## FEATURES

## 5 V multichannel audio system

Accepts 16-/18-/20-/24-bit data
Supports 24-bit and 96 kHz sample rate
Multibit $\Sigma-\Delta$ modulators with data directed scrambling
Differential output for optimum performance
ADCs: -92 dB THD + N, 105 dB SNR and dynamic range
DACs: -95 dB THD + N, 108 dB SNR and dynamic range
On-chip volume control with "auto-ramp" function
Programmable gain amplifier for ADC input
Hardware and software controllable clickless mute
Digital de-emphasis processing
Supports $256 \times f_{s}, 512 \times f_{s}$, or $768 \times f_{s}$ master clock
Power-down mode plus soft power-down mode
Flexible serial data port with right justified, left justified, $I^{2} S$
compatible, and DSP serial port modes
TDM interface mode supports 8 in/8 out using a single
SHARC ${ }^{\circledR}$ SPORT
52-lead MQFP (PQFP) plastic package

## APPLICATIONS

## Home theater systems

Automotive audio systems
DVD recorders
Set-top boxes
Digital audio effects processors

## PRODUCT OVERVIEW

The AD1836A is a high performance, single-chip codec that provides three stereo DACs and two stereo ADCs using ADI's patented multibit $\Sigma-\Delta$ architecture. An $\mathrm{SPI}^{\oplus}$ port is included, allowing a microcontroller to adjust volume and many other parameters. The AD1836A operates from a 5 V supply, with provision for a separate output supply to interface with low voltage external circuitry. The AD1836A is available in a 52 -lead MQFP (PQFP) package.


Figure 1.

Rev. 0
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## AD1836A

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## REVISION HISTORY

Revision 0: Initial Version

## AD1836A—SPECIFICATIONS

Table 1. Test conditions, unless otherwise noted. Performance of all channels is identical (exclusive of the Interchannel Gain Mismatch and Interchannel Phase Deviation specifications).

| Parameter | Rating |
| :--- | :--- |
| Supply Voltages (AVDD, DVDD) | 5 V |
| Ambient Temperature | $25^{\circ} \mathrm{C}$ |
| Master Clock | $12.288 \mathrm{MHz}\left(48 \mathrm{kHz} \mathrm{fs}, 256 \times \mathrm{f}_{\mathrm{s}}\right.$ Mode) |
| Input Signal | $1.000 \mathrm{kHz}, 0 \mathrm{dBFS}$ (Full Scale) |
| Input Sample Rate | 48 kHz |
| Measurement Bandwidth | 20 Hz to 20 kHz |
| Word Width | 24 Bits |
| Load Capacitance (Digital Output) | 100 pF |
| Load Impedance (Digital Output) | 2.5 kJ |
| Input Voltage HI | 2.4 V |
| Input Voltage LO | 0.8 V |

## AD1836A

Table 2. Analog Performance

| Parameter |  | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ANALOG-TO-DIGITAL CONVERTERS | ADC Resolution (all ADCs) |  | 24 |  | Bits |
|  | Dynamic Range ( 20 Hz to $20 \mathrm{kHz},-60 \mathrm{~dB}$ Input) ${ }^{1,2}$ |  |  |  |  |
|  | No Filter (RMS), AD1836AAS | 97 | 102 |  | dB |
|  | With A-Weighted Filter (RMS), AD1836AAS | 100 | 105 |  | dB |
|  | No Filter (RMS), AD1836ACS | 94 | 99 |  | dB |
|  | With A-Weighted Filter (RMS), AD1836ACS | 97 | 102 |  | dB |
|  | Total Harmonic Distortion + Noise (-1 dBFS) ${ }^{1}$ |  | -92 | -83 | dB |
|  | Full-Scale Input Voltage (Differential) |  | 2.18 (6.16) |  | V rms (V pp) |
|  | Gain Error | -5.0 |  | +5.0 | \% |
|  | Interchannel Gain Mismatch | -0.1 |  | +0.1 | dB |
|  | Offset Error | -10 | 0 | +10 | mV |
|  | Gain Drift |  | 100 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
|  | Interchannel Isolation |  | -110 |  | dB |
|  | Programmable Input Gain |  | 12 |  | dB |
|  | Gain Step Size |  | 3 |  | dB |
|  | CMRR, Direct Input, 100 mV RMS , 1 kHz |  | -77 | -63 | dB |
|  | CMRR, Direct Input, 100 mV RMS, 20 kHz |  | -72 | -60 | dB |
|  | CMRR, PGA Differential Input, 100 mV RMS, 1 kHz |  | -57 | -39 | dB |
|  | CMRR, PGA Differential Input, $100 \mathrm{mV} \mathrm{RMS}, 20 \mathrm{kHz}$ |  | -57 | -39 | dB |
|  | Input Resistance | 10 |  |  | k $\Omega$ |
|  | Input Capacitance |  |  | 15 | pF |
|  | Common-Mode Input Volts |  | 2.25 |  | V |
| DIGITAL-TO-ANALOG CONVERTERS | Dynamic Range ( 20 Hz to $20 \mathrm{kHz},-60 \mathrm{~dB}$ Input) ${ }^{1,2}$ |  |  |  |  |
|  | No Filter (RMS), AD1836AAS | 102 | 105 |  | dB |
|  | With A-Weighted Filter (RMS), AD1836AAS | 105 | 108 |  | dB |
|  | No Filter (RMS), AD1836ACS | 99 | 102 |  | dB |
|  | With A-Weighted Filter (RMS), AD1836ACS | 102 | 105 |  | dB |
|  | Total Harmonic Distortion + Noise (0 dBFS) ${ }^{1}$ |  | -95 | -85 | dB |
|  | Full-Scale Output Voltage (Differential) |  | 2.0 (5.6) |  | V rms (V pp) |
|  | Gain Error | -6.0 |  | +6.0 | \% |
|  | Interchannel Gain Mismatch | -0.3 |  | +0.3 | dB |
|  | Offset Error | 15 | 55 | 95 | mV |
|  | Gain Drift |  | 150 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
|  | Interchannel Isolation |  | -110 |  | dB |
|  | Interchannel Phase Deviation |  | $\pm 0.1$ |  | Degrees |
|  | Volume Control Step Size (1023 Linear Steps) |  | 0.098 |  | \% |
|  | Volume Control Range (Max Attenuation) |  | 60 |  | dB |
|  | Max Attenuation |  | -100 |  | dB |
|  | De-emphasis Gain Error |  |  | $\pm 0.1$ | dB |
|  | Output Resistance at Each Pin |  | 115 |  | $\Omega$ |
|  | $V_{\text {REF }}$ (FILTR), Common-Mode Output | 2.2 | 2.25 | 2.3 | V |

[^0]${ }^{2}$ Measured with Audio Precision System Two Cascade in rms mode. Averaging mode will show approximately 2 dB better performance.

Table 3. Digital I/O

| Parameter | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Input Voltage $\mathrm{HI}\left(\mathrm{V}_{\mathbf{H}}\right)$ | 2.2 |  |  | V |
| Input Voltage LO ( $\mathrm{V}_{\mathrm{IL}}$ ) |  |  | 0.8 | $\checkmark$ |
| Input Leakage ( $\mathrm{I}_{\mathbb{H}} @ \mathrm{~V}_{\mathbb{H}}=2.4 \mathrm{~V}$ ) |  |  | 10 | $\mu \mathrm{A}$ |
| Input Leakage ( $\mathrm{IL}^{\text {@ }}$ V $\mathrm{IL}=0.8 \mathrm{~V}$ ) |  |  | 10 | $\mu \mathrm{A}$ |
| High Level Output Voltage (VОН) $\mathrm{l}_{\mathrm{OH}}=2 \mathrm{~mA}$ | ODVDD - 0.4 |  |  | V |
| Low Level Output Voltage ( $\mathrm{V}_{\circ}$ ) $\mathrm{loL}=2 \mathrm{~mA}$ |  |  | 0.5 | V |
| Input Capacitance |  |  | 20 | pF |

Table 4. Power Supplies

| Parameter | Voltage, DVDD and AVDD | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Supplies | Voltage, ODVDD | 4.75 | 5 | 5.25 | V |
|  | Analog Current | 3.0 | $3.3 / 5$ | 5.25 | V |
|  | Analog Current—Power-Down |  | 108 | mA |  |
|  | Digital Current | 47 | mA |  |  |
|  | Digital Current—Power-Down | 78 | mA |  |  |
| Dissipation | Operation—Both Supplies | 1.5 | mA |  |  |
|  | Operation—Analog Supplies | 930 | mW |  |  |
|  | Operation—Digital Supplies | 540 | mW |  |  |
|  | Power-Down—Both Supplies | 390 | mW |  |  |

Table 5. Temperature Range

| Parameter | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- |
| Specifications Guaranteed |  | 25 |  | ${ }^{\circ} \mathrm{C}$ |
| Functionality Guaranteed | -40 |  | +85 | ${ }^{\circ} \mathrm{C}$ Ambient |
| Storage | -40 |  | +110 | ${ }^{\circ} \mathrm{C}$ Case |

Table 6. Digital Filter @ 44.1 kHz

| Parameter |  | Min | Typ | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ADC DECIMATION FILTER | Pass Band |  | 20 |  | kHz |
|  | Pass-Band Ripple |  | $\pm 0.0001$ |  | dB |
|  | Transition Band |  | 22 |  | kHz |
|  | Stop Band |  | 24 |  | kHz |
|  | Stop-Band Attenuation |  | 120 |  | dB |
|  | Group Delay |  | 990.20 |  | $\mu \mathrm{S}$ |
| DAC INTERPOLATION FILTER | Pass Band |  | 20 |  | kHz |
|  | Pass-Band Ripple |  | $\pm 0.01$ |  | dB |
|  | Transition Band |  | 22 |  | kHz |
|  | Stop Band |  | 24 |  | kHz |
|  | Stop-Band Attenuation |  | 70 |  | dB |
|  | Group Delay |  | 446.35 |  | $\mu \mathrm{s}$ |

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Table 7. Timing Specifications

| Parameter |  |  | Comments | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MASTER CLOCK AND RESET | $\mathrm{t}_{\text {м }}$ | MCLK High | $512 \times \mathrm{f}_{5}$ Mode | 18 |  | ns |
|  | $\mathrm{tmL}^{\text {L }}$ | MCLK Low | $512 \times$ fs Mode | 18 |  | ns |
|  | $\mathrm{t}_{\text {maLk }}$ | MCLK Period | $512 \times \mathrm{f}_{5}$ Mode | 36 |  | ns |
|  | fmaLk | MCLK Frequency | $512 \times \mathrm{f}_{5}$ Mode |  | 27 | MHz |
|  | tPDR | PD/RST Low |  | 5 |  | ns |
|  | $\mathrm{t}_{\text {PDRR }}$ | $\overline{\text { PD/RST }}$ Recovery | Reset to Active Output | 4500 |  | $\mathrm{t}_{\text {MCLK }}$ |
| SPI PORT | tснн | CCLK High |  | 10 |  | ns |
|  | tchl | CCLK Low |  | 10 |  | ns |
|  | tcos | CDATA Setup | To CCLK Rising | 5 |  | ns |
|  | tcon | CDATA Hold | From CCLK Rising | 5 |  | ns |
|  | tcıs | $\overline{\text { CLATCH Setup }}$ | To CCLK Rising | 5 |  | ns |
|  | tcLH | $\overline{\text { CLATCH Hold }}$ | From CCLK Falling | 5 |  | ns |
|  | tcode | COUT Enable | From CCLK Falling |  | 10 | ns |
|  | tcod | COUT Delay | From CCLK Falling |  | 10 | ns |
|  | tcor | COUT Hold | From CCLK Falling | 0 |  | ns |
|  | tcots | COUT Three-State | From CCLK Falling |  | 10 | ns |
| DAC SERIAL PORT (Normal Modes) | tठвн | DBCLK High |  | 15 |  | ns |
|  | tbBL | DBCLK Low |  | 15 |  | ns |
|  | $\mathrm{f}_{\mathrm{DB}}$ | DBCLK Frequency |  | $64 \times \mathrm{fs}^{\text {s }}$ |  | ns |
|  | tıLs | DLRCLK Setup | To DBCLK Rising | 0 |  | ns |
|  | tıLH | DLRCLK Hold | From DBCLK Rising | 10 |  | ns |
|  | tdds | DSDATA Setup | To DBCLK Rising | 0 |  | ns |
|  | $\mathrm{t}_{\text {DDH }}$ | DSDATA Hold | From DBCLK Rising | 20 |  | ns |
| DAC SERIAL PORT <br> (Packed 128 Mode, Packed 256 Mode) | tDBH | DBCLK High |  | 15 |  | ns |
|  | tbBL | DBCLK Low |  | 15 |  | ns |
|  | $\mathrm{f}_{\mathrm{DB}}$ | DBCLK Frequency |  | $256 \times \mathrm{f}_{\text {s }}$ |  | ns |
|  | tols | DLRCLK Setup | To DBCLK Rising | 0 |  | ns |
|  | toLH | DLRCLK Hold | From DBCLK Rising | 10 |  | ns |
|  | $\mathrm{t}_{\mathrm{DDS}}$ | DSDATA Setup | To DBCLK Rising | 0 |  | ns |
|  | $\mathrm{t}_{\text {DDH }}$ | DSDATA Hold | From DBCLK Rising | 20 |  | ns |
| ADC SERIAL PORT (Normal Modes) | $\mathrm{t}_{\text {ABD }}$ | ABCLK Delay | From MCLK Transition, $256 \times \mathrm{f}_{\mathrm{s}}$ Mode From MCLK Rising, $512 \times$ fs Mode |  | 15 | ns |
|  |  |  |  |  |  |  |
|  | $\mathrm{t}_{\text {ALS }}$ | LRCLK Skew | From ABCLK Falling | -2 | +2 | ns |
|  | $\mathrm{t}_{\text {ABDD }}$ | ASDATA Delay | From ABCLK Falling |  | 5 | ns |
| ADC SERIAL PORT <br> (Packed 128 Mode, Packed 256 Mode) | $\mathrm{t}_{\text {ABD }}$ | ABCLK Delay | From MCLK Transition, $256 \times \mathrm{f}_{\mathrm{s}}$ Mode From MCLK Rising, $512 \times$ fs Mode |  | 15 | ns |
|  |  |  |  |  |  |  |
|  | $\mathrm{t}_{\text {ALS }}$ | LRCLK Skew | From ABCLK Falling | -2 | +2 | ns |
|  | $\mathrm{t}_{\text {ABDD }}$ | ASDATA Delay | From ABCLK Falling |  | 5 | ns |
| ADC SERIAL PORT (TDM Packed AUX) | $\mathrm{t}_{\text {ABD }}$ | ABCLK Delay | From MCLK Transition, $256 \times \mathrm{f}_{\mathrm{s}}$ Mode From MCLK Rising, $512 \times$ fs Mode |  | 15 | ns |
|  |  |  |  |  |  |  |
|  | $\mathrm{t}_{\text {ALS }}$ | LRCLK Skew | From ABCLK Falling | -2 | +2 | ns |
|  | $\mathrm{t}_{\text {ABDD }}$ | ASDATA Delay | From ABCLK Falling |  | 5 | ns |
|  | $\mathrm{t}_{\mathrm{DDS}}$ | DSDATA1 Hold | To ABCLK Rising | 0 |  | ns |
|  | todh | DSDATA1 Hold | From ABCLK Rising | 7 |  | ns |
| AUXILIARY INTERFACE | taxDs | AAUXDATA Setup | To AUXBCLK Rising | 7 |  | ns |
|  | $\mathrm{t}_{\text {AXDH }}$ | AAUXDATA Hold | From AUXBCLK Rising | 10 |  | ns |
|  | toxDD | DAUXDATA Delay | From AUXBCLK Falling |  | 25 | ns |

Table 7. Timing Specifications (Continued)

| Parameter |  |  | Comments | Min | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| AUXILIARY INTERFACE <br> (Master Mode) | txBD | AUXBCLK Delay | From MCLK Transition, $256 \times$ fs Mode From MCLK Rising, $512 \times \mathrm{f}_{\mathrm{s}}$ Mode |  | 15 | ns |
|  |  |  |  |  |  |  |
|  | txıs | AUXLRCLK Skew | From AUXBCLK Falling | -3 | +3 | ns |
| AUXILIARY INTERFACE (Slave Mode) | tхвн | AUXBCLK High |  | 60 |  | ns |
|  | tхBL | AUXBCLK Low |  | 60 |  | ns |
|  | fxB | AUXBCLK Frequency |  | $64 \times \mathrm{f}_{5}$ |  | ns |
|  | tols | AUXLRCLK Setup | To AUXBCLK Rising | 5 |  | ns |
|  | toLH | AUXLRCLK Hold | From AUXBCLK Rising | 15 |  | ns |

## AD1836A

## ABSOLUTE MAXIMUM RATINGS

Table 8. AD1836A Absolute Maximum Ratings

| Parameter | Min | Max | Unit |
| :--- | :--- | :--- | :--- |
| Analog (AVDD) | -0.3 | +6 | V |
| Digital (DVDD) | -0.3 | +6 | V |
| Input Current (Except Supply Pins) |  | $\pm 20$ | mA |
| Analog Input Voltage (Signal Pins) | -0.3 | $\mathrm{AVDD}+0.3$ | V |
| Digital Input Voltage (Signal Pins) | -0.3 | DVDD +0.3 | V |
| Ambient Temperature (Operating) | -40 | +85 | ${ }^{\circ} \mathrm{C}$ |

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other condition $s$ above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

Table 9. Package Characteristics

| Parameter | Min | Typ | Max | Unit |
| :--- | :--- | :--- | :--- | :--- |
| $\theta_{\text {JA }}($ Thermal Resistance [Junction to Ambient]) |  | 45 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |
| $\theta_{\text {」 }}($ Thermal Resistance [Junction to Case]) |  | 18 |  | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## PIN CONFIGURATION AND PIN FUNCTIONAL DESCRIPTIONS



Figure 2. 52-Lead MQFP

Table 10. Pin Function Descriptions-52-Lead MQFP

| Pin No. | In/Out | Mnemonic | Description |
| :---: | :---: | :---: | :---: |
| 1 | I | DVDD | Digital Power Supply. Connect to digital 5 V supply. |
| 2 | I | CDATA | Serial Control Input. |
| 3 | 1 | PD/RST | Power-Down Reset (Active Low). |
| 4 | 0 | OUTLP3 | DAC 3 Left Positive Output. |
| 5 | 0 | OUTLN3 | DAC 3 Left Negative Output. |
| 6 | 0 | OUTLP2 | DAC 2 Left Positive Output. |
| 7 | 0 | OUTLN2 | DAC 2 Left Negative Output. |
| 8 | 0 | OUTLP1 | DAC 1 Left Positive Output. |
| 9 | O | OUTLN1 | DAC 1 Left Negative Output. |
| 10 | I | AVDD | Analog Power Supply. Connect to analog 5 V . |
| 11 | I | AGND | Analog Ground. |
| 12 | I | FILTD | Filter Capacitor Connection. Bypass with $10 \mu \mathrm{~F} \\| 100 \mathrm{nF}$ to AGND. |
| 13 | I | FILTR | Voltage Reference Filter Capacitor Connection. Bypass with $10 \mu \mathrm{~F} \\| 100 \mathrm{nF}$ to AGND. |
| 14 | I | AGND | Analog Ground. |
| 15 | 1 | AVDD | Analog Power Supply. Connect to analog 5 V supply. |
| 16 | 1 | ADC1INLP | ADC1 Left Positive Input. |
| 17 | I | ADC1INLN | ADC1 Left Negative Input. |
| 18 | I | ADC1INRP | ADC1 Right Positive Input. |
| 19 | I | ADC1INRN | ADC1 Right Negative Input. |
| 20 | 1 | ADC2INLP/CAPL2 | ADC2 Left Positive Input (Direct Mode)/ADC2 Left Decoupling Cap (MUX/PGA and PGA Differential Mode). |
| 21 | I | ADC2INLN/CAPL1 | ADC2 Left Negative Input (Direct Mode)/ADC2 Left Decoupling Capacitor (MUX/PGA and PGA Differential Mode). |
| 22 | I | ADC2INL1 | ADC2 Left Input 1 (MUX/PGA Mode)/Left Positive Input (PGA Differential Mode). |
| 23 | I | ADC2INL2 | ADC2 Left Input 2 (MUX/PGA Mode)/Left Negative Input (PGA Differential Mode). |
| 24 | I | ADC2INR2 | ADC2 Right Input 2 (MUX/PGA Mode)/Right Negative Input (PGA Differential Mode). |
| 25 | I | ADC2INR1 | ADC2 Right Input 1 (MUX/PGA Mode)/Right Positive Input (PGA Differential Mode). |
| 26 | I | ADC2INRN/CAPR1 | ADC2 Right Negative Input (Direct Mode)/ADC2 Right Decoupling Capacitor (MUX/PGA and PGA Differential Mode). |

## AD1836A

| Pin No. | In/Out | Mnemonic | Description |
| :---: | :---: | :---: | :---: |
| 27 | I | ADC2INRP/CAPR2 | ADC2 Right Positive Input (Direct Mode)/ADC2 Right Decoupling Capacitor (MUX/PGA and PGA Differential Mode). |
| 28 | 1 | AGND | Analog Ground. |
| 29 | 1 | AGND | Analog Ground. |
| 30 | 0 | OUTRN1 | DAC 1 Right Negative Output. |
| 31 | 0 | OUTRP1 | DAC 1 Right Positive Output. |
| 32 | 0 | OUTRN2 | DAC 2 Right Negative Output. |
| 33 | 0 | OUTRP2 | DAC 2 Right Positive Output. |
| 34 | 0 | OUTRN3 | DAC 3 Right Negative Output. |
| 35 | 0 | OUTRP3 | DAC 3 Right Positive Output. |
| 36 | I/O | DLRCLK | LR Clock for DACs. |
| 37 | I/O | DBCLK | Bit Clock for DACs. |
| 38 | 1 | DSDATA1 | DAC Input 1 (Input to DAC 1 L and R). |
| 39 | 1 | DGND | Digital Ground. |
| 40 | I | DVDD | Digital Power Supply. Connect to digital 5 V supply. |
| 41 | 1 | DSDATA2 | DAC Input 2 (Input to DAC 2 L and R). |
| 42 | 1 | DSDATA3 | DAC Input 3 (Input to DAC 3 L and R). |
| 43 | 0 | ABCLK | Bit Clock for ADCs. |
| 44 | 0 | ALRCLK | LR Clock for ADCs. |
| 45 | I | MCLK | Master Clock Input. |
| 46 | 1 | ODVDD | Digital Output Driver Power Supply. Connect to 3.3 V or 5 V logic supply. |
| 47 | 0 | ASDATA1 | ADC Serial Data Output 1 (ADC 1 L and R). |
| 48 | 0 | ASDATA2 | ADC Serial Data Output 2 (ADC 2 L and R). |
| 49 | O | COUT | Output for Control Data. |
| 50 | 1 | $\overline{\text { CLATCH }}$ | Latch Input for Control Data. |
| 51 | 1 | CCLK | Control Clock Input for Control Data. |
| 52 | 1 | DGND | Digital Ground. |

## FUNCTIONAL OVERVIEW

## ADCs

There are four ADC channels in the AD1836A configured as two independent stereo pairs. One stereo pair is the primary ADC and has fully differential inputs. The second pair can be programmed to operate in one of three possible input modes (programmed via SPI ADC Control Register 3). The ADC section may also operate at a sample rate of 96 kHz with only the two primary channels active. The ADCs include an on-board digital decimation filter with 120 dB stop-band attenuation and linear phase response, operating at an oversampling ratio of 128 (for 4 -channel 48 kHz operation) or 64 (for 2 -channel 96 kHz operation).

The primary ADC pair should be driven from a differential signal source for best performance. The input pins of the primary ADC connect directly to the internal switched capacitors. To isolate the external driving op amp from the "glitches" caused by the internal switched capacitors, each input pin should be isolated by using a series-connected external $100 \Omega$ resistor together with a 1 nF capacitor connected from each input to ground. This capacitor must be of high quality, for example, ceramic NPO or polypropylene film.

The secondary input pair can operate in one of three modes:

- Direct differential inputs (driven the same way as the primary ADC inputs described above).
- PGA mode with differential inputs. In this mode, the PGA amplifier can be programmed using the SPI port to give an input gain of 0 dB to 12 dB in steps of 3 dB . External capacitors are used after the PGA to supply filtering for the switched capacitor inputs.
- Single-ended MUX/PGA mode. In this mode, two singleended stereo inputs are provided that can be selected using the SPI port. Input gain can be programmed from 0 dB to 12 dB in steps of 3 dB . External capacitors are used to supply filtering for the switched capacitor inputs.

Peak level information for each ADC may be read from the SPI port through Registers 12 to 15 . The data is supplied as a 10 -bit word with a maximum range of 0 dB to -60 dB and a resolution of 1 dB . The registers hold peak information until read; after reading, the registers are reset so that new peak information can be acquired. Refer to the register descriptions for the details on this format.

A digital high-pass filter can be switched in line with the ADCs under SPI control to remove residual dc offsets. It has a 1.3 Hz , 6 dB per octave cutoff at a 44.1 kHz sample rate. The cutoff frequency will scale directly with sample frequency. Note that it does not remove these offsets from the peak level measurement.

The voltage at the $V_{\text {ref }}$ pin, FILTR ( $\sim 2.25 \mathrm{~V}$ ), can be used to bias external op amps that buffer the input signals. See the Power Supply and Voltage Reference section.

## DACs

The AD1836A has six DAC channels arranged as three independent stereo pairs, with six fully differential analog outputs for improved noise and distortion performance. Each channel has its own independently programmable attenuator, adjustable in 1024 linear steps. Digital inputs are supplied through three serial data input pins (one for each stereo pair) and a common frame (DLRCLK) and bit (DBCLK) clock. Alternatively, one of the "packed data" modes may be used to access all six channels on a single TDM data pin.

Each set of differential output pins sits at the dc level of $\mathrm{V}_{\text {ReF }}$ and swings $\pm 1.4 \mathrm{~V}$ for a 0 dB digital input signal. A single op amp third order external low-pass filter is recommended to remove high frequency noise present on the output pins, as well as to provide differential-to-single-ended conversion. Note that the use of op amps with low slew rate or low bandwidth may cause high frequency noise and tones to fold down into the audio band; care should be exercised in selecting these components.

The voltage at the $V_{\text {ref }}$ pin, FILTR ( $\sim 2.25 \mathrm{~V}$ ), can be used to bias the external op amps that buffer the output signals. See the Power Supply and Voltage Reference section.

## CLOCK SIGNALS

The master clock frequency can be selected for 256,512 , or 768 times the sample rate. The default at power-up is $256 \times \mathrm{f}$. For operation at 96 kHz , the master clock frequency should stay at the same absolute frequency. For example, if the AD1836A is programmed in $256 \times \mathrm{fs}, 48 \mathrm{kHz}$ mode, the frequency of the master clock would be $256 \times 48 \mathrm{kHz}=12.288 \mathrm{MHz}$. If the AD1836A is then switched to 96 kHz operation (via writing to the SPI port), the frequency of the master clock should remain at 12.288 MHz (which is now $128 \times \mathrm{f}_{\mathrm{s}}$ ).

The internal clock used in the AD1836A is $512 \times \mathrm{fs}(48 \mathrm{kHz}$ mode) or $256 \times \mathrm{ff}_{\mathrm{S}}$ ( 96 kHz mode). A clock doubler is used to generate this internal master clock from the external clock in the $256 \times \mathrm{f}_{\mathrm{S}}$ and $768 \times \mathrm{f}_{\mathrm{s}}$ modes.

To maintain the highest performance possible, it is recommended that the clock jitter of the master clock signal be limited to less than 300 ps rms , measured using the edge-toedge technique. Even at these levels, extra noise or tones may appear in the DAC outputs if the jitter spectrum contains large spectral peaks. It is highly recommended that an independent crystal oscillator generate the master clock. In addition, it is especially important that the clock signal should not be passed
through an FPGA or other large digital chip before being applied to the AD1836A. In most cases, this will induce clock jitter due to the fact that the clock signal is sharing common power and ground connections with other unrelated digital output signals.

The six DAC channels use a common serial bit clock to clock in the serial data and a common left-right framing clock. The four ADC channels output a common serial bit clock and a left-right framing clock. The clock signals are all synchronous with the sample rate.

## RESET AND POWER-DOWN

Reset will power down the chip and set the control registers to their default settings. After reset is de-asserted, an initialization routine will run inside the AD1836A to clear all memories to zero. This initialization lasts for approximately 4500 MCLKs.

The power-down bit in the DAC Control Register 1 and ADC Control Register 1 will power down the respective digital section. The analog circuitry does not power down. All other register settings are retained.

To avoid possible synchronization problems, if MCLK is $512 \mathrm{f}_{\mathrm{s}}$ or $768 \mathrm{f}_{\mathrm{s}}$, the clock rate should be set in ADC Control Register 3 within the first 3072 MCLK cycles after reset, or DLRCLK and DBCLK should be withheld until after the internal initialization completes (see above).

## SERIAL CONTROL PORT

The AD1836A has an SPI compatible control port that permits programming the internal control registers for the ADCs and DACs and for reading the ADC signal level from the internal peak detectors. The DAC output levels may be independently programmed by means of an internal digital attenuator adjustable in 1024 linear steps.

The SPI control port is a 4 -wire serial control port. The format is similar to the Motorola SPI format except the input data-word is 16 bits wide. The maximum serial bit clock frequency is 8 MHz and may be completely asynchronous to the sample rate of the ADCs and DACs. Figure 3 shows the format of the SPI signal.

All control registers are write-only. They cannot be read back. The ADC peak registers are read-only. They are reset to zero each time they are read and are updated at the next sample time.

Due to an anomaly in the SPI interface, when a write to a DAC control register follows after a read or a write to an ADC
register, it may not be executed properly. Any such write should be performed twice.


Figure 3. Format of SPI Signal

## POWER SUPPLY AND VOLTAGE REFERENCE

The AD1836A is designed for 5 V supplies. Separate power supply pins are provided for the analog and digital sections. These pins should be bypassed with 100 nF ceramic chip capacitors, as close to the pins as possible, to minimize noise pickup. A bulk aluminum electrolytic capacitor of at least $22 \mu \mathrm{~F}$ should also be provided on the same PC board as the codec. For critical applications, improved performance will be obtained with separate supplies for the analog and digital sections. If this is not possible, it is recommended that the analog and digital supplies be isolated by means of a ferrite bead in series with each supply. It is important that the analog supply be as clean as possible.

For ease in interfacing to various logic families, the digital output drivers are supplied from the ODVDD pin. For CMOS logic, this should be connected to the 5 V digital supply. For 3.3 V logic, it should be connected to the 3.3 V supply. For TTL levels, it can be tied to either. All digital inputs are compatible with TTL and CMOS levels.

The internal voltage reference $V_{\text {ref }}$ is brought out on Pin 13 (FILTR) and should be bypassed as close as possible to the chip, with a parallel combination of $10 \mu \mathrm{~F}$ and 100 nF . The reference voltage may be used to bias external op amps to the commonmode voltage of the input and output signal pins. The current drawn should be limited to less than $50 \mu \mathrm{~A}$. This source can be connected directly to op amp inputs but should be buffered if it is required to drive resistive networks.

The FILTD pin should be connected to an external grounded capacitor. This pin is used to reduce the noise of the internal DAC bias circuitry, thereby reducing the DAC output noise. In some cases, this capacitor may be eliminated with little effect on performance.

## SERIAL DATA PORTS—DATA FORMAT

The ADC serial data output mode defaults to the popular $I^{2}$ S format, where the data is delayed by 1 BCLK interval from the edge of the LRCLK. By programming Bits 8 and 9 in ADC Control Register 2, the serial mode can be changed to right justified (RJ), left justified DSP (DSP), left justified (LJ), Packed Mode 128, or Packed Mode 256. In the RJ mode, it is necessary to set Bits 6 and 7 to define the width of the data-word.

The DAC serial data input mode defaults to $I^{2} S$. By programming Bits 5, 6, and 7 in DAC Control Register 1, the mode can be changed to RJ, DSP, LJ, Packed Mode 128, or Packed Mode 256. The word width defaults to 24 bits but can be changed by programming Bits 3 and 4 in DAC Control Register 1. The packed modes accept six channels of data at the DSDATA1 input pin, which is routed independently to each of the six internal DACs.

A special "auxiliary mode" is provided to allow two external stereo ADCs and one external stereo DAC to be interfaced with the AD1836A to provide $8 \mathrm{in} / 8$ out operation. In addition, this mode supports glueless interface to a single SHARC DSP serial port, allowing a SHARC DSP to access all eight channels of analog I/O. In this special mode, many pins are redefined; see Table 11 for a list of redefined pins. Two versions of this mode are available. In the master mode, the AD1836A provides the LRCLK and BCLK signals for the external ADCs and DAC. In the slave mode, external ADC1 provides the LRCLK and BCLK signals (which must be divided down properly from the external master clock), and the AD1836A will sync to these external clocks. In the absence of the external ADC clocks in slave mode, the ALRCLK and ABCLK outputs of the AD1836A (TDM frame sync and bit clock) will default to be the same as in master mode. See
Figure 9 through Figure 11 for details of these modes. Figure 12 shows the internal signal flow diagram of the auxiliary mode.

The following figures show the serial mode formats.


Figure 4. Stereo Serial Modes


Figure 5. ADC Packed Mode 128


Figure 6. ADC Packed Mode 256


Figure 7. DAC Packed Mode 128


Figure 8. DAC Packed Mode 256

## AD1836A



NOTE
AUX BCLK FREQUENCY IS $64 \times$ FRAME RATE; TDM BCLK FREQUENCY IS $256 \times$ FRAME RATE. FSTDM FOLLOWS AUX LRCLK BY $31 / 2 \pm 1 / 2$ TDM BCLK IN BOTH MASTER AND SLAVE MODES.

Figure 9. AUX Mode Timing (Note that the Clocks Are Not to Scale)


Figure 10. AUX Mode Connection to SHARC (Master Mode)


Figure 11. AUX Mode Connection to SHARC (Slave Mode)

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Table 11. Pin Function Changes in AUX Mode

| Pin Name ( ${ }^{2} \mathbf{S} /$ AUX Mode) | $\mathrm{I}^{2}$ S Mode | AUX Mode |
| :---: | :---: | :---: |
| ASDATA1(0) | $\mathrm{I}^{2}$ 'S Data Out, Internal ADC1 | TDM Data Out, to SHARC |
| ASDATA2(O)/DAUXDATA(O) | $1^{2}$ S Data Out, Internal ADC2 | AUX-12S Data Out (to External DAC) |
| DSDATA1(I) | $1^{2} 5$ Data In, Internal DAC1 | TDM Data In, from SHARC |
| DSDATA2(I)/AAUXDATA(I) | $1^{2} \mathrm{~S}$ Data In, Internal DAC2 | AUX-12S Data In 1 (to External ADC) |
| DSDATA3(I)/AAUXDATA2(I) | $1^{2} \mathrm{~S}$ Data In, Internal DAC3 | AUX-12S Data In 2 (to External ADC) |
| ALRCLK(O) | LRCLK for Internal ADC1, ADC2 | TDM Frame Sync Out, to SHARC |
| ABCLK(O) | BCLK for Internal ADC1, ADC2 | TDM BCKL Out, to SHARC |
| DLRCLK(I)/AUXLRCLK(I/O) | LRCLK In/Out Internal DACs | AUX LRCLK In/Out, Driven by External IRCLK from ADC (in slave mode). In master mode, driven by internal MCLK/512. |
| DBCLK(I)/AUXBCLK(I/O) | BCLK In/Out Internal DACs | AUX BCLK In/Out, Driven by External BCLK from ADC (in slave mode). In master mode, driven by internal MCLK/8. |



Figure 12. Extended TDM Mode (Internal Flow Diagram)

## SPI CONTROL REGISTERS

Note that all control registers default to zero at power-up.
Table 12. Serial SPI Word Format

| Register Address | Read/Write | Reserved | Data Field |
| :--- | :--- | :--- | :--- |
| $\mathbf{1 5 : 1 2}$ | $\mathbf{1 1}$ | $\mathbf{1 0}$ | $\mathbf{9 : 0}$ |
| 4 Bits | $1=$ Read | 0 | 10 Bits |
|  | $0=$ Write |  |  |

Table 13. Register Addresses and Functions

| Register Address |  |  | RD/WR | Reserved | Function |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Bit 15 | Bit 14 | Bit 13 | Bit 12 | Bit 11 | Bit 10 | Bits 9:0 |
| 0 | 0 | 0 | 0 | 0 | 0 | DAC Control 1 |
| 0 | 0 | 0 | 1 | 0 | 0 | DAC Control 2 |
| 0 | 0 | 1 | 0 | 0 | 0 | DAC1L Volume |
| 0 | 0 | 1 | 1 | 0 | 0 | DAC1R Volume |
| 0 | 1 | 0 | 0 | 0 | 0 | DAC2L Volume |
| 0 | 1 | 0 | 1 | 0 | 0 | DAC2R Volume |
| 0 | 1 | 1 | 0 | 0 | 0 | DAC3L Volume |
| 0 | 1 | 1 | 1 | 0 | 0 | ADC3R Volume |
| 1 | 0 | 0 | 0 | 0 | 0 | ADC1R—Peak Level (Read-Only) |
| 1 | 0 | 0 | 1 | 0 | 0 | ADC2L—Peak Level (Read-Only) (Read-Only) |
| 1 | 0 | 1 | 0 | 0 | 0 | ADC2R—Peak Level (Read-Only) |
| 1 | 0 | 1 | 1 | 0 | 0 | ADC Control 1 2 |
| 1 | 1 | 0 | 0 | 0 | 0 | ADC Control 3 |
| 1 | 1 | 0 | 1 | 0 | 0 | Reserved |
| 1 | 1 | 1 | 0 | 0 | 0 | 0 |
| 1 | 1 | 1 | 1 | 0 | 0 |  |

Table 14. DAC Control Register 1
Packed Mode: Eight channels are "packed" in DSDATA1 serial input. Packed Mode 128: Refer to Figure 7. Packed Mode 256: Refer to Figure 8.

| Address | RD/WR | Reserved | Function |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | De-emphasis | Serial Mode | Data-Word Width | Power-Down | Interpolator Mode | Reserved |
| 15, 14, 13, 12 | 11 | 10 | 9,8 | 7,6,5 | 4,3 | 2 | 1 | 0 |
| 0000 | 0 | 0 | $\begin{aligned} & 00=\text { None } \\ & 01=44.1 \mathrm{kHz} \\ & 10=32.0 \mathrm{kHz} \\ & 11=48.0 \mathrm{kHz} \end{aligned}$ | $\begin{aligned} & 000=R^{2} S \\ & 001=\text { RJ } \\ & 010=\text { DSP } \\ & 011=L J \\ & 100=\text { Packed Mode } 256 \\ & 101=\text { Packed Mode } 128 \\ & 110=\text { Reserved } \\ & 111=\text { Reserved } \end{aligned}$ | $\begin{aligned} & 00=24 \text { Bits } \\ & 01=20 \text { Bits } \\ & 10=16 \text { Bits } \\ & 11=\text { Reserved } \end{aligned}$ | $\begin{aligned} & 0=\text { Normal } \\ & 1=\text { PWRDWN } \end{aligned}$ | $\begin{aligned} & 0=8 \times(48 \mathrm{kHz}) \\ & 1=4 \times(96 \mathrm{kHz}) \end{aligned}$ | 0 |

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Table 15. DAC Control Register 2

| Address | RD/WR | Reserved | DAC Mute |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | DAC3R | DAC3L | DAC2R | DAC2L | DAC1R | DAC1L |
| 15, 14, 13, 12 | 11 | 10, 9, 8, 7, 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 0001 | 0 | 00000 | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ |

Table 16. DAC Volume Registers

|  |  |  | Function |
| :--- | :--- | :--- | :--- |
| Address | RD/WR | Reserved | Volume |
| $\mathbf{1 5 , 1 4 , 1 3 , 1 2}$ | $\mathbf{1 1}$ | $\mathbf{1 0}$ | $\mathbf{9 : 0}$ |
| 0010: DAC1L | 0 | 0 | 0 to 1023 in 1024 Linear Steps |
| 0011: DAC1R |  |  |  |
| 0100: DAC2L |  |  |  |
| 0101: DAC2R |  |  |  |
| 0110: DAC3L |  |  |  |
| 0111: DAC3R |  |  |  |

Table 17. ADC Control Register 1

| Address | RD/WR | Reserved | Function |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Filter | Power-Down | Sample Rate | Left Gain | Right Gain |
| 15, 14, 13, 12 | 11 | 10,9 | 8 | 7 | 6 | 5, 4, 3 | 2,1,0 |
| 1100 | 0 | 00 | $\begin{aligned} & 0=\text { DC } \\ & 1=\text { High Pass } \end{aligned}$ | $\begin{aligned} & \hline 0=\text { Normal } \\ & 1=\text { PWRDWN } \end{aligned}$ | $\begin{aligned} & 0=48 \mathrm{kHz} \\ & 1=96 \mathrm{kHz} \end{aligned}$ | $\begin{aligned} & \hline 000=0 \mathrm{~dB} \\ & 001=3 \mathrm{~dB} \\ & 010=6 \mathrm{~dB} \\ & 011=9 \mathrm{~dB} \\ & 100=12 \mathrm{~dB} \\ & 101=\text { Reserved } \\ & 110=\text { Reserved } \\ & 111=\text { Reserved } \end{aligned}$ | $\begin{aligned} & \hline 000=0 \mathrm{~dB} \\ & 001=3 \mathrm{~dB} \\ & 010=6 \mathrm{~dB} \\ & 011=9 \mathrm{~dB} \\ & 100=12 \mathrm{~dB} \\ & 101=\text { Reserved } \\ & 110=\text { Reserved } \\ & 111=\text { Reserved } \end{aligned}$ |

Table 18. ADC Control Register 2
Packed Mode: Eight channels are "packed" in ASDATA1 serial output. Packed Mode 128: Refer to Figure 5. Packed Mode 256: Refer to Figure 6. Packed Mode AUX: Refer to Figure 9 to Figure 11. Note that Packed AUX mode affects the entire chip, including the DAC serial mode.

| Address | RD/WR | Reserved | Master/Slave AUX Mode | SOUT Mode | Word Width | ADC Mute |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  | ADC2R | ADC2L | ADC1R | ADC1L |
| 15, 14, 13, 12 | 11 | 10 | 9 | 8,7,6 | 5,4 | 3 | 2 | 1 | 0 |
| 1101 | 0 | 0 | $\begin{aligned} & 0=\text { Slave } \\ & 1=\text { Master } \end{aligned}$ | $\begin{aligned} & 000=I^{2} \mathrm{~S} \\ & 001=\text { RJ } \\ & 010=\text { DSP } \\ & 011=\sqcup \\ & 100=\text { Packed Mode } 256 \\ & 101=\text { Packed Mode } 128 \\ & 110=\text { Packed Mode AUX } \end{aligned}$ | $\begin{aligned} & 00=24 \text { Bits } \\ & 01=20 \text { Bits } \\ & 10=16 \text { Bits } \\ & 11=\text { Reserved } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ | $\begin{aligned} & 0=\text { On } \\ & 1=\text { Mute } \end{aligned}$ |

Table 19. ADC Control Register 3
When changing clock mode, other SPI bits that are written during the same SPI transaction may be lost. Therefore, it is recommended that these be set separately.

|  |  |  |  | Function |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Address | RD/WR | Reserved | Clock Mode | Left Differential I/P Select | Right Differential I/P Select | Left MUX/PGA Enable | Left MUX I/P Select | Right MUX/PGA Enable | Right MUX I/P Select |
| $\begin{aligned} & \text { 15, 14, } \\ & 13,12 \end{aligned}$ | 11 | 10, 9, 8 | 7,6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 1110 | 0 | 000 | $\begin{aligned} & 00=256 \times \mathrm{f}_{\mathrm{s}} \\ & 01=512 \times \mathrm{f}_{\mathrm{s}} \\ & 10=768 \times \mathrm{fs}_{\mathrm{s}} \end{aligned}$ | $0=$ Differential PGA Mode <br> 1 = PGA/MUX <br> Mode (Single- <br> Ended Input) | $0=$ Differential PGA Mode <br> 1 = PGA/MUX <br> Mode (Single- <br> Ended Input) | $\begin{aligned} & 0=\text { Direct } \\ & 1=\text { MUX/PGA } \end{aligned}$ | $\begin{aligned} & 0=1 / P 0 \\ & 1=I / P 1 \end{aligned}$ | $\begin{aligned} & 0=\text { Direct } \\ & 1=\text { MUX/PGA } \end{aligned}$ | $\begin{aligned} & \hline 0=1 / P 0 \\ & 1=1 / P 1 \end{aligned}$ |

Table 20. ADC Peak Level Data Registers

| Address | RD/WR | Reserved | Peak Level Data (10 Bits) |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | 6 Data Bits | 4 Fixed Bits |
| 15, 14, 13, 12 | 11 | 10 | 9:4 | 3:0 |
| $\begin{aligned} & 1000=A D C 1 L \\ & 1001=A D C 1 R \\ & 1010=A D C 2 L \\ & 1011=A D C 2 R \end{aligned}$ | 1 | 0 | $\begin{aligned} & 000000=0.0 \mathrm{dBFS} \\ & 000001=-1.0 \mathrm{dBFS} \\ & 000010=-2.0 \mathrm{dBFS} \\ & 000011=-3.0 \mathrm{dBFS} \end{aligned}$ $111100=-60 \mathrm{dBFS} \text { Min }$ | $0000$ <br> The 4 LSBs are always zero. |

## AD1836A



ноте
ADC2 SINGLE-ENDED MUX PGA INPUT MODE-LEFT CHANNEL ONLY SHOWN. CONTROL REGISTER 3 CONTENTS: 6 LSBs: SELECT INPUT NO. 1: 111010 SELECT INPUT NO. 2: 111111

Figure 13. Single-Ended MUX/PGA Mode


Figure 14. Differential PGA Mode

## OUTLINE DIMENSIONS



Figure 15. 52-Lead Plastic Quad Flat Package [MQFP]
(S-52A)
Dimensions shown in millimeters

## ESD CAUTION

ESD (electrostatic discharge) sensitive device. Electrostatic charges as high as 4000 V readily accumulate on the human body and test equipment and can discharge without detection. Although this product features proprietary ESD protection circuitry, permanent damage may occur on devices subjected to high energy electrostatic discharges. Therefore, proper ESD precautions are recommended to avoid performance degradation or loss of functionality.

ORDERING GUIDE

| AD1836A Products | Temperature Package | Package Description | Package Option |
| :--- | :--- | :--- | :--- |
| AD1836AAS | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ Ambient | 52 -Lead MQFP | S-52A |
| AD1836AASRL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ Ambient | 52 -Lead MQFP | S-52A on 13" Reels |
| AD1836ACS | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ Ambient | 52 -Lead MQFP | S-52A |
| AD1836ACSRL | $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ Ambient | 52 -Lead MQFP | S-52A on 13" Reels |
| EVAL-AD1836AEB |  | Evaluation Board |  |

## AD1836A

## NOTES

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Компания «Life Electronics» занимается поставками электронных компонентов импортного и отечественного производства от производителей и со складов крупных дистрибьюторов Европы, Америки и Азии.

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- Работу по проектам и поставку образцов.
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- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.


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> Email: org@lifeelectronics.ru


[^0]:    ${ }^{1}$ Total harmonic distortion + noise and dynamic range typical specifications are for two channels active, $\mathrm{max} / \mathrm{min}$ are all channels active.

