High Speed Dual MOSFET Driver

Features

- 6ns rise and fall time with 1000pF load
- ▶ 2.0A peak output source/sink current
- ▶ 1.2V to 5V input CMOS compatible
- 4.5V to 13V total supply voltage
- Smart Logic threshold
- Low jitter design
- Two matched channels
- Outputs can swing below ground
- Low inductance package
- Thermally-enhanced package

Applications

- Medical ultrasound imaging
- Piezoelectric transducer drivers
- Nondestructive evaluation
- PIN diode driver
- CCD Clock driver/buffer
- High speed level translator

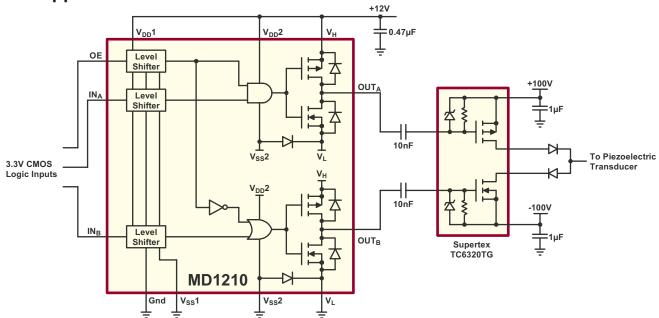
General Description

The Supertex MD1210 is a high speed, dual MOSFET driver. It is designed to drive high voltage P and N-channel MOSFET transistors for medical ultrasound and other applications requiring a high output current for a capacitive load. The high-speed input stage of the MD1210 can operate from 1.2V to 5.0V logic interface with an optimum operating input signal range of 1.8V to 3.3V. An adaptive threshold circuit is used to set the level translator switch threshold to the average of the input logic 0 and logic 1 levels. The input logic levels may be ground referenced, even though the driver is putting out bipolar signals. The level translator uses a proprietary circuit, which provides DC coupling together with high-speed operation.

 $\rm V_{DD1},\ V_{DD2},\ and\ V_{H}\ should\ be\ connected\ to\ the\ positive\ supply\ voltage,\ and\ V_{SS1},\ V_{SS2},\ and\ V_{L}\ should\ be\ connected\ to\ 0V\ or\ to\ Ground.$ The GND pin is the logic control input signal digital ground. The output stage is capable of peak currents of up to $\pm 2.0A,$ depending on the supply voltages used and load capacitance present.

The OE pin serves a dual purpose. First, its logic H level is used to compute the threshold voltage level for the channel input level translators. Secondly, when OE is low, the outputs are disabled, with the A output high and the B output low. This assists in properly pre-charging the AC coupling capacitors that may be used in series in the gate drive circuit of an external PMOS and NMOS transistor pair.

Typical Application Circuit



Ordering Information

DEVICE	Package Option
DEVICE	12-Lead 4x4x0.8pitch QFN
MD1210	MD1210K6-G

⁻G indicates package is RoHS compliant ('Green')

Absolute Maximum Ratings

Parameter	Value
$V_{\rm DD1}, V_{\rm DD2}, V_{\rm H}$ - supply voltage	-0.5V to +13.5V
V_{SS1}, V_{SS2}, V_L - supply voltage	0V
Logic input levels	-0.5V to 7.0V
Maximum junction temperature	+125°C
Storage temperature	-65°C to 150°C
Operating temperature	-20°C to 85°C

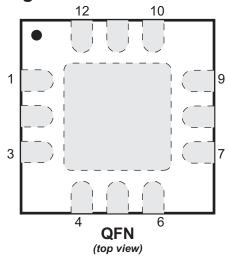
Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied. Continuous operation of the device at the absolute rating level may affect device reliability. All voltages are referenced to device ground.

P





Pin Configuration



DC Electrical Characteristics

(Over operating conditions unless otherwise specified, $V_H = V_{DD1} = V_{DD2} = 12V$, $V_L = V_{SS1} = V_{SS2} = 0V$, $V_{OE} = 3.3V$, $T_J = 25$ °C)

Sym	Parameter	Min	Тур	Max	Units	Conditions	
V_{DD1}, V_{DD2}	Supply voltage	4.5	-	13	V		
V _H	Output high supply voltage	V _{SS} + 2.0	ı	V _{DD}	V		
V_L	Output low supply voltage	-	ı	V _{DD} - 2.0	V		
I _{DD1Q}	V _{DD1} quiescent current	-	0.55	-	mA		
I _{DD2Q}	V _{DD2} quiescent current	-	-	10	μA	No input transitions	
l _{HQ}	V _H quiescent current	-	-	10	μA		
I _{DD1}	V _{DD1} average current	-	0.88	-	mA	0	
l _{DD2}	V _{DD2} average current	-	6.6	-	mA	One channel on at 5.0Mhz, No load	
I _H	V _H average current	-	23	-	mA	110 1000	
V _{IH}	Input logic voltage high	V _{OE} - 0.3	-	5.0	V		
V _{IL}	Input logic voltage low	0	-	0.3	V	For logic inputs INA and INB	
I _{IH}	Input logic current high	-	-	1.0	μA	For logic inputs INA and INB	
I _{IL}	Input logic current low	-	-	1.0	μA		
V _{IH}	OE Input logic voltage high	1.2	-	5.0	V		
V _{IL}	OE Input logic voltage low	0	-	0.3	V	For logic input OE	
R _{IN}	Input logic impedance to GND	12	20	30	ΚΩ		
C _{IN}	Logic input capacitance	-	5.0	10	pF	All inputs	
θ_{JA}	Thermal resistance to air	-	47	-	°C/W	1oz. 4-layer 3x4" PCB with thermal pad and thermal via array	
$\theta_{ extsf{JC}}$	Thermal resistance to case	-	7.0	-	°C/W		

Outputs $(V_H = V_{DD1} = V_{DD2} = 12V, V_L = V_{SS1} = V_{SS2} = 0V, V_{OE} = 3.3V, T_J = 25^{\circ}\text{C})$

Sym	Parameter	Min	Тур	Max	Units	Conditions
R _{SINK}	Output sink resistance	-	-	12.5	Ω	I _{SINK} = 50mA
R _{SOURCE}	Output source resistance	-	-	12.5	Ω	I _{SOURCE} = 50mA
I _{SINK}	Peak output sink current	-	2.0	-	А	
SOURCE	Peak output source current	-	2.0	-	Α	

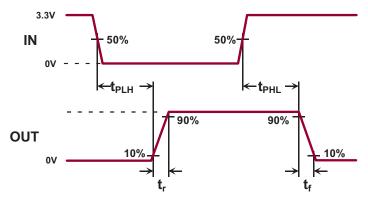
AC Electrical Characteristics $(V_H = V_{DD1} = V_{DD2} = 12V, V_I = V_{SS1} = V_{SS2} = 0V, V_{OF} = 3.3V, T_J = 25^{\circ}C)$

Sym	Parameter	Min	Typ	Max		Conditions	
t _{irf}	Inputs or OE rise & fall time	-	-	10	ns	Logic input edge speed requirement	
t _{PLH}	Propagation delay when output is from low to high	-	7.0	-	ns		
t _{PHL}	Propagation delay when output is from high to low	-	7.0	-	ns	C _{LOAD} = 1000pF, see timing diagram Input signal rise/fall time of 2ns	
t _{POE}	Propagation delay OE to outputs	-	9.0	-	ns		
t _r	Output rise time	-	6.0	-	ns		
t,	Output fall time	-	6.0	-	ns		
It _r - t _f I	Rise and fall time matching	-	1.0	-	ns		
I t _{PLH} -t _{PHL} I	Propagation low to high and high to low matching	-	1.0	-	ns	For each channel	
Δt_{dm}	Propagation delay match	-	±2.0	-	ns	Device to device delay match	

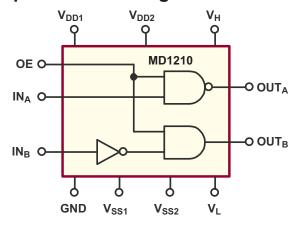
Logic Truth Table

	Logic Inputs	Out	put	
OE	INA	INB	OUTA	OUTB
Н	L	L	V _H	V _H
Н	L	Н	V _H	V _L
Н	Н	L	$V_{\scriptscriptstyle L}$	V _H
Н	Н	Н	V _L	V _L
L	X	X	V _H	V _L

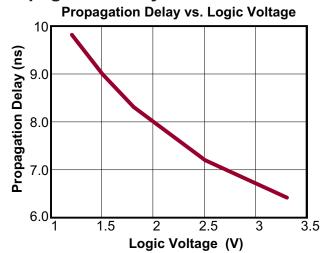
Timing Diagram



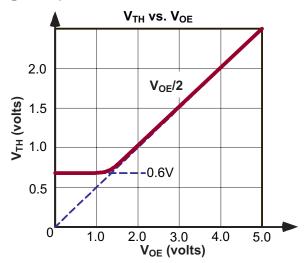
Simplified Block Diagram



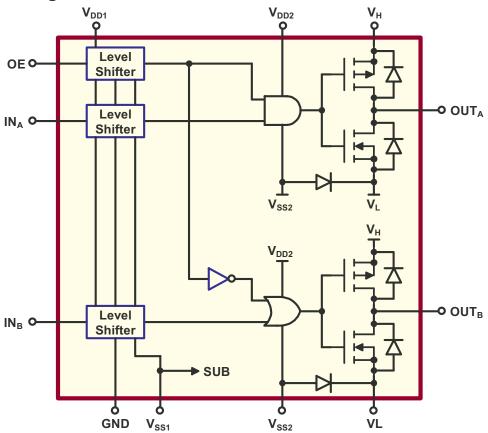
Propagation Delay



Logic Input Threshold



Detailed Block Diagram



Application Information

For proper operation of the MD1210, low inductance bypass capacitors should be used on the various supply pins. The GND input pin should be connected to the digital ground. The IN_A, IN_B, and OE pins should be connected to their logic source with a swing of GND to logic level high, which is 1.2V to 5.0V. Good trace practices should be followed corresponding to the desired operating speed. The internal circuitry of the MD1210 is capable of operating up to 100MHz, with the primary speed limitation being the loading effects of the load capacitance. Because of this speed and the high transient currents that result with capacitive loads, the bypass capacitors should be as close to the chip pins as possible. The V_{sst} , V_{ss2}, and V₁ pins should have low inductance feed-through connections directly to a ground plane. The power connections $V_{\rm DD1}$ and $V_{\rm DD2}$ should have a ceramic bypass capacitor to the ground plane with short leads and decoupling components to prevent resonance in the power leads. A common capacitor and voltage source may be used for these two pins, which should always have the same DC voltage applied. For applications sensitive to jitter and noise, separate decoupling networks may be used for V_{DD1} and V_{DD2} .

The V_H and V_L pins can draw fast transient currents of up to 2.0A, so they should be provided with an appropriate bypass capacitor located next to the chip pins. A ceramic capacitor of up to 1.0 μ F may be appropriate, with a series ferrite bead to prevent resonance in the power supply lead coming to the capacitor. Pay particular attention to minimizing trace lengths and using sufficient trace width to reduce inductance. Surface mount components are highly recommended. Since the output impedance of this driver is very low, in some cases it may be desirable to add a small series resistor in series with the output signal to obtain better waveform integrity at the load terminals. This will of course reduce the output voltage slew rate at the terminals of a capacitive load.

Pay particular attention to the parasitic coupling from the driver output to the input signal terminals. This feedback may cause oscillations or spurious waveform shapes on the edges of signal transitions. Since the input operates with signals down to 1.2V even small coupled voltages may cause problems. Use of a solid ground plane and good power and signal layout practices will prevent this problem. Be careful that the circulating ground return current from a capacitive load cannot react with common inductance to cause noise voltages in the input logic circuitry.

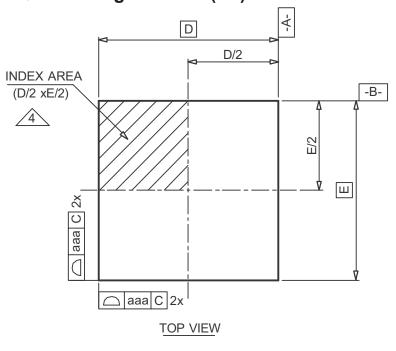
Pin Description

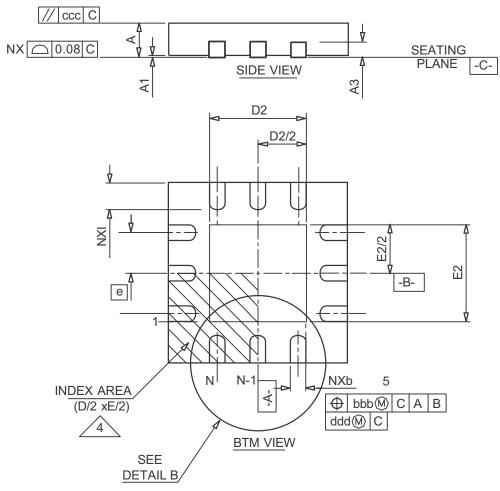
Pin #	Name	Description
1	IN _A	Logic input. Controls OUT_A when OE is high. Input logic high will cause the output to swing to V_L . Input logic low will cause the output to swing to V_H .
2	V _L	Supply voltage for N-channel output stage.
3	IN _B	Logic input. Controls OUT_B when OE is high. Input logic high will cause the output to swing to V_L . Input logic low will cause the output to swing to V_H .
4	GND	Logic input ground reference.
5	V _{SS1}	Low side analog circuit and level shifter supply voltage. Should be at the same potential as V _{SS2} .
6	V _{SS2}	Low side gate drive supply voltage.
7	OUT _B	Output driver. Swings from V_H to V_L . Intended to drive the gate of an external N-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUT_B will swing to V_L turning off the external N-channel MOSFET.
8	V _H	Supply voltage for P-channel output stage.
9	OUT _A	Output driver. Swings from V_H to V_L . Intended to drive the gate of an external P-channel MOSFET via a series capacitor. When OE is low, the output is disabled. OUT_A will swing to V_H turning off the external P-channel MOSFET.
10	$V_{_{\mathrm{DD2}}}$	High side gate drive supply voltage.
11	V _{DD1}	High side analog circuit and level shifter supply voltage. Should be at the same potential as V _{DD2} .
12	OE	Output-enable logic input. When OE is high, $(V_{OE} + V_{GND})/2$ sets the threshold transition between logic level high and low for IN_A and IN_B . When OE is low, OUT_A is at V_H and OUT_B is at V_L regardless of IN_A and IN_B

Note: 1.Thermal Pad and Pin#5 (V_{SS1}) must be connected externally.

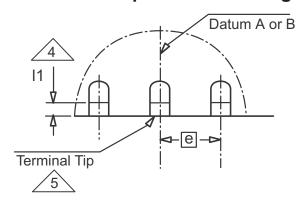
2. Index Pad and Thermal Pad are connected internally

12-Lead 4x4x0.8pitch QFN Package Outline (K6)

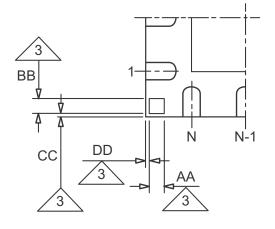




12-Lead 4x4x0.8pitch QFN Package Outline (K6)



Symbol	Height Dimensions			
	Min Nom		Max	
D BSC		4.0		
E BSC		4.0		
е		0.80		
D2	2.0	2.15	2.25	
E2	2.0	2.15	2.25	
b	0.25	0.30	0.35	
I	0.45	0.55	0.65	
А	0.80	0.90	1.0	
A1	0.00	0.02	0.05	
А3		0.20 ref		
L1	0.03		0.15	
Issue	А			



Bottom ID Dimensions			
AA BB CC DD			
.434	.434	.181	.181

Tolerance of Form & Position			
aaa	0.15		
bbb	0.10		
ccc	0.10		
ddd	0.05		
Issue	A		

Notes:

- 1. Dimensioning and tolerancing conform to ASME Y14.5m 1994.
- 2. All dimensions are in millimeters, all angles are in degrees (°).
- 3. The terminal #1 identifier and terminal numbering convention shall conform to JEDEC publication 95, SPP-002. Details of terminal #1 identifier are optional, but must be located within the zone indicated. The terminal #1identifier may be either a mold or marked feature.
- 4. Depending on the method of lead termination at the edge of the package, pull back (L1) may be present. L minus L1 to be equal to or greater than 0.33mm.
- 5. Dimension B applies to metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip. If the terminal has the optional radius on the other end of the terminal, the dimension B should not be measured in that radius area.

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to http://www.supertex.com/packaging.html.)

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ИНН 7805602321 КПП 780501001 P/C 40702810122510004610 ФАКБ "АБСОЛЮТ БАНК" (ЗАО) в г.Санкт-Петербурге К/С 3010181090000000703 БИК 044030703

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