



3.0 V to 5.5 V, ± 12 kV IEC ESD Protected, 500 kbps/50 Mbps RS-485 Transceivers

Data Sheet

ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E

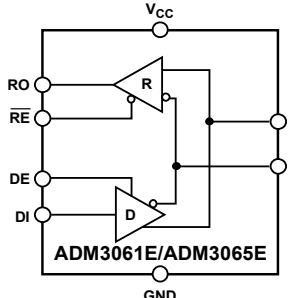
FEATURES

- TIA/EIA RS-485 compliant over full supply range
- 3.0 V to 5.5 V operating voltage range on V_{CC}
- 1.62 V to 5.5 V V_{IO} logic supply option available
- ESD protection on the bus pins
 - IEC 61000-4-2 ± 12 kV contact discharge
 - IEC 61000-4-2 $\geq \pm 12$ kV air discharge
 - HBM: $\geq \pm 30$ kV
- Full hot swap support (glitch free power-up/power-down)
- High speed 50 Mbps data rate (ADM3065E/ADM3066E/
ADM3067E/ADM3068E)
- Low speed 500 kbps data rate for long cables (ADM3061E/
ADM3062E/ADM3063E)
- Full receiver short-circuit, open circuit, and bus idle fail-safe
- Extended temperature range up to 125°C
- PROFIBUS compliant at $V_{CC} \geq 4.5$ V
- Half duplex and full duplex models available
- Allows connection of up to 128 transceivers onto the bus
- Space-saving package options
 - 10-lead, 3 mm × 3 mm LFCSP
 - 8-lead and 10-lead, 3 mm × 3 mm MSOP
 - 8-lead and 14-lead, narrow body SOIC

APPLICATIONS

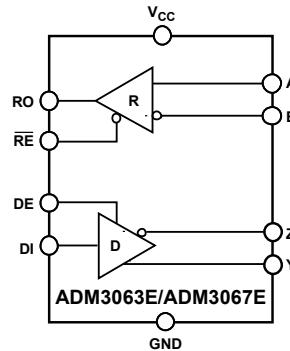
- Industrial fieldbuses
- Process control
- Building automation
- PROFIBUS networks
- Motor control servo drives and encoders

FUNCTIONAL BLOCK DIAGRAMS



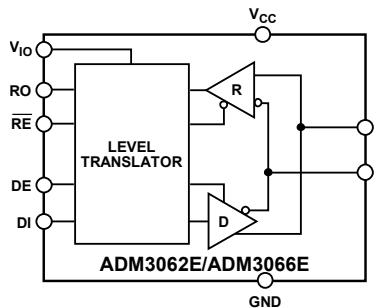
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Figure 1. ADM3061E/ADM3065E Functional Block Diagram



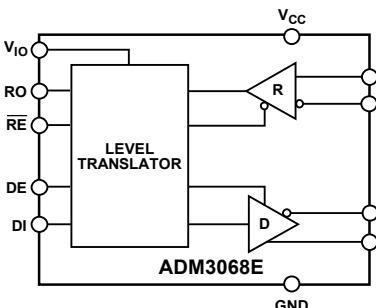
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Figure 2. ADM3063E/ADM3067E Functional Block Diagram



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Figure 3. ADM3062E/ADM3066E Functional Block Diagram



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Figure 4. ADM3068E Functional Block Diagram

Rev. F

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TABLE OF CONTENTS

Features	1
Applications	1
Functional Block Diagrams.....	1
Revision History	2
General Description	4
Specifications.....	5
Timing Specifications	7
Absolute Maximum Ratings.....	11
Thermal Resistance	11
ESD Caution.....	11
Pin Configurations and Function Descriptions	12
Typical Performance Characteristics	16
Test Circuits.....	20
Theory of Operation	21
REVISION HISTORY	
6/2019—Rev. E to Rev. F	
Added ADM3068E	Universal
Added Figure 2; Renumbered Sequentially	1
Changes to Features Section and Figure 4.....	1
Changes to Table 1 Title.....	4
Changes to Table 2.....	5
Changes to ADM3061E/ADM3062E/ADM3063E Section and Table 3	7
Changes to ADM3065E/ADM3066E/ADM3067E/ADM3068E Section and Table 4.....	8
Changes to Figure 5 and Figure 7	9
Changes to Figure 8	10
Added Endnote 1 to Digital Input and Output Voltage (DE, RE, DI, and RO) Parameter, Table 5.....	11
Changes to Digital Input and Output Voltage (DE, RE, DI, and RO) Parameter, Table 5	11
Changes to Table 8.....	13
Added Figure 14 and Table 10; Renumbered Sequentially	15
Changes to Figure 19 Caption.....	16
Changes to Figure 43 and Figure 44.....	20
Changes to Isolated High Speed RS-485 Node Section.....	24
Changes to Ordering Guide	27
4/2019—Rev. D to Rev. E	
Added ADM3063E	Universal
Change to Features Section	1
Changes to Figure 3	1
Changes to Table 1.....	4
Changes to Table 2.....	5
Added Endnote 1, Table 2; Renumbered Sequentially	5
Change to Table 7	11
Changes to Table 8.....	12
Changes to Figure 12 and Table 9.....	13
IEC ESD Protected RS-485	21
High Driver Differential Output Voltage.....	21
IEC 61000-4-2 ESD Protection	21
Truth Tables.....	22
Receiver Fail-Safe	22
Hot Swap Capability.....	22
128 Transceivers on the Bus.....	22
Driver Output Protection.....	22
Applications Information	23
Isolated High Speed RS-485 Node	24
Outline Dimensions	25
Ordering Guide	27

Changes to Figure 14, Figure 15, Figure 16, and Figure 18 Captions	14
Changes to Figure 19, Figure 20, Figure 23, and Figure 24 Captions	15
Change to Figure 25 Caption	14
Changes to Figure 39	18
Changes to IEC ESD Protected RS-485 Section	19
Changes to Truth Tables Section, Table 11, Table 12, and Receiver Fail-Safe Section	20
Added Table 10; Renumbered Sequentially	20
Changes to Isolated High Speed RS-485 Node Section and Figure 47	22
Changes to Ordering Guide.....	25

3/2019—Rev. C to Rev. D

Added ADM3067E and 14-Lead SOIC_N, R-14	Universal
Changes to Feature Section	1
Added Figure 3; Renumbered Sequentially	1
Moved Table 1 to	4
Changes to Table 2	5
Changes to ADM3065E/ADM3066E/ADM3067E Section	7
Change to Pin 3, Description Column, Table 7	11
Changes to Figure 10, Figure 11, and Table 8	12
Added Figure 12 and Table 9; Renumbered Sequentially	13
Changes to Figure 14	14
Moved Test Circuits to	18
Changes to Table 10 and Table 11	20
Updated Outline Dimensions	26
Changes to Ordering Guide	27

1/2018—Rev. B to Rev. C

Added ADM3062E	Universal
Changes to Figure 2 and Table 1	1

Changes to ADM3061E/ADM3062E Timing Specifications Section and Figure 3	6
Changes to Figure 5 and Figure 6	7
Changes to Figure 9 and Figure 10	11
Changes to Figure 16 and Figure 17	12
Changes to Figure 44	21
Changes to Figure 45	22
Changes to Ordering Guide	25

12/2017—Rev. A to Rev. B

Added ADM3061E.....	Universal
Changes to Product Title, Features Section, Figure 1, and Table 1 ...	1
Changes to General Description Section	3
Changes to Table 2	4
Added ADM3061E Timing Specification Section and Table 3; Renumbered Sequentially	6
Moved Figure 3	6
Moved Figure 4, Figure 5, and Figure 6	7
Changes to ADM3065E/ADM3066E Timing Specification Section Title	8
Added 10-Lead MSOP Parameter and 10-Lead LFCSP Parameter, Table 5.....	9
Changes to Operating Temperature Range Parameter, Table 5 and Table 6	9
Changes to Figure 7, Figure 8, and Table 7	10
Changes to Table 8	11
Changes to Figure 11	12
Added Figure 23; Renumbered Sequentially	13
Added Figure 24, Figure 25, Figure 26, Figure 27, and Figure 28 ..	14

Changed High Speed IEC ESD Protected RS-485 Section to IEC ESD Protected RS-485 Section	17
Changes to IEC ESD Protected RS-485 Section	17
Added Endnote 4, Table 9	18
Changes to Table 10	18
Changes to Figure 44	21
Changes to Figure 45	22
Changes to Ordering Guide	25

5/2017—Rev. 0 to Rev. A

Added ADM3066E.....	Universal
Changes to Features Section, Figure 1, and Table 1	1
Added Figure 2; Renumbered Sequentially	1
Moved General Description Section	3
Changes to General Description Section	3
Changes to Specifications Section and Table 2	4
Changes to Timing Specifications Section and Figure 3	5
Changes to Figure 4, Figure 5, and Figure 6	6
Added V _{IO} to GND Parameter, Table 4	7
Changes to Thermal Resistance Section and Table 5	7
Added Figure 8	8
Changes to Table 6	8
Added Figure 9 and Figure 10	9
Added Table 7; Renumbered Sequentially	9
Changes to Figure 14, Figure 16, and Figure 17	10
Changes to Table 8 and Table 9	15
Added Figure 42 and Figure 43	20
Changes to Ordering Guide	21

3/2017—Revision 0: Initial Version

GENERAL DESCRIPTION

The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E are 3.0 V to 5.5 V, IEC electrostatic discharge (ESD) protected RS-485 transceivers, allowing the devices to withstand ± 12 kV contact discharges on the transceiver bus pins without latch-up or damage. The ADM3062E/ADM3066E/ADM3068E feature a V_{IO} logic supply pin that allow a flexible digital interface capable of operating as low as 1.62 V.

The ADM3065E/ADM3066E/ADM3067E/ADM3068E are suitable for high speed, 50 Mbps, bidirectional data communication on multipoint bus transmission lines. The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E feature a 1/4 unit load input impedance that allows up to 128 transceivers on a bus. The ADM3061E/ADM3062E/ADM3063E models offer all of the same features as the ADM3065E/ADM3066E/ADM3067E/ADM3068E models at a low 500 kbps data rate that is suitable for operation over long cable runs.

The ADM3061E/ADM3062E/ADM3065E/ADM3066E are half-duplex RS-485 transceivers, fully compliant to the PROFIBUS® standard with increased 2.1 V bus differential voltage at $V_{CC} \geq 4.5$ V. The ADM3063E/ADM3067E/ADM3068E are full duplex RS-485 transceiver options.

The RS-485 transceivers are available in a number of space-saving packages, such as the 10-lead, 3 mm \times 3 mm lead frame chip-scale package (LFCSP), the 8-lead or 10-lead, 3 mm \times 3 mm mini small outline package (MSOP), and the 8-lead or 14-lead, narrow body standard small outline packages (SOIC_N). Models with operating temperature ranges of -40°C to $+125^{\circ}\text{C}$ and -40°C to $+85^{\circ}\text{C}$ are available.

Excessive power dissipation caused by bus contention or by output shorting is prevented by a thermal shutdown circuit. If a significant temperature increase is detected in the internal driver circuitry during fault conditions, this feature forces the driver output into a high impedance state.

The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E guarantee a logic high receiver output when the receiver inputs are shorted, open, or connected to a terminated transmission line with all drivers disabled.

Table 1 presents an overview of the ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E data rate capability across temperature, power supply, and package options. Refer to the Ordering Guide for model numbering.

Table 1. Summary of the ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E Operating Conditions—Data Rate Capability Across Temperature, Power Supply, and Package

Maximum Data Rate ¹	Maximum V_{CC} (V)	Maximum Temperature	Package Description
50 Mbps	5.5	-40°C to $+125^{\circ}\text{C}$	10-lead LFCSP
50 Mbps	5.5	-40°C to $+105^{\circ}\text{C}$	8-lead SOIC_N, 8-lead MSOP, 10-lead MSOP, and 14-lead SOIC_N
50 Mbps	3.6	-40°C to $+125^{\circ}\text{C}$	8-lead SOIC_N, 8-lead MSOP, 10-lead MSOP, and 14-lead SOIC_N
500 kbps	5.5	-40°C to $+125^{\circ}\text{C}$	8-lead SOIC_N, 8-lead MSOP, 10-lead MSOP, 10-lead LFCSP, and 14-lead SOIC_N

¹ The ADM3065E/ADM3066E/ADM3067E/ADM3068E data input (DI) is transmitting 50 Mbps (or 500 kbps for the ADM3061E/ADM3062E/ADM3063E) clock data, and the ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E driver enable (DE) is enabled for 50% of the DI transmit time

SPECIFICATIONS

V_{CC} = 3.0 V to 5.5 V, V_{IO} = 1.62 V to V_{CC} (ADM3062E/ADM3066E/ADM3068E), T_A = T_{MIN} (-40°C) to T_{MAX} (+125°C), unless otherwise noted. All typical specifications are at T_A = 25°C, V_{IO} = V_{CC} = 3.3 V, unless otherwise noted.

Table 2.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
POWER SUPPLY						
No Load Supply Current	I_{CC}		3.5	7.5	mA	$DE = V_{IO}$ ¹ , $\overline{RE} = 0$ V
			3.5	7.5	mA	$DE = V_{IO}$, $\overline{RE} = V_{IO}$
		3	4.5		mA	$DE = 0$ V, $\overline{RE} = 0$ V
ADM3065E/ADM3066E/ADM3067E/ ADM3068E Supply Current, Data Rate = 50 Mbps	I_{CC}	107	172		mA	Load resistance (R_L) = 54 Ω, $DE = V_{IO}$, $\overline{RE} = 0$ V ($V_{CC} \geq 4.5$ V)
ADM3061E/ADM3062E/ADM3063E Supply Current, Data Rate = 500 kbps	I_{CC}	67	75		mA	$R_L = 54$ Ω, $DE = V_{IO}$, $\overline{RE} = 0$ V ($V_{CC} = 3.0$ V)
		100	165		mA	$R_L = 54$ Ω, $DE = V_{IO}$, $\overline{RE} = 0$ V ($V_{CC} \geq 4.5$ V)
Supply Current in Shutdown Mode	I_{SHDN}	56	74		mA	$R_L = 54$ Ω, $DE = V_{IO}$, $\overline{RE} = 0$ V ($V_{CC} = 3.0$ V)
V_{IO} Shutdown Current ²	I_{IOSHDN}	210	450		μA	$DE = 0$ V, $\overline{RE} = V_{IO}$
		1	50		μA	$DE = 0$ V, $\overline{RE} = V_{IO}$
DRIVER						
Differential Outputs						
Output Voltage, Loaded	$ V_{OD2} $	2.0	2.5	V_{CC}	V	$V_{CC} \geq 3.0$ V, $R_L = 50$ Ω, see Figure 38
	$ V_{OD2} $	1.5	2.1	V_{CC}	V	$V_{CC} \geq 3.0$ V, $R_L = 27$ Ω (RS-485), see Figure 38
	$ V_{OD2} $	2.1	3.5	V_{CC}	V	$V_{CC} \geq 4.5$ V, $R_L = 50$ Ω, see Figure 38
	$ V_{OD2} $	2.1	3	V_{CC}	V	$V_{CC} \geq 4.5$ V, $R_L = 27$ Ω (RS-485), see Figure 38
	$ V_{OD3} $	1.5	2.1	V_{CC}	V	$V_{CC} \geq 3.0$ V, -7 V ≤ common-mode voltage (V_{CM}) ≤ +12 V, see Figure 39
	$ V_{OD3} $	2.1	3	V_{CC}	V	$V_{CC} \geq 4.5$ V, -7 V ≤ V_{CM} ≤ +12 V, see Figure 39
Change in Differential Input Voltage for Complementary Output States	$\Delta V_{OD} $		0.2		V	$R_L = 27$ Ω or 50 Ω, see Figure 38
Common-Mode Output Voltage	V_{OC}		1.6	3.0	V	$R_L = 27$ Ω or 50 Ω, see Figure 38
Change in Common Mode Voltage for Complementary Output States	$\Delta V_{OC} $		0.2		V	$R_L = 27$ Ω or 50 Ω, see Figure 38
Output Short-Circuit Current	I_{OS}	-250	+250		mA	-7 V < output voltage (V_{OUT}) < +12 V
ADM3063E/ADM3067E Output Leakage (Y, Z)	I_O		+100		μA	$DE = 0$ V, $\overline{RE} = 0$ V, $V_{CC} = 0$ V or 3.6 V, input voltage (V_{IN}) = 12 V
		-100			μA	$DE = 0$ V, $\overline{RE} = 0$ V, $V_{CC} = 0$ V or 3.6 V, $V_{IN} = -7$ V
Logic Inputs (DE, \overline{RE} , DI)						
Input Voltage						
Low	V_{IL}			$0.33 \times V_{IO}$	V	$DE, \overline{RE}, DI, 1.62$ V ≤ V_{IO} ≤ 5.5 V
High	V_{IH}		$0.67 \times V_{IO}$		V	$DE, \overline{RE}, DI, 1.62$ V ≤ V_{IO} ≤ 5.5 V
Input Current	I_I	-2	+2		μA	$DE, \overline{RE}, DI, 1.62$ V ≤ V_{IO} ≤ 5.5 V, 0 V ≤ V_{IN} ≤ V_{IO}

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
RECEIVER						
Differential Inputs						
Differential Input Threshold Voltage	V _{TH}	-200	-125	-30	mV	-7 V < V _{CM} < +12 V
Input Voltage Hysteresis	V _{HYS}		30		mV	-7 V < V _{CM} < +12 V
Input Current (A, B)	I _I		0.1	0.25	mA	DE = 0 V, V _{CC} = powered/unpowered, V _{IN} = 12 V
		-0.20	-0.1		mA	DE = 0 V, V _{CC} = powered/unpowered, V _{IN} = -7 V
Line Input Resistance	R _{IN}	48	96		kΩ	-7 V ≤ V _{CM} ≤ +12 V
Logic Outputs						
Output Voltage						
Low	V _{OL}			0.4	V	V _{IO} = 3.6 V, output current (I _{OUT}) = 2 mA, V _{ID} ³ ≤ -0.2 V
				0.4	V	V _{IO} = 2.7 V, I _{OUT} = 1 mA, V _{ID} ≤ -0.2 V ²
				0.2	V	V _{IO} = 1.95 V, I _{OUT} = +500 μA, V _{ID} ≤ -0.2 V ²
High	V _{OH}	2.4			V	V _{IO} = 3.0 V, I _{OUT} = -2 mA, V _{ID} ≥ -0.03 V
		2.0			V	V _{IO} = 2.3 V, I _{OUT} = -1 mA, V _{ID} ≥ -0.03 V ²
		V _{IO} - 0.2			V	V _{IO} = 1.65 V, I _{OUT} = -500 μA, V _{ID} ≥ -0.03 V ²
Short-Circuit Current				85	mA	V _{OUT} = GND or V _{IO}
Three-State Output Leakage	I _{OZR}			±2	μA	RO pin = 0 V or V _{IO}

¹ V_{IO} = V_{CC} for ADM3061E/ADM3063E/ADM3065E/ADM3067E.² ADM3062E/ADM3066E/ADM3068E only.³ V_{ID} is the receiver input differential voltage.

TIMING SPECIFICATIONS**ADM3061E/ADM3062E/ADM3063E**

V_{CC} = 3.0 V to 5.5 V, V_{IO} = 1.62 V to V_{CC} (ADM3062E), $T_A = T_{MIN}$ (-40°C) to T_{MAX} (+125°C), unless otherwise noted. All typical specifications are at $T_A = 25^\circ\text{C}$, $V_{IO} = V_{CC} = 3.3$ V, unless otherwise noted.

Table 3.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
DRIVER						
Maximum Data Rate ¹		500			kbps	
Propagation Delay	t_{DPLH}, t_{DPHL}		220	800	ns	R_{LDIFF} capacitor = 54 Ω, C_{L1} capacitor = C_{L2} capacitor = 100 pF, see Figure 5 and Figure 40
Skew	t_{DSKEW}		5	100	ns	R_{LDIFF} = 54 Ω, $C_{L1} = C_{L2} = 100$ pF, see Figure 5 and Figure 40
Rise/Fall Times	t_{DR}, t_{DF}	120	300	800	ns	R_{LDIFF} = 54 Ω, $C_{L1} = C_{L2} = 100$ pF, see Figure 5 and Figure 40
Enable to Output High	t_{DZH}		100	1000	ns	$R_L = 110 \Omega$, $C_L = 50$ pF, see Figure 6 and Figure 41
Enable to Output Low	t_{DZL}		100	1000	ns	$R_L = 110 \Omega$, $C_L = 50$ pF, see Figure 6 and Figure 41
Disable Time from Low	t_{DLZ}		350	2000	ns	$R_L = 110 \Omega$, $C_L = 50$ pF, see Figure 6 and Figure 41
Disable Time from High	t_{DHZ}		600	2000	ns	$R_L = 110 \Omega$, $C_L = 50$ pF, see Figure 6 and Figure 41
Enable Time from Shutdown to High	$t_{DZH(SHDN)}^2$		550	2000	ns	$R_L = 110 \Omega$, $C_L = 50$ pF, see Figure 6 and Figure 41
Enable Time from Shutdown to Low	$t_{DZL(SHDN)}^2$		550	2000	ns	$R_L = 110 \Omega$, $C_L = 50$ pF, see Figure 6 and Figure 41
RECEIVER						
Maximum Data Rate		500			kbps	
Propagation Delay	t_{RPLH}, t_{RPHL}		200		ns	$C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, $V_{CM} = 1.5$ V, see Figure 7 and Figure 42
Skew/Pulse Width Distortion	t_{RSKEW}		50		ns	$C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, $V_{CM} = 1.5$ V, see Figure 7 and Figure 42
Enable to Output High	t_{RZH}		10	50	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, DE high, see Figure 8 and Figure 44
Enable to Output Low	t_{RZL}		10	50	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, DE high, see Figure 8 and Figure 44
Disable Time from Low	t_{RLZ}		10	50	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, see Figure 8 and Figure 44
Disable Time from High	t_{RHZ}		10	50	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, see Figure 8 and Figure 44
Enable from Shutdown to High	$t_{RZH(SHDN)}^3$			2000	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, see Figure 8 and Figure 43
Enable from Shutdown to Low	$t_{RZL(SHDN)}^3$			2000	ns	$R_L = 1 \text{ k}\Omega$, $C_L = 15$ pF, $ V_{ID} \geq 1.5$ V, see Figure 8 and Figure 43
TIME TO SHUTDOWN	t_{SHDN}^4	40			ns	

¹ Maximum data rate assumes a ratio of $t_{DR}:t_{BIT}:t_{DF}$ equal to 1:0.5:1.

² $t_{DZH(SHDN)}$ and $t_{DZL(SHDN)}$ refer to the time for the device to enable when DE changes from 0 V to V_{CC} . $\overline{RE} = V_{CC}$ for this condition.

³ $t_{RZH(SHDN)}$ and $t_{RZL(SHDN)}$ refer to the time for the device to enable when \overline{RE} changes from V_{CC} to 0 V. DE = 0 V for this condition.

⁴ Minimum time required to put the device into shutdown: DE and \overline{RE} must be disabled for more than 40 ns for the device to go into shutdown.

ADM3065E/ADM3066E/ADM3067E/ADM3068E

V_{CC} = 3.0 V to 5.5 V, V_{IO} = 1.62 V to V_{CC} (ADM3066E/ADM3068E), T_A = T_{MIN} (-40°C) to T_{MAX} (+125°C), unless otherwise noted. All typical specifications are at T_A = 25°C, V_{IO} = V_{CC} = 3.3 V, unless otherwise noted.

Table 4.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
DRIVER						
Maximum Data Rate ¹		50			Mbps	
Propagation Delay	t_{DPLH}, t_{DPHL}		9	15	ns	$R_{LDIFF} = 54 \Omega, C_{L1} = C_{L2} = 100 \text{ pF}$, see Figure 5 and Figure 40
Skew	t_{DSKEW}		1	2	ns	$R_{LDIFF} = 54 \Omega, C_{L1} = C_{L2} = 100 \text{ pF}$, see Figure 5 and Figure 40
Rise/Fall Times	t_{DR}, t_{DF}		4	6.7	ns	$R_{LDIFF} = 54 \Omega, C_{L1} = C_{L2} = 100 \text{ pF}$, see Figure 5 and Figure 40
Enable to Output High	t_{DZH}		10	30	ns	$R_L = 110 \Omega, C_L = 50 \text{ pF}$, see Figure 6 and Figure 41
Enable to Output Low	t_{DZL}		10	30	ns	$R_L = 110 \Omega, C_L = 50 \text{ pF}$, see Figure 6 and Figure 41
Disable Time from Low	t_{DLZ}		10	30	ns	$R_L = 110 \Omega, C_L = 50 \text{ pF}$, see Figure 6 and Figure 41
Disable Time from High	t_{DHZ}		10	30	ns	$R_L = 110 \Omega, C_L = 50 \text{ pF}$, see Figure 6 and Figure 41
Enable Time from Shutdown to High	$t_{DZH(SHDN)}^2$		550	2000	ns	$R_L = 110 \Omega, C_L = 50 \text{ pF}$, see Figure 6 and Figure 41
Enable Time from Shutdown to Low	$t_{DZL(SHDN)}^2$		550	2000	ns	$R_L = 110 \Omega, C_L = 50 \text{ pF}$, see Figure 6 and Figure 41
RECEIVER						
Maximum Data Rate		50			Mbps	
Propagation Delay	t_{RPLH}, t_{RPHL}		20	35	ns	$C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}, V_{CM} = 1.5 \text{ V}$, see Figure 7 and Figure 42
Skew/Pulse Width Distortion	t_{RSKEW}		1	3	ns	$C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}, V_{CM} = 1.5 \text{ V}$, see Figure 7 and Figure 42
Enable to Output High	t_{RZH}		10	35	ns	$R_L = 1 \text{ k}\Omega, C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}, DE \text{ high}$, see Figure 8 and Figure 44
Enable to Output Low	t_{RZL}		10	35	ns	$R_L = 1 \text{ k}\Omega, C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}, DE \text{ high}$, see Figure 8 and Figure 44
Disable Time from Low	t_{RLZ}		10	35	ns	$R_L = 1 \text{ k}\Omega, C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}$, see Figure 8 and Figure 44
Disable Time from High	t_{RHZ}		10	35	ns	$R_L = 1 \text{ k}\Omega, C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}$, see Figure 8 and Figure 44
Enable from Shutdown to High	$t_{RZH(SHDN)}^3$		450	2000	ns	$R_L = 1 \text{ k}\Omega, C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}$, see Figure 8 and Figure 43
Enable from Shutdown to Low	$t_{RZL(SHDN)}^3$		450	2000	ns	$R_L = 1 \text{ k}\Omega, C_L = 15 \text{ pF}, V_{ID} \geq 1.5 \text{ V}$, see Figure 8 and Figure 43
TIME TO SHUTDOWN	t_{SHDN}^4	40			ns	

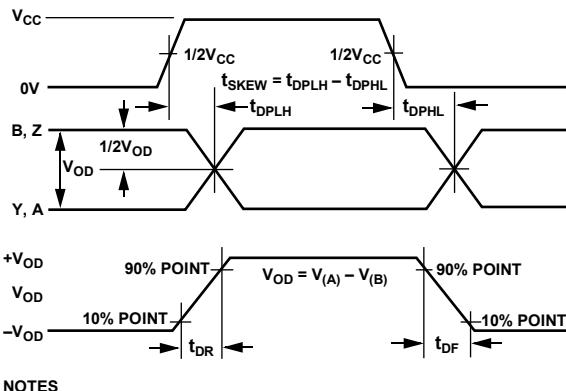
¹ Maximum data rate assumes a ratio of $t_{DR}:t_{BIR}:t_{DF}$ equal to 1:1:1.

² $t_{DZH(SHDN)}$ and $t_{DZL(SHDN)}$ refer to the time for the device to enable when DE changes from 0 V to V_{CC} . $\overline{RE} = V_{CC}$ for this condition.

³ $t_{RZH(SHDN)}$ and $t_{RZL(SHDN)}$ refer to the time for the device to enable when \overline{RE} changes from V_{CC} to 0 V. $DE = 0$ V for this condition.

⁴ Minimum time required to put the device into shutdown: DE and \overline{RE} must be disabled for more than 40 ns for the device to go into shutdown.

Timing Diagrams

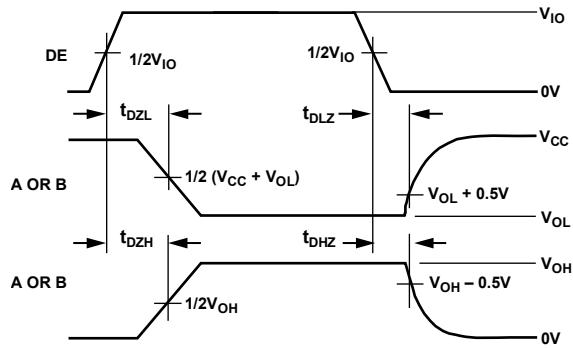


NOTES

1. V_{OD} IS THE DIFFERENCE BETWEEN A AND B, WITH $+V_{OD}$ BEING THE MAXIMUM POINT OF V_{OD} , AND $-V_{OD}$ BEING THE MINIMUM POINT OF V_{OD} .
2. $V_{CC} = V_{IO}$ FOR ADM3062E/ADM3066E/ADM3068E.

14666-003

Figure 5. Driver Propagation Delay Rise and Fall Timing Diagram

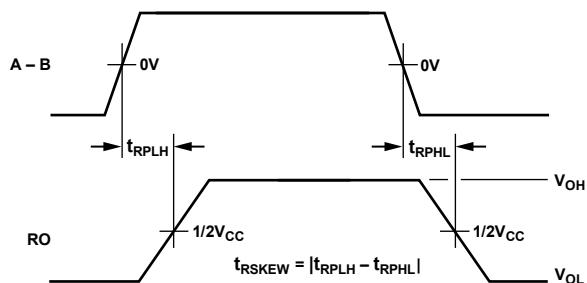


NOTES

1. $V_{IO} = V_{CC}$ FOR ADM3061E/ADM3063E/ADM3065E/ADM3067E
2. A = Y, B = Z FOR ADM3063E/ADM3067E/ADM3068E

14666-004

Figure 6. Driver Enable and Disable Timing Diagram



NOTES

1. $V_{CC} = V_{IO}$ FOR ADM3062E/ADM3066E/ADM3068E.

14666-005

Figure 7. Receiver Propagation Delay Timing Diagram

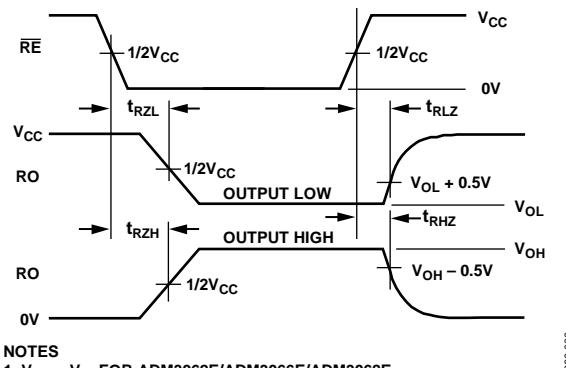


Figure 8. Receiver Enable and Disable Timing Diagram

14666-008

ABSOLUTE MAXIMUM RATINGS

Table 5.

Parameter	Rating
V _{CC} to GND	6 V
V _{IO} to GND	-0.3 V to +6 V
Digital Input and Output Voltage (DE, RE, DI, and RO)	-0.3 V to V _{IO} ¹ + 0.3 V
Driver Output and Receiver Input Voltage	-9 V to +14 V
Operating Temperature Ranges	-40°C to +85°C -40°C to +125°C
Storage Temperature Range	-65°C to +150°C
Continuous Total Power Dissipation	
8-Lead SOIC_N	0.225 W
8-Lead MSOP	0.151 W
10-Lead MSOP	0.151 W
10-Lead LFCSP	0.450 W
14-Lead SOIC_N	0.239 W
Maximum Junction Temperature (T _J)	150°C
Lead Temperature	
Soldering (10 sec)	300°C
Vapor Phase (60 sec)	215°C
Infrared (15 sec)	220°C
ESD on the Bus Pins (A, B, Y, Z)	
IEC 61000-4-2 Contact Discharge	±12 kV
IEC 61000-4-2 Air Discharge	
10 Positive and 10 Negative Discharges	±12 kV
Three Positive or Three Negative Discharges	±15 kV
ESD Human Body Model (HBM)	
On the Bus Pins (A, B, Y, Z)	≥±30 kV
All Other Pins	±8 kV

¹ V_{IO} = V_{CC} on the ADM3061E/ADM3063E/ADM3065E/ADM3067E.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

θ_{JA} is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure.

θ_{JC} is the junction to case thermal resistance.

Table 6. Thermal Resistance

Package Type	θ _{JA} ¹	θ _{JC} ¹	Unit
R-8	110.88	58.63	°C/W
RM-8	165.69	49.61	°C/W
RM-10	165.69	49.61	°C/W
R-14	104.5	42.90	°C/W
CP-10-9	55.65	33.22	°C/W

¹ Thermal impedance simulated values are based on JEDEC 2S2P thermal test board with no bias. See JEDEC JESD-51.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

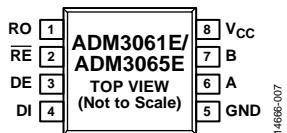


Figure 9. ADM3061E/ADM3065E 8-Lead Narrow Body SOIC_N Pin Configuration

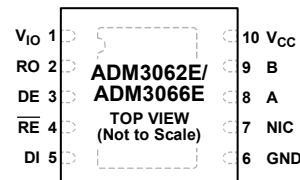


Figure 10. ADM3061E/ADM3065E 8-Lead MSOP Pin Configuration

Table 7. ADM3061E/ADM3065E Pin Function Descriptions

Pin No.	Mnemonic	Description
1	RO	Receiver Output Data. This output is high when $(A - B) \geq -30$ mV and low when $(A - B) \leq -200$ mV. This output is tristated when the receiver is disabled; that is, when RE is driven high.
2	RE	Receiver Enable Input. This is an active low input. Driving this input low enables the receiver and driving it high disables the receiver.
3	DE	Driver Enable. A high level on this pin enables the driver differential outputs, A and B. A low level places the driver output into a high impedance state.
4	DI	Transmit Data Input. Data to be transmitted by the driver is applied to this input.
5	GND	Ground.
6	A	Noninverting Driver Output and Receiver Input. When the driver is disabled, or when V_{CC} is powered down, Pin A is put into a high impedance state to avoid overloading the bus.
7	B	Inverting Driver Output and Receiver Input. When the driver is disabled, or when V_{CC} is powered down, Pin B is put into a high impedance state to avoid overloading the bus.
8	V_{CC}	3.0 V to 5.5 V Power Supply. Adding a 0.1 μ F decoupling capacitor between the V_{CC} pin and the GND pin is recommended.

Data Sheet ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E

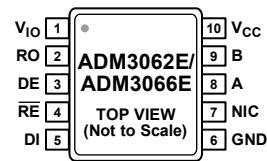


NOTES

1. NIC = NO INTERNAL CONNECTION. THIS PIN IS NOT INTERNALLY CONNECTED.
2. EXPOSED PAD. THE EXPOSED PAD MUST BE CONNECTED TO GROUND.

14686-009

Figure 11. ADM3062E/ADM3066E 10-Lead LFCSP Pin Configuration



1. NIC = NO INTERNAL CONNECTION. THIS PIN IS NOT INTERNALLY CONNECTED.

14686-010

Figure 12. ADM3062E/ADM3066E 10-Lead MSOP Pin Configuration

Table 8. ADM3066E/ADM3062E Pin Function Descriptions

Pin No.	Mnemonic	Description
1	V _{IO}	1.62 V to 5.5 V Logic Supply. Adding a 0.1 μ F decoupling capacitor between the V _{IO} pin and the GND pin is recommended.
2	RO	Receiver Output Data. This output is high when (A – B) \geq –30 mV and is low when (A – B) \leq –200 mV. This output is tristated when the receiver is disabled; that is, when RE is driven high.
3	DE	Driver Enable. A high level on this pin enables the driver differential outputs, A and B. A low level places the driver output into a high impedance state.
4	RE	Receiver Enable Input. This is an active low input. Driving this input low enables the receiver, and driving it high disables the receiver.
5	DI	Transmit Data Input. Data to be transmitted by the driver is applied to this input.
6	GND	Ground.
7	NIC	No Internal Connection. This pin is not internally connected.
8	A	Noninverting Driver Output and Receiver Input. When the driver is disabled, or when V _{CC} is powered down, Pin A is put into a high impedance state to avoid overloading the bus.
9	B	Inverting Driver Output and Receiver Input. When the driver is disabled, or when V _{CC} is powered down, Pin B is put into a high impedance state to avoid overloading the bus.
10	V _{CC}	3.0V to 5.5 V Power Supply. Adding a 0.1 μ F decoupling capacitor between the V _{CC} pin and the GND pin is recommended.
	EPAD	Exposed Pad. The exposed pad must be connected to ground.

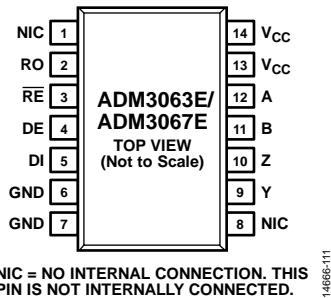


Figure 13. ADM3063E/ADM3067E 14-Lead SOIC Pin Configuration

Table 9. ADM3063E/ADM3067E Pin Function Descriptions

Pin No.	Mnemonic	Description
1, 8	NIC	No Internal Connection. This pin is not internally connected.
2	RO	Receiver Output Data. This output is high when $(A - B) \geq -30$ mV and is low when $(A - B) \leq -200$ mV. This output is tristated when the receiver is disabled; that is, when RE is driven high.
3	RE	Receiver Enable Input. This is an active low input. Driving this input low enables the receiver and driving it high disables the receiver.
4	DE	Driver Enable. A high level on this pin enables the driver differential outputs, Y and Z. A low level places the driver output into a high impedance state.
5	DI	Transmit Data Input. Data to be transmitted by the driver is applied to this input.
6, 7	GND	Ground.
9	Y	Driver Noninverting Output. When the driver is disabled, or when V _{CC} is powered down, Pin Y is put into a high impedance state to avoid overloading the bus.
10	Z	Driver Inverting Output. When the driver is disabled, or when V _{CC} is powered down, Pin Z is put into a high impedance state to avoid overloading the bus.
11	B	Inverting Receiver Input.
12	A	Noninverting Receiver Input.
13, 14	V _{CC}	3.0 V to 5.5 V Power Supply. Adding a 0.1 μ F decoupling capacitor between the V _{CC} pin and the GND pin is recommended.

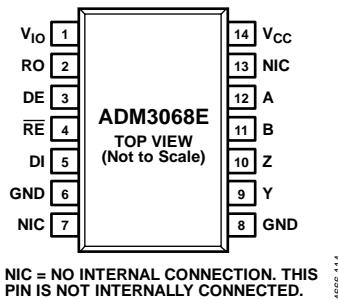
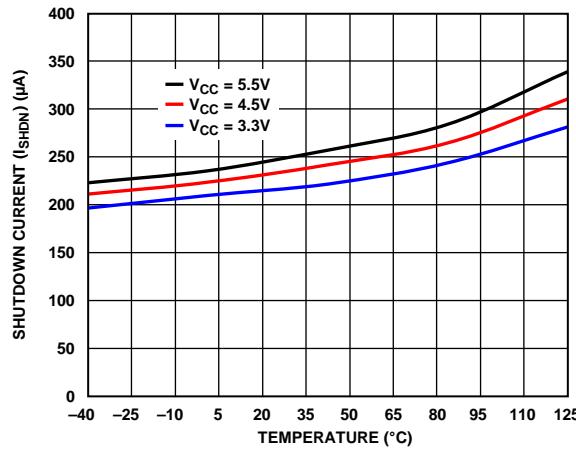


Figure 14. ADM3068E 14-Lead SOIC Pin Configuration

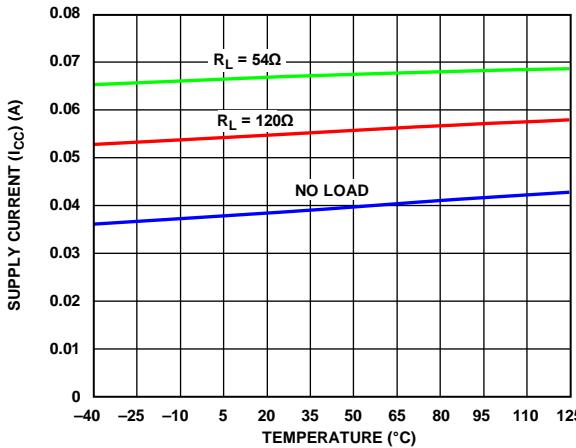
Table 10. ADM3068E Pin Function Descriptions

Pin No.	Mnemonic	Description
1	V _{IO}	1.62 V to 5.5 V Logic Supply. Adding a 0.1 μ F decoupling capacitor between the V _{IO} pin and the GND pin is recommended.
2	RO	Receiver Output Data. This output is high when (A – B) \geq –30 mV and is low when (A – B) \leq –200 mV. This output is tristated when the receiver is disabled; that is, when RE is driven high.
3	DE	Driver Enable. A high level on this pin enables the driver differential outputs, Y and Z. A low level places the driver output into a high impedance state.
4	RE	Receiver Enable Input. This is an active low input. Driving this input low enables the receiver and driving it high disables the receiver.
5	DI	Transmit Data Input. Data to be transmitted by the driver is applied to this input.
6, 8	GND	Ground.
7, 13	NIC	No Internal Connection. This pin is not internally connected.
9	Y	Driver Noninverting Output. When the driver is disabled, or when V _{CC} is powered down, Pin Y is put into a high impedance state to avoid overloading the bus.
10	Z	Driver Inverting Output. When the driver is disabled, or when V _{CC} is powered down, Pin Z is put into a high impedance state to avoid overloading the bus.
11	B	Inverting Receiver Input.
12	A	Noninverting Receiver Input.
14	V _{CC}	3.0 V to 5.5 V Power Supply. Adding a 0.1 μ F decoupling capacitor between the V _{CC} pin and the GND pin is recommended.

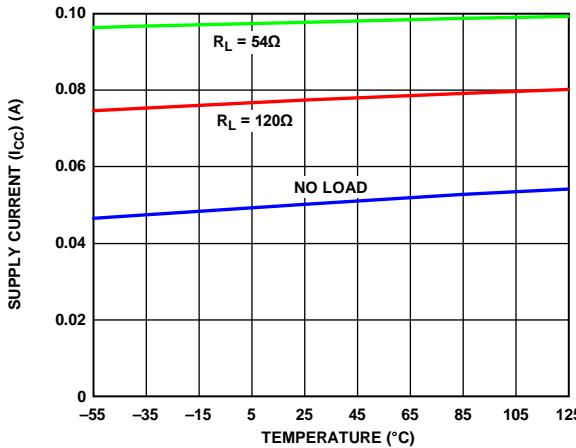
TYPICAL PERFORMANCE CHARACTERISTICS

Figure 15. Shutdown Current (I_{SHDN}) vs. Temperature

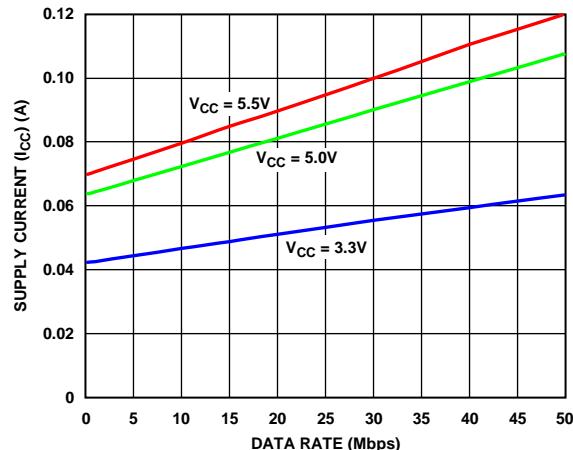
14666-019

Figure 16. Supply Current (I_{cc}) vs. Temperature, Data Rate = 50 Mbps, 50 Mbps Models, $V_{CC} = 3.3V$

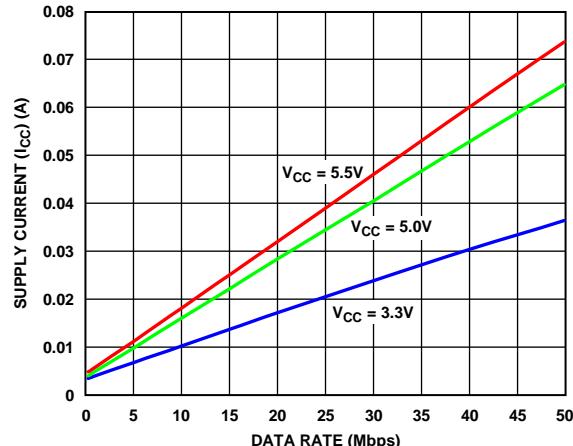
14666-019

Figure 17. Supply Current (I_{cc}) vs. Temperature, Data Rate = 50 Mbps, 50 Mbps Models, $V_{CC} = 5.0V$

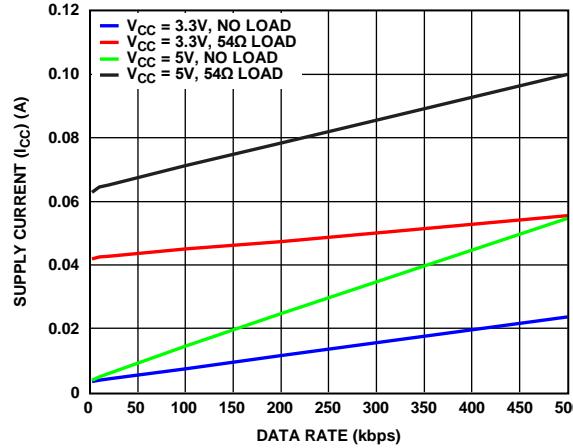
14666-020

Figure 18. Supply Current (I_{cc}) vs. Data Rate with 54 Ω Load Resistance, 50 Mbps Models

14666-021

Figure 19. Supply Current (I_{cc}) vs. Data Rate with No Load Resistance, 50 Mbps Models

14666-022

Figure 20. Supply Current (I_{cc}) vs. Data Rate with 54 Ω Load Resistance and No Load Resistance, 500 kbps Models

14666-023

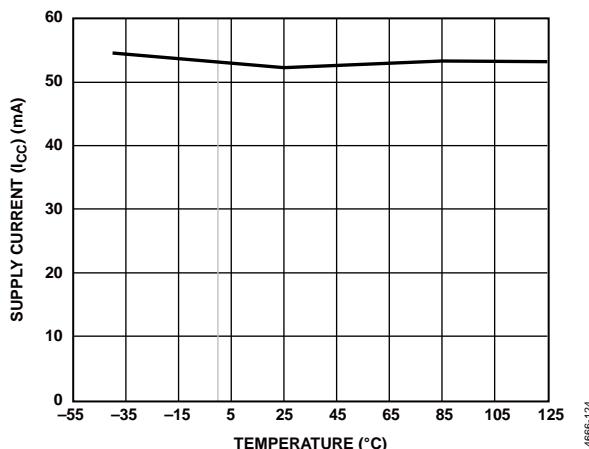


Figure 21. Supply Current (I_{CC}) vs. Temperature, Data Rate = 500 kbps, 500 kbps Models, $V_{CC} = 3.0\text{ V}$

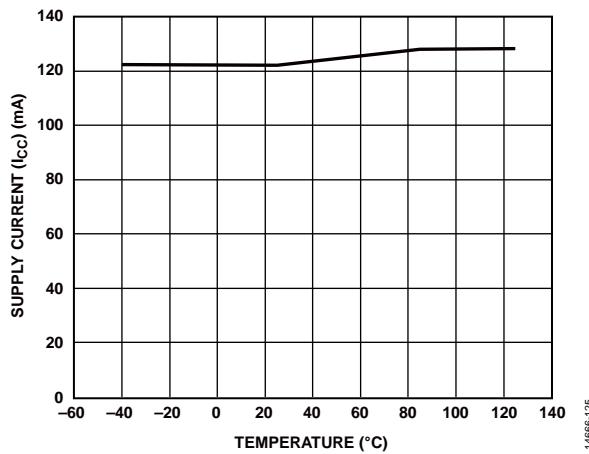


Figure 22. Supply Current (I_{CC}) vs. Temperature, Data Rate = 500 kbps, 500 kbps Models, $V_{CC} = 5.5\text{ V}$

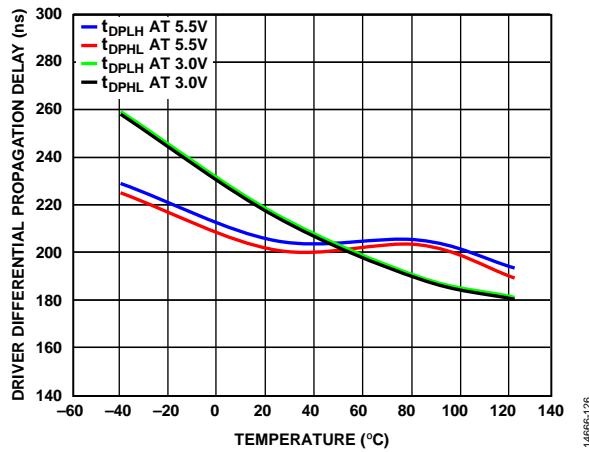


Figure 23. Driver Differential Propagation Delay vs. Temperature, 500 kbps Models

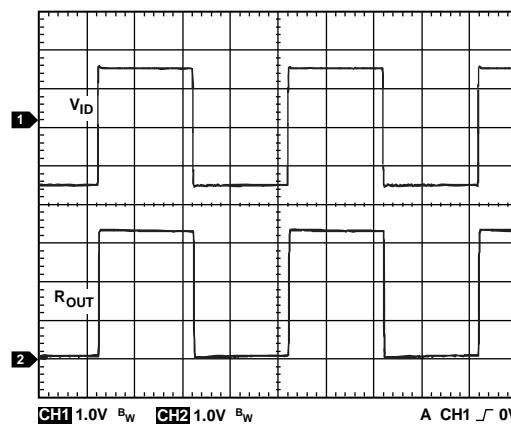


Figure 24. Receiver Propagation Delay (Oscilloscope Plot), Data Rate = 500 kbps, $V_{ID} \geq 1.5\text{ V}$

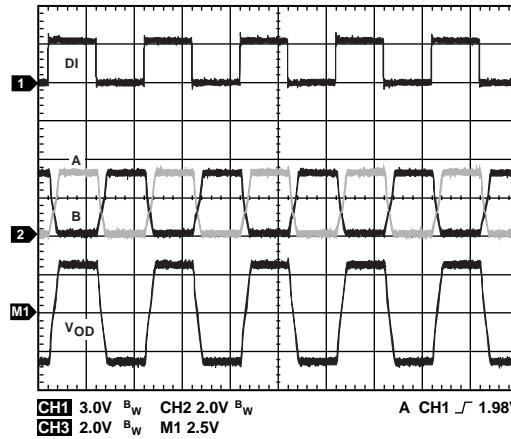


Figure 25. Driver Propagation Delay (Oscilloscope Plot), Data Rate = 500 kbps, 500 kbps Models

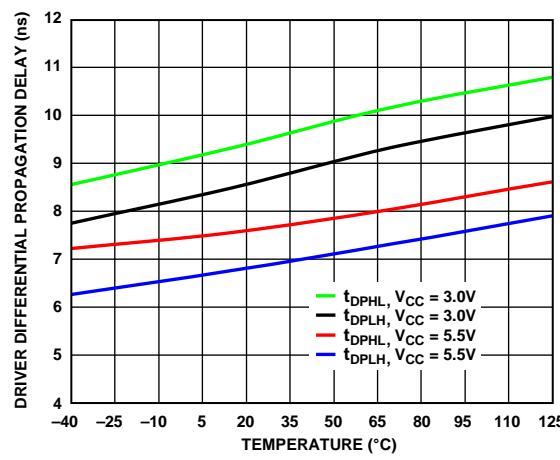
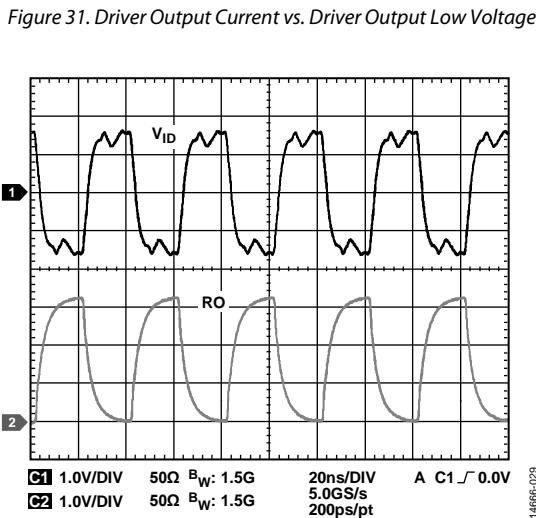
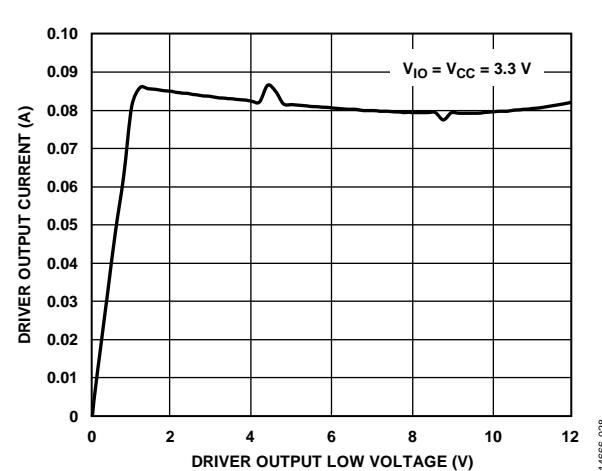
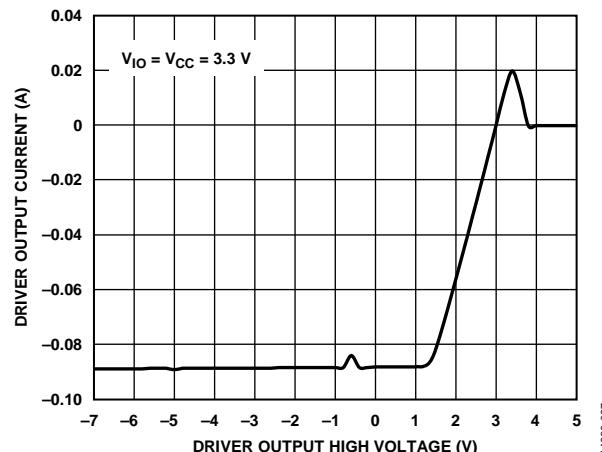
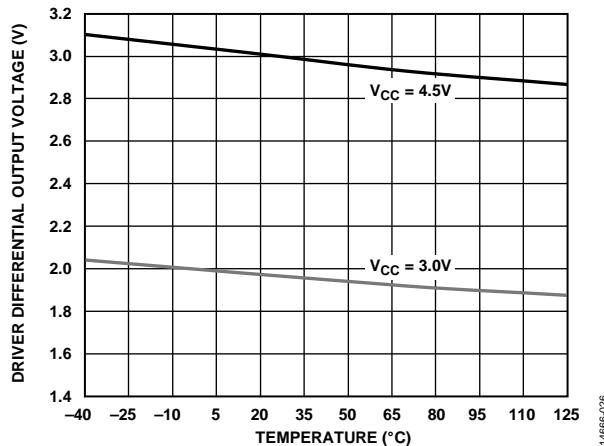
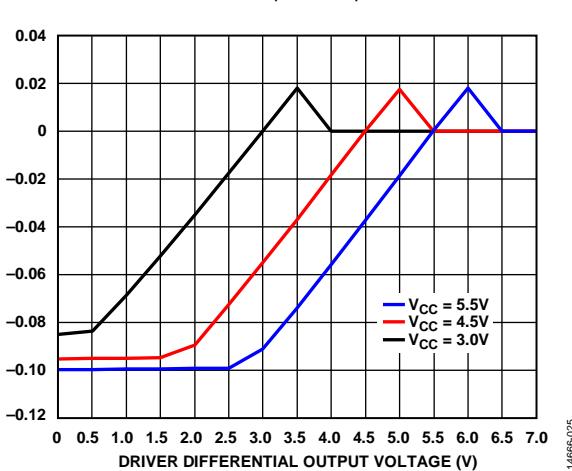
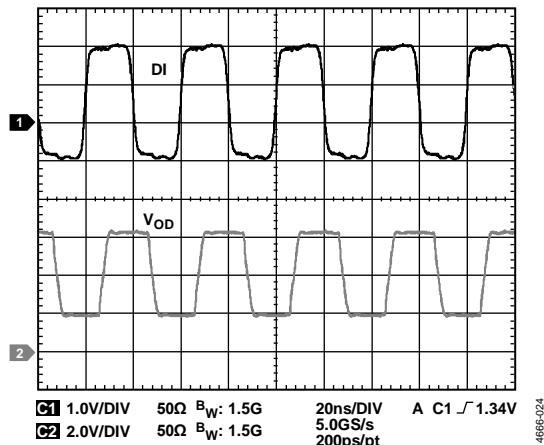
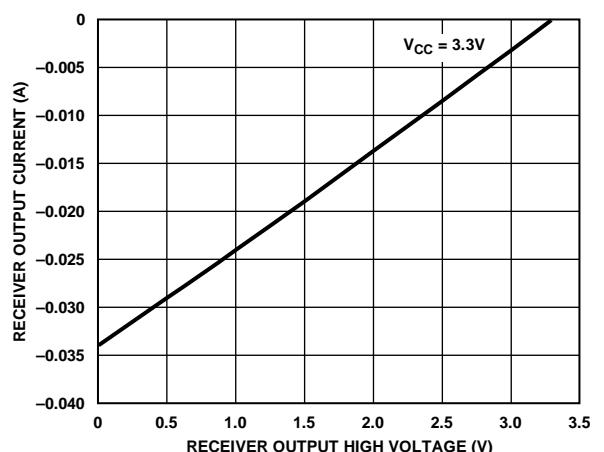
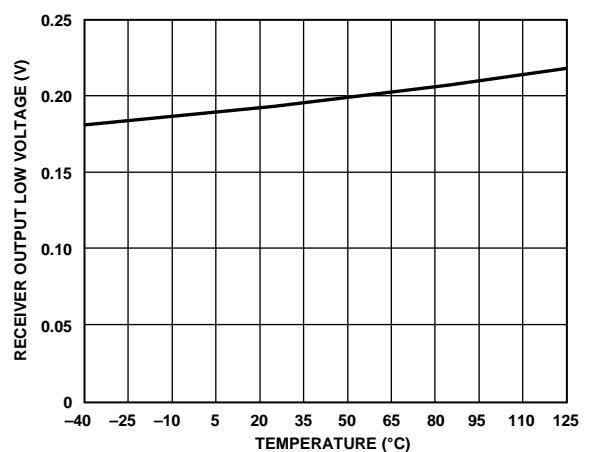
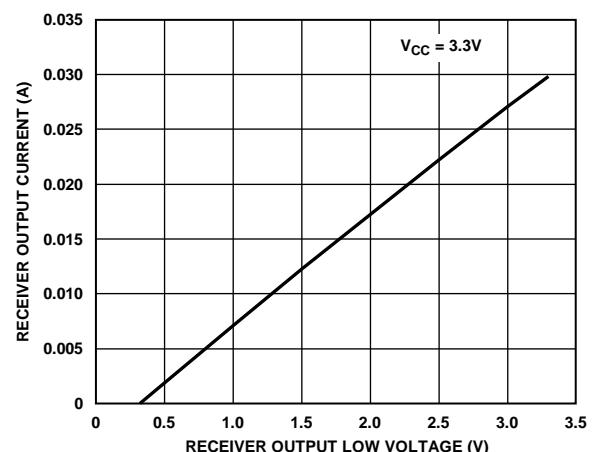
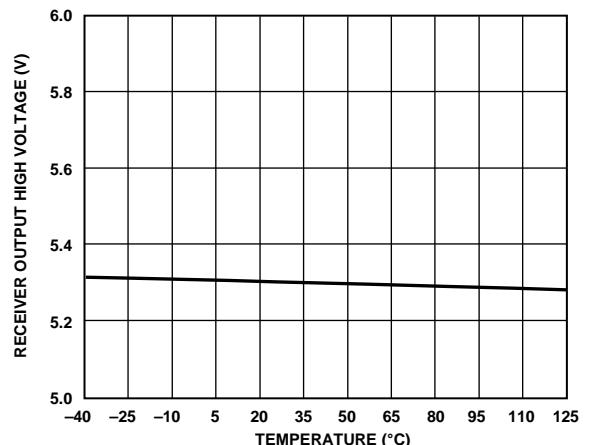
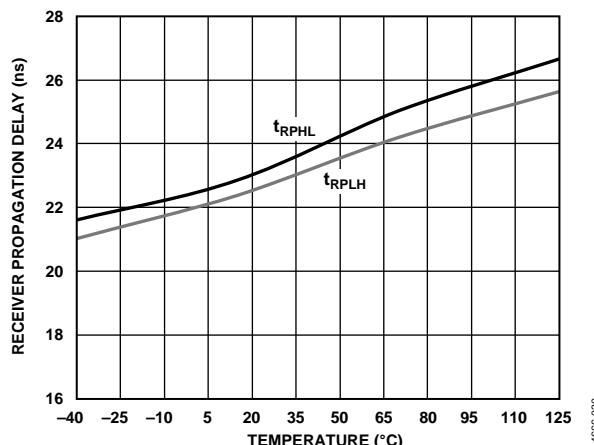


Figure 26. Driver Differential Propagation Delay vs. Temperature, 50 Mbps Models





TEST CIRCUITS

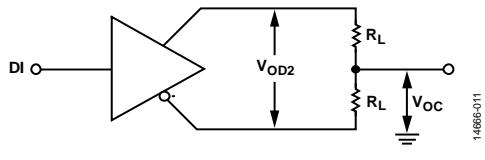


Figure 38. Driver Voltage Measurements

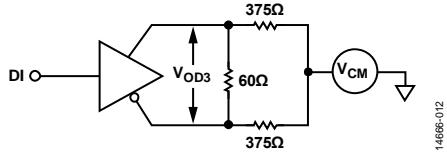


Figure 39. Driver Voltage Measurements over Common-Mode Range

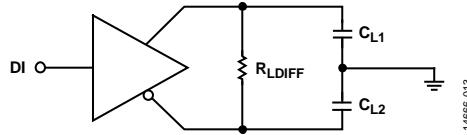
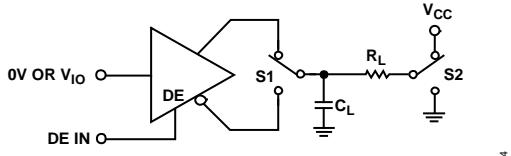


Figure 40. Driver Propagation Delay



NOTES
1. $V_{IO} = V_{CC}$ FOR ADM3061E/ADM3063E/ADM3065E/ADM3067E.

Figure 41. Driver Enable/Disable

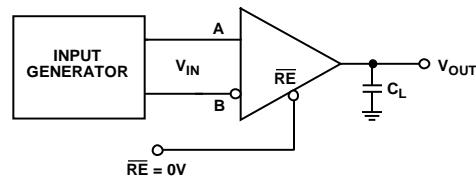


Figure 42. Receiver Propagation Delay/Skew

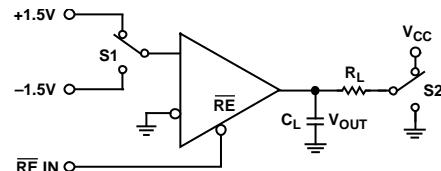
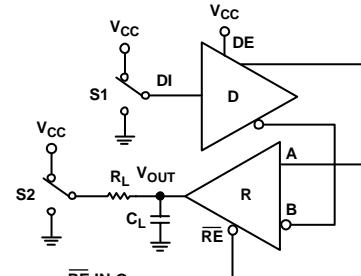


Figure 43. Receiver Enable/Disable from Shutdown



NOTES
1. $V_{CC} = V_{IO}$ FOR ADM3062E/ADM3066E/ADM3068E.

Figure 44. Receiver Enable/Disable

THEORY OF OPERATION

IEC ESD PROTECTED RS-485

The ADM3065E/ADM3066E/ADM3067E/ADM3068E are 3.0 V to 5.5 V, 50 Mbps RS-485 transceivers with IEC 61000-4-2 Level 4 ESD protection on the bus pins. The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E can withstand up to ± 12 kV contact discharge on transceiver bus pins (A, B, Y, and Z) without latch-up or damage. The ADM3061E/ADM3062E/ADM3063E has the same robust IEC 61000-4-2 ESD protection as the ADM3065E/ADM3066E/ADM3067E/ADM3068E models and operate at a lower, 500 kbps data rate.

HIGH DRIVER DIFFERENTIAL OUTPUT VOLTAGE

The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E have characteristics that are optimized for use in PROFIBUS applications. When powered at $V_{CC} \geq 4.5$ V, the ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E driver output differential voltage meets or exceeds the PROFIBUS requirements of 2.1 V with a 54 Ω load.

IEC 61000-4-2 ESD PROTECTION

ESD is the sudden transfer of electrostatic charge between bodies at different potentials caused either by near contact or induced by an electric field. It has the characteristics of high current in a short time period. The primary purpose of the IEC 61000-4-2 test is to determine the immunity of systems to external ESD events outside the system during operation. IEC 61000-4-2 describes testing using two coupling methods: contact discharge and air discharge. Contact discharge implies a direct contact between the discharge gun and the equipment under test (EUT). During air discharge testing, the charged electrode of the discharge gun is moved toward the EUT until a discharge occurs as an arc across the air gap. The discharge gun does not make direct contact with the EUT. A number of factors affect the results and repeatability of the air discharge test, including humidity, temperature, barometric pressure, distance, and rate of approach to the EUT. This method is a more accurate representation of an actual ESD event but is not as repeatable. Therefore, contact discharge is the preferred test method.

During testing, the data port is subjected to at least 10 positive and 10 negative single discharges. Selection of the test voltage is dependent on the system end environment.

Figure 45 shows the 8 kV contact discharge current waveform as described in the IEC 61000-4-2 specification. Some of the key waveform parameters are rise times of less than 1 ns and pulse widths of approximately 60 ns.

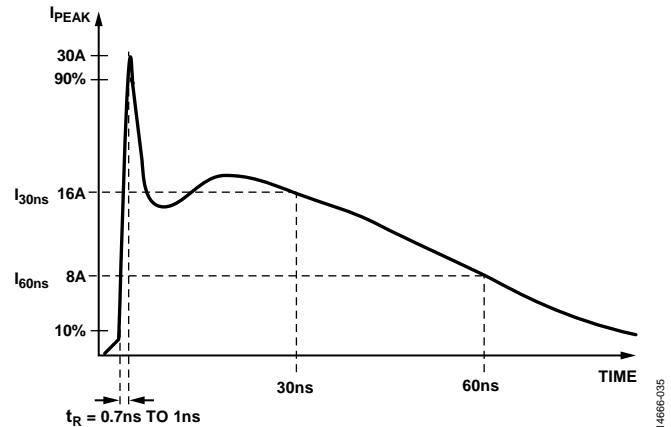


Figure 45. IEC 61000-4-2 ESD Waveform (8 kV)

14686-035

Figure 46 shows the 8 kV contact discharge current waveform from the IEC 61000-4-2 standard compared to the HBM ESD 8 kV waveform. Figure 46 shows that the two standards specify a different waveform shape and peak current. The peak current associated with an IEC 61000-4-2 8 kV pulse is 30 A, whereas the corresponding peak current for HBM ESD is more than five times less, at 5.33 A. The other difference is the rise time of the initial voltage spike, with the IEC 61000-4-2 ESD waveform having a much faster rise time of 1 ns, compared to the 10 ns associated with the HBM ESD waveform. The amount of power associated with an IEC ESD waveform is much greater than that of an HBM ESD waveform. The HBM ESD standard requires the EUT to be subjected to three positive and three negative discharges, whereas the IEC ESD standard requires 10 positive and 10 negative discharge tests.

The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E with IEC 61000-4-2 ESD ratings is better suited for operation in harsh environments compared to other RS-485 transceivers that state varying levels of HBM ESD protection.

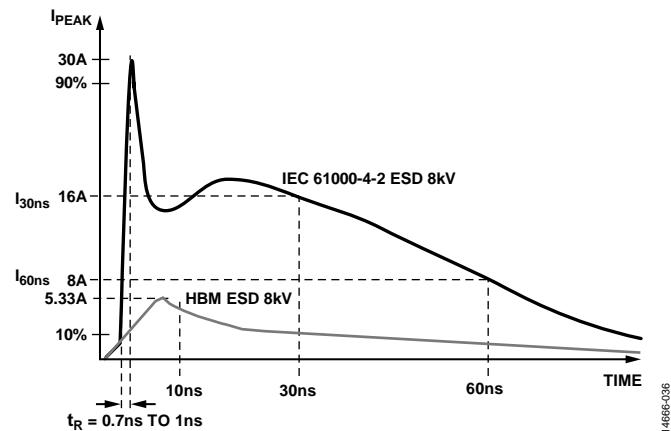


Figure 46. IEC 61000-4-2 ESD Waveform 8 kV Compared to HBM ESD Waveform 8 kV

14686-036

TRUTH TABLES

Table 12 and Table 13 use the abbreviations shown in Table 11.

Table 11. Truth Table Abbreviations

Letter	Description
H	High level
I	Indeterminate
L	Low level
X	Any state
Z	High impedance (off)

Table 12. Transmitting Truth Table

Supply Status		Inputs			Outputs	
V _{IO} ¹	V _{CC}	RE	DE	DI	A/Y	B/Z
On	On	X	H	H	H	L
On	On	X	H	L	L	H
On	On	X	L	X	Z	Z
Off	On	X	X	X	I	I
On	Off	X	X	X	Z	Z
Off	Off	X	X	X	Z	Z

¹ For the ADM3061E, ADM3063E, ADM3065E, and ADM3067E, the V_{IO} pin is not applicable.

Table 13. Receiving Truth Table

Supply Status		Inputs		Outputs	
V _{IO} ¹	V _{CC}	A – B	RE	DE	RO
On	On	>–0.03 V	L	X	H
On	On	<–0.2 V	L	X	L
On	On	–0.2 V ≤ A – B ≤ –0.03 V	L	X	I
On	On	Inputs open/shorted	L	X	H
On	On	X	H	X	Z
On	Off	X	L	X	I
On	Off	X	H	X	Z
Off	On	X	L	X	I
Off	Off	X	X	X	I

¹ For the ADM3061E, ADM3063E, ADM3065E, and ADM3067E, the V_{IO} pin is not applicable.

RECEIVER FAIL-SAFE

The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E guarantee a logic high receiver output when the receiver inputs are shorted, open, or connected to a terminated transmission line with all drivers disabled. This receiver output is achieved by setting the receiver input threshold between –30 mV and –200 mV. If the differential receiver input voltage (A – B) is greater than or equal to –30 mV, the RO pin is logic high.

If the (A – B) input is less than or equal to –200 mV, RO is logic low. In the case of a shorted, open circuit or terminated bus with all transmitters disabled, the receiver differential input voltage is pulled to 0 V, resulting in a logic high with a 30 mV minimum noise margin.

HOT SWAP CAPABILITY

When a circuit board is inserted into a powered (or hot) backplane, differential disturbances to the data bus can lead to data errors. During this period, processor logic output drivers are high impedance and are unable to drive the DE and RE inputs of the RS-485 transceivers to a defined logic level. Leakage currents up to $\pm 10 \mu\text{A}$ from the high impedance state of the processor logic drivers can cause standard complementary metal-oxide semiconductor (CMOS) enable inputs of a transceiver to drift to an incorrect logic level. Additionally, parasitic circuit board capacitance can cause coupling of V_{CC} or GND to the enable inputs. Without the hot swap capability, these factors can improperly enable the driver or receiver of the transceiver. When V_{CC} or V_{IO} rises, an internal pull-down circuit holds DE low and RE high. After the initial power-up sequence, the pull-down circuit becomes transparent, resetting the hot swap tolerable input.

128 TRANSCEIVERS ON THE BUS

The standard RS-485 receiver input impedance is $12 \text{ k}\Omega$ (one unit load), and the standard driver can drive up to 32 unit loads. The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E transceivers have a one fourth unit load receiver input impedance ($48 \text{ k}\Omega$), allowing up to 128 transceivers to be connected in parallel on one communication line. Any combination of these devices and other RS-485 transceivers with a total of 32 unit loads or fewer can be connected to the line.

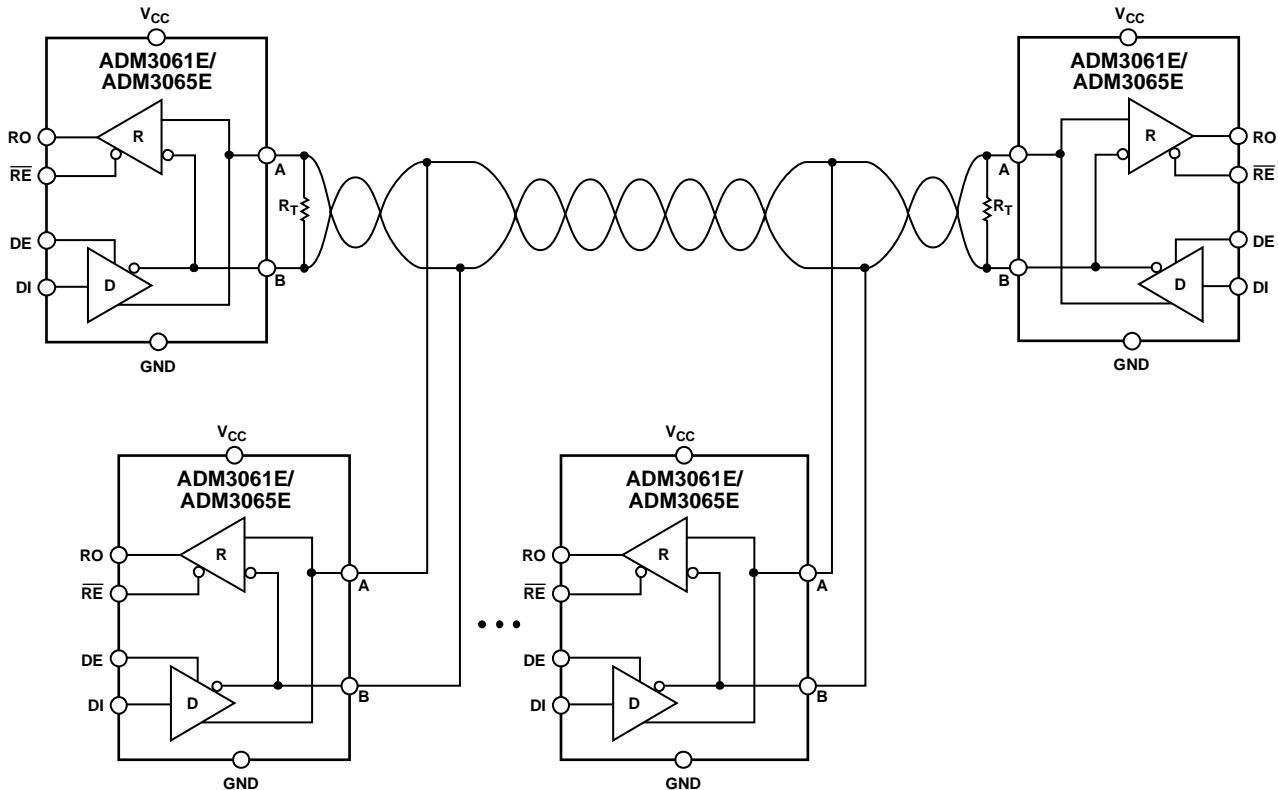
DRIVER OUTPUT PROTECTION

The ADM3061E/ADM3062E/ADM3063E/ADM3065E/ADM3066E/ADM3067E/ADM3068E feature two methods to prevent excessive output current and power dissipation caused by faults or by bus contention. Current-limit protection on the output stage provides immediate protection against short circuits over the whole common-mode voltage range. In addition, a thermal shutdown circuit forces the driver outputs into a high impedance state if the die temperature rises excessively. This circuitry is designed to disable the driver outputs when a die temperature of 150°C is reached. As the device cools, the drivers are reenabled at a temperature of 140°C.

APPLICATIONS INFORMATION

The ADM3061E/ADM3065E transceiver is designed for bidirectional data communications on multipoint bus transmission lines. Figure 47 shows a typical network applications circuit.

To minimize reflections, terminate the line at both ends with a termination resistor (the value of the termination resistor must be equal to the characteristic impedance of the cable used) and keep stub lengths off the main line as short as possible.



NOTES

1. THE MAXIMUM NUMBER OF NODES IS 128.
2. R_T IS EQUAL TO THE CHARACTERISTIC IMPEDANCE OF THE CABLE USED.

Figure 47. ADM3061E/ADM3065E Typical Half-Duplex RS-485 Communications Network

ISOLATED HIGH SPEED RS-485 NODE

Galvanic isolation, with reinforced insulation and 5 kV rms transient withstand voltage, can be added to the ADM3065E using Analog Devices, Inc., *iCoupler*[®] and *isoPower*[®] technology. The ADuM6401 provides the required quad channels of 5 kV rms signal isolation, operating at rates up to 25 Mbps, together with an integrated dc-to-dc converter. The ADuM6401 combines with the ADM3065E (shown in Figure 48) with the V_{ISO} pin configured for 3.3 V by connecting the V_{SEL} pin to GND_{ISO} and a 5 V supply connected to V_{DD1}. Operation at 3.3 V ensures the ADM3065E remains within the load capability of ADuM6401 even at 25 Mbps.

The dc-to-dc converter in the ADuM6401 *isoPower* device provides regulated, isolated power to the ADM3065E (and the ADuM241D). These *isoPower* devices use high frequency

switching elements to transfer power through the transformers. Take care during PCB layout to meet emissions standards. See the AN-0971 Application Note for PCB layout recommendations.

Galvanic isolation of the ADM3065E at the full data rate, up to 50 Mbps, can be implemented using the ADuM241D quad-channel digital isolator and the ADuM6028 isolated dc-to-dc converter, as shown in Figure 49. The ADuM6028 is an 8-pin device that contains a 300 mW dc-to-dc converter optimized to meet emissions standards on a 2-layer PCB using two ferrite beads. The ADuM241D operates at a data rate up to 150 Mbps, and offers the precise timing required to fully support the ADM3065E at 50 Mbps.

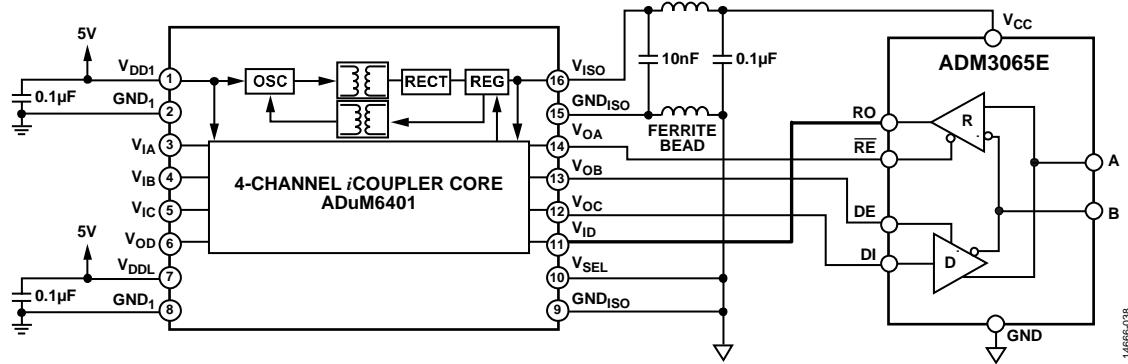


Figure 48. Signal and Power Isolated 25 Mbps RS-485 Solution (Simplified Diagram—All Connections Not Shown)

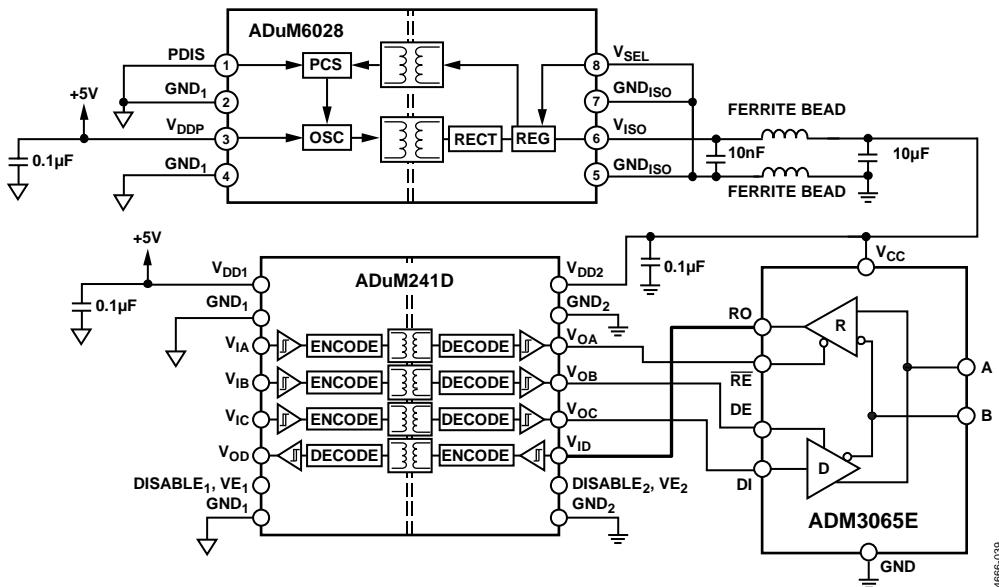
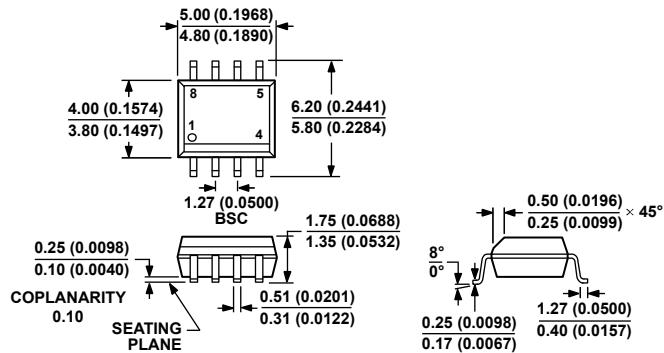


Figure 49. Signal and Power Isolated 50 Mbps RS-485 Solution (Simplified Diagram—All Connections Not Shown)

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MS-012-AA
CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS
(IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR
REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

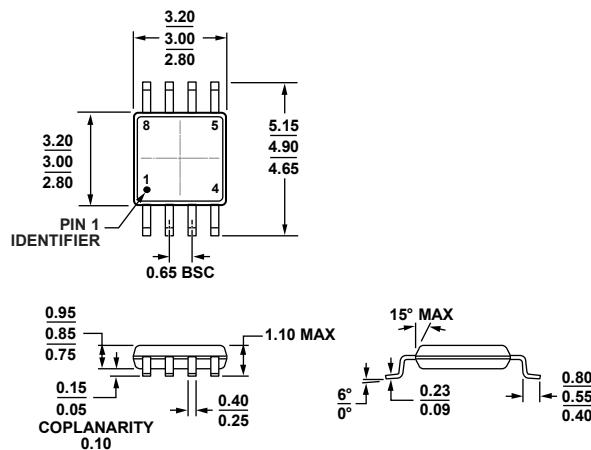
012407-A

Figure 50. 8-Lead Standard Small Outline Package [SOIC_N]

Narrow Body

(R-8)

Dimensions shown in millimeters and (inches)



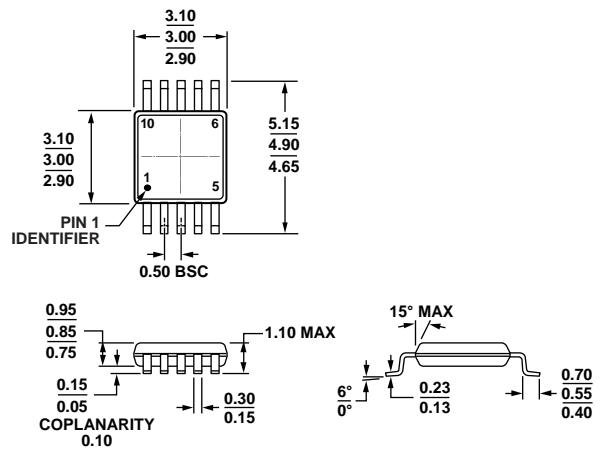
10-07-2009-B

COMPLIANT TO JEDEC STANDARDS MO-187-AA

Figure 51. 8-Lead Mini Small Outline Package [MSOP]

(RM-8)

Dimensions shown in millimeters

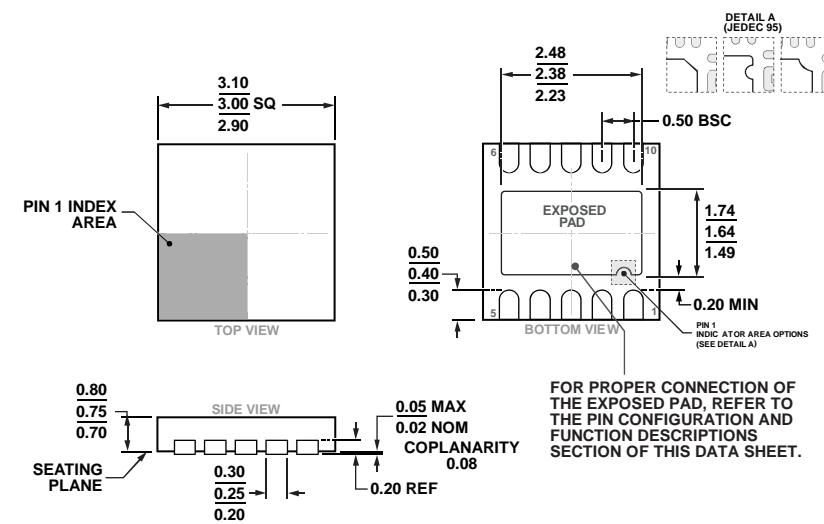


COMPLIANT TO JEDEC STANDARDS MO-187-BA

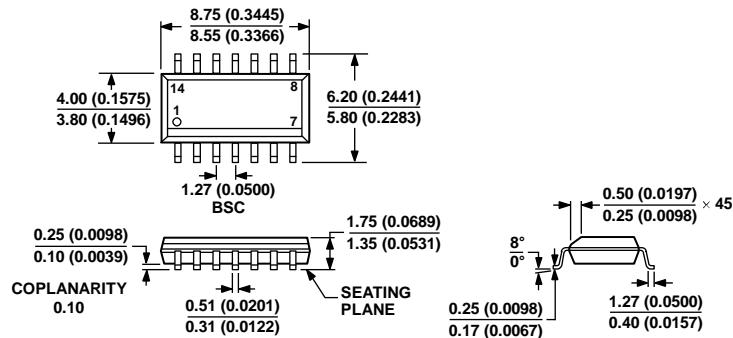
*Figure 52. 10-Lead Mini Small Outline Package [MSOP]
(RM-10)*

Dimensions shown in millimeters

091709-V



**Figure 53. 10-Lead Lead Frame Chip Scale Package [LFCSP]
 3 mm × 3 mm Body and 0.75 mm Package Height
 (CP-10-9)**
Dimensions shown in millimeters



COMPLIANT TO JEDEC STANDARDS MS-012-AB
**CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS
(IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR
REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.**

60606-A

*Figure 54. 14-Lead Standard Small Outline Package [SOIC_N]
 Narrow Body
 (R-14)
 Dimensions shown in millimeters and (inches)*

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option	Marking Code
ADM3061EARZ	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3061EARZ-R7	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3061EBRZ	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3061EBRZ-R7	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3061EARMZ	-40°C to +85°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MBY
ADM3061EARMZ-R7	-40°C to +85°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MBY
ADM3061EBRMZ	-40°C to +125°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MC0
ADM3061EBRMZ-R7	-40°C to +125°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MC0
ADM3062EACPZ	-40°C to +85°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MCC
ADM3062EACPZ-R7	-40°C to +85°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MCC
ADM3062EBCPZ	-40°C to +125°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MCD
ADM3062EBCPZ-R7	-40°C to +125°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MCD
ADM3062EARMZ	-40°C to +85°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC7
ADM3062EARMZ-R7	-40°C to +85°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC7
ADM3062EBRMZ	-40°C to +125°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC8
ADM3062EBRMZ-R7	-40°C to +125°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC8
ADM3063EARZ	-40°C to +85°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3063EARZ-R7	-40°C to +85°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3063EBRZ	-40°C to +125°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3063EBRZ-R7	-40°C to +125°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3065EARZ	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3065EARZ-R7	-40°C to +85°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3065EBRZ	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3065EBRZ-R7	-40°C to +125°C	8-Lead Standard Small Outline Package [SOIC_N]	R-8	
ADM3065EARMZ	-40°C to +85°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MC1
ADM3065EARMZ-R7	-40°C to +85°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MC1
ADM3065EBRMZ	-40°C to +125°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MC2
ADM3065EBRMZ-R7	-40°C to +125°C	8-Lead Mini Small Outline Package [MSOP]	RM-8	MC2

Model ¹	Temperature Range	Package Description	Package Option	Marking Code
ADM3066EACPZ	-40°C to +85°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MC9
ADM3066EACPZ-R7	-40°C to +85°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MC9
ADM3066EBCPZ	-40°C to +125°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MCA
ADM3066EBCPZ-R7	-40°C to +125°C	10-Lead Lead Frame Chip Scale Package [LFCSP]	CP-10-9	MCA
ADM3066EARMZ	-40°C to +85°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC4
ADM3066EARMZ-R7	-40°C to +85°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC4
ADM3066EBRMZ	-40°C to +125°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC5
ADM3066EBRMZ-R7	-40°C to +125°C	10-Lead Mini Small Outline Package [MSOP]	RM-10	MC5
ADM3067EARZ	-40°C to +85°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3067EARZ-R7	-40°C to +85°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3067EBRZ	-40°C to +125°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3067EBRZ-R7	-40°C to +125°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3068EARZ	-40°C to +85°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3068EARZ-R7	-40°C to +85°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3068EBRZ	-40°C to +125°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
ADM3068EBRZ-R7	-40°C to +125°C	14-Lead Standard Small Outline Package [SOIC_N]	R-14	
EVAL-ADM3061EEBZ		8-Lead SOIC_N Evaluation Board		
EVAL-ADM3061EEB1Z		8-Lead MSOP Evaluation Board		
EVAL-ADM3062EEBZ		10-Lead MSOP Evaluation Board		
EVAL-ADM3062EEB1Z		10-Lead LFCSP Evaluation Board		
EVAL-ADM3063EEBZ		14-Lead SOIC_N Evaluation Board		
EVAL-ADM3065EEBZ		8-Lead SOIC_N Evaluation Board		
EVAL-ADM3065EEB1Z		8-Lead MSOP Evaluation Board		
EVAL-ADM3066EEBZ		10-Lead MSOP Evaluation Board		
EVAL-ADM3066EEB1Z		10-Lead LFCSP Evaluation Board		
EVAL-ADM3067EEBZ		14-Lead SOIC_N Evaluation Board		
EVAL-ADM3068EEBZ		14-Lead SOIC_N Evaluation Board		

¹ Z = RoHS Compliant Part.

ООО "ЛайфЭлектроникс"

"LifeElectronics" LLC

ИНН 7805602321 КПП 780501001 Р/С 40702810122510004610 ФАКБ "АБСОЛЮТ БАНК" (ЗАО) в г.Санкт-Петербурге К/С 30101810900000000703 БИК 044030703

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С конца 2013 года компания активно расширяет линейку поставок компонентов по направлению коаксиальный кабель, кварцевые генераторы и конденсаторы (керамические, пленочные, электролитические), за счёт заключения дистрибуторских договоров

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- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
- Поставку компонентов, требующих военную и космическую приемку.
- Входной контроль качества.
- Наличие сертификата ISO.

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- Регистрацию проекта у производителя компонентов.
- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



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