SC28L92

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter (DUART)

Rev. 07 – 19 December 2007 Product data sheet

1. General description

The SC28L92 is a pin and function replacement for the SCC2692 and SC26C92 operating at 3.3 V or 5 V supply with added features and deeper FIFOs. Its configuration on power-up is that of the SC26C92. Its differences from the SCC2692 and SC26C92 are: 16 character receiver, 16 character transmit FIFOs, watchdog timer for each receiver, mode register 0 is added, extended baud rate and overall faster speeds, programmable receiver and transmitter interrupts. (Neither the SC26C92 nor the SCC2692 is being discontinued.)

Pin programming will allow the device to operate with either the Motorola or Intel bus interface. The bit 3 of the MR0A register allows the device to operate in an 8 byte FIFO mode if strict compliance with the SC26C92 FIFO structure is required.

The NXP Semiconductors SC28L92 Dual Universal Asynchronous Receiver/Transmitter (DUART) is a single-chip CMOS-LSI communications device that provides two full-duplex asynchronous receiver/transmitter channels in a single package. It interfaces directly with microprocessors and may be used in a polled or interrupt driven system with modem and DMA interface.

The operating mode and data format of each channel can be programmed independently. Additionally, each receiver and transmitter can select its operating speed as one of 28 fixed baud rates; a 16× clock derived from a programmable counter/timer, or an external $1\times$ or 16 \times clock. The baud rate generator and counter/timer can operate directly from a crystal or from external clock inputs. The ability to independently program the operating speed of the receiver and transmitter make the DUART particularly attractive for dual-speed channel applications such as clustered terminal systems.

Each receiver and transmitter is buffered by 8 or 16 character FIFOs to minimize the potential of receiver overrun, transmitter underrun and to reduce interrupt overhead in interrupt driven systems. In addition, a flow control capability is provided via RTS/CTS signaling to disable a remote transmitter when the receiver buffer is full.

Also provided on the SC28L92 are a multipurpose 7-bit input port and a multipurpose 8-bit output port. These can be used as general purpose I/O ports or can be assigned specific functions (such as clock inputs or status/interrupt outputs) under program control.

The SC28L92 is available in three package versions: PLCC44, QFP44, and HVQFN48.

2. Features

- Member of IMPACT family: 3.3 V to 5.0 V, -40 °C to +85 °C and 68xxx or 80xxx bus interface for all devices
- Dual full-duplex independent asynchronous receiver/transmitters
- 16 character FIFOs for each receiver and transmitter
- Pin programming selects 68xxx or 80xxx bus interface
- Programmable data format
	- ◆ 5 data to 8 data bits plus parity
	- ◆ Odd, even, no parity or force parity
	- \blacklozenge 1 stop, 1.5 stop or 2 stop bits programmable in $\frac{1}{16}$ -bit increments
- 16-bit programmable counter/timer
- Programmable baud rate for each receiver and transmitter selectable from:
	- \triangle 28 fixed rates: 50 kBd to 230.4 kBd
		- ◆ Other baud rates to 1 MHz at 16 \times
		- ◆ Programmable user-defined rates derived from a programmable counter/timer
	- \blacklozenge External 1 \times or 16 \times clock
- Parity, framing, and overrun error detection
- False start bit detection
- Line break detection and generation
- Programmable channel mode
	- ◆ Normal (full-duplex)
	- ◆ Automatic echo
	- ◆ Local loopback
	- ◆ Remote loopback
	- ◆ Multi-drop mode (also called wake-up or 9-bit)
- Multi-function 7-bit input port (includes IACKN)
	- ◆ Can serve as clock or control inputs
	- ◆ Change of state detection on four inputs
	- \triangle Inputs have typically > 100 k Ω pull-up resistors
	- ◆ Change of state detectors for modem control
- Multi-function 8-bit output port
	- ◆ Individual bit set/reset capability
	- ◆ Outputs can be programmed to be status/interrupt signals
	- ◆ FIFO status for DMA interface
- Versatile interrupt system
	- ◆ Single interrupt output with eight maskable interrupting conditions
	- ◆ Output port can be configured to provide a total of up to six separate interrupt outputs that may be wire ORed
	- ◆ Each FIFO can be programmed for four different interrupt levels
	- ◆ Watchdog timer for each receiver
- Maximum data transfer rates: 1×-1 Mbit/s, 16×-1 Mbit/s
- Automatic wake-up mode for multi-drop applications
- Start-end break interrupt/status
- Detects break which originates in the middle of a character

- On-chip crystal oscillator
- Power-down mode
- Receiver time-out mode
- Single 3.3 V or 5 V power supply
- Powers up to emulate SC26C92

3. Ordering information

Table 1. Ordering information

 V_{CC} = 3.3 V ± 10 % or V_{CC} = 5.0 V ± 10 %; T_{amb} = -40 °C to +85 °C

NXP Semiconductors SC28L92

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter

4. Block diagram

NXP Semiconductors SC28L92

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter

Fig 2. Block diagram (68xxx mode)

5. Pinning information

5.1 Pinning

NXP Semiconductors SC28L92

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter

NXP Semiconductors SC28L92

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter

5.2 Pin description

Table 2. Pin description for 80xxx bus interface (Intel) continued

Table 2. Pin description for 80xxx bus interface (Intel) continued

[1] HVQFN48 package die supply ground is connected to both GND pin and exposed center pad. GND pin must be connected to supply ground for proper device operation. For enhanced thermal, electrical, and board level performance, the exposed pad needs to be soldered to the board using a corresponding thermal pad on the board and for proper heat conduction through the board, thermal vias need to be incorporated in the PCB in the thermal pad region.

Table 3. Pin description for 68xxx bus interface (Motorola)

Table 3. Pin description for 68xxx bus interface (Motorola) continued

Table 3. Pin description for 68xxx bus interface (Motorola) …continued

[1] HVQFN48 package die supply ground is connected to both GND pin and exposed center pad. GND pin must be connected to supply ground for proper device operation. For enhanced thermal, electrical, and board level performance, the exposed pad needs to be soldered to the board using a corresponding thermal pad on the board and for proper heat conduction through the board, thermal vias need to be incorporated in the PCB in the thermal pad region.

6. Functional description

6.1 Block diagram

The SC28L92 DUART consists of the following eight major sections: data bus buffer, operation control, interrupt control, timing, communications channels A and B, input port and output port. Refer to [Section 4 "Block diagram"](#page-3-1).

6.1.1 Data bus buffer

The data bus buffer provides the interface between the external and internal data buses. It is controlled by the operation control block to allow read and write operations to take place between the controlling CPU and the DUART.

6.1.2 Operation control

The operation control logic receives operation commands from the CPU and generates appropriate signals to internal sections to control device operation. It contains address decoding and read and write circuits to permit communications with the microprocessor via the data bus.

6.1.3 Interrupt control

A single active LOW interrupt output (INTRN) is provided which is activated upon the occurrence of any of eight internal events. Associated with the interrupt system are the Interrupt Mask Register (IMR) and the Interrupt Status Register (ISR). The IMR can be programmed to select only certain conditions to cause INTRN to be asserted. The ISR can be read by the CPU to determine all currently active interrupting conditions. Outputs OP3 to OP7 can be programmed to provide discrete interrupt outputs for the transmitter,

receivers, and counter/timer. When OP3 to OP7 are programmed as interrupts, their output buffers are changed to the open-drain active LOW configuration. The OP pins may be used for DMA and modem control as well (see [Section](#page-48-0) 7.4).

6.1.4 FIFO configuration

Each receiver and transmitter has a 16 byte FIFO. These FIFOs may be configured to operate at a fill capacity of either 8 bytes or 16 bytes. This feature may be used if it is desired to operate the SC28L92 in strict compliance with the SC26C92. The 8 byte or 16 byte mode is controlled by the MR0A[3] bit. A logic 0 value for this bit sets the 8-bit mode (the default); a logic 1 sets the 16 byte mode. MR0A bit 3 sets the FIFO size for both channels.

The FIFO fill interrupt level automatically follow the programming of the MR0A[3] bit. See [Table](#page-26-0) 25 and [Table](#page-26-1) 26.

6.1.5 68xxx mode

When the I/M pin is connected to GND (ground), the operation of the SC28L92 switches to the bus interface compatible with the Motorola bus interfaces. Several of the pins change their function as follows:

- IP6 becomes IACKN input
- RDN becomes DACKN
- WRN becomes R/WN

The interrupt vector is enabled and the interrupt vector will be placed on the data bus when IACKN is asserted LOW. The interrupt vector register is located at address 0xC. The contents of this register are set to 0x0F on the application of RESETN.

The generation of DACKN uses two positive edges of the X1 clock as the DACKN delay from the falling edge of CEN. **If the CEN is withdrawn before two edges of the X1 clock occur, the generation of DACKN is terminated**. Systems not strictly requiring DACKN may use the 68xxx mode with the bus timing of the 80xxx mode greatly decreasing the bus cycle time.

6.2 Timing circuits

6.2.1 Crystal clock

The timing block consists of a crystal oscillator, a baud rate generator, a programmable 16-bit counter/timer, and four clock selectors. The crystal oscillator operates directly from a crystal connected across the X1/CLK and X2 inputs. If an external clock of the appropriate frequency is available, it may be connected to X1/CLK. The clock serves as the basic timing reference for the Baud Rate Generator (BRG), the counter/timer, and other internal circuits. A clock signal within the limits specified in [Section 10 "Dynamic](#page-53-0) [characteristics"](#page-53-0) must always be supplied to the DUART. If an external clock is used instead of a crystal, X1 should be driven using a configuration similar to the one in [Figure](#page-61-0) 17. Nominal crystal rate is 3.6864 MHz. Rates up to 8 MHz may be used.

6.2.2 Baud rate generator

The baud rate generator operates from the oscillator or external clock input at the X1 input and is capable of generating 28 commonly used data communications baud rates ranging from 50 kBd to 38.4 kBd. Programming bit 0 of MR0 to a logic 1 gives additional baud rates of 57.6 kBd, 115.2 kBd and 230.4 kBd (500 kHz with X1 at 8.0 MHz). Note that the MR0A[2:0] control this change and that the change applies to both channels. MR0B[2:0] are reserved.

The baud rates are based on an input frequency of 3.6864 MHz. Changing the X1 frequency will change all baud rates by ratio of 3.6864 MHz to the new frequency. All rates generated by the BRG will be in the $16\times$ mode. The clock outputs from the BRG are at $16\times$ the actual baud rate.

The counter/timer can be used as a timer to produce a $16\times$ clock for any other baud rate by counting down the crystal clock or an external clock. The four clock selectors allow the independent selection, for each receiver and transmitter, of any of these baud rates or external timing signal. The use of the counter/timer also requires the generation of a frequency $16\times$ of the baud rate. See [Section](#page-14-0) 6.2.3.

6.2.3 Counter/timer

The Counter/timer is a 16-bit programmable divider that operates in one of three modes: counter, timer and time-out. In the timer mode it generates a square wave. In the counter mode it generates a time delay. In the time-out mode it monitors the time between received characters. The C/T uses the numbers loaded into the Counter/Timer Lower Register (CTLR) and the Counter/Timer Upper Register (CTUR) as its divisor.

The counter/timer clock source and mode of operation (counter or timer) is selected by the Auxiliary Control Register bits 6 to 4 (ACR[6:4]). The output of the counter/timer may be used for a baud rate and/or may be output to the OP pins for some external function that may be totally unrelated to data transmission. The counter/timer also sets the counter/timer ready bit in the Interrupt Status Register (ISR) when its output transitions from logic 1 to logic 0. A register read address (see [Table](#page-22-0) 4) is reserved to issue a start counter/timer command and a second register read address is reserved to issue a stop command. The value of D[7:0] is ignored. The START command always loads the contents of CTUR, CTLR to the counting registers. The STOP command always resets the ISR[3] bit in the interrupt status register.

6.2.4 Timer mode

In the timer mode a symmetrical square wave is generated whose half period is equal in time to division of the selected counter/timer clock frequency by the 16-bit number loaded in the CTLR CTUR. Thus, the frequency of the counter/timer output will be equal to the counter/timer clock frequency divided by twice the value of the CTUR CTLR. While in the timer mode the ISR bit 3 (ISR[3]) will be set each time the counter/timer transitions from logic 1 to logic 0 (HIGH-to-LOW). This continues regardless of issuance of the stop counter command. ISR[3] is reset by the stop counter command.

Note: Reading of the CTU and CTL registers in the timer mode is not meaningful. When the C/T is used to generate a baud rate and the C/T is selected through the CSR then the receivers and/or transmitter will be operating in the 16× mode. Calculation for the number n to program the counter/timer upper and lower registers is shown in [Equation](#page-15-0) 1. The value of the divisor n is

ⁿ counter/*timer input clock* $\overline{2 \times 16 \times (desired\; baud\; rate)}$

Often this division will result in a non-integer number; 26.3 for example. One may only program integer numbers to a digital divider. Therefore 26 (0x1A) would be chosen. If 26.7 were the result of the division, then 27 (0x1B) would be chosen. This gives a baud rate error of 0.3/26.3 or 0.3/26.7 that yields a percentage error of 1.14 % or 1.12 % respectively, well within the ability of the asynchronous mode of operation. Higher input frequency to the counter reduces the error effect of the fractional division.

6.2.5 Counter mode

In the counter mode the counter/timer counts the value of the CTLR CTUR down to zero and then sets the ISR[3] bit and sets the counter/timer output from 1 to 0. It then rolls over to 65,365 and continues counting with no further observable effect. Reading the C/T in the counter mode outputs the present state of the C/T. If the C/T is not stopped, a read of the C/T may result in changing data on the data bus.

6.2.6 Time-out mode

The time-out mode uses the received data stream to control the counter. The time-out mode forces the C/T into the timer mode. Each time a received character is transferred from the shift register to the Rx FIFO, the counter is restarted. If a new character is not received before the counter reaches zero count, the counter ready bit is set, and an interrupt can be generated. This mode can be used to indicate when data has been left in the Rx FIFO for more than the programmed time limit. If the receiver has been programmed to interrupt the CPU when the receive FIFO is full, and the message ends before the FIFO is full, the CPU will not be interrupted for the remaining characters in the Rx FIFO.

By programming the C/T such that it would time-out in just over one character time, the above situation could be avoided. The processor would be interrupted any time the data stream had stopped for more than one character time. **Note**: This is very similar to the watchdog time of MR0. The difference is in the programmability of the delay time and that the watchdog timer is restarted by either a receiver load to the Rx FIFO or a system read from it.

This mode is enabled by writing the appropriate command to the command register. Writing 0xA to CRA or CRB will invoke the time-out mode for that channel. Writing 0xC to CRA or CRB will disable the time-out mode. Only one receiver should use this mode at a time. However, if both are on, the time-out occurs after both receivers have been inactive for the time-out period. The start of the C/T will be on the logic OR of the two receivers.

The time-out mode disables the regular start counter or stop counter commands and puts the C/T into counter mode under the control of the received data stream. Each time a received character is transferred from the shift register to the Rx FIFO, the C/T is stopped after one C/T clock, reloaded with the value in CTUR and CTLR and then restarted on the next C/T clock. If the C/T is allowed to end the count before a new character has been received, the counter ready bit, ISR[3], will be set. If IMR[3] is set, this will generate an interrupt. Since receiving a character restarts the C/T, the receipt of a character after the C/T has timed out will clear the counter ready bit, ISR[3], and the interrupt. Invoking the Set Time-out Mode On command, $CRx = 0xA$, will also clear the counter ready bit and stop the counter until the next character is received. The counter/timer is controlled with six commands: Start/Stop C/T, Read/Write Counter/Timer lower register and Read/Write

(1)

Counter/Timer upper register. These commands have slight differences depending on the mode of operation. Please see the detail of the commands in [Section 7.3.3 "Command](#page-34-0) [registers".](#page-34-0)

6.2.7 Time-out mode caution

When operating in the special time-out mode, it is possible to generate what appears to be a false interrupt, i.e., an interrupt without a cause. This may result when a time-out interrupt occurs and then, **before** the interrupt is serviced, another character is received, i.e., the data stream has started again. (The interrupt latency is longer than the pause in the data stream.) In this case, when a new character has been received, the counter/timer will be restarted by the receiver, thereby withdrawing its interrupt. If, at this time, the interrupt service begins for the previously seen interrupt, a read of the ISR will show the Counter Ready bit not set. If nothing else is interrupting, this read of the ISR will return a 0x00 character. This action may present the appearance of a spurious interrupt.

6.2.8 Communications channels A and B

Each communications channel of the SC28L92 comprises a full-duplex asynchronous receiver/transmitter (UART). The operating frequency for each receiver and transmitter can be selected independently from the baud rate generator, the counter/timer, or from an external input. The transmitter accepts parallel data from the CPU, converts it to a serial bit stream, inserts the appropriate start, stop, and optional parity bits and outputs a composite serial stream of data on the TxD output pin. The receiver accepts serial data on the RxD pin, converts this serial input to parallel format, checks for start bit, stop bit, parity bit (if any), or break condition and sends an assembled character to the CPU via the receive FIFO. Three status bits (break received, framing and parity errors) are also FIFOed with each data character.

6.2.9 Input port

The inputs to this unlatched 7-bit (6-bit for 68xxx mode) port can be read by the CPU by performing a read operation at address 0xD. A HIGH input results in a logic 1 while a LOW input results in a logic 0. D7 will always read as a logic 1. The pins of this port can also serve as auxiliary inputs to certain portions of the DUART logic, modem and DMA.

Four change-of-state detectors are provided which are associated with inputs IP3, IP2, IP1 and IP0. A HIGH-to-LOW or LOW-to-HIGH transition of these inputs, lasting longer than 25 µs to 50 µs, will set the corresponding bit in the input port change register. The bits are cleared when the register is read by the CPU. Any change of state can also be programmed to generate an interrupt to the CPU.

The input port change of state detection circuitry uses a 38.4 kHz sampling clock derived from one of the baud rate generator taps. This results in a sampling period of slightly more than 25 µs (this assumes that the clock input is 3.6864 MHz). The detection circuitry, in order to guarantee that a true change in level has occurred, requires two successive samples at the new logic level be observed. As a consequence, the minimum duration of the signal change is 25 µs if the transition occurs coincident with the first sample pulse. The 50 µs time refers to the situation in which the change of state is just missed and the first change of state is not detected until 25 µs later.

6.2.10 Output port

The output ports are controlled from six places: the OPCR, OPR, MR, Command, SOPR and ROPR registers. The OPCR register controls the source of the data for the output ports OP2 through OP7. The data source for output ports OP0 and OP1 is controlled by the MR and CR registers. When the OPR is the source of the data for the output ports, the data at the ports is inverted from that in the OPR register. The content of the OPR register is controlled by the set output port bits command and the reset output bits command. These commands are at 0xE and 0xF, respectively. When these commands are used, action takes place only at the bit locations where ones exist. For example, a one in bit location 5 of the data word used with the set output port bits command will result in OPR5 being set to one. The OP5 would then be set to zero (V_{SS}). Similarly, a one in bit position 5 of the data word associated with the reset output ports bits command would set OPR5 to zero and, hence, the pin OP5 to a one (V_{DD}) .

These pins along with the IP pins and their change-of-state detectors are often used for modem and DMA control.

6.3 Operation

6.3.1 Transmitter

The SC28L92 is conditioned to transmit data when the transmitter is enabled through the command register. The SC28L92 indicates to the CPU that it is ready to accept a character by setting the TxRDY bit in the status register. This condition can be programmed to generate an interrupt request at OP6 or OP7 and INTRN. When the transmitter is initially enabled the TxRDY and TxEMPT bits will be set in the status register. When a character is loaded to the transmit FIFO the TxEMPT bit will be reset. The TxEMPT will not set until: 1) the transmit FIFO is empty and the transmit shift register has finished transmitting the stop bit of the last character written to the transmit FIFO, or 2) the transmitter is disabled and then re-enabled. The TxRDY bit is set whenever the transmitter is enabled and the Tx FIFO is not full. Data is transferred from the holding register to transmit shift register when it is idle or has completed transmission of the previous character. Characters cannot be loaded into the Tx FIFO while the transmitter is disabled.

The transmitter converts the parallel data from the CPU to a serial bit stream on the TxD output pin. It automatically sends a start bit followed by the programmed number of data bits, an optional parity bit, and the programmed number of stop bits. The least significant bit is sent first. Following the transmission of the stop bits, if a new character is not available in the Tx FIFO, the TxD output remains HIGH and the TxEMT bit in the Status Register (SR) will be set to 1. Transmission resumes and the TxEMT bit is cleared when the CPU loads a new character into the Tx FIFO.

If the transmitter is disabled it continues operating until the character currently being transmitted and any characters in the Tx FIFO, including parity and stop bits, have been transmitted. New data cannot be loaded to the Tx FIFO when the transmitter is disabled.

When the transmitter is reset it stops sending data immediately.

The transmitter can be forced to send a break (a continuous LOW condition) by issuing a START BREAK command via the CR register. The break is terminated by a STOP BREAK command or a transmitter reset.

If CTS option is enabled ($MR2[4] = 1$), the CTS input at IP0 or IP1 must be LOW in order for the character to be transmitted. The transmitter will check the state of the CTS input at the beginning of each character transmitted. If it is found to be HIGH, the transmitter will delay the transmission of any following characters until the CTS has returned to the LOW state. CTS going HIGH during the serialization of a character will not affect that character.

The transmitter can also control the RTSN outputs, OP0 or OP1 via MR2[5]. When this mode of operation is set, the meaning of the OP0 or OP1 signals will usually be end of message. See description of bit MR2[5] in [Table 30 "MR2A - Mode Register 2 channel A](#page-28-0) [\(address 0x0\) bit description"](#page-28-0) for more detail. This feature may be used to automatically turn around a transceiver in simplex systems.

6.3.2 Receiver

The SC28L92 is conditioned to receive data when enabled through the command register. The receiver looks for a HIGH-to-LOW (mark-to-space) transition of the start bit on the RxD input pin. If a transition is detected, the state of the RxD pin is sampled each $16\times$ clock for 7 clocks to $\frac{1}{2}$ clocks (16 \times clock mode) or at the next rising edge of the bit time clock (1× clock mode). If RxD is sampled HIGH, the start bit is invalid and the search for a valid start bit begins again. If RxD is still LOW, a valid start bit is assumed and the receiver continues to sample the input at one bit time intervals at the theoretical center of the bit, until the proper number of data bits and parity bit (if any) have been assembled, and one stop bit has been detected. The least significant bit is received first. The data is then transferred to the receive FIFO and the RxRDY bit in the SR is set to a 1. This condition can be programmed to generate an interrupt at OP4 or OP5 and INTRN. If the character length is less than 8 bits, the most significant unused bits in the Rx FIFO are set to zero.

After the stop bit is detected, the receiver will immediately look for the next start bit. However, if a non-zero character was received without a stop bit (framing error) and RxD remains LOW for one half of the bit period after the stop bit was sampled, then the receiver operates as if a new start bit transition had been detected at that point (one-half bit time after the stop bit was sampled).

The parity error, framing error and overrun error (if any) are strobed into the SR from the next byte to be read from the Rx FIFO. If a break condition is detected (RxD is LOW for the entire character including the stop bit), a character consisting of all zeros will be loaded into the Rx FIFO and the received break bit in the SR is set to 1. The RxD input must return to HIGH for two (2) clock edges of the X1 crystal clock for the receiver to recognize the end of the break condition and begin the search for a start bit.

This will usually require a HIGH time of one X1 clock period or 3 X1 edges since the clock of the controller is not synchronous to the X1 clock.

6.3.3 Transmitter reset and disable

Note the difference between transmitter disable and reset. A transmitter reset stops transmitter action immediately, clears the transmitter FIFO and returns the idle state. A transmitter disable withdraws the transmitter interrupts but allows the transmitter to continue operation until all bytes in its FIFO and shift register have been transmitted including the final stop bits. It then returns to its idle state.

6.3.4 Receiver FIFO

The Rx FIFO consists of a First-In-First-Out (FIFO) stack with a capacity of 8 or 16 characters. Data is loaded from the receive shift register into the topmost empty position of the FIFO. The RxRDY bit in the status register is set whenever one or more characters are available to be read, and a FFULL status bit is set if all 8 or 16 stack positions are filled with data. Either of these bits can be selected to cause an interrupt. A read of the Rx FIFO outputs the data at the top of the FIFO. After the read cycle, the data FIFO and its associated status bits (see [Section](#page-19-0) 6.3.5) are popped thus emptying a FIFO position for new data.

A disabled receiver with data in its FIFO may generate an interrupt (see [Section](#page-19-0) 6.3.5). Its status bits remain active and its watchdog, if enabled, will continue to operate.

6.3.5 Receiver status bits

In addition to the data word, three status bits (parity error, framing error and received break) are also appended to each data character in the FIFO. The overrun error, MR1[5], is not FIFOed.

Status can be provided in two ways, as programmed by the error mode control bit in the mode register. In the character mode, status is provided on a character-by-character basis; the status applies only to the character at the top of the FIFO. In the block mode, the status provided in the SR for these three bits is the logic OR of the status for all characters coming to the top of the FIFO since the last reset error from the command register was issued. In either mode reading the SR does not affect the FIFO. The FIFO is popped only when the Rx FIFO is read. Therefore the status register should be read prior to reading the FIFO.

If the FIFO is full when a new character is received, that character is held in the receive shift register until a FIFO position is available. If an additional character is received while this state exits, the contents of the FIFO are not affected; the character previously in the shift register is lost and the overrun error status bit (SR[4]) will be set upon receipt of the start bit of the new (overrunning) character.

The receiver can control the deactivation of RTS. If programmed to operate in this mode, the RTSN output will be negated when a valid start bit was received and the FIFO is full. When a FIFO position becomes available, the RTSN output will be reasserted (set LOW) automatically. This feature can be used to prevent an overrun, in the receiver, by connecting the RTSN output to the CTSN input of the transmitting device.

If the receiver is disabled, the FIFO characters can be read. However, no additional characters can be received until the receiver is enabled again. If the receiver is reset, the FIFO and all of the receiver status, and the corresponding output ports and interrupt are reset. No additional characters can be received until the receiver is enabled again.

6.3.6 Receiver reset and disable

Receiver disable stops the receiver immediately. Data being assembled in the receiver shift register is lost. Data and status in the FIFO is preserved and may be read. A re-enable of the receiver after a disable will cause the receiver to begin assembling characters at the next start bit detected.

A receiver reset will discard the present shift register date, reset the receiver ready bit (RxRDY), clear the status of the byte at the top of the FIFO and realign the FIFO read/write pointers.

6.3.7 Watchdog

A watchdog timer is associated with each receiver. Its interrupt is enabled by MR0[7]. The purpose of this timer is to alert the control processor that characters are in the Rx FIFO which have not been read. This situation may occur at the end of a transmission when the last few characters received are not sufficient to cause an interrupt.

This counter times out after 64 bit times. It is reset each time a character is transferred from the receiver shift register to the Rx FIFO or a read of the Rx FIFO is executed.

6.3.8 Receiver time-out mode

In addition to the watchdog timer described in [Section](#page-20-0) 6.3.7, the counter/timer may be used for a similar function. Its programmability, of course, allows much greater precision of time-out intervals.

The time-out mode uses the received data stream to control the counter. Each time a received character is transferred from the shift register to the Rx FIFO, the counter is restarted. If a new character is not received before the counter reaches zero count, the counter ready bit is set, and an interrupt can be generated. This mode can be used to indicate when data has been left in the Rx FIFO for more than the programmed time limit. Otherwise, if the receiver has been programmed to interrupt the CPU when the receive FIFO is full, and the message ends before the FIFO is full, the CPU may not know there is data left in the FIFO. The CTU and CTL value would be programmed for just over one character time, so that the CPU would be interrupted as soon as it has stopped receiving continuous data. This mode can also be used to indicate when the serial line has been marking for longer than the programmed time limit. In this case, the CPU has read all of the characters from the FIFO, but the last character received has started the count. If there is no new data during the programmed time interval, the counter ready bit will get set, and an interrupt can be generated.

The time-out mode is enabled by writing the appropriate command to the command register. Writing 0xA to CRA or CRB will invoke the time-out mode for that channel. Writing 0xC to CRA or CRB will disable the time-out mode. The time-out mode should only be used by one channel at once, since it uses the C/T. If, however, the time-out mode is enabled from both receivers, the time-out will occur only when **both** receivers have stopped receiving data for the time-out period. CTU and CTL must be loaded with a value greater than the normal receive character period. The time-out mode disables the regular start counter or stop counter commands and puts the C/T into counter mode under the control of the received data stream. Each time a received character is transferred from the shift register to the Rx FIFO, the C/T is stopped after one C/T clock, reloaded with the value in CTU and CTL and then restarted on the next C/T clock. If the C/T is allowed to end the count before a new character has been received, the counter ready bit, ISR[3], will be set. If IMR[3] is set, this will generate an interrupt. Receiving a character after the C/T has timed out will clear the counter ready bit, ISR[3], and the interrupt. Invoking the set time-out mode on command, CRx = 0xA, will also clear the counter ready bit and stop the counter until the next character is received.

6.3.9 Time-out mode caution

When operating in the special time-out mode, it is possible to generate what appears to be a false interrupt, i.e., an interrupt without a cause. This may result when a time-out interrupt occurs and then, **before** the interrupt is serviced, another character is received, i.e., the data stream has started again. (The interrupt latency is longer than the pause in the data stream.) In this case, when a new character has been receiver, the counter/timer will be restarted by the receiver, thereby withdrawing its interrupt. If, at this time, the interrupt service begins for the previously seen interrupt, a read of the ISR will show the counter ready bit not set. If nothing else is interrupting, this read of the ISR will return a 0x00 character.

6.3.10 Multi-drop mode (9-bit or wake-up)

The DUART is equipped with a wake-up mode for multi-drop applications. This mode is selected by programming bits MR1A[4:3] or MR1B[4:3] to 11 for channels A and B, respectively. In this mode of operation, a master station transmits an address character followed by data characters for the addressed slave station. The slave stations, with receivers that are normally disabled, examine the received data stream and wake-up the CPU (by setting RxRDY) only upon receipt of an address character. The CPU compares the received address to its station address and enables the receiver if it wishes to receive the subsequent data characters. Upon receipt of another address character, the CPU may disable the receiver to initiate the process again.

A transmitted character consists of a start bit, the programmed number of data bits, and Address/Data (A/D) bit, and the programmed number of stop bits. The polarity of the transmitted A/D bit is selected by the CPU by programming bit MR1A[2]/MR1B[2]. $MR1A[2]/MR1B[2] = 0$ transmits a zero in the A/D bit position, which identifies the corresponding data bits as data while MR1A[2]/MR1B[2] = 1 transmits a one in the A/D bit position, which identifies the corresponding data bits as an address. The CPU should program the mode register prior to loading the corresponding data bits into the Tx FIFO.

In this mode, the receiver continuously looks at the received data stream, whether it is enabled or disabled. If disabled, it sets the RxRDY status bit and loads the character into the Rx FIFO if the received A/D bit is a one (address tag), but discards the received character if the received A/D bit is a zero (data tag). If enabled, all received characters are transferred to the CPU via the Rx FIFO. In either case, the data bits are loaded into the data FIFO while the A/D bit is loaded into the status FIFO position normally used for parity error (SRA[5] or SRB[5]). Framing error, overrun error, and break detect operate normally whether or not the receive is enabled.

7. Programming

7.1 Register overview

The operation of the DUART is programmed by writing control words into the appropriate registers. Operational feedback is provided via status registers which can be read by the CPU. The addressing of the registers is described in [Table](#page-22-0) 4.

The contents of certain control registers are initialized to zero on RESET. Care should be exercised if the contents of a register are changed during operation, since certain changes may cause operational problems.

For example, changing the number of bits per character while the transmitter is active may cause the transmission of an incorrect character. In general, the contents of the MR, the CSR, and the OPCR should only be changed while the receiver(s) and transmitter(s) are not enabled, and certain changes to the ACR should only be made while the C/T is stopped.

Each channel has three mode registers (MR0, MR1 and MR2) which control the basic configuration of the channel. Access to these registers is controlled by independent MR address pointers. These pointers are set to 0x0 or 0x1 by MR control commands in the command register Miscellaneous Commands. Each time the MR registers are accessed the MR pointer increments, stopping at MR2. It remains pointing to MR2 until set to 0x0 or 0x1 via the miscellaneous commands of the command register. The pointer is set to 0x1 on reset for compatibility with previous Philips Semiconductors UART software.

Mode, command, clock select, and status registers are duplicated for each channel to provide total independent operation and control. Refer to [Section](#page-23-0) 7.2 for register bit overview. The reserved registers at addresses 0x2 and 0xA should never be read during normal operation since they are reserved for internal diagnostics.

Table 4. SC28L92 register addressing READ (RDN = 0), WRITE (WRN = 0)[\[1\]](#page-22-1)

[1] The three MR registers are accessed via the MR Pointer and Commands 0x1n and 0xBn (where n = represents receiver and transmitter enable bits).

Table 5. Registers for channels A and B

Table 5. Registers for channels A and B continued

Table 6. Registers supporting both channels

7.2 Condensed register bit formats

Table 21. ROPR - Reset Output Port bits Register (ROPR)

Table 22. OPCR - Output Port Configuration Register

OP1 and OP0 are the RTSN output and are controlled by the MR register

7.3 Register descriptions

7.3.1 Mode registers

7.3.1.1 Mode Register 0 channel A (MR0A)

Table 23. MR0A - Mode Register 0 channel A (address 0x0) bit allocation

MR0 is accessed by setting the MR pointer to logic 0 via the command register command B.

Table 25. Receiver FIFO interrupt fill level[\[1\]](#page-26-2)

[1] Interrupt fill level must be set when the transmit and receive FIFOs are empty, otherwise the new level takes effect only after a read or a write to the FIFO.

Table 26. Transmitter FIFO interrupt fill level[\[1\]](#page-26-3)

[1] Interrupt fill level must be set when the transmit and receive FIFOs are empty, otherwise the new level takes effect only after a read or a write to the FIFO.

7.3.1.2 Mode Register 1 channel A (MR1A)

Table 27. MR1A - Mode Register 1 channel A (address 0x0) bit allocation

MR1A is accessed when the channel A MR pointer points to MR1. The pointer is set to MR1 by RESET or by a set pointer command applied via CR command 1. After reading or writing MR1A, the pointer will point to MR2A^{[\[1\]](#page-27-0)}.

[1] In block error mode, block error conditions must be cleared by using the error reset command (command 0x4) or a receiver reset.

Table 28. MR1A - Mode Register 1 channel A (address 0x0) bit description

7.3.1.3 Mode Register 2 channel A (MR2A)

Table 29. MR2A - Mode Register 2 channel A (address 0x0) bit allocation

MR2A is accessed when the channel A MR pointer points to MR2, which occurs after any access to MR1A. Accesses to MR2A do not change the pointer.

Table 31. DUART mode description

Table 32. Stop bit length

Table 32. Stop bit length …continued

\sim	
MR2A[3:0] (hexadecimal)	Stop bit length ^[1]
В	1.750
C	1.813
D	1.875
F	1.938
	2.000

[1] Add 0.5 to values shown for 0 to 7 if channel is programmed for 5 bit per character

7.3.1.4 Mode Register 0 channel B (MR0B)

MR0B (address 0x8) is accessed when the channel B MR pointer points to MR1. The pointer is set to MR0 by RESET or by a set pointer command applied via CRB. After reading or writing MR0B, the pointer will point to MR1B.

The bit definitions for this register are identical to MR0A, except the FIFO size bit and that all control actions apply to the channel B receiver, transmitter, the corresponding inputs and outputs. MR0B[2:0] are reserved.

7.3.1.5 Mode Register 1 channel B (MR1B)

MR1B (address 0x8) is accessed when the channel B MR pointer points to MR1. The pointer is set to MR1 by RESET or by a set pointer command applied via CRB. After reading or writing MR1B, the pointer will point to MR2B.

The bit definitions for this register are identical to MR1A, except that all control actions apply to the channel B receiver and transmitter and the corresponding inputs and outputs.

7.3.1.6 Mode Register 2 channel B (MR2B)

MR2B (address 0x8) is accessed when the channel B MR pointer points to MR2, which occurs after any access to MR1B. Accesses to MR2B do not change the pointer.

The bit definitions for mode register are identical to the bit definitions for MR2A, except that all control actions apply to the channel B receiver and transmitter and the corresponding inputs and outputs.

7.3.2 Clock select registers

Table 35. Baud rate (based on a 3.6864 MHz crystal clock) See [Table](#page-33-0) 36 for bit rate characteristics.

Table 36. Bit rate generator characteristic[s\[1\]](#page-33-1) Crystal or clock = 3.6864 MHz.

[1] Duty cycle of $16 \times$ clock is 50 % \pm 1 %.

7.3.2.2 Clock Select Register channel B (CSRB)

7.3.3 Command registers

7.3.3.1 Command Register channel A (CRA)

CRA is a register used to supply commands to channel A. Multiple commands can be specified in a single write to CRA as long as the commands are non-conflicting, e.g., the enable transmitter and reset transmitter commands cannot be specified in a single command word.

Table 39. CRA - Command register channel A (address 0x2) bit description

Table 40. Miscellaneous commands

Table 40. Miscellaneous commands …continued

7.3.3.2 Command Register channel B (CRB)

CRB is a register used to supply commands to channel B. Multiple commands can be specified in a single write to CRB as long as the commands are non-conflicting, e.g., the enable transmitter and reset transmitter commands cannot be specified in a single command word.

The bit definitions for this register are identical to the bit definitions for CRA, with the exception of miscellaneous commands 0xE and 0xF which are used for Power-down mode. These two commands are not used in CRB. All other control actions that apply to CRA also apply to CRB.
7.3.4 Status registers

7.3.4.1 Status Register channel A (SRA)

[1] These status bits are appended to the corresponding data character in the receive FIFO. A read of the status provides these bits [7:5] from the top of the FIFO together with bits [4:0]. These bits are cleared by a reset error status command. In character mode they are discarded when the corresponding data character is read from the FIFO. In block error mode, the error-reset command (command 0x4 or receiver reset) must used to clear block error conditions.

Table 42. SRA - Status register channel A (address 0x1) bit description

7.3.4.2 Status Register channel B (SRB)

Table 43. SRB - Status register channel B (address 0x9) bit allocation

[1] These status bits are appended to the corresponding data character in the receive FIFO. A read of the status provides these bits [7:5] from the top of the FIFO together with bits [4:0]. These bits are cleared by a reset error status command. In character mode they are discarded when the corresponding data character is read from the FIFO. In block error mode, the error-reset command (command 0x4 or receiver reset) must used to clear block error conditions.

The bit definitions for this register are identical to the bit definitions for SRA, except that all status applies to the channel B receiver and transmitter and the corresponding inputs and outputs.

7.3.5 Output Configuration Control Register (OPCR)

This register controls the signal presented by the OP[7:2] pins. The signal presented by the OP[1:0] pins is controlled by the Rx, Tx, and the command register. The default condition of the OP pins is to drive the complement of the data in the OPR[7:0] register.

When OP[7:2] pins drive DMA or interrupt type signals, they switch to open-drain configuration. Otherwise, they drive strong logic 0 or logic 1 levels.

Table 44. OPCR - Output configuration control register (address 0xD) bit allocation

	Table ++. Of OR - Output comiquiation control register (address vAD) bit anocation						
configure OP ₇	configure OP6	configure OP ₅	configure OP ₄	configure OP3		configure OP2	
OPCR - Output configuration control register (address 0xD) bit description Table 45.							

this output is not masked by the contents of the IMR.

Table 45. OPCR - Output configuration control register (address 0xD) bit description

7.3.6 Set Output Port bits Register (SOPR)

Ones in the byte written to this register will cause the corresponding bit positions in the OPR to set to 1. Zeros have no effect. This allows software to set individual bits without keeping a copy of the OPR bit configuration.

Table 47. SOPR - Set output port bits register (address 0xE) bit description

7.3.7 Reset Output Port bits Register (ROPR)

Ones in the byte written to the ROPR will cause the corresponding bit positions in the OPR to set to 0. Zeros have no effect. This allows software to reset individual bits without keeping a copy of the OPR bit configuration.

Table 49. ROPR - Reset output port bits register (address 0xF) bit description

7.3.8 Output Port Register (OPR)

Table 50. OPR - Output port register (no address) bit allocation

The output pins (OP pins) drive the complement of the data in this register as controlled by SOPR and ROPR.

7.3.9 Auxiliary Control Register (ACR)

Table 52. ACR - Auxiliary control register (address 0x4) bit allocation

[1] The timer mode generates a square wave.

7.3.10 Input Port Change Register (IPCR)

Table 55. IPCR - Input port change register (address 0x4) bit allocation

7.3.11 Interrupt Status Register (ISR)

This register provides the status of all potential interrupt sources. The contents of this register are masked by the Interrupt Mask Register (IMR). If a bit in the ISR is a logic 1 and the corresponding bit in the IMR is also a logic 1, the INTRN output will be asserted (LOW). If the corresponding bit in the IMR is a zero, the state of the bit in the ISR has no effect on the INTRN output. Note that the IMR does not mask the reading of the ISR. The true status will be provided regardless of the contents of the IMR. The contents of this register are initialized to 0x0 when the DUART is reset.

Table 57. ISR - Interrupt status register (address 0x5) bit allocation

Table 58. ISR - Interrupt status register (address 0x5) bit description

programming of ACR[3:0]. The bit is cleared when the CPU reads the IPCR.

7.3.12 Interrupt Mask Register (IMR)

The programming of this register selects which bits in the ISR causes an interrupt output. If a bit in the ISR is a logic 1 and the corresponding bit in the IMR is also a logic 1 the INTRN output will be asserted. If the corresponding bit in the IMR is a zero, the state of the bit in the ISR has no effect on the INTRN output. Note that the IMR does not mask the programmable interrupt outputs OP3 to OP7 or the reading of the ISR.

7.3.13 Interrupt Vector Register (IVR; 68xxx mode) or General Purpose register (GP; 80xxx mode)

This register stores the Interrupt Vector. It is initialized to 0x0F on hardware reset and is usually changed from this value during initialization of the SC28L92. The contents of this register will be placed on the data bus when IACKN is asserted LOW or a read of address 0xC is performed.

When not operating in the 68xxx mode, this register may be used as a general purpose one byte storage register. A convenient use could be to store a shadow of the contents of another SC28L92 register (IMR, for example).

7.3.14 Counter/timer registers

The CTPU and CTPL hold the eight MSBs and eight LSBs, respectively, of the value to be used by the counter/timer in either the counter or timer modes of operation. The minimum value which may be loaded into the CTPU/CTPL registers is 0x0002. Note that these registers are write only and cannot be read by the CPU.

In the timer mode, the C/T generates a square wave whose period is twice the value (in C/T clock periods) of the CTPU and CTPL. The waveform so generated is often used for a data clock. The formula for calculating the divisor n to load to the CTPU and CTPL for a particular $1 \times$ data clock is shown in [Equation](#page-47-0) 2.

```
n = \frac{counter/timer \ clock \ frequency}{2 \times 16 \times (desired \ band \ rate)}
```
(2)

Often this division will result in a non-integer number; 26.3, for example. One can only program integer numbers in a digital divider. Therefore, 26 would be chosen. This gives a baud rate error of 0.3/26.3 which is 1.14 %; well within the ability asynchronous mode of operation.

The C/T will not be running until it receives an initial start counter command (read at address A3 to A0 = 1110). After this, while in timer mode, the C/T will run continuously. Receipt of a start counter command (read with A3 to $A0 = 1110$) causes the counter to terminate the current timing cycle and to begin a new cycle using the values in CTPU and CTPL. If the value in CTPU and CTPL is changed, the current half-period will not be affected, but subsequent half periods will be affected.

The counter ready status bit (ISR[3]) is set once each cycle of the square wave. The bit is reset by a stop counter command (read with A3 to $A0 = 1111$). The command however, does not stop the C/T. The generated square wave is output on OP3 if it is programmed to be the C/T output. In the counter mode, the value C/T loaded into CTPU and CTPL by the CPU is counted down to 0. Counting begins upon receipt of a start counter command. Upon reaching terminal count 0x0000, the counter ready interrupt bit (ISR[3]) is set. The counter continues counting past the terminal count until stopped by the CPU. If OP3 is programmed to be the output of the C/T, the output remains HIGH until terminal count is reached, at which time it goes LOW. The output returns to the HIGH state and ISR[3] is cleared when the counter is stopped by a stop counter command. The CPU may change the values of CTPU and CTPL at any time, but the new count becomes effective only on the next start counter commands. If new values have not been loaded, the previous count values are preserved and used for the next count cycle.

In the counter mode, the current value of the upper and lower 8 bits of the counter (CTU, CTL) may be read by the CPU. It is recommended that the counter be stopped when reading to prevent potential problems which may occur if a carry from the lower 8 bits to the upper 8 bits occurs between the times that both halves of the counter are read. However, note that a subsequent start counter command will cause the counter to begin a new count cycle using the values in CTPU and CTPL.

When the C/T clock divided by 16 is selected, the maximum divisor becomes 1,048,575.

7.4 Output port notes

The output ports are controlled from four places: the OPCR register, the OPR register, the MR registers and the command register (except the SCC2681 and SCC68681). The OPCR register controls the source of the data for the output ports OP2 to OP7. The data source for output ports OP0 and OP1 is controlled by the MR and CR registers. When the OPR is the source of the data for the output ports, the data at the ports is inverted from that in the OPR register.

The content of the OPR register is controlled by the Set Output Port bits command and the Reset Output Port bits command. These commands are at 0xE and 0xF, respectively. When these commands are used, action takes place only at the bit locations where ones exist. For example, a logic 1 in bit location 5 of the data word used with the Set Output Port bits command will result in OPR5 being set to one. The OP5 would then be set to logic 0 ($V_{\rm SS}$). Similarly, a logic 1 in bit position 5 of the data word associated with the Reset Output Ports bits command would set OPR5 to logic 0 and, hence, the pin OP5 to a logic 1 (V_{DD}) .

7.5 The CTS, RTS, CTS enable Tx signals

Clear To Send (CTS) is usually meant to be a signal to the transmitter meaning that it may transmit data to the receiver. The CTS input is on pin IP0 for TxA and on IP1 for TxB. The CTS signal is active LOW; thus, it is called CTSAN for TxA and CTSBN for TxB. RTS is usually meant to be a signal from the receiver indicating that the receiver is ready to receive data. It is also active LOW and is, thus, called RTSAN for RxA and RTSBN for RxB. RTSAN is on pin OP0 and RTSBN is on OP1. A receiver's RTS output will usually be connected to the CTS input of the associated transmitter. Therefore, one could say that RTS and CTS are different ends of the same wire.

MR2[4] is the bit that allows the transmitter to be controlled by the CTS pin (IP0 or IP1). When this bit is set to one AND the CTS input is driven HIGH, the transmitter will stop sending data at the end of the present character being serialized. It is usually the RTS output of the receiver that will be connected to the transmitter's CTS input. The receiver will set RTS HIGH when the receiver FIFO is full AND the start bit of the 9th or 17th character is sensed. Transmission then stops with 9 or 17 valid characters in the receiver. When MR2[4] is set to one, CTSN must be at zero for the transmitter to operate. If MR2[4] is set to zero, the IP pin will have no effect on the operation of the transmitter. MR1[7] is the bit that allows the receiver to control OP0. When OP0 (or OP1) is controlled by the receiver, the meaning of that pin will be.

8. Limiting values

Table 64. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

[1] For operation at elevated temperatures, the device must be derated based on 150 °C maximum junction temperature.

[2] Parameters are valid over specified temperature range.

[3] This product includes circuitry specifically designed for the protection of its internal devices from damaging effects of excessive static charge. Nonetheless, it is suggested that conventional precautions be taken to avoid applying any voltages larger than the rated maxima.

9. Static characteristics

Table 65. Static characteristics, 5 V operation[\[1\]](#page-50-0)

 V_{CC} = 5 V ± 10 %; T_{amb} = -40 °C to +85 °C; unless otherwise specified.

[1] The following conditions apply:

a) Parameters are valid over specified temperature and voltage range.

b) All voltage measurements are referenced to ground. For testing, all inputs swing between 0.4 V and 3.0 V with a transition time of 5 ns maximum. For X1/CLK this swing is between 0.4 V and 0.8V_{CC}. All time measurements are referenced at input voltages of 0.8 V and 2.0 V, and output voltages of 0.8 V and 2.0 V, as appropriate.

c) Typical values are at 25 \degree C, typical supply voltages, and typical processing parameters.

[2] Test conditions for outputs: $C_L = 125$ pF, except open-drain outputs. Test conditions for open-drain outputs: $C_L = 125$ pF, constant current source = 2.6 mA.

[3] Input port pins have active pull-up transistors that will source a typical 2 μ A from V_{CC} when the input pins are at V_{SS}. Input port pins at V_{CC} source 0.0 μ A.

[4] All outputs are disconnected. Inputs are switching between CMOS levels of V_{CC} − 0.2 V and V_{SS} + 0.2 V.

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter

Table 66. Static characteristics, 3.3 V operatio[n\[1\]](#page-52-0)

[1] The following conditions apply:

b) All voltage measurements are referenced to ground. For testing, all inputs swing between 0.4 V and 3.0 V with a transition time of 5 ns maximum. For X1/CLK this swing is between 0.4 V and $0.8V_{CC}$. All time measurements are referenced at input voltages of 0.8 V and 2.0 V, and output voltages of 0.8 V and 2.0 V, as appropriate.

c) Typical values are at 25 °C, typical supply voltages, and typical processing parameters.

[2] Test conditions for outputs: C_L = 125 pF, except interrupt outputs. Test conditions for interrupt outputs: C_L = 125 pF, constant current source = 2.6 mA.

[3] Input port pins have active pull-up transistors that will source a typical 2 μ A from V_{CC} when they are at V_{SS}. Input port pins at V_{CC} source $0.0 \mu \dot{A}$.

[4] All outputs are disconnected. Inputs are switching between CMOS levels of V_{CC} − 0.2 V and V_{SS} + 0.2 V.

a) Parameters are valid over specified temperature and voltage range.

10. Dynamic characteristics

Table 67. Dynamic characteristics, 5 V operation[\[1\]](#page-54-0)

 V_{CC} = 5.0 V \pm 10 %, T_{amb} = -40 °C to +85 °C, unless otherwise specified.

Table 67. Dynamic characteristics, 5 V operation^[1] ... continued

 V_{CC} = 5.0 V \pm 10 %, T_{amb} = -40 °C to +85 °C, unless otherwise specified.

[1] The following conditions apply:

a) Parameters are valid over specified temperature and voltage range.

b) All voltage measurements are referenced to ground. For testing, all inputs swing between 0.4 V and 3.0 V with a transition time of 5 ns maximum. For X1/CLK this swing is between 0.4 V and $0.8V_{CC}$. All time measurements are referenced at input voltages of 0.8 V and 2.0 V, and output voltages of 0.8 V and 2.0 V, as appropriate.

c) Test conditions for outputs: $C_L = 125$ pF, except open-drain outputs. Test conditions for open-drain outputs: $C_1 = 125$ pF, constant current source = 2.6 mA.

d) Typical values are the average values at $+25$ °C and 5 V.

[2] Timing is illustrated and referenced to the WRN and RDN Inputs. Also, CEN may be the strobing input. CEN and RDN (also CEN and WRN) are ORed internally. The signal asserted last initiates the cycle and the signal negated first terminates the cycle.

[3] Guaranteed by characterization of sample units.

[4] If CEN is used as the strobing input, the parameter defines the minimum HIGH times between one CEN and the next. The RDN signal must be negated for t_{RWD} to guarantee that any status register changes are valid.

[5] Minimum frequencies are not tested but are quaranteed by design.

[6] Clocks for $1 \times$ mode should maintain a 60/40 duty cycle or better.

[7] Minimum DACKN time is $((t_{DCR}$ or $t_{DCW})$ t_{CSC} + 2 X1 edges + rise time over 5 ns). Two X1 edges is 273 ns at 3.6864 MHz. For faster bus cycles, the 80xxx bus timing may be used while in the 68xxx mode. It is not necessary to wait for DACKN to insure the proper operation of the SC26C92. In all cases the data will be written to the SC28L92 on the falling edge of DACKN or the rise of CEN. The fall of CEN initializes the bus cycle. The rise of CEN ends the bus cycle. DACKN LOW or CEN HIGH completes the write cycle.

Table 68. Dynamic characteristics, 3.3 V operatio[n\[1\]](#page-56-0)

 V_{CC} = 3.3 V ± 10 %, T_{amb} = -40 °C to +85 °C, unless otherwise specified.

Table 68. Dynamic characteristics, 3.3 V operation^[1] ...continued

 V_{CC} = 3.3 V \pm 10 %, T_{amb} = -40 °C to +85 °C, unless otherwise specified.

[1] The following conditions apply:

a) Parameters are valid over specified temperature and voltage range.

b) All voltage measurements are referenced to ground. For testing, all inputs swing between 0.4 V and 3.0 V with a transition time of 5 ns maximum. For X1/CLK this swing is between 0.4 V and $0.8V_{CC}$. All time measurements are referenced at input voltages of 0.8 V and 2.0 V, and output voltages of 0.8 V and 2.0 V, as appropriate.

c) Test conditions for outputs: $C_L = 125$ pF, except open-drain outputs. Test conditions for open-drain outputs: $C_1 = 125$ pF, constant current source = 2.6 mA.

d) Typical values are the average values at +25 °C and 3.3 V.

[2] Timing is illustrated and referenced to the WRN and RDN Inputs. Also, CEN may be the strobing input. CEN and RDN (also CEN and WRN) are ORed internally. The signal asserted last initiates the cycle and the signal negated first terminates the cycle.

[3] Guaranteed by characterization of sample units.

[4] If CEN is used as the strobing input, the parameter defines the minimum HIGH times between one CEN and the next. The RDN signal must be negated for t_{RWD} to guarantee that any status register changes are valid.

[5] Minimum frequencies are not tested but are quaranteed by design.

[6] Clocks for $1 \times$ mode should maintain a 60/40 duty cycle or better.

[7] Minimum DACKN time is $((t_{DCR}$ or $t_{DCW})$ t_{CSC} + 2 X1 edges + rise time over 5 ns). Two X1 edges is 273 ns at 3.6864 MHz. For faster bus cycles, the 80xxx bus timing may be used while in the 68xxx mode. It is not necessary to wait for DACKN to insure the proper operation of the SC26C92. In all cases the data will be written to the SC28L92 on the falling edge of DACKN or the rise of CEN. The fall of CEN initializes the bus cycle. The rise of CEN ends the bus cycle. DACKN LOW or CEN HIGH completes the write cycle.

11. Timing diagrams

3.3 V/5.0 V Dual Universal Asynchronous Receiver/Transmitter

12. Test information

13. Package outline

Fig 24. Package outline SOT187-2 (PLCC44)

Fig 25. Package outline SOT307-2 (QFP44)

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HVQFN48: plastic thermal enhanced very thin quad flat package; no leads; 48 terminals; body 6 x 6 x 0.85 mm

Fig 26. Package outline SOT778-4 (HVQFN48)

14. Soldering

This text provides a very brief insight into a complex technology. A more in-depth account of soldering ICs can be found in Application Note AN10365 "Surface mount reflow soldering description".

14.1 Introduction to soldering

Soldering is one of the most common methods through which packages are attached to Printed Circuit Boards (PCBs), to form electrical circuits. The soldered joint provides both the mechanical and the electrical connection. There is no single soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and Surface Mount Devices (SMDs) are mixed on one printed wiring board; however, it is not suitable for fine pitch SMDs. Reflow soldering is ideal for the small pitches and high densities that come with increased miniaturization.

14.2 Wave and reflow soldering

Wave soldering is a joining technology in which the joints are made by solder coming from a standing wave of liquid solder. The wave soldering process is suitable for the following:

- **•** Through-hole components
- **•** Leaded or leadless SMDs, which are glued to the surface of the printed circuit board

Not all SMDs can be wave soldered. Packages with solder balls, and some leadless packages which have solder lands underneath the body, cannot be wave soldered. Also, leaded SMDs with leads having a pitch smaller than ~0.6 mm cannot be wave soldered, due to an increased probability of bridging.

The reflow soldering process involves applying solder paste to a board, followed by component placement and exposure to a temperature profile. Leaded packages, packages with solder balls, and leadless packages are all reflow solderable.

Key characteristics in both wave and reflow soldering are:

- **•** Board specifications, including the board finish, solder masks and vias
- **•** Package footprints, including solder thieves and orientation
- **•** The moisture sensitivity level of the packages
- **•** Package placement
- **•** Inspection and repair
- **•** Lead-free soldering versus PbSn soldering

14.3 Wave soldering

Key characteristics in wave soldering are:

- **•** Process issues, such as application of adhesive and flux, clinching of leads, board transport, the solder wave parameters, and the time during which components are exposed to the wave
- **•** Solder bath specifications, including temperature and impurities

14.4 Reflow soldering

Key characteristics in reflow soldering are:

- **•** Lead-free versus SnPb soldering; note that a lead-free reflow process usually leads to higher minimum peak temperatures (see Figure 27) than a PbSn process, thus reducing the process window
- **•** Solder paste printing issues including smearing, release, and adjusting the process window for a mix of large and small components on one board
- **•** Reflow temperature profile; this profile includes preheat, reflow (in which the board is heated to the peak temperature) and cooling down. It is imperative that the peak temperature is high enough for the solder to make reliable solder joints (a solder paste characteristic). In addition, the peak temperature must be low enough that the packages and/or boards are not damaged. The peak temperature of the package depends on package thickness and volume and is classified in accordance with Table 69 and 70

Table 69. SnPb eutectic process (from J-STD-020C)

Table 70. Lead-free process (from J-STD-020C)

Moisture sensitivity precautions, as indicated on the packing, must be respected at all times.

Studies have shown that small packages reach higher temperatures during reflow soldering, see Figure 27.

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For further information on temperature profiles, refer to Application Note AN10365 "Surface mount reflow soldering description".

15. Abbreviations

16. Revision history

17. Legal information

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19. Contents

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