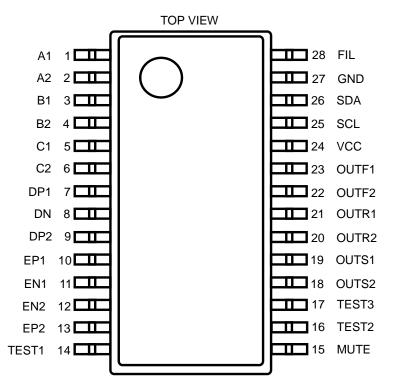


OProduct structure : Silicon monolithic integrated circuit OThis product has no designed protection against radioactive rays

Typical Application Circuit



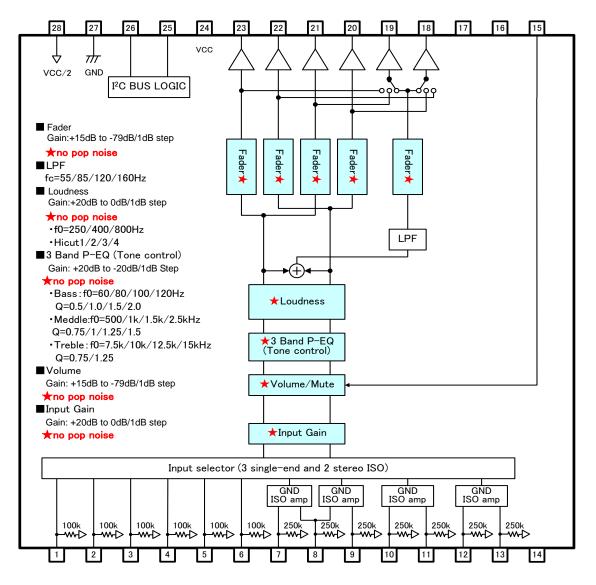
Pin Configuration



Pin Descriptions

Pin No.	Pin Name	Description	Pin No.	Pin Name	Description
1	A1	A input terminal of 1ch	15	MUTE	External compulsory mute terminal
2	A2	A input terminal of 2ch	16	TEST2	Test pin
3	B1	B input terminal of 1ch	17	TEST3	Test pin
4	B2	B input terminal of 2ch	18	OUTS2	Subwoofer output terminal of 2ch
5	C1	C input terminal of 1ch	19	OUTS1	Subwoofer output terminal of 1ch
6	C2	C input terminal of 2ch	20	OUTR2	Rear output terminal of 2ch
7	DP1	D positive input terminal of 1ch	21	OUTR1	Rear output terminal of 1ch
8	DN	D negative input terminal	22	OUTF2	Front output terminal of 2ch
9	DP2	D positive input terminal of 2ch	23	OUTF1	Front output terminal of 1ch
10	EP1	E positive input terminal of 1ch	24	VCC	Power supply terminal
11	EN1	E negative input terminal of 1ch	25	SCL	I ² C Communication clock terminal
12	EN2	E negative input terminal of 2ch	26	SDA	I ² C Communication data terminal
13	EP2	E positive input terminal of 2ch	27	GND	GND terminal
14	TEST1	Test pin	28	FIL	VCC/2 terminal

Block Diagram



Absolute Maximum Ratings (Ta=25°C)

Parameter	Symbol	Rating	Unit
Power supply Voltage	Vcc	10.0	V
Input voltage	Vin	Vcc+0.3 to GND-0.3	V
Power Dissipation	Pd	1.06 ^(Note 1)	W
Storage Temperature	Tstg	-55 to +150	°C

 (Note 1) When mounted on ROHM Standard board(70x70x1.6 (mm³), derate by 8.5mW/°C for Ta=25°C or more. Thermal resistance θja = 117.6(°C/W) Material : A FR4 grass epoxy board(3% or less of copper foil area)
 Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Conditions

Parameter	Symbol	Limit	Unit
Power Supply Voltage	Vcc	7.0 to 9.5	V
Temperature	Topr	-40 to +85	°C

Electrical Characteristics

(Unless otherwise noted, Ta=25°C, Vcc=8.5V, f=1kHz, VIN=1Vrms, Rg=600Ω, RL=10kΩ, A1 input, Input gain 0dB, Mute OFF, Volume 0dB, Tone control 0dB, Loudness 0dB, LPF OFF, Fader 0dB)

X				Limit	,		
BLOCK	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
	Circuit Current	lq	-	38	48	mA	No signal
	Voltage Gain	Gv	-1.5	0	+1.5	dB	Gv=20log(Vout/VIN)
	Channel Balance	CB	-1.5	0	+1.5	dB	$CB = G_{V1} - G_{V2}$
	Total Harmonic Distortion 1 (FRONT,REAR)	THD+N1	-	0.001	0.05	%	V _{OUT} =1Vrms BW=400Hz-30KHz
	Total Harmonic Distortion 2 (SUBWOOFER)	THD+N2	-	0.002	0.05	%	V _{OUT} =1Vrms BW=400Hz-30KHz
AL	Output Noise Voltage 1 (FRONT,REAR) *	V _{NO1}	-	3.8	15	μVrms	Rg = 0Ω BW = IHF-A
GENERAL	Output Noise Voltage 2 (SUBWOOFER) *	V_{NO2}	-	4.8	15	μVrms	Rg = 0Ω BW = IHF-A
В	Desidual Output Naiss Valtage						Fader = -∞dB
	Residual Output Noise Voltage	VNOR	-	1.8	10	μVrms	Rg = 0Ω BW = IHF-A
	Crosstalk Between Channels*	СТС	-	-100	-90	dB	$\begin{array}{l} Rg = 0\Omega \\ CTC = 20 log(V_{OUT} / V_{IN}) \\ BW = IHF-A \end{array}$
	Ripple Rejection	RR	-	-70	-40	dB	f=1kHz V _{RR} =100mVrms RR=20log(Vcc IN/Vouт)
	Input Impedance (A, B, C)	R _{IN_s}	70	100	130	kΩ	
	Input Impedance (D, E)	Rin_d	175	250	325	kΩ	
TOR	Maximum Input Voltage	VIM	2.1	2.3	-	Vrms	V _{IM} at THD+N(V _{OUT})=1% BW=400Hz-30KHz
T SELECTOR	Crosstalk Between Selectors *	CTS	-	-100	-90	dB	$\begin{array}{l} \text{Rg} = 0\Omega \\ \text{CTS} = 20 \text{log}(V_{\text{OUT}}/V_{\text{IN}}) \\ \text{BW} = \text{IHF-A} \end{array}$
INPUT	Common Mode Rejection Ratio * (D, E)	CMRR	50	65	-	dB	XP1 and XN input XP2 and XN input CMRR=20log(V_{IN}/V_{OUT}) BW = IHF-A,[*X···D,E]

Electrical Characteristics – continued

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BLOCK	Parameter	Symbol	Min	Тур	Max	Unit	Conditions
GAIN	Minimum Input Gain	Gin_min	-2	0	+2	dB	Input gain 0dB V _{IN} =100mVrms G _{IN} =20log(V _{OUT} /V _{IN})
INPUT	Maximum Input Gain	GIN_MAX	+18	+20	+22	dB	Input Gain +20dB V _{IN} =100mVrms G _{IN} =20log(V _{OUT} /V _{IN})
_	Gain Set Error	GIN_ERR	-2	0	+2	dB	Gain=+20dB to +1dB
MUTE	Mute Attenuation *	G _{MUTE}	-	-105	-85	dB	Mute ON G _{MUTE} =20log(V _{OUT} /V _{IN}) BW = IHF-A
	Maximum Gain	Gv_max	13	15	17	dB	Volume = $15dB$ V _{IN} =100mVrms Gv=20log(V _{OUT} /V _{IN})
VOLUME	Maximum Attenuation *	Gv_min	-	-100	-85	dB	$ Volume = -\infty dB \\ Gv=20log(V_{OUT}/V_{IN}) \\ BW = IHF-A $
-	Attenuation Set Error 1	Gv_err1	-2	0	+2	dB	GAIN & ATT=+15dB to -15dB
	Attenuation Set Error 2	GV_ERR2	-3	0	+3	dB	ATT=-16dB to -47dB
	Attenuation Set Error 3	Gv_err3	-4	0	+4	dB	ATT=-48dB to -79dB
(0)	Maximum Boost Gain	Gb_bst	18	20	22	dB	Gain=+20dB f=100Hz V _{IN} =100mVrms GB=20log (V _{OUT} /V _{IN})
BASS	Maximum Cut Gain	Gb_cut	-22	-20	-18	dB	Gain=-20dB f=100Hz V _{IN} =2Vrms GB=20log (V _{OUT} /V _{IN})
	Gain Set Error	G_{B_ERR}	-2	0	+2	dB	Gain=+20dB to -20dB f=100Hz
OLE	Maximum Boost Gain	Gm_bst	18	20	22	dB	Gain=+20dB f=1KHz V _{IN} =100mVrms GM=20log (V _{OUT} /V _{IN})
MIDDLE	Maximum Cut Gain	G м_сит	-22	-20	-18	dB	Gain=-20dB f=1kHz V _{IN} =2Vrms GM=20log (V _{OUT} /V _{IN})
	Gain Set Error	Gm_err	-2	0	+2	dB	Gain=+20dB to -20dB f=1kHz
BLE	Maximum Boost Gain	Gt_bst	18	20	22	dB	Gain=+20dB f=10kHz V _{IN} =100mVrms GT=20log (V _{OUT} /V _{IN})
TREBI	Maximum Cut Gain	Gt_cut	-22	-20	-18	dB	Gain=-20dB f=10kHz V _{IN} =2Vrms GT=20log (V _{OUT} /V _{IN})
	Gain Set Error	Gt_err	-2	0	+2	dB	Gain=+20dB to -20dB f=10kHz
ШК	Maximum Boost Gain	G _{F_BST}	13	15	17	dB	Fader=15dB V _{IN} =100mVrms G _F =20log(V _{OUT} /V _{IN})
FADER / SUBWOOFER	Maximum Attenuation *	$G_{F_{MIN}}$	-	-100	-90	dB	Fader = -∞dB GF=20log(V _{OUT} /V _{IN}) BW = IHF-A
UB	Gain Set Error	G _{F_ERR}	-2	0	+2	dB	Gain=+15dB to +1dB
N/	Attenuation Set Error 1	GF_ERR1	-2	0	+2	dB	ATT=-1dB to -15dB
Ř	Attenuation Set Error 2	GF_ERR2	-3	0	+3	dB	ATT=-16dB to -47dB
B	Attenuation Set Error 3	G _{F_ERR3}	-4	0	+4	dB	ATT=-48dB to -79dB
FA	Output Impedance	Rout	-	-	50	Ω	V _{IN} =100mVrms THD+N=1%
	Maximum Output Voltage	V _{OM}	2	2.2	-	Vrms	BW=400Hz-30KHz
LOUDNESS	Maximum Gain	Gl_max	17	20	23	dB	Gain 20dB V _{IN} =100mVrms GL=20log(V _{OUT} /V _{IN})
LOUD	Gain Set Error	Gl_err	-2	0	+2	dB	GAIN=+20dB to +1dB

VP-9690A (Average value detection, effective value display) filter by Matsushita Communication is used for * measurement. Phase between input / output is same.

Typical Performance Curves

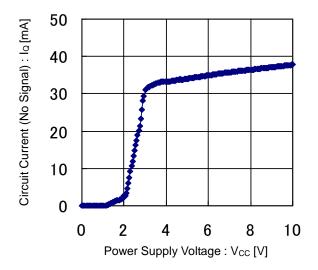


Figure 1. Circuit Current (No Signal) vs Power Supply Voltage

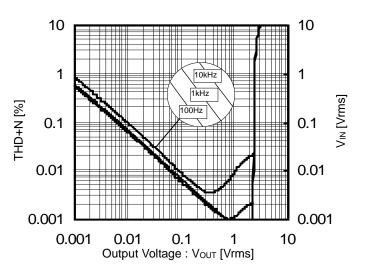


Figure 2. Total Harmonic Distortion vs Output Voltage

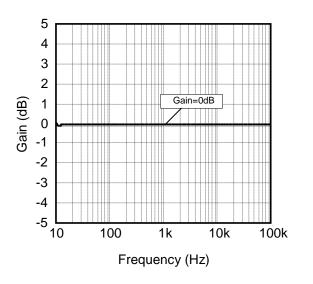


Figure 3. Gain vs Frequency

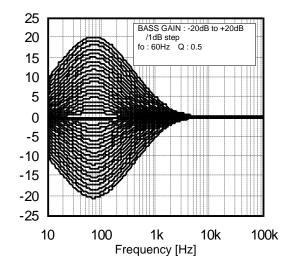
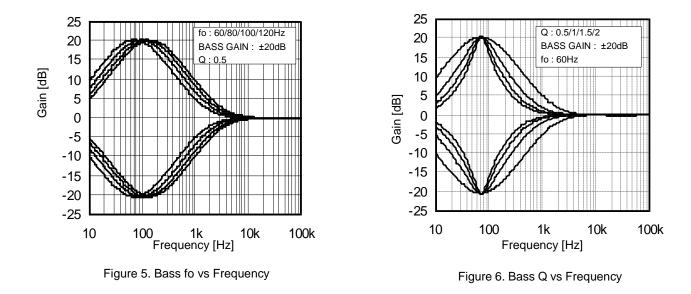
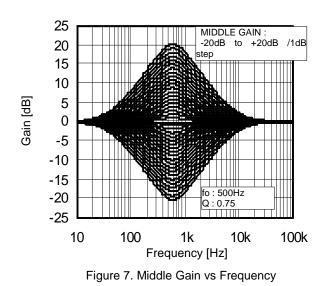


Figure 4. Bass Gain vs Frequency

Gain[dB]





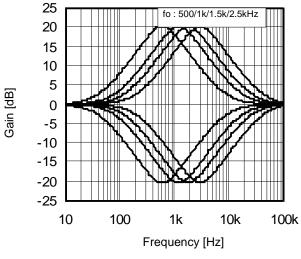
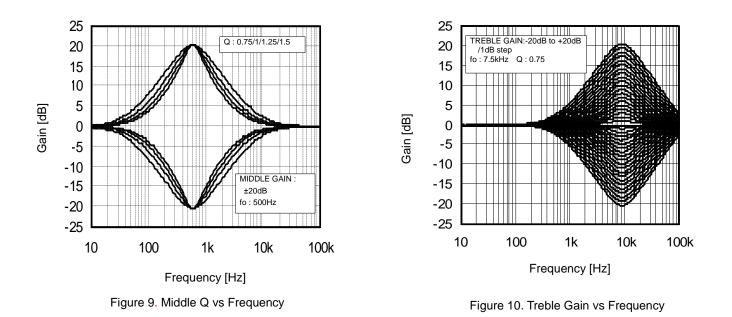


Figure 8. Middle fo vs Frequency



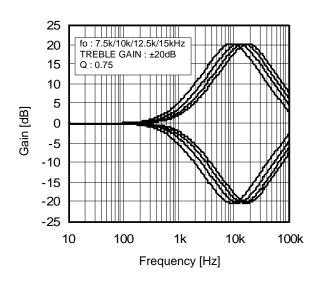


Figure 11. Treble fo vs Frequency

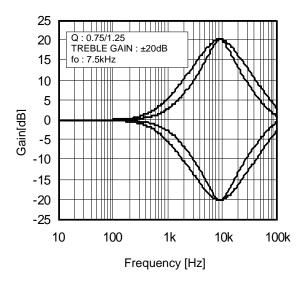


Figure 12. Treble Q vs Frequency

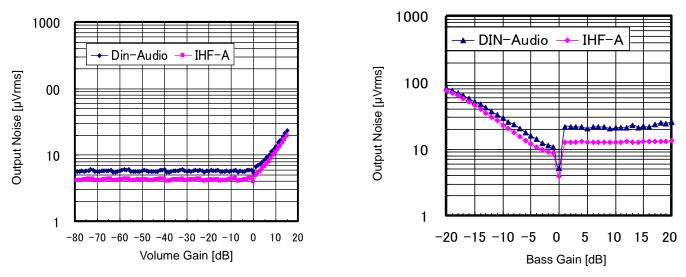
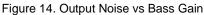


Figure 13. Output Noise vs Volume Gain



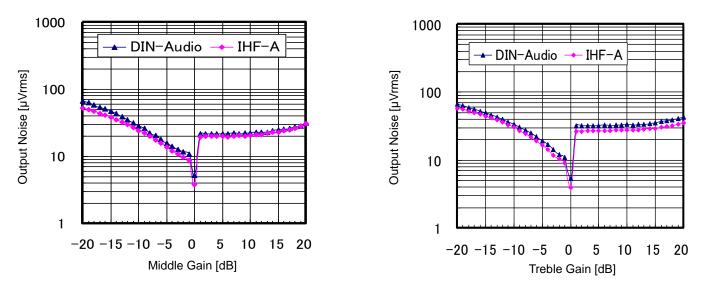


Figure 15. Output Noise vs Middle Gain

Figure 16. Output Noise vs Treble Gain

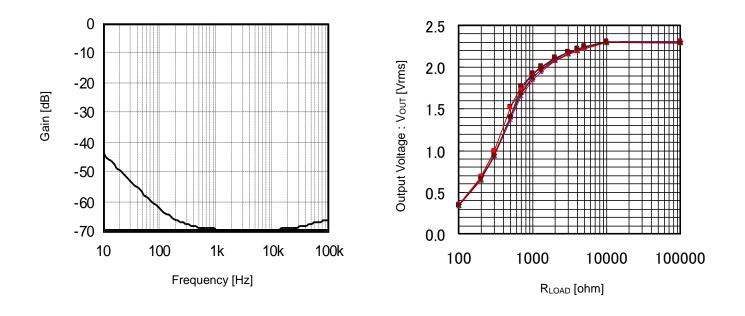


Figure 17. CMRR vs Frequency

Figure 18. Output Voltage vs RLOAD

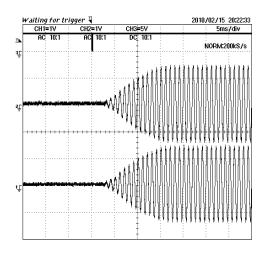


Figure 19. Advanced Switch 1

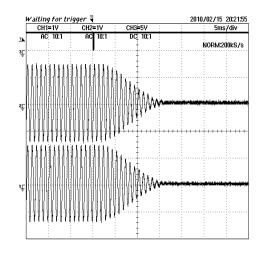


Figure 20. Advanced Switch 2

Timing Chart

CONTROL SIGNAL SPECIFICATION

(1) Electrical Specifications and Timing for Bus Lines and I/O Stages

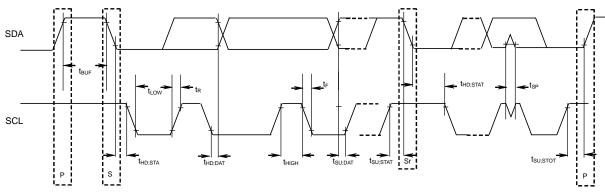


Figure 21. Definition of Timing on the I²C-bus

Table 1 Characteristics of the SDA and SCL bus lines for I²C-bus devices (Unless specified particularly, Ta= 25° C, V_{CC}=8.5V)

	Parameter	Symbol	Fast-mod	e l²C-bus	Unit
	Faldineter	Symbol	Min	Max	Unit
1	SCL clock frequency	f _{SCL}	0	400	kHz
2	Bus free time between a STOP and START condition	t BUF	1.3	-	μS
3	Hold time (repeated) START condition. After this period, the first clock	t	0.6		
3	pulse is generated	t hd;sta	0.0	-	μS
4	LOW period of the SCL clock	t _{LOW}	1.3	-	μS
5	HIGH period of the SCL clock	tніgн	0.6	-	μS
6	Set-up time for a repeated START condition	tsu;sta	0.6	-	μS
7	Data hold time:	t _{HD;DAT}	0.06 ^(Note)	-	μS
8	Data set-up time	tsu;dat	120	-	ns
9	Set-up time for STOP condition	tsu;sто	0.6	-	μS

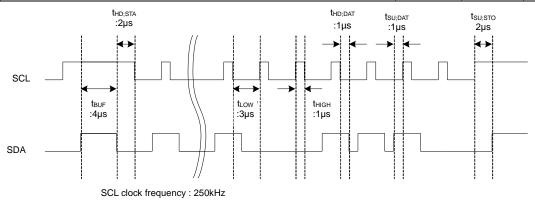
All values referred to VIH Min and VIL Max Levels (see Table 2).

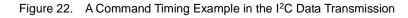
(Note) The device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the VIH Min of the SCL signal) in order to bridge the undefined region of the falling edge of SCL.
About 7 (t_{HD,DAT}), 8(t_{SU,DAT}), make the setup in which the margin is fully in .

About 7 (thd;dat), 8(tsu;dat), make the setup in which the margin is fully in .

Table 2 Characteristics of the SDA and SCL I/O stages for I²C-bus devices

	Parameter	Sumbol	Fast-mode	Unit	
	Parameter	Symbol	Min	Max	Unit
10	LOW level input voltage:	VIL	-0.3	+1	V
11	HIGH level input voltage:	VIH	2.3	5	V
12	Pulse width of spikes which must be suppressed by the input filter.	tsp	0	50	ns
13	LOW level output voltage: at 3mA sink current	Vol1	0	0.4	V
14	Input current each I/O pin with an input voltage between 0.4V and 4.5V.	lı lı	-10	+10	μA





(2) <u>I²C BUS FORMAT</u>

		MSB I	SB	MSB		LSB		LSB						
	S	Slave Addres	s A	A Select Addre		Α		Data		Р				
	1bit	8bit	1bit	1bit 8bit 1bit 8bit 1bit										
		S	= S	 Start conditions (Recognition of start bit) 										
		Slave Addr	ess = R	ecognition of s	lave addr	ess. 7	' bits in up	per order are	volunta	ary.				
			TI	The least significant bit is "L" for write mode.										
		A	= A	= ACKNOWLEDGE bit (Recognition of acknowledgement)										
		Select Add	ess = S	= Select address for volume, bass and treble.										
		Data	= D	= Data on every volume and tone.										
		Р	= S	= Stop condition (Recognition of stop bit)										
(3) <u>l</u> ²	² C BU	S Interface Protoc	ol		-									

(a) Basic form

10.7									
S	Slave Addre	ess	Α	Select Add	Α	Data	Α	Р	
	MSB	LSB		MSB	LSB	N	ISB	LSB	

(b) Automatic increment (Select Address increases (+1) according to the number of da
--

ſ	S Slave Address			Select Address	А	Data1	А	D	ata2	А		DataN	А	Ρ
MSB LSB MSB LSB MSB LSB MSB LSB MSB LSB														
((Example) ①Data1 shall be set as data of address specified by Select Address.													
	2 Data2 shall be set as data of address specified by Select Address +1.													

③DataN shall be set as data of address specified by Select Address +N-1.

(c) Configuration unavailable for transmission (In this case, only Select Address1 is set.

S	Sla	lave Address A Select Address1 A				Α	Da	ita	А	Selec	t Addres	ss 2	А	Dat	a	4	Ρ
MS	SΒ	LSB		MSB	LSB	N	ISB	LS	В	MSB		LSB	M	SB	LSB		
	(Note) If any data is transmitted as Select Address 2 next to data, it is recognized																
		as data, not as Select Address 2.															

(4) Slave Address

MSB							LSB	_
A6	A5	A4	A3	A2	A1	A0	R/W	1
1	0	0	0	0	0	0	0	80H

(5) Select Address & Data

Items	Select Address	MSB			Da	ata			LSB	
nems	(hex)	D7	D6	D5	D4	D3	D2	D1	D0	
Initial setup 1	01	Advanced switch ON/OFF	0	Advanced switch time of Input Gain/Volume Tone/Fader/Loudness		0	1		switch time Mute	
Initial setup 2	02	LPF Phase	0		er Output lect	0	Subwoofer LPF fc			
Initial setup 3	03	0	0	0	Loudn	ess fo	0	0	1	
Input Selector	05	Full-diff Type	0	0		Input selector				
Input gain	06	Mute ON/OFF	0	0			Input Gain			
Volume gain	20			١	/olume Gain	/ Attenuatio	n			
Fader 1ch Front	28				Fader Gain	/ Attenuatior	า			
Fader 2ch Front	29		Fader Gain / Attenuation							
Fader 1ch Rear	2A	Fader Gain / Attenuation								
Fader 2ch Rear	2B		Fader Gain / Attenuation							
Fader Subwoofer	2C				Fader Gain	/ Attenuatior	า			
Test Mode	30	1	1	1	1	1	1	1	1	
Bass setup	41	0	0	Bas	s fo	0	0	Bass Q		
Middle setup	44	0	0	Mido	lle fo	0	0	Mide	dle Q	
Treble setup	47	0	0	Treb	le fo	0	0	0	Treble Q	
Bass gain	51	Bass Boost/ Cut	0	0			Bass Gain			
Middle gain	54	Middle Boost/ Cut	0	0 Middle Gain						
Treble gain	57	Treble Boost/ Cut	0	0	0 Treble Gain					
Loudness Gain	75	0	Loudn	ess Hicut	Loudness Gain					
System Reset	FE	1	0	0	0	0	0	0	1	

Advanced switch

Note

- 1. The Advanced Switch works in the latch part while changing from one function to another..
- 2. Upon continuous data transfer, the Select Address rolls over because of the automatic increment function, as shown below.

$$\rightarrow 01 \rightarrow 02 \rightarrow 03 \rightarrow 05 \rightarrow 06 \rightarrow 20 \rightarrow 28 \rightarrow 29 \rightarrow 2A \rightarrow 2B \rightarrow 2C$$

$$\rightarrow 30 \rightarrow 41 \rightarrow 44 \rightarrow 47 \rightarrow 51 \rightarrow 54 \rightarrow 57 \rightarrow 75$$

- 3. Advanced Switch is not used for the functions of input selector and subwoofer output select etc. Please turn on MUTE when changing the settings of this side of the set.
- 4. When using Mute function of this IC at the time of changing input selector, please switch mute ON/OFF while waiting for advanced-mute time.

Select address 01 (hex	()									
Time	MSB	Ac	Advanced switch time of Mute LS							
Time	D7	D6	D5	D4	D3	D2	D1	D0		
0.6msec	Advanced		Advonced	awitah tima			0	0		
1.0msec	Advanced Switch	0		d switch time	0	1	0	1		
1.4msec	ON/OFF	U		ain/Volume r/Loudness	0	I	1	0		
3.2msec	UN/OFF		Tone/Faue				1	1		

Time	MSB	Advanced switch time of Input gain/Volume/Tone/Fader/Loudness								
	D7	D6	D5	D4	D3	D2	D1	D0		
4.7 msec	Advisionand		0	0		1	Advanced switch			
7.1 msec	Advanced Switch	0	0	1						
11.2 msec	ON/OFF	0	1	0	0	I	Time o	of Mute		
14.4 msec			1	1	1					

Mode	MSB		Advanced switch ON/OFF								
Mode	D7	D6	D5	D4	D3	D2	D1	D0			
OFF	0	0	0 Advanced switch time of Input gain/Volume Tone/Fader/Loudness		0	1	Advanced switch Time of Mute				
ON	1	_			-						

Select address 02(hex)										
fo	MSB	MSB Subwoofer LPF fc								
fc	D7	D6	D5	D4	D3	D2	D1	D0		
OFF						0	0	0		
55Hz						0	0	1		
85Hz		0	Subwoof	er Output	0	0	1	0		
120Hz	LPF Phase	0	Se	lect	0	0	1	1		
160Hz						1	0	0		
Prohibition]			

Mode	MSB		Subwo	oofer C	Dutput	Select	t	LSB	
Mode	D7	D6	D5	D4	D3	D2	D1	D0	
LPF			0	0					
Front		0	0	1			- ,		
Rear	LPF Phase	0	1	0	0	Su	= fc		
Prohibition			1	1					

Phase	MSB			LSB					
Filase	D7	D6	D5	D4	D3	D2	D1	D0	
0°	0	0	Subwoofer output select		0	Su	ubwoofer LPF fc		
180°	1	0			0	50			

Select address 03(hex)

fO	MSB			Loudr	ness fo)		LSB
10	D7	D6	D5	D4	D3	D2	D1	D0
250Hz				0	0			
400Hz	0	0	0	0	1	0	0	4
800Hz	0	0	0	1	0	0	0	I
Prohibition				1	1			

: Initial condition

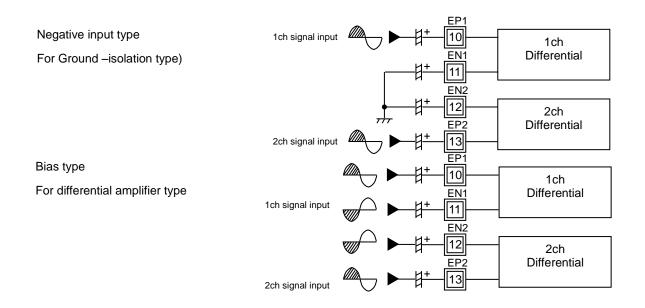
Select address 05(hex)

Mada			MSB		Ir	nput S	Select	or		LSB
Mode	OUTF1	OUTF2	D7	D6	D5	D4	D3	D2	D1	D0
A	A1	A2				0	0	0	0	0
В	B1	B2				0	0	0	0	1
С	C1	C2				0	0	0	1	0
D single	DP1	DP2				0	0	0	1	1
E1 single	EP1	EN1	F II I:44			0	1	0	1	0
E2 single	EN2	EP2	Full-diff	0	0	0	1	0	1	1
A diff	A1	B1	bias type select	0	0	0	1	1	1	1
C diff	B2	C2	Select			1	0	0	0	0
D diff	DP1	DP2				0	0	1	1	0
E full diff	EP1	EP2				0	1	0	0	0
Inp	ut SHORT					0	1	0	0	1
P	Prohibition							Other setting)	
			tance of oac	h innut torr	ninal ia lau	orod from			/	

Input SHORT : The input impedance of each input terminal is lowered from $100k\Omega(Typ)$ to $6 k\Omega(Typ)$. (For quick charge of coupling capacitor)

Mode	MSB	MSB Full-diff Bias Type Select								
	D7	D6	D5	D4	D3	D2	D1	D0		
Negative Input	0	0	0			nput Selecto	r			
Bias	1	0	0			nput Selecto	11			

: Initial condition



Select address 06 (he)				-	<u> </u>					
Gain	MSB		1		Gain			LSB		
	D7	D6	D5	D4	D3	D2	D1	D0		
0dB				0	0	0	0	0		
1dB				0	0	0	0	1		
2dB			-		0	0	0	1	0	
3dB				0	0	0	1	1		
4dB				0	0	1	0	0		
5dB				0	0	1	0	1		
6dB				0	0	1	1	0		
7dB				0	0	1	1	1		
8dB				0	1	0	0	0		
9dB			0			0	1	0	0	1
10dB				0	1	0	1	0		
11dB	Mute	0		0	1	0	1	1		
12dB	ON/OFF	0		0	0	1	1	0	0	
13dB				0	1	1	0	1		
14dB				0	1	1	1	0		
15dB				0	1	1	1	1		
16dB				1	0	0	0	0		
17dB				1	0	0	0	1		
18dB				1	0	0	1	0		
19dB]			1	0	0	1	1		
20dB				1	0	1	0	0		
]			1	1	0	1	1		
Prohibition	Prohibition		:	:	• •	:	:			
				1	1	1	1	1		

Mode	MSB		Γ	/lute C	LSB			
Mode	D7	D6	D5	D4	D3	D2	D1	D0
OFF	0	0	0			Innut Coin		
ON	1	0	0			Input Gain		

: Initial condition

Gain & ATT	MSB		ol, Fad	er Gai	n / Atte	enuatio	on	LSB
Gaill & ATT	D7	D6	D5	D4	D3	D2	D1	D0
	0	0	0	0	0	0	0	0
	0	0	0	0	0	0	0	1
Prohibition	:	:	:	:	:	:	:	:
	0	1	1	1	0	0	0	0
15dB	0	1	1	1	0	0	0	1
14dB	0	1	1	1	0	0	1	0
13dB	0	1	1	1	0	0	1	1
:	:	:	:	:	:	:	:	:
-77dB	1	1	0	0	1	1	0	1
-78dB	1	1	0	0	1	1	1	0
-79dB	1	1	0	0	1	1	1	1
	1	1	0	1	0	0	0	0
Prohibition	:	:	•	:	:	:	:	:
	1	1	1	1	1	1	1	0
-∞dB	1	1	1	1	1	1	1	1

Select address 20, 28, 29, 2A, 2B, 2C (hex)

Select address 41(hex)

Q factor	MSB		Ba	ass C	LSB			
QIACIOI	D7	D6	D5	D4	D3	D2	D1	D0
0.5			0 Bass fo		0		0	0
1.0	0	0				0	0	1
1.5	0	0	Das	510	0	0	1	0
2.0							1	1

fo	MSB			Bass	LSB			
to	D7	D6	D5	D4	D3	D2	D1	D0
60Hz			0	0				
80Hz	0	0	0	1	0	0	Ba	ass actor
100Hz	0	0	1	0	0	0	Q fa	actor
120Hz			1	1				

Select address 44(hex)

Q factor	MSB		Mic	ddle	LSB			
	D7	D6	D5	D4	D3	D2	D1	D0
0.75							0	0
1.0	0	0	Mida	dle fe	0	0	0	1
1.25	0	0	WILCO	dle fo	0	0	1	0
1.5							1	1

fo	MSB			Middle	LSB			
to	D7	D6	D5	D4	D3	D2	D1	D0
500Hz			0	0				
1kHz	0	0	0	1	0	0	Mic	ldle
1.5kHz	0	0	1	0	0	0	Middle Q factor	
2.5kHz			1	1				

: Initial condition

Select address 47 (hex)

Q factor	MSB		Tre	eble	Q facto	or	LSB		
QTACIO	D7	D6	D5	D4	D3	D2	D1	D0	
0.75	0	0	Trob	le fo	0	0	0	0	
1.25	0	0	Her	ie iu	0	0	0	1	
fo	MSB	Treble			∋ fo			LSB	
10	D7	D6	D5	D4	D3	D2	D1	D0	
7.5kHz			0	0					
10kHz	0	0	0	1	0	0	0	Treble	
12.5kHz	0	0	1	0	0	0	0	Q factor	
15kHz			1	1					

Select address 51, 54, 57 (hex)

Gain	MŚB	E	Bass/N	1iddle/	Treble	Gain		LSB	
Gaili	D7	D6	D5	D4	D3	D2	D1	D0	
0dB				0	0	0	0	0	
1dB				0	0	0	0	1	
2dB				0	0	0	1	0	
3dB				0	0	0	1	1	
4dB				0	0	1	0	0	
5dB				0	0	1	0	1	
6dB				0	0	1	1	0	
7dB				0	0	1	1	1	
8dB				0	1	0	0	0	
9dB				0	1	0	0	1	
10dB	Bass/			0	1	0	1	0	
11dB	Middle/		0		0	1	0	1	1
12dB	Treble	0		0	1	1	0	0	
13dB	Boost			0	1	1	0	1	
14dB	/cut			0	1	1	1	0	
15dB				0	1	1	1	1	
16dB				1	0	0	0	0	
17dB				1	0	0	0	1	
18dB				1	0	0	1	0	
19dB				1	0	0	1	1	
20dB				1	0	1	0	0	
				1	0	1	0	1	
Prohibition	Prohibition			:	:	:	:	:	
				1	1	1	1	0	
				1	1	1	1	1	
	MOD	D = 1		-ll - /T			4		
Mode	MSB	1	ss/Mid				1	LSB	
	D7	D6	D5	D4	D3	D2	D1	D0	

Bass/Middle/Treble Gain

: Initial condition

Boost

Cut

0

1

0

0

BD37532FV

Select address 75 (he)	<)									
Mode	MSB	MSB Loudness Hicut								
Mode	D7	D6	D5	D4	D3	D2	D1	D0		
Hicut1		0	0							
Hicut2	0	0	1			oudness Gai	in			
Hicut3	0	1	0]	L	Joudness Ga	111			
Hicut4		1	1							

Gain	MSB		L	oudne	ss Ga	in		LSB
Gain	D7	D6	D5	D4	D3	D2	D1	D0
0dB			•	0	0	0	0	0
1dB				0	0	0	0	1
2dB				0	0	0	1	0
3dB				0	0	0	1	1
4dB				0	0	1	0	0
5dB				0	0	1	0	1
6dB				0	0	1	1	0
7dB				0	0	1	1	1
8dB				0	1	0	0	0
9dB				0	1	0	0	1
10dB				0	1	0	1	0
11dB				0	1	0	1	1
12dB	0	Loudne	ss Hicut	0	1	1	0	0
13dB				0	1	1	0	1
14dB				0	1	1	1	0
15dB				0	1	1	1	1
16dB				1	0	0	0	0
17dB				1	0	0	0	1
18dB				1	0	0	1	0
19dB				1	0	0	1	1
20dB				1	0	1	0	0
				1	0	1	0	1
Prohibition				:	:	:	:	:
				1	1	1	1	1

: Initial Condition

(6) About Power ON Reset

Built-in IC initialization is made during power ON of the supply voltage. Please send initial data to all addresses at supply voltage on. Also, please turn ON MUTE at the set side until initial data is sent.

Parameter	Symbol		Limit		Unit	Conditions
Falameter	Symbol	Min	Тур	Max	Unit	Conditions
Rise Time of VCC	trise	33	-	-	µsec	V_{CC} rise time from 0V to 5V
VCC Voltage of Release Power ON Reset	V _{POR}	-	4.1	-	V	

(7) About External Compulsory Mute Terminal

It is possible to forcibly set Mute from the outside by setting input voltage at the MUTE terminal.

Mute Voltage Condition	Mode
GND to 1.0V	MUTE ON
2.3V to Vcc	MUTE OFF

Establish the voltage of MUTE in the condition to be defined.

Application Information

1. Function and Specifications

Function	Specifications									
	· (Stereo input)									
	Single-End/Diff/Full-Diff									
	(Possible to set the number of single-end/diff/full-diff as follows)									
land	Mada 4	Single-End	Differential	Full-Differential						
Input selector	Mode 1 Mode 2	0	3	1						
	Mode 3	3	1	1						
	Mode 4	4	0	1						
	Mode 5 Mode 6	<u>5</u> 6	1	0						
		-	ombination of inp	÷						
Input	・+20dB to	0dB (1dB step))							
gain	Possible	to use "Advanc	ed switch" for pre	evention of switching	noise.					
Mute	· Possible	to use "Advanc	ed switch" for pre	evention of switching	noise.					
Volume	・+15dB to	-79dB (1dB ste	ep), -∞dB							
volume	· Possible	to use "Advanc	ed switch" for pre	evention of switching	noise.					
	• +20dB to -20dB (1dB step)									
Dees	· Q=0.5, 1, 1.5, 2									
Bass	• fo=60, 80, 100, 120Hz									
	Possible to use "Advanced switch" when changing gain									
	+20dB to -20dB (1dB step)									
Middle	• Q=0.75, ²	, 1.25, 1.5								
Middle	・ fo=500, 1k, 1.5k 2.5kHz									
	Possible to use "Advanced switch" when changing gain									
	• +20dB to -20dB (1dB step)									
Taskla	· Q=0.75, 1.25									
Treble	• fo=7.5k, 10k, 12.5k, 15kHz									
	Possible to use "Advanced switch" when changing gain									
	 +15dB to -79dB(1dB step), -∞dB 									
Fader	· Possible	to use "Advanc	ed switch" for pre	evention of switching	noise.					
		0dB(1dB step)		-						
Loudness	• fo=250/40									
	· Possible	to use "Advanc	ed switch" for pre	evention of switching	noise.					
	• fc=55/85/	120/160Hz, pas	SS	-						
LPF	 Phase sh 	-								
L		. ,								

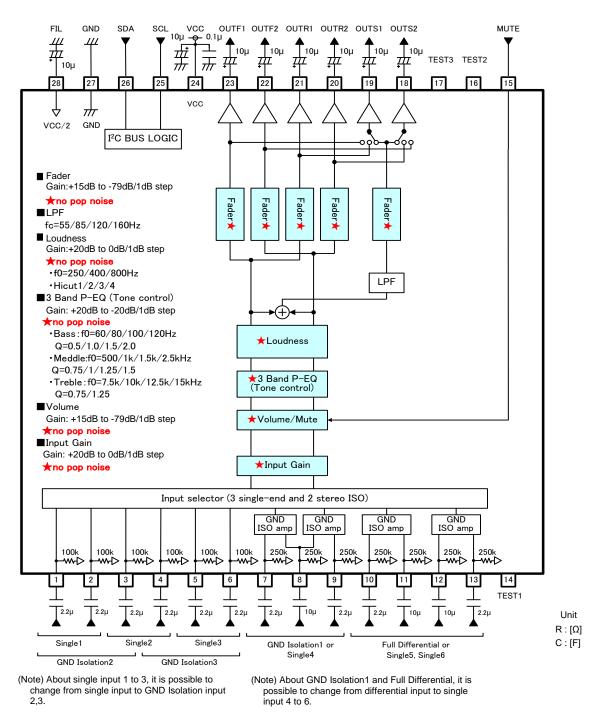
(dB) D7 D6 D5 D4 D3 D2 D1 D0 +15 0 1 1 1 0 0 1 .33 1 0 1 0 1 1 0 1 1 0 1 1 0 1 1 0 1 1 1 0 1 1 0 1 1 1 1 1 1 1 1 1 1 0 1	2. \	/olume	/ гас	aer vo	Jume	Allei	iualic	on Da	la										
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	(C	lB)	D7	D6	D5	D4	D3	D2	D1	D0	(dB)	D7	D6	D5	D4	D3	D2	D1	D0
$\begin{array}{cccccccccccccccccccccccccccccccccccc$	+	15	0	1	1	1	0	0	0	1	-33	1	0	1	0	0	0	0	1
$\begin{array}{cccccccccccccccccccccccccccccccccccc$				1			-	-	-				-		-	-	-	-	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $							-	_					-			-	-		
+1101110101 $+10$ 011101101 $+9$ 01110111 $+8$ 01111000 $+7$ 01111000 $+7$ 0111101 $+6$ 0111101 $+4$ 0111101 $+4$ 0111101 $+2$ 0111110 $+3$ 0111110 $+1$ 0111110 $+2$ 0111111 0 1000001 $+2$ 10000011 -1 00000011 -1 00000011 -1 0000111 -1 0001111 -1 0001111 -1 0001111 -1 </td <td></td> <td></td> <td>0</td> <td>1</td> <td>1</td> <td>1</td> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td></td> <td>1</td> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td>1</td> <td>0</td> <td>0</td>			0	1	1	1	0	1	0	0		1	0	1	0	0	1	0	0
+100110110 $+9$ 01110110 $+8$ 0111011 $+8$ 0111001 $+7$ 01111001 $+7$ 01111001 $+6$ 01111001 $+4$ 0111101 $+4$ 0111101 $+3$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 -2 10000011 -2 1000111 -2 1000111 -2 1001 </td <td>-</td> <td></td> <td>0</td> <td>1</td> <td>1</td> <td>1</td> <td>0</td> <td>1</td> <td>0</td> <td>1</td> <td></td> <td>1</td> <td>0</td> <td>1</td> <td>0</td> <td>0</td> <td>1</td> <td>0</td> <td>-</td>	-		0	1	1	1	0	1	0	1		1	0	1	0	0	1	0	-
+901110111 $+8$ 01111000 $+7$ 01111000 $+6$ 01111001 $+6$ 01111010 $+5$ 0111101 $+4$ 0111101 $+2$ 0111110 $+3$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0111111 $+2$ 0000000 $+1$ 0000000 $+1$ 0000000 $+1$ 0000000 $+1$ 000000 $+1$ 000011 $+1$ 000011 $+1$ 00011 $+1$ <	-		-				-		-				0		-	-		-	
+801111000+70111101+60111101+50111101+40111101+40111101+40111110+30111110+41010101+20111111-41000001-41000010-11000011-21000011-41000111-41000111-41000101-41000111-41000111-41000111-41001111-41000111-7100 <t< td=""><td></td><td></td><td>-</td><td>1</td><td>1</td><td></td><td>-</td><td>1</td><td></td><td>-</td><td></td><td></td><td>0</td><td></td><td>-</td><td>-</td><td></td><td></td><td>-</td></t<>			-	1	1		-	1		-			0		-	-			-
+70111001 $+6$ 01110101 $+4$ 0111101 $+4$ 0111101 $+4$ 0111101 $+4$ 0111101 $+2$ 01111101 $+2$ 01111111 0 1000000 $+1$ 0111111 0 1000001 $+1$ 0111111 0 1000001 -1 0000011 -1 0000011 -1 000011 -4 100011 -4 100011 -4 100011 -4 100011 -4 100011 -4 100011 -6 1001 <td></td> <td></td> <td></td> <td></td> <td></td> <td></td> <td>-</td> <td>-</td> <td></td> <td></td> <td></td> <td></td> <td>-</td> <td></td> <td>-</td> <td>-</td> <td>-</td> <td></td> <td></td>							-	-					-		-	-	-		
+601111010+501111011+401111011+30111101+201111101+201111110+10111110101000000+10000001-21000011-21000011-41000101-41000101-41000101-41000101-51000101-61001011-6100101-7100011-10100111-11100111-12100111-14100 <td></td> <td></td> <td>-</td> <td></td> <td></td> <td></td> <td></td> <td>-</td> <td>-</td> <td>-</td> <td>_</td> <td></td> <td>-</td> <td></td> <td></td> <td>-</td> <td>-</td> <td>-</td> <td>-</td>			-					-	-	-	_		-			-	-	-	-
+501111011 $+4$ 01111011 $+4$ 0111101 $+3$ 01111100 $+3$ 01111101 $+2$ 01111110 $+1$ 01111110 $+2$ 01111110 -43 10101010 -43 10101110 -1 10000000 -1 100001100 -3 1000011100 -5 100011101100 -5 1000111011001 -6 1000111110110 -7 10001111101111 -6 100111			-					_					-						
+4 0 1	-							-					-		-		-		
+301111101 $+2$ 011111111 $+2$ 01111111111 $+1$ 01000000-47101011111 0 10000000-47101011 <td>-</td> <td></td> <td>-</td> <td></td> <td>-</td> <td></td> <td>-</td> <td>-</td> <td></td> <td>-</td> <td></td> <td></td> <td>-</td> <td>-</td> <td>-</td> <td></td> <td>-</td> <td></td> <td></td>	-		-		-		-	-		-			-	-	-		-		
+201111111 $+1$ 01111111 $+1$ 01111111 0 10000000 -1 10000000 -2 100000111000 -3 10000110110001 -44 10000101100011 -44 10000101100011 -44 100001101100011 -44 10000110110011001100110011100111001110111111111111111111111111111111111 </td <td></td> <td></td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td></td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td>-</td> <td></td> <td>-</td> <td></td>			-	-	-	-	-	-	-	-		-	-	-	-	-		-	
+101111111101000000000000 -1 100000011100000 -2 10000011100001 -2 100001110110001 -4 1000011100110010 -5 100001110110001110011001110001110011100100111001111110011<	-								-				-					-	
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-91001001 -10 10010101 -11 10001011 -12 1000111 -12 1000110 -13 1000110 -13 1000110 -13 1000111 -16 1001111 -16 1001111 -16 1001111 -16 1001111 -16 1001111 -16 1001111 -16 1001111 -16 1001111 -16 1000000 -17 1001001 -17 100101 -20 100111 -22 100111 -24 100111 -22	-			-	-	-	-		-				-			-	-		
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2. Volume / Fader Volume Attenuation Data

: Initial condition

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3. Application Circuit



Notes on wiring

- \oplus Please connect the decoupling capacitor of the power supply in the shortest possible distance to GND.
- GND lines should be one-point connected.
- ③ Wiring pattern of Digital shall be away from that of analog unit and crosstalk should not be acceptable.
- ④ If possible, SCL and SDA lines of I²C BUS should not be in parallel.
- The lines should be shielded, if they are adjacent to each other.
- (5) If possible, analog input lines should not be in parallel. The lines should be shielded, if they are adjacent to each other.
- 6 TEST pins (Pin 14,15,16) should be OPEN.

Power Dissipation

About the thermal design of the IC

Characteristics of an IC are greatly affected by the temperature at which it is used. Exceeding absolute maximum ratings may degrade and destroy the device. Careful consideration must be given to the heat of the IC from the two standpoints of immediate damage and long-term reliability of operation.

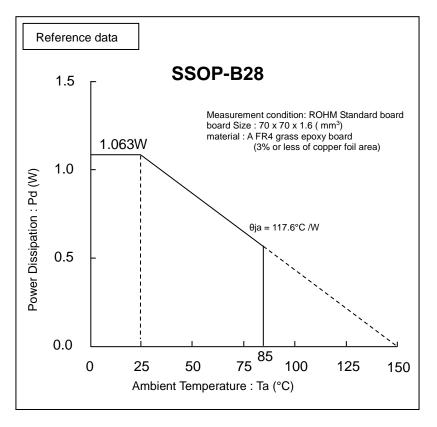


Figure 23. Temperature Derating Curve

(Note) Values are actual measurements and are not guaranteed.

Power dissipation values vary according to the board on which the IC is mounted.

I/O Equivalent Circuits

Terminal	Terminal	Terminal		
No.	Name	Voltage	Equivalent Circuit	Terminal Description
1 2 3 4 5 6	A1 A2 B1 B2 C1 C2	4.25	VCC	A terminal for signal input. The input impedance is 100kΩ (Typ).
7 8 9 10 11 12 13	DP1 DN DP2 EP1 EN1 EN2 EP2	4.25	VCC	Input terminal available to single/Differential mode. The input impedance is 250kΩ (Typ).
15	MUTE	-	VCC A BY BY I.65V GND	A terminal for external compulsory mute. If terminal voltage is High level, the mute is OFF. And if the terminal voltage is Low level, the mute is ON.
18 19 20 21 22 23	OUTS2 OUTS1 OUTR2 OUTR1 OUTF2 OUTF1	4.25	VCC A GND GND	A terminal for fader and Subwoofer output.

Values in the pin explanation and input/output equivalent circuit are for reference purposes only. It is not a guaranteed value.

I/O Equivalent Circuits - continued

Terminal No.	Terminal Name	Terminal Voltage	Equivalent Circuit	Terminal Description
24	VCC	8.5		Power supply terminal.
25	SCL	-	VCC	A terminal for clock input of I ² C BUS communication.
26	SDA	-	VCC	A terminal for data input of I ² C BUS communication.
27	GND	0		Ground terminal.
28	FIL	4.25		Voltage for reference bias of analog signal system. The simple pre-charge circuit and simple discharge circuit for an external capacitor are built in.
14 16 17	TEST	-		TEST terminal

Values in the pin explanation and input/output equivalent circuit are for reference purposes only. It is not a guaranteed value.

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes – continued

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

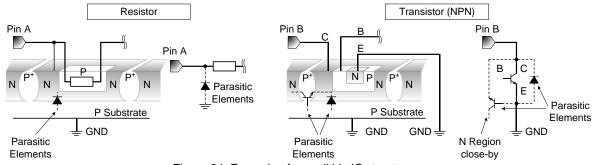
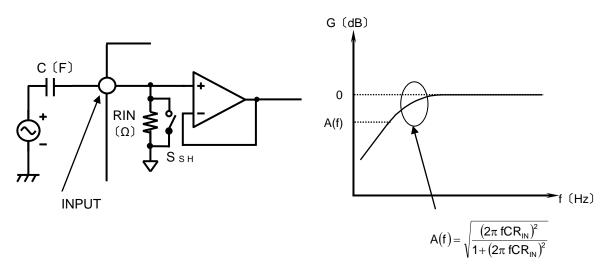


Figure 24. Example of monolithic IC structure

13. About a Signal Input Part

(a) About Input Coupling Capacitor Constant Value The constant value of input coupling capacitor C(F) is decided with respect to the input impedance $R_{IN}(\Omega)$ at the input signal terminal of the IC that would be sufficient to form an RC characterized HPF.



(b) About the Input Selector SHORT

SHORT mode is the command which makes switch S_{SH} =ON of input selector part so that the input impedance R_{IN} of all terminals becomes small. Switch S_{SH} is OFF when SHORT command is not selected. The constant time brought about by the small resistance inside and the capacitor outside the LSI becomes small when this command is used. The charge time of the capacitor becomes short. Since SHORT mode turns ON the switch of S_{SH} and makes it low impedance, please use it at no signal condition.

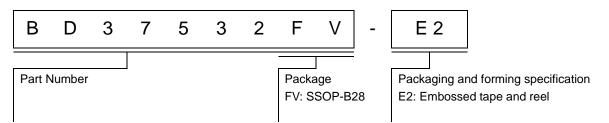
14. About Mute Terminal (Pin 15) when Power Supply is OFF

There should be no applied voltage to Mute terminal (Pin 15) when power-supply is OFF. If in case voltage is supplied to Mute terminal, please insert a series resistor (about $2.2k\Omega$) to Mute terminal. (Please refer to Application Circuit Diagram.)

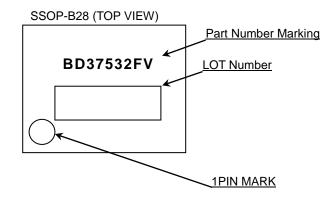
15. About TEST Pin

TEST Pin should be OPEN. Pin 14. 16, 17 are TEST Pins

Ordering Information

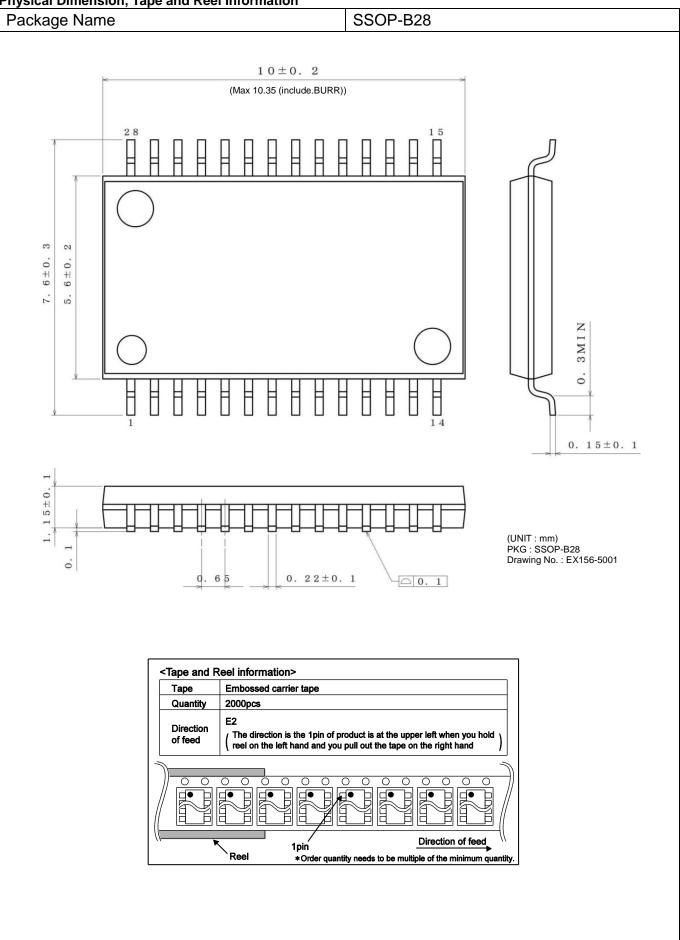


Marking Diagram



Physical Dimension, Tape and Reel Information

Datasheet



Revision History

Date	Revision	Changes
16.Dec.2015	001	New Release

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CLASSⅢ	CLASSⅢ	CLASS II b	
CLASSⅣ	CLASSII	CLASSⅢ	CLASSII

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 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation depending on ambient temperature. When used in sealed area, confirm that it is the use in the range that does not exceed the maximum junction temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

Precaution for Mounting / Circuit board design

- 1. When a highly active halogenous (chlorine, bromine, etc.) flux is used, the residue of flux may negatively affect product performance and reliability.
- 2. In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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- 1. If change is made to the constant of an external circuit, please allow a sufficient margin considering variations of the characteristics of the Products and external components, including transient characteristics, as well as static characteristics.
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This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

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 - [c] the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- 4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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Тел: +7 (812) 336 43 04 (многоканальный) Email: org@lifeelectronics.ru

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