



### Features

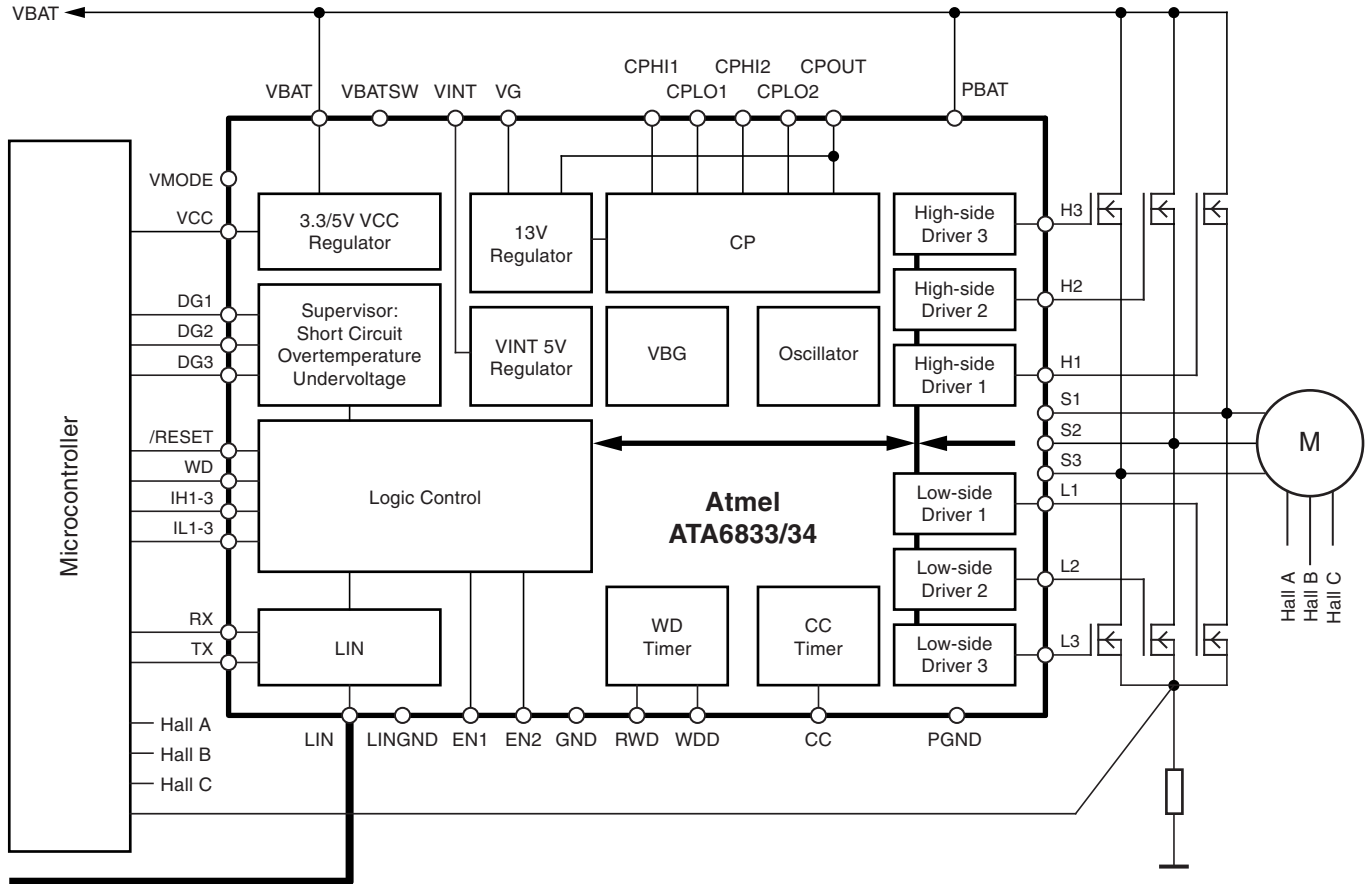
- Atmel ATA6833 temperature range  $T_J = 150^{\circ}\text{C}$
- Atmel ATA6834 extended temperature range  $T_J = 200^{\circ}\text{C}$
- Direct driving of 6 external NMOS transistors with a maximum switching frequency of 50kHz
- Integrated charge pump to provide gate voltages for high-side drivers and to supply the gate of the external battery reverse protection NMOS
- Built-in 5V/3.3V voltage regulator with current limitation
- Reset signal for the microcontroller
- Sleep Mode with supply current of typically  $< 45\mu\text{A}$
- Wake-up via LIN bus or high voltage input
- Programmable window watchdog
- Battery overvoltage protection and battery undervoltage management
- Overtemperature warning and protection (shutdown)
- 200mA peak current for each output driver
- LIN transceiver conformal to LIN 2.1 and SAEJ2602-2 with outstanding EMC and ESD performance
- QFN48 package  $7\text{mm} \times 7\text{mm}$

# 1. Description

The Atmel® ATA6833 and Atmel ATA6834 are system basis chips for three-phase brushless DC motor controllers designed in Atmel's state-of-the-art 0.8µm SOI technology SMART-I.S.™ 1. In combination with a microcontroller and six discrete power MOSFETs, the system basis chip forms a BLDC motor control unit for automotive applications. In addition, the circuits provide a 3.3V/5V linear regulator and a window watchdog.

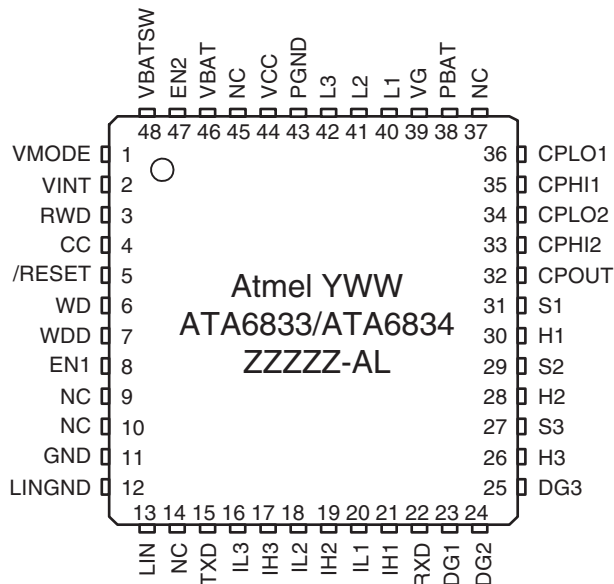
The circuit includes various control and protection functions like overvoltage and overtemperature protection, short circuit detection, and undervoltage management. Thanks to these function blocks, the driver fulfils a maximum of safety requirements and offers a high integration level to save cost and space in various applications. The target applications are most suitable for the automotive market due to the robust technology and the high qualification level. Atmel ATA6834, in particular, is designed for applications in a high-temperature environment.

Figure 1-1. Block Diagram



## 2. Pin Configuration

Figure 2-1. Pinning QFN48



Note: YWW Date code (Y = Year - above 2000, WW = week number)  
 ATA683x Product name  
 ZZZZZ Wafer lot number  
 AL Assembly sub-lot number

Table 2-1. Pin Description

Pin	Symbol	I/O	Function
1	VMODE	I	Selector for $V_{CC}$ and interface logic voltage level
2	VINT	I/O	Blocking capacitor
3	RWD	I	Resistor defining the watchdog interval
4	CC	I/O	RC combination to adjust cross conduction time
5	/RESET	O	Reset signal for microcontroller
6	WD	I	Watchdog trigger signal
7	WDD	I	Enable and disable the watchdog
8	EN1	I	Microcontroller output to switch system in Sleep Mode
9	N.C.		Connect to GND
10	N.C.		Connect to GND
11	GND	I	Ground
12	LINGND		Ground for LIN connected to GND
13	LIN	I/O	LIN-bus terminal
14	NC		Connect to GND
15	TXD	I	Transmit signal to LIN bus from microcontroller
16	IL3	I	Control Input for output L3
17	IH3	I	Control Input for output H3

**Table 2-1. Pin Description**

Pin	Symbol	I/O	Function
18	IL2	I	Control Input for output L2
19	IH2	I	Control Input for output H2
20	IL1	I	Control Input for output L1
21	IH1	I	Control Input for output H1
22	RXD	O	Receive signal from LIN bus for microcontroller
23	DG1	O	Diagnostic output 1
24	DG2	O	Diagnostic output 2
25	DG3	O	Diagnostic output 3
26	H3	O	Gate voltage high-side 3
27	S3	I/O	Voltage at half bridge 3
28	H2	O	Gate voltage high-side 2
29	S2	I/O	Voltage at half bridge 2
30	H1	O	Gate voltage high-side 1
31	S1	I/O	Voltage at half bridge 1
32	CPOUT	I/O	Charge pump output capacitor
33	CPHI2	I	Charge pump capacitor 2
34	CPLO2	O	Charge pump capacitor 2
35	CPHI1	I	Charge pump capacitor 1
36	CPLO1	O	Charge pump capacitor 1
37	NC		Connect to GND
38	PBAT	I	Power supply (after reverse protection) for charge pump and gate drivers
39	VG	I/O	Blocking capacitor
40	L1	O	Gate voltage H-bridge, low-side 1
41	L2	O	Gate voltage H-bridge, low-side 2
42	L3	O	Gate voltage H-bridge, low-side 3
43	PGND	I	Power ground for H-bridge and charge pump
44	VCC	O	5V/100mA supply for microcontroller
45	NC		Connect to GND
46	VBAT	I	Supply voltage for IC core (after reverse protection)
47	EN2	I	High voltage enable input
48	VBATSW	O	100Ω PMOS switch from $V_{BAT}$

## 3. Functional Description

### 3.1 Power Supply Unit with Supervisor Functions

#### 3.1.1 Power Supply

The IC has to be supplied by a reverse-protected battery voltage. To prevent damage to the IC, proper external protection circuitry has to be added. It is recommended to use at least one capacitor combination of storage and RF capacitors behind the reverse protection circuitry, which is connected close to the VBAT and GND pins of the IC.

A fully integrated low-power and low-drop regulator (VINT regulator), stabilized by an external blocking capacitor, provides the necessary low-voltage supply needed for the wake-up process. A trimmed low-power band gap is used as reference for the VINT regulator as well as for the VCC regulator. All internal blocks are supplied by VINT regulator. VINT regulator must not be used for any external supply purposes.

Nothing inside the IC except the logic interface to the external microcontroller is supplied by the 5V/3.3V VCC regulator.

Both voltage regulators are checked by a “power-good comparator”, which keeps the whole chip in reset as long as the internal supply voltage (VINT regulator output) is too low and generates a reset for the external microcontroller if the output voltage of the VCC regulator is not sufficient.

#### 3.1.2 VBatt Switch

This high-voltage switch provides the battery voltage at pin VBATSW for various purposes. It is switched ON after power on reset when the IC transits to Active Mode and it will only turn OFF when the IC changes to Sleep Mode. Watchdog resets do not have an effect on the switch. The switch can be used for measuring purposes as well as to switch on external voltage regulators.

#### 3.1.3 Voltage Supervisor

This function is implemented to protect the IC and the external power MOS transistors from damage due to overvoltage on PBAT input. In the event of overvoltage ( $V_{THOV}$ ) or undervoltage ( $V_{THUV}$ ), the external NMOS motor driver transistors will be switched off. The failure state will be flagged on DG2 pin. It is recommended to block PBAT with an external RF capacitor to suppress high frequency disturbances.

#### 3.1.4 Temperature Supervisor

An integrated temperature sensor prevents the IC from overheating. If the temperature is above the overtemperature pre-warning threshold  $T_{JPW\ set}$ , the diagnostic pin DG3 will be switched to HIGH to signal this event to the external microcontroller. The microcontroller should take actions to reduce the power dissipation in the IC. If the temperature rises above the overtemperature shutdown threshold  $T_{J\ switch\ off}$ , the VCC regulator and all output drivers together with the LIN transceiver will be switched OFF immediately and the /RESET signal will go LOW. Both thresholds have a built-in hysteresis to avoid oscillations. The IC will return to normal operation (Active Mode) when it has cooled down below the shutdown threshold. When the junction temperature drops below the pre-warning threshold, bit DG3 will be switched LOW.

## 3.2 Active Mode and Sleep Mode

The IC has two modes: Sleep Mode and Active Mode. Switching between the modes is described below. By default the IC starts in Active Mode (which means normal operation) after power-on. A *Go to Sleep* procedure switches the IC from Active Mode to Sleep Mode (standby). A *Go to Active* procedure brings the IC back from Sleep Mode to Active Mode. When in Sleep Mode the internal 5V supply (VINT regulator), the EN2 pin input structure, and a certain part of the LIN receiver remain active to ensure a proper startup of the system. The VCC regulator is turned off.

The *Go to Sleep* and *Go to Active* procedures are implemented as follows:

### Go to Sleep

Pin EN1 is a low-voltage input supplied by the VCC regulator. It is ESD protected by diodes against VCC and GND. Thus the input voltage at pin EN1 must not go below GND or exceed the output voltage of the VCC regulator. A transition from HIGH to LOW followed by a permanent LOW signal for a minimum time period  $t_{\text{gotosleep}}$  (typical 10 $\mu\text{s}$ ) at pin EN1 switches the IC to Sleep Mode as the EN1 is edge triggered. VCC is switched off in Sleep Mode. It is recommended to keep EN1 LOW during normal operation.

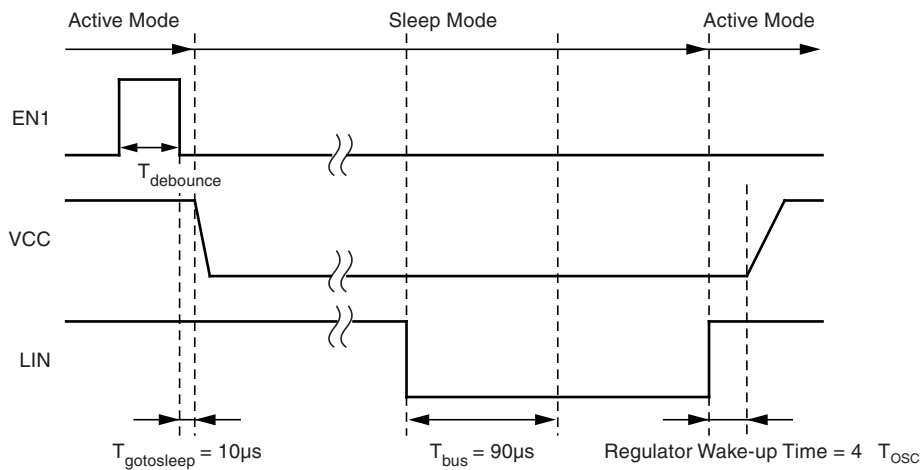
### Go to Active Using Pin EN2

Pin EN2 is a high-voltage input for external wake-up signals. Its input structure consists of a comparator with a built-in hysteresis. It is ESD-protected by diodes against GND and  $V_{\text{BAT}}$ , B, and for this reason the applied input voltage must not go below GND or exceed  $V_{\text{BAT}}$ . Pulling EN2 up to  $V_{\text{BAT}}$  switches the IC to Active Mode. EN2 is debounced and edge triggered.

### Go to Active Using the LIN Interface

Using the LIN interface provides a second possibility to wake-up the IC (see Figure 3-1). A falling edge at pin LIN followed by a dominant bus level maintained for a minimum time period ( $T_{\text{bus}}$ ) and ending with a rising edge leads to a remote wake-up request. The device switches from Sleep Mode to Active Mode. The VCC regulator is activated and the internal LIN slave termination resistor is switched on.

Figure 3-1. Wake-up Using the LIN Interface



### 3.3 5V/3.3V VCC Regulator

The 5V/3.3V regulator is fully integrated. It requires an external electrolytic capacitor in the range of 2.2 $\mu$ F up to 10 $\mu$ F and with an ESR in the range from 2 $\Omega$  to 15 $\Omega$  for stability (see Figure 3-2). The output voltage can be configured as either 5V or 3.3V by connecting pin VMODE to either pin VINT or GND. Since the regulator is not designed to be switched between both output voltages during operation, it is advisable to hard-wire VMODE pin. The logic levels of the microcontroller interface are adapted to the VCC regulator output voltage. The maximum output current ( $I_{OS1}$ ) of the regulator is 100mA. For  $T_J > 150^\circ\text{C}$  the  $I_{OS1}$  of Atmel<sup>®</sup> ATA6834 is reduced to 80mA. The VCC regulator has a built-in short circuit protection. A comparator checks the output voltage of the VCC regulator and keeps the external microcontroller in reset as long as the voltage is below the lower operation minimum (shown in Figure 3-3).

Figure 3-2. ESR versus Load Current for External Capacitors with Different Values

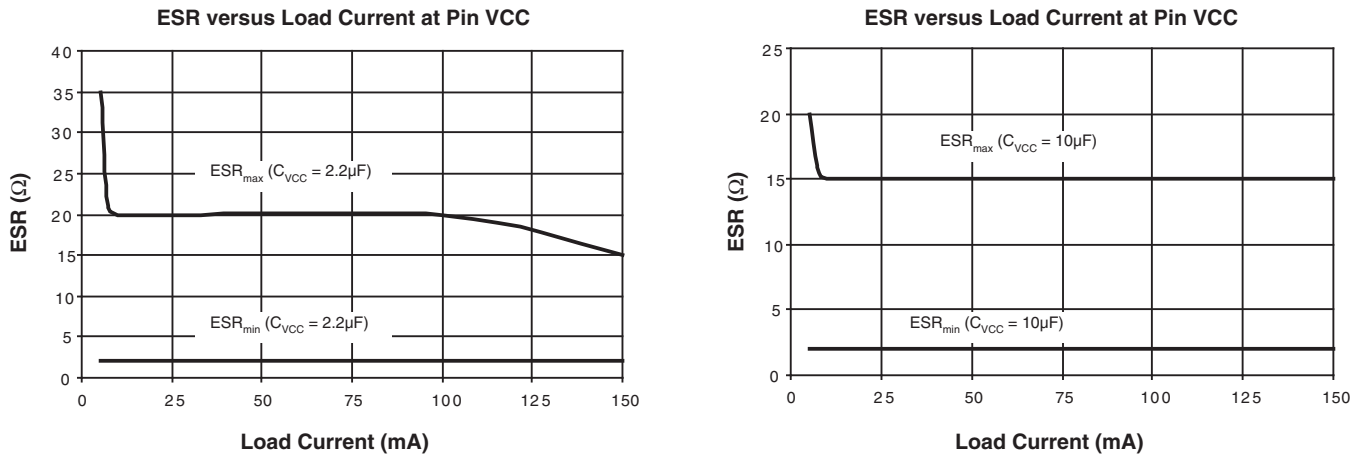
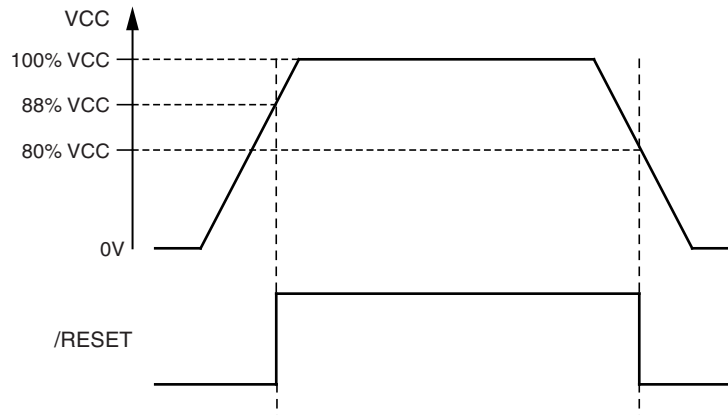


Figure 3-3. /RESET as Function of the VCC Output Voltage

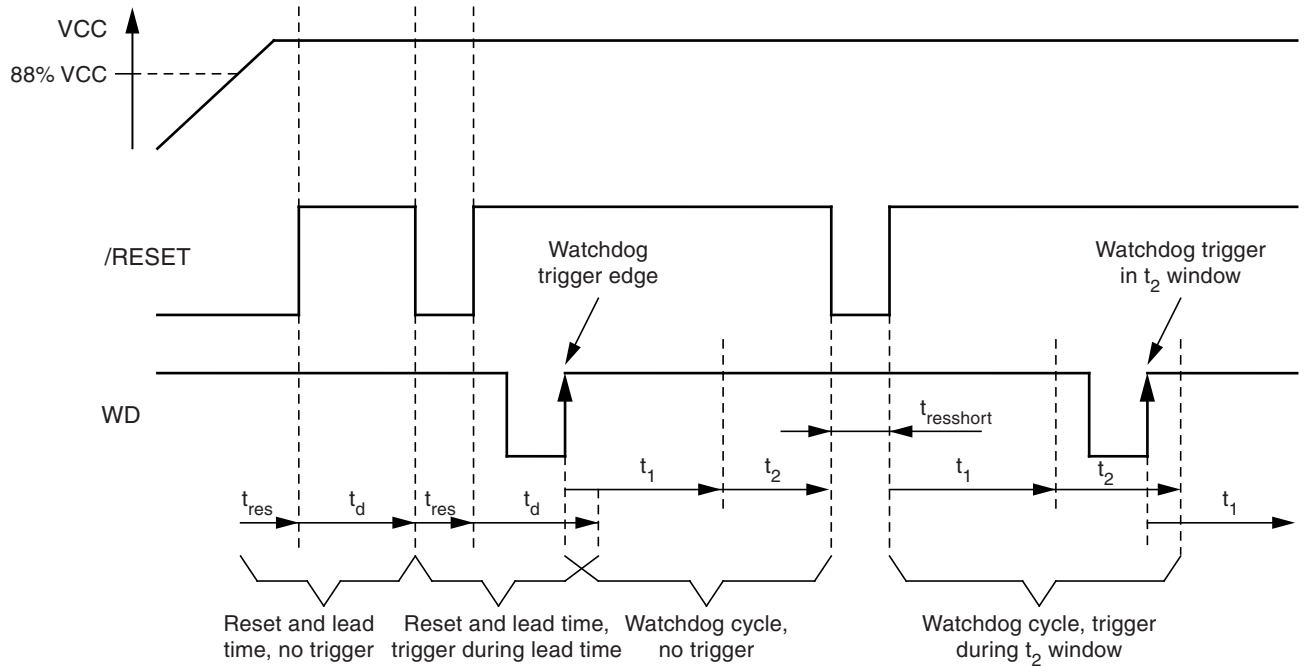


### 3.4 Reset and Watchdog Management

The watchdog timing is based on the trimmed internal watchdog oscillator. Its period time  $T_{OSC}$  is determined by the external resistor  $R_{WD}$ . A HIGH signal on WDD pin enables the watchdog function; a LOW signal disables it. Since WDD pin is equipped with an internal pull-up resistor the watchdog is enabled by default. In order to keep the current consumption as low as possible the watchdog is switched off during Sleep Mode.

The timing diagram in Figure 3-4 shows the watchdog and external reset timing.

**Figure 3-4. Timing Diagram of the Watchdog in Conjunction with the /RESET Signal**



After power-up of the VCC regulator (VCC output exceeds 88% of its nominal value) /RESET output stays LOW for the timeout period  $t_{res}$  (typical 10ms). Subsequently /RESET output switches to HIGH. During the following time  $t_d$  (typical 500ms) a rising edge at the input WD is expected otherwise another external reset will be triggered.

When the watchdog has been correctly triggered for the first time, normal watch-dog operation begins. A normal watchdog cycle consists of two time sections  $t_1$  and  $t_2$  followed by a short pulse for the time  $t_{resshort}$  at /RESET if no valid trigger has been applied at pin WD during  $t_2$ . Rising edges on WD pin during  $t_1$  also cause a short pulse on /RESET. Start for such a cycle is always the time of the last rising edge either on WD pin or on /RESET pin.

If the watchdog is disabled (WDD = LOW), only the initial reset for the time  $t_{res}$  after power-up will be generated.

Additional resets will be generated if the VCC output voltage drops below 80% of its nominal value.

The following example demonstrates how to calculate the timing scheme for valid watchdog trigger pulses, which the external microcontroller has to provide in order to prevent undesired resets.

**Example:**

Using an external resistor  $R_{WD} = 33 \text{ k}\Omega \pm 1\%$  results in typical parameters as follows:

$$T_{OSC} = 12.4 \mu\text{s}$$

$$t_1 = 980 \times T_{OSC} = 12.1 \text{ms} \pm 10\%$$

$$t_2 = 780 \times T_{OSC} = 9.6 \text{ms} \pm 10\%$$

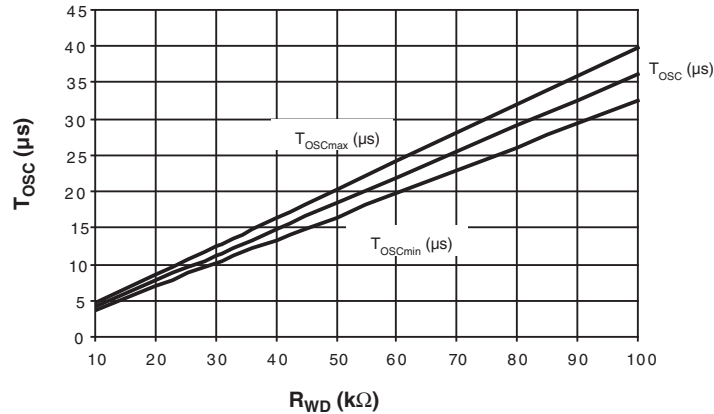
$$t_1 + t_2 = 21.7 \text{ms} \pm 10\%$$

Hence, the minimum time the external microcontroller has to wait before pin WD can be triggered is in worst case  $t_{min} = 1.1 \times t_1 = 13.3 \text{ms}$ . The maximum time for the watchdog trigger on WD pin is  $t_{max} = 0.9 \times (t_1 + t_2) = 19.5 \text{ms}$ . Thus watchdog trigger input must remain within  $t_{max} - t_{min} = 6.2 \text{ms}$ .

Other values can be set up by picking a different resistor value for  $R_{WD}$ . The dependency of  $T_{OSC}$  on the value of  $R_{WD}$  is shown in Figure 3-5.



Figure 3-5.  $T_{OSC}$  versus  $R_{WD}$



The tolerance of  $T_{OSC}$  is  $\pm 10\%$  for resistors  $R_{WD}$  with maximum  $\pm 1\%$  in tolerance.

### 3.5 Charge Pump

A charge pump has been implemented in order to provide sufficient voltage to operate the external high-side power-NMOS transistors and the VG regulator, which drives the low-side Power-NMOS transistors. The charge pump output voltage at CPOUT pin is controlled to settle typically about 15V above the voltage at pin PBAT. A built-in supervisor circuit checks if the output voltage is sufficient to operate the VG regulator and external Power-NMOS transistors. The output voltage is accepted as good when it rises above  $VCP_{CPGOOD}$ . A charge pump failure is flagged at DG2 if this minimum can not be reached or if the output voltage drops below the lower threshold of  $VCP_{CPGOOD}$  due to overloading.

The two shuffle capacitors should have the same value. The value of the reservoir capacitor should be at least twice the value of one shuffle capacitor. Two external shuffle capacitors and an external reservoir capacitor have to be provided. The typical values for the two shuffle capacitor is 100nF, and for the reservoir capacitor is 470nF. All capacitors should be ceramic. It is advisable to pick a reservoir capacitor with twice or three-times the size of the two equally-sized shuffle capacitors. The greater the capacitors, the greater the output current capability.

### 3.6 VG Regulator

The VG regulator provides a stable voltage to supply the low-side gate drivers and to deliver sufficient voltage for the external low-side Power-NMOS transistors. Typically the output voltage is 12V. In order to guarantee reliable operation even with a low battery voltage, the VG regulator is supplied by the charge pump output. For stability, an external ceramic capacitor of typically 470nF has to be provided. There is no internal supervision of the VG output voltage.

### 3.7 Output Drivers and Control Inputs IL1-IL3, IH1-IH3

This IC offers six push-pull output drivers for the external low-side and high-side power-NMOS transistors. To guarantee reliable operation, the low-side drivers are supplied by the VG regulator while the high-side drivers are supplied directly by the charge pump. All drivers are designed to operate at switching frequencies in the range of DC up to 50kHz. The maximum gate charge that can be delivered to each external Power-NMOS transistor at 50kHz is 100nC.

The output drivers are directly controlled by the digital input pins IL1 to IL3 and IH1 to IH3 (see Table 3-1). All pins are equipped with an internal pull-down resistor. To operate the output drivers properly the following requirements have to be fulfilled:

1. Device is in Active Mode.
2. In case of watchdog is enabled, at least one valid watchdog trigger has been accepted.
3. The voltage at pin PBAT lies within its operation range. Neither undervoltage nor overvoltage is present.
4. The charge pump output voltage has been accepted as good, thus it exceeded  $VCP_{CPGOOD}$ .
5. No overtemperature shutdown has occurred.

If a short circuit is detected by one of the sense inputs S1 to S3, the output drivers will be switched off after a debounce time of 6  $\mu s$  and the output DG1 will be flagged (see also Section 3.8 "Short Circuit Detection" on page 10). The output drivers will be enabled again and DG1 will be cleared with a rising edge at one of the control inputs (IL1 to IL3, IH1 to IH3).

Additional logic prevents short circuits due to switching on one power-NMOS transistor while the opposite one in the same branch is switched on already.

**Table 3-1. Status of the Output Drivers Depending on the Control Inputs**

Mode	Control Inputs IL(1..3)	Control Inputs IH(1..3)	Driver Stage for External Power MOS L(1..3), H(1..3)	Comments
Sleep	X	X	OFF	Sleep Mode
Active	0	0	OFF	
Active	1	0	L(1..3) ON, H(1..3) OFF	
Active	0	1	H(1..3) ON, L(1..3) OFF	
Active	1	1	OFF	Shoot-through protection

### 3.8 Short Circuit Detection

Short circuits in the motor bridge circuitry are sensed by S1 to S3 inputs. Internal comparators monitor the voltage differences between the drain and the source terminals of the external power-NMOS transistors. If one transistor switches on and its drain-source voltage exceeds  $V_{SC}$  threshold (typically 4V) after a blanking time  $t_{SC}$  (typically 6 $\mu$ s, see Figure 3-6), a short circuit in this branch will be detected. In this case, all output drivers will be switched off immediately and pin DG1 will be set to HIGH. With a rising edge at any of the pins IL1 to IL3 or IH1 to IH3, the diagnostic output DG1 will be reset and the drivers can be switched on again.

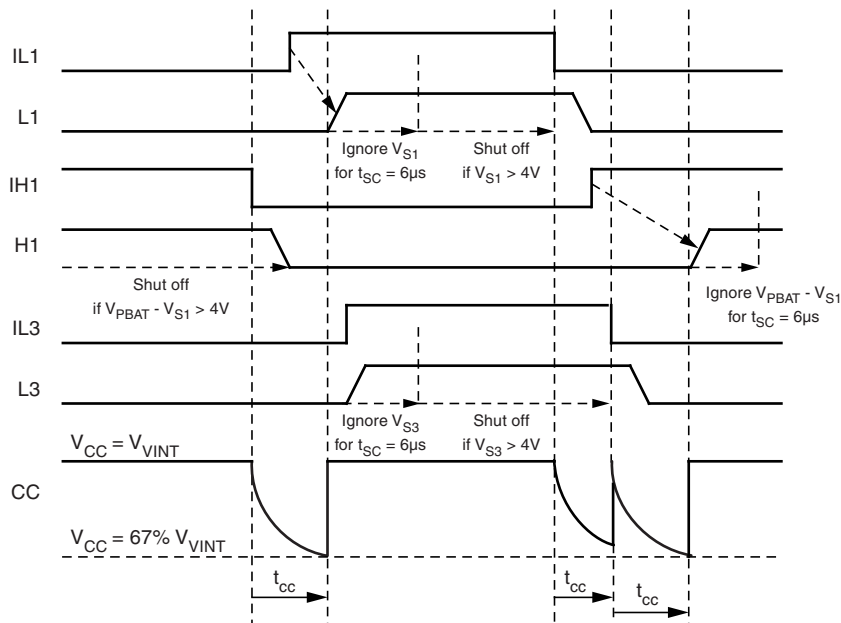
### 3.9 Cross Conduction Timer

In order to prevent damage of the motor bridge due to peak currents a non-overlapping phase for switching the power-NMOS transistors is mandatory. Therefore, a cross conduction timer has been implemented to prevent switching on any output driver for a time  $t_{CC}$  after any other driver has been switched off. This also accounts for toggling any other driver after a short circuit was detected. An external RC parallel combination defines the value for  $t_{CC}$  and can be estimated as follows:

$$t_{CC} = K_{CC} \times R_{CC} \text{ (k}\Omega\text{)} \times C_{CC} \text{ (nF)}, K_{CC} \text{ is specified in Section 8. "Electrical Characteristics" on page 15.}$$

The RC combination is connected between CC and GND pins. When one of the drivers has been switched off the RC combination is charged to 5V ( $V_{INT}$ ) and discharged with its time constant. Any low to high transition at the control inputs will be masked out at the driver outputs until the voltage at CC pin drops below 67% of its initial value ( $V_{INT}$ ). The timer will be re-triggered at any time by any falling edge at the control inputs. This is shown in the following figure.

**Figure 3-6. Interaction of Short Circuit Detection and Cross Conduction Timer**



At least 5kW minimum and 5nF at maximum should be used as values for the RC combination. 10kW is recommended. If the non-overlapping phase is controlled by the external microcontroller, it is possible to do without the external capacitor. The minimum time  $t_{CC}$  is defined by the parasitic capacitance at CC pin.

### 3.10 Diagnostic Outputs D1 - D3

As mentioned in the sections above, the diagnostic outputs DG1 to DG3 are used to signal failures. This is summarized in the following table.

Note: This is only valid for  $V_{CC} > V_{THRESHLow}$ . Otherwise all diagnostic outputs will be tristated.

**Table 3-2. Status of the Diagnostic Outputs (Normal Operation)**

Device Status					Diagnostic Outputs			Comments
CPOK	OT1	OV	UV	SC	DG1	DG2	DG3	
0	X	X	X	X	–	1	–	Charge pump failure
X	1	X	X	X	–	–	1	Overtemperature prewarning
X	X	1	X	X	–	1	–	Overvoltage
X	X	X	1	X	–	1	–	Undervoltage
X	X	X	X	1	1	–	–	Short circuit

Note: X represents: no effect)  
 OT1: overtemperature warning  
 OV: overvoltage of PBAT  
 UV: undervoltage of PBAT  
 SC: short circuit  
 CPOK: charge pump OK

In order to differentiate between LIN and EN2 wake-up, DG1 output will be set to LOW or HIGH respectively. LOW indicates wake-up by LIN, HIGH indicates wake-up by EN2. DG1 output will be cleared by the first valid watchdog trigger after wake-up or by the first rising edge at one of the control inputs (IL1 to IL3 and IH1 to IH3) if the watchdog is disabled.

**Table 3-3. Indicating Wake-up Source**

Diagnostic Outputs			Wake-up Source
DG1	DG2	DG3	
1	–	–	EN2
0	–	–	LIN

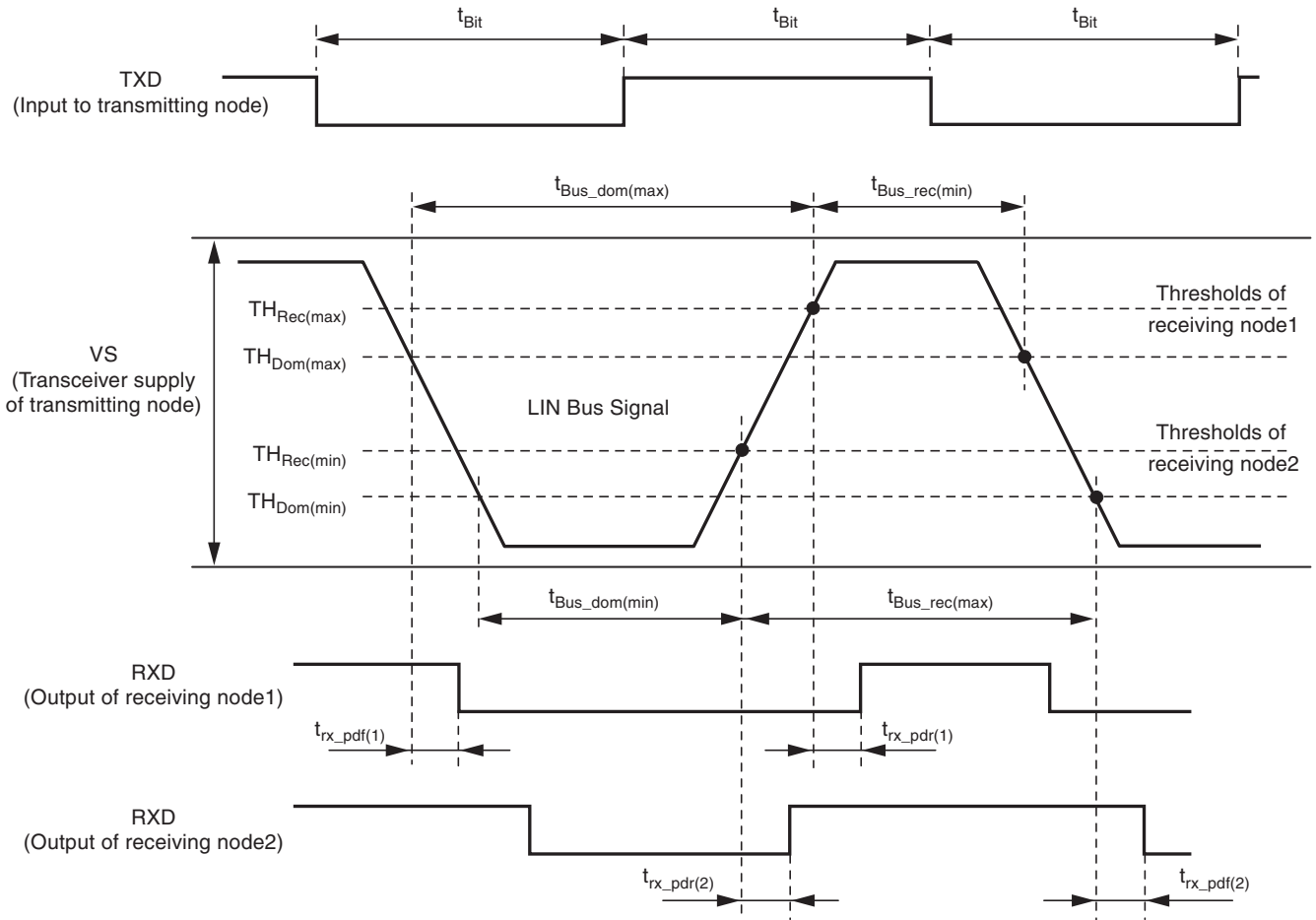
### 3.11 LIN Transceiver

Atmel® ATA6833 and Atmel ATA6834 include a fully integrated LIN transceiver complying with LIN specification 2.1 and SAEJ2602 2. The transceiver consists of a low-side driver with slew rate control, wave shaping, current limiting, and a high voltage comparator followed by a debouncing unit in the receiver.

During transmission, the data applied at pin TXD will be transferred to the bus driver to generate a bus signal on LIN pin. TXD input has an internal pull-up resistor.

To minimize the electromagnetic emission of the bus line, the bus driver has a built-in slew rate control and wave-shaping unit. The transmission will be aborted by a thermal shutdown or by a transition to Sleep Mode.

**Figure 3-7. Definition of Bus Timing Parameters**



The recessive BUS level is generated from the integrated 30kΩ pull-up resistor in series with an active diode. This diode protects against reverse currents on the bus line in case of a voltage difference between the bus line and VSUP ( $V_{BUS} > V_{SUP}$ ). No additional termination resistor is necessary to use the IC as a LIN slave. If this IC is used as a LIN master, the LIN pin is terminated by an external 1kΩ resistor in series with a diode to VBAT.

As PWM communication directly over the LIN transceiver in both directions is possible, there is no TXD timeout feature implemented in the LIN transceiver.

## 4. Absolute Maximum Ratings

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

All voltages are referenced to pin GND. (xxx) Values for the Atmel® ATA6834.

Parameters	Pin	Symbol	Min.	Max.	Unit
Input voltage	PGND	$V_{PGND}$	-0.3	+0.3	V
Negative input current	VBAT	IVBAT	-15		mA
Negative input current	PBAT	IPBAT	-20		mA
Supply voltage	VBAT	$V_{VBAT}$		+40	V
Supply voltage	PBAT	$V_{PBAT}$		+40	V
Logic output voltage	/RESET, DG1, DG2, DG3, RXD	$V_{/RESET}, V_{DG1}, V_{DG2}, V_{DG3}, V_{RXD}$	-0.3	$V_{VCC} + 0.3$	V
Logic input voltage	IL1-3, IH1-3, WD, WDD, EN1, TXD	$V_{IL1-3}, V_{IH1-3}, V_{WD}, V_{EN1}, V_{TXD}$	-0.3	$V_{VCC} + 0.3$	V
Output voltage	VINT, VCC	$V_{INT}, V_{VCC}$	-0.3	+5.5	V
Analog input voltage	RWD, CC	$V_{RWD}$	-0.3	$V_{VCC} + 0.3$	V
Digital input voltage	EN2	$V_{EN2}$	-0.3	$V_{VBAT} + 0.3$	V
Digital input voltage	VMODE	$V_{VMODE}$	-0.3	$V_{VINT} + 0.3$	V
Output voltage	VG	$V_{VG}$		+16	V
Input voltage	LIN	$V_{VLIN}$	-27	$V_{VBAT} + 2$	V
Output voltage	S1, S2, S3	$V_{S1}, V_{S2}, V_{S3}$	(-6)	+30	V
Output voltage	L1, L2, L3	$V_{L1}, V_{L2}, V_{L3}$	$V_{PGND} - 0.3$	$V_{VG} + 0.3$	V
Output voltage	H1, H2, L3	$V_{H1}, V_{H2}, V_{H3}$	$V_{S1, 2, 3} - 1$	$V_{S1, 2, 3} + 16$	V
Charge pump	CPLO1, 2	$V_{CPLO1}, V_{CPLO2}$		$V_{PBAT} + 0.3$	V
Charge pump	CPHI1, 2	$V_{CPHI1}, V_{CPHI2}$		$V_{CPOUT} + 0.3$	V
Output voltage	CPOUT	$V_{CPOUT}$		+40	V
Output voltage	VBATSW	$V_{VBATSW}$	-0.3	$V_{VBAT} + 0.3$	V
Storage temperature		$T_{Storage}$	-55	+150	°C
Reverse current	CPLOx, CPHIx, VG, CPOUT, Sx	$I_{CPLOx\_R}, I_{CPHIx\_R}, I_{VG\_R}, I_{CPOUT\_R}, I_{Sx\_R}$	-2		mA
	Lx, Hx	$I_{Lx\_R}, I_{Hx\_R}$	-1		mA

Note: Estimated values take  $T_J > 150^\circ\text{C}$  into account.

## 5. Thermal Resistance

Parameters	Symbol	Value	Unit
Thermal resistance junction to heat slug	$R_{thjc}$	< 5	K/W
Thermal resistance junction to ambient when heat slug is soldered to PCB	$R_{thja}$	25	K/W

## 6. Operating Range

The operating conditions define the limits for functional operation and parametric characteristics of the device. Functionality outside these limits is not implied unless otherwise stated explicitly. (xxx) Values for the Atmel® ATA6834

Parameters	Symbol	Min	Max	Unit
Operating supply voltage <sup>(1)</sup>	$V_{VBAT}$	5.5	$V_{THOV}^{(4)}$	V
Operating supply voltage <sup>(2)</sup>	$V_{VBAT}$	4.3	5.5	V
Operating supply voltage <sup>(3)</sup>	$V_{VBAT}$	$V_{THOV}$	40	V
Ambient temperature range	$T_A$	-40	+150	°C
Junction temperature range	$T_J$	-40	+150 (200)	°C

- Notes:
1. Full functionality
  2. Output drivers are switched off, extended range for parameters for voltage regulators
  3. Output drivers and charge pump are switched off

## 7. Noise and Surge Immunity, ESD and Latch-up

Parameters	Standard and Test Conditions	Value
Conducted interferences	ISO 7637-1	Level 4 <sup>(1)</sup>
Conducted disturbances	CISP25	Level 5
ESD according to IBEE LIN EMC - Pins LIN, PBAT, VBAT - Pin EN2 (33kΩ serial resistor)	Test specification 1.0 following IEC 61000-4-2	±6 kV ±5 kV
ESD HBM with 1.5kΩ/100pF	ESD- STM5.1-2001 JESD22-A114E 2007 CEI/IEC 60749-26: 2006 AEC-Q100-002-Ref_D	±2 kV
ESD HBM with 1.5kΩ/100pF Pins EN2, LIN, PBAT, VBAT against GND	ESD- STM5.1-2001 JESD22-A114E 2007 CEI/IEC 60749-26: 2006 AEC-Q100-002-Ref_D	±8 kV
ESD CDM (field induced method)	ESD STM5.3.1 - 1999	±500V

Note: 1. Test pulse 5:  $V_{bat\ max} = 40V$

Static latch-up tested according to AEC-Q100-004 and JESD78.

- 3 to 6 samples, 0 failures
- Electrical post-stress testing at room temperature

In test, the voltage at the pins VBAT, LIN, CP, VBATSW, Hx, and Sx must not exceed 45V when not able to drive the specified current.

## 8. Electrical Characteristics

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel® ATA6834.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
<b>1 Power Supply and Supervisor Functions</b>									
1.1	Current consumption $V_{VBAT}$	$V_{VBAT} = 13.5V^{(1)}$	VBAT	$I_{VBAT}$			7	mA	A
1.3	Current consumption $V_{VBAT}$ in Standby Mode	$V_{VBAT} = 13.5V$	VBAT	$I_{VBAT}$			65	$\mu A$	A
1.4	Current consumption $V_{VBAT}$ in Standby Mode	$V_{PBAT} = 13.5V$	PBAT	$I_{VPBAT}$	9.0		20.0	$\mu A$	A
1.5	Internal power supply	$V_{VBAT} > 7V$	VINT	$V_{VINT}$	4.7	5.0	5.3	V	A
1.6	Overvoltage lock-out threshold		PBAT	$V_{THOVLO}$	20.0		23.0	V	A
1.7	Overvoltage hysteresis		PBAT	$V_{TOVhys}$	1.0		1.5	V	A
1.8	Undervoltage lock-out threshold		PBAT	$V_{THUVRC}$	4.75		5.25	V	A
1.9	Undervoltage threshold hysteresis		PBAT	$V_{TUVhys}$	0.2		0.4	V	A
1.10	$R_{DSON}$ VBAT-Switch switch	$V_{VBAT} = 13.5V$ , $I_{VBATSW} = -15mA$	VBATSW	$R_{ON\_VBATSW}$			100	$\Omega$	A
1.11	Thermal prewarning set			$T_{JPW\ set}$	120 (170)	145 (195)	170 (220)	$^{\circ}C$	B
1.12	Thermal prewarning reset			$T_{JPW\ reset}$	105 (155)	130 (180)	155 (205)	$^{\circ}C$	B
1.13	Thermal prewarning hysteresis			$\Delta T_{JPW}$		15		$^{\circ}C$	B
1.14	Thermal shutdown off			$T_{J\ switch\ off}$	150 (200)	175 (225)	200 (250)	$^{\circ}C$	B
1.15	Thermal shutdown on			$T_{J\ switch\ on}$	135 (185)	160 (210)	185 (235)	$^{\circ}C$	B
1.16	Thermal shutdown hysteresis			$\Delta T_{J\ switch\ off}$		15		$^{\circ}C$	B
1.17	Ratio thermal shutdown off/thermal prewarning set			$\frac{T_{J\ switch\ off}}{T_{JPW\ set}}$	1.05	1.15			B
1.18	Ratio thermal shutdown on/thermal prewarning reset			$\frac{T_{J\ switch\ on}}{T_{JPW\ reset}}$	1.05	1.15			B
<b>2 5V/3.3V Regulator</b>									
2.1	Regulated output voltage	$V_{MODE} = V_{INT}, 7V < V_{BAT} < 40V$ $V_{MODE} = GND, 5.5V < V_{BAT} < 40V$ $I_{Load} = 0$ to 100mA	$V_{CC}$	$V_{VCC}$	4.85 3.20		5.15 3.40	V	A
2.2	Regulated output voltage	$V_{MODE} = V_{INT}, 7V < V_{BAT} < 40V$ $V_{MODE} = GND, 5.5V < V_{BAT} < 40V$ $I_{Load} = 0$ to 80mA $150^{\circ}C < T_J < 200^{\circ}C$	$V_{CC}$	$V_{VCC}$	4.85 3.20		5.15 3.40	V	A
2.3	Regulated output voltage	$V_{MODE} = V_{INT}, 5.5V < V_{BAT} < 7V$ $V_{MODE} = GND, 5V < V_{BAT} < 5.5V$ $I_{Load} = 0$ to 60mA	$V_{CC}$	$V_{VCC}$	4.50 2.97		5.15 3.40	V	A

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

## 8. Electrical Characteristics (Continued)

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel® ATA6834.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
2.4	Regulated output voltage	$V_{MODE} = V_{INT}$ , $5.5V < V_{BAT} < 7V$ $V_{MODE} = GND$ , $5V < V_{BAT} < 5.5V$ $I_{Load} = 0$ to 50mA $150^{\circ}C < T_J < 200^{\circ}C$	V <sub>CC</sub>	V <sub>VCC</sub>	4.50 2.97		5.15 3.40	V	A
2.5	Line regulation	$V_{MODE} = V_{INT}$ , $7V < V_{BAT} < 40V$ $V_{MODE} = GND$ , $5.5V < V_{BAT} < 40V$ $I_{Load} = 50mA$ , $-40^{\circ}C < T_J < 150^{\circ}C$	V <sub>CC</sub>				50 50	mV	A
2.6	Load regulation	$V_{MODE} = V_{INT}$ , $V_{BAT} > 7V$ $V_{MODE} = GND$ , $V_{BAT} > 5.5V$ $I_{Load} = 0$ to 100mA $I_{Load} = 0$ to 80mA, $150^{\circ}C < T_J < 200^{\circ}C$	V <sub>CC</sub>				50 50	mV	A
2.7	Output current limit	$V_{MODE} = V_{INT}$ , $V_{BAT} > 7V$ $V_{MODE} = GND$ , $V_{BAT} > 5.5V$ $I_{Load} @$ RESET	V <sub>CC</sub>	I <sub>OS1</sub>	100 100		360 360	mA	A
2.8	Output current limit	$V_{MODE} = V_{INT}$ , $V_{BAT} > 7V$ $V_{MODE} = GND$ , $V_{BAT} > 5.5V$ $I_{Load} @$ RESET, $150^{\circ}C < T_J < 200^{\circ}C$	V <sub>CC</sub>	I <sub>OS1</sub>	70 70		360 360	mA	C
2.12	HIGH threshold V <sub>MODE</sub>			V <sub>VMODE H</sub>			4.0	V	A
2.13	LOW threshold V <sub>MODE</sub>			V <sub>VMODE L</sub>	0.7			V	A
<b>3 Reset and Watchdog</b>									
3.1	V <sub>CC</sub> threshold voltage level for /RESET	V <sub>MODE</sub> = V <sub>INT</sub> (V <sub>MODE</sub> = GND)		V <sub>thRESHLow</sub>	3.8 2.5		4.2 2.8	V	A B
3.2	Hysteresis	V <sub>MODE</sub> = V <sub>INT</sub> (V <sub>MODE</sub> = GND)		HYS <sub>REStH</sub>	0.2 0.13		0.6 0.4	V	A B
3.3	Length of pulse at /RESET			t <sub>res</sub>	8		12	ms	A
3.4	Length of short pulse at /RESET			t <sub>resshort</sub>	1.6		2.4	ms	A
3.5	Wait for the first WD trigger			t <sub>d</sub>	400		600	ms	A
3.6	Time for V <sub>CC</sub> < V <sub>thRESL</sub> before activating /RESET			t <sub>delayRESL</sub>			2	μs	C
3.8	Watchdog oscillator period	R <sub>RWD</sub> = 33kΩ ± 1%		T <sub>OSC</sub>	11.09		13.55	μs	A
3.12	Close window			t1		980 × T <sub>OSC</sub>			A
3.13	Open window			t2		780 × T <sub>OSC</sub>			A
3.14	Output low-level at pin /RESET	I <sub>OLRES</sub> = 1mA		V <sub>OLRES</sub>			0.4	V	A
3.15	Internal pull-up resistor at pin /RESET			R <sub>PURES</sub>	5	10	15	kΩ	D

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter



## 8. Electrical Characteristics (Continued)

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel® ATA6834.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
4	LIN Transceiver								
4.1	Low-level output current	Normal mode; $V_{LIN} = 0V$ , $V_{RXD} = 0.4V$		$I_{L_{RXD}}$	2			mA	D
4.2	High-level output current	Normal mode; $V_{LIN} = V_{BAT}$ $V_{RXD} = V_{CC} - 0.4V$		$I_{H_{RXD}}$			-2	mA	D
4.3	Driver recessive output voltage	$V_{TXD} = V_{CC}$ ; $I_{LIN} = 0mA$		$V_{BUS_{rec}}$	$0.9 \times V_{BAT}$			V	A
4.4	Driver dominant voltage $V_{BUS_{dom\_DRV\_LoSUP}}$	$V_{VBAT} = 7.3V$ $R_{load} = 500\Omega$		$V_{LoSUP}$			1.2	V	A
4.5	Driver dominant voltage $V_{BUS_{dom\_DRV\_HiSUP}}$	$V_{VBAT} = 18V$ $R_{load} = 500\Omega$		$V_{HiSUP}$			2	V	A
4.6	Driver dominant voltage $V_{BUS_{dom\_DRV\_LoSUP}}$	$V_{VBAT} = 7.3V$ $R_{load} = 1000\Omega$		$V_{LoSUP\_1k}$	0.6			V	A
4.7	Driver dominant voltage $V_{BUS_{dom\_DRV\_HiSUP}}$	$V_{VBAT} = 18V$ $R_{load} = 1000\Omega$		$V_{HiSUP\_1k}$	0.8			V	A
4.8	Pull up resistor to VS	serial diode required		$R_{LIN}$	20		47	k $\Omega$	A
4.9	Current limitation	$V_{BUS} = V_{BAT\_max}$		$I_{BUS\_LIM}$	50		200	mA	A
4.10	Input leakage current at the receiver including pull-up resistor as specified	Input leakage current driver off $V_{BUS} = 0V$ $V_{BAT} = 12V$		$I_{BUS\_PAS\_dom}$	-1			mA	A
4.11	Leakage current LIN recessive	Driver off $8V < V_{BAT} < 18V$ $8V < V_{BUS} < 18V$ $V_{BUS} = V_{BAT}$		$I_{BUS\_PAS\_rec}$			20	$\mu A$	A
4.12	Leakage current at ground loss Control unit disconnected from ground Loss of local ground must not affect communication in the residual network	$GND_{Device} = VS$ $V_{BAT} = 12V$ $0V < V_{BUS} < 18V$		$I_{BUS\_NO\_gnd}$	-1		+1	mA	A
4.13	Node has to sustain the current that can flow under this condition. Bus must remain operational under this condition	$V_{BAT}$ disconnected $V_{SUP\_Device} = GND$ $0V < V_{BUS} < 18V$		$I_{BUS}$			100	$\mu A$	A
4.14	Center of receiver threshold	$V_{BUS\_CNT} = (V_{th\_dom} + V_{th\_rec})/2$		$V_{BUS\_CNT}$	$0.475 \times V_{VBAT}$	$0.5 \times V_{VBAT}$	$0.525 \times V_{VBAT}$	V	A
4.15	Receiver dominant state	$V_{EN} = 5V$		$V_{BUS_{dom}}$			$0.4 \times V_{VBAT}$	V	A
4.16	Receiver recessive state	$V_{EN} = 5V$		$V_{BUS_{rec}}$	$0.6 \times V_{VBAT}$			V	A
4.17	Receiver input hysteresis	$V_{HYS} = V_{th\_rec} - V_{th\_dom}$		$V_{BUS_{hys}}$			$0.175 \times V_{VBAT}$	V	A

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

## 8. Electrical Characteristics (Continued)

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel® ATA6834.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
4.18	Duty cycle 1	$7V < V_{VBAT} < 18V$ $TH_{rec(max)} = 0.744 \times V_{VBAT}$ $TH_{Dom(max)} = 0.581 \times V_{VBAT}$ $t_{Bit} = 50\mu s$ $D1 = t_{Bus\_rec(min)} / (2 \times t_{Bit})$ Load1: 1nF + 1 kΩ Load2: 10nF + 500Ω		D1	0.396				A
4.19	Duty cycle 2	$7V < V_{VBAT} < 18V$ $TH_{rec(min)} = 0.422 \times V_{VBAT}$ $TH_{Dom(min)} = 0.284 \times V_{VBAT}$ $t_{Bit} = 50\mu s$ $D2 = t_{Bus\_rec(max)} / (2 \times t_{Bit})$ Load1: 1nF + 1 kΩ Load2: 10nF + 500Ω		D2			0.581		A
4.20	Duty cycle 3	$7V < V_{VBAT} < 18V$ $TH_{rec(max)} = 0.778 \times V_{VBAT}$ $TH_{Dom(max)} = 0.616 \times V_{VBAT}$ $t_{Bit} = 96\mu s$ $D3 = t_{Bus\_rec(min)} / (2 \times t_{Bit})$ Load1: 1nF + 1 kΩ Load2: 10nF + 500Ω		D3	0.417				A
4.21	Duty cycle 4	$7V < V_{VBAT} < 18V$ $TH_{rec(max)} = 0.389 \times V_{VBAT}$ $TH_{Dom(max)} = 0.251 \times V_{VBAT}$ $t_{Bit} = 96\mu s$ $D4 = t_{Bus\_rec(min)} / (2 \times t_{Bit})$ Load1: 1nF + 1kΩ Load2: 10nF + 500Ω		D4			0.590		A
4.22	Receiver propagation delay	$7V < V_{VBAT} < 18V$ $t_{rec\_pd} = \max(t_{rx\_pdr}, t_{rx\_pdf})$		$t_{rx\_pd}$			6	μs	A
4.23	Symmetry of receiver propagation delay rising edge minus falling edge	$7V < V_{VBAT} < 18V$ $t_{rx\_sym} = t_{rx\_pdr} - t_{rx\_pdf}$		$t_{rx\_sym}$	-2		+2	μs	
4.24	Dominant time for wake-up via LIN-bus	$V_{LIN} = 0V$		$T_{BUS}$	30	90	150	μs	A
4.25	Capacitance on LIN pin to GND			$C_{LIN}$			10	pF	D
<b>5</b>	<b>Control Inputs EN1, IL1-3, IH1-3, WD, TX, WDD</b>								
5.1	Input low-level threshold			$V_{IL}$			$0.3 \times V_{VCC}$	V	A
5.2	Input high-level threshold			$V_{IH}$	$0.7 \times V_{VCC}$			V	A
5.3	Hysteresis			HYS	0.3				C
5.4	Pull-down resistor	EN1, IL1-3, IH1-3, WD		$R_{PD}$	25	50	100	kΩ	A
5.5	Pull-up resistor	TXD, WDD		$R_{PU}$	25	50	100	kΩ	A
5.7	Debounce time EN1			$t_{gotosleep}$	9	10	11	μs	A

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

## 8. Electrical Characteristics (Continued)

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel® ATA6834.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
<b>6 Charge Pump</b>									
6.1	Charge pump voltage	$V_{VBAT} > 7V$ $I_{LoadCPOUT} = 0A$ $I_{LoadVG} = 0A$ $C_{CP1,2} = 47nF$ $C_{CPOUT} = 220nF$	CPOUT	$V_{CPOUT}$	$V_{VBAT} + 11V$		$V_{VBAT} + 18$	V	A
6.2	Charge pump voltage	$V_{VBAT} > 7V$ $I_{LoadCPOUT} = 7.5mA,$ $I_{LoadVG} = 0A$ $C_{CP1,2} = 47nF$ $C_{CPOUT} = 220nF$	CPOUT	$V_{CPOUT}$	$V_{VBAT} + 10V$			V	A
6.3	Period charge pump oscillator			$T_{CP}$		2.5		$\mu s$	B
6.4	Charge pump output voltage for active drivers		CPOUT	$V_{CPCPGOOD}$	5.25		8.0	V	A
<b>7 VG Regulator</b>									
7.1	VG Regulator Output Voltage	$V_{BAT} = 13.5V$ $V_{CPOUT} = 20V$ $I_{LoadVG} = 7.5mA$	VG	$V_{VG}$	11	12.5	14	V	A
7.2	VG Regulator Line Regulation	$V_{BAT} = 13.5V$ $V_{CPOUT1} = 20V, V_{CPOUT2} = 35V$ $I_{LoadVG} = 7.5mA$	VG	$\Delta V_{VG\_Line}$			100	mV	A
7.3	VG Regulator Load Regulation	$V_{BAT} = 13.5V$ $V_{CPOUT} = 25V$ $I_{LoadVG1} = 1mA, I_{LoadVG2} = 60mA$	VG	$\Delta V_{VG\_Load}$			100	mV	A
<b>8 H-bridge Driver</b>									
8.1	Low-side driver HIGH output voltage			$V_{LxH}$			$V_{VG}$	V	D
8.2	ON-resistance of sink stage of pins Lx	$I_{LX} = 100mA$		$R_{DSON\_LxL}$			20	$\Omega$	A
8.3	ON-resistance of source stage of pins Lx	$I_{LX} = 100mA$		$R_{DSON\_LxH}$			20	$\Omega$	A
8.4	Output peak current at pins Lx switched to LOW	$V_{Lx} = 3V$		$I_{LxL}$	-100			mA	D
8.5	Output peak current at pins Lx switched to HIGH	$V_{Lx} = 3V$		$I_{LxH}$			100	mA	D
8.6	Sink resistance between Lx and GND		Lx to GND	$R_{Lxsink}$	45	75	115	k $\Omega$	A
8.7	ON-resistance of sink stage of pins Hx	$V_{Sx} = 0V$		$R_{DSON\_HxL}$			20	$\Omega$	A
8.8	ON-resistance of source stage of pins Hx	$V_{Sx} = V_{VBAT}$ $I_{Hx} = 100mA$		$R_{DSON\_HxH}$			20	$\Omega$	A
8.9	Output peak current at pins Hx (switched from low to high)	$V_{Hx} - V_{Sx} = 0V;$ $V_{VBAT} = 7V - 20V$ $C = 10nF$ $R = 1\Omega$		$I_{HxH}$			-200	mA	C

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

## 8. Electrical Characteristics (Continued)

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel® ATA6834.

No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
8.10	Output peak current at pins Hx (switched from high to low)	$V_{Hx} - V_{Sx} = 10V$ ; $V_{VBAT} = 7 - 20V$ $C = 10nF$ $R = 1\Omega$		$I_{HxL}$	200			mA	C
8.11	Output peak current at pins Lx (switched from low to high)	$V_{Lx} = 0V$ ; $V_{VBAT} = 7 - 20V$ $C = 10nF$ $R = 1\Omega$		$I_{LxH}$			-200	mA	C
8.12	Output peak current at pins Lx (switched from high to low)	$V_{Lx} = 10V$ ; $V_{VBAT} = 7 - 20V$ $C = 10nF$ $R = 1\Omega$		$I_{LxL}$	200			mA	C
8.13	Output voltage low level pins Hx	$V_{Sx} = 0V$ $I_{Hx} = 1mA$		$V_{HxL}$			0.3	V	A
8.14	Output voltage high level pins Hx	$I_{Hx} = -100\mu A$		$V_{HxHstat}$	$V_{VCPOUT} - 1V$		$V_{VCPOUT}$	V	A
8.15	Sink resistance between Hx and Sx			$R_{Hxsink}$	45	75	115	k $\Omega$	A
8.16	Sink resistance between Sx and GND		Sx to GND	$R_{Sxsink}$		1		M $\Omega$	D
Dynamic Parameters									
8.17	Propagation delay time, low-side driver from high to low			$t_{LxHL}$			0.9	$\mu s$	A
8.18	Propagation delay time, low-side driver from low to high			$t_{LxLH}$			0.9	$\mu s$	A
8.19	Fall time low-side driver	$V_{VBAT} = 13.5V$ $C_{Gx} = 5nF$		$t_{Lxf}$			0.3	$\mu s$	A
8.20	Rise time low-side driver	$V_{VBAT} = 13.5V$ $C_{Gx} = 5nF$		$t_{Lxr}$			0.3	$\mu s$	A
8.21	Propagation delay time, high-side driver from high to low			$t_{HxHL}$			0.9	$\mu s$	A
8.22	Propagation delay time, high-side driver from low to high			$t_{HxLH}$			0.9	$\mu s$	A
8.23	Fall time high-side driver	$V_{VBAT} = 13.5V$ , $C_{Gx} = 5nF$		$t_{Hxf}$			0.3	$\mu s$	A
8.24	Rise time high-side driver	$V_{VBAT} = 13.5V$ , $C_{Gx} = 5nF$		$t_{Hxr}$			0.3	$\mu s$	A
8.25	Short circuit detection voltage			$V_{SC}$	3.5	4	4.5	V	A
8.26	Short circuit blanking time			$t_{SC}$	5.4	6	6.6	$\mu s$	A

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

## 8. Electrical Characteristics (Continued)

All parameters given are valid for  $5.5V \leq V_{VBAT} \leq 18V$  and for  $-40^{\circ}C \leq T_J \leq 150^{\circ}C$  ( $200^{\circ}C$ ) unless stated otherwise. All values refer to PIN GND. (xxx) Values for the Atmel<sup>®</sup> ATA6834.

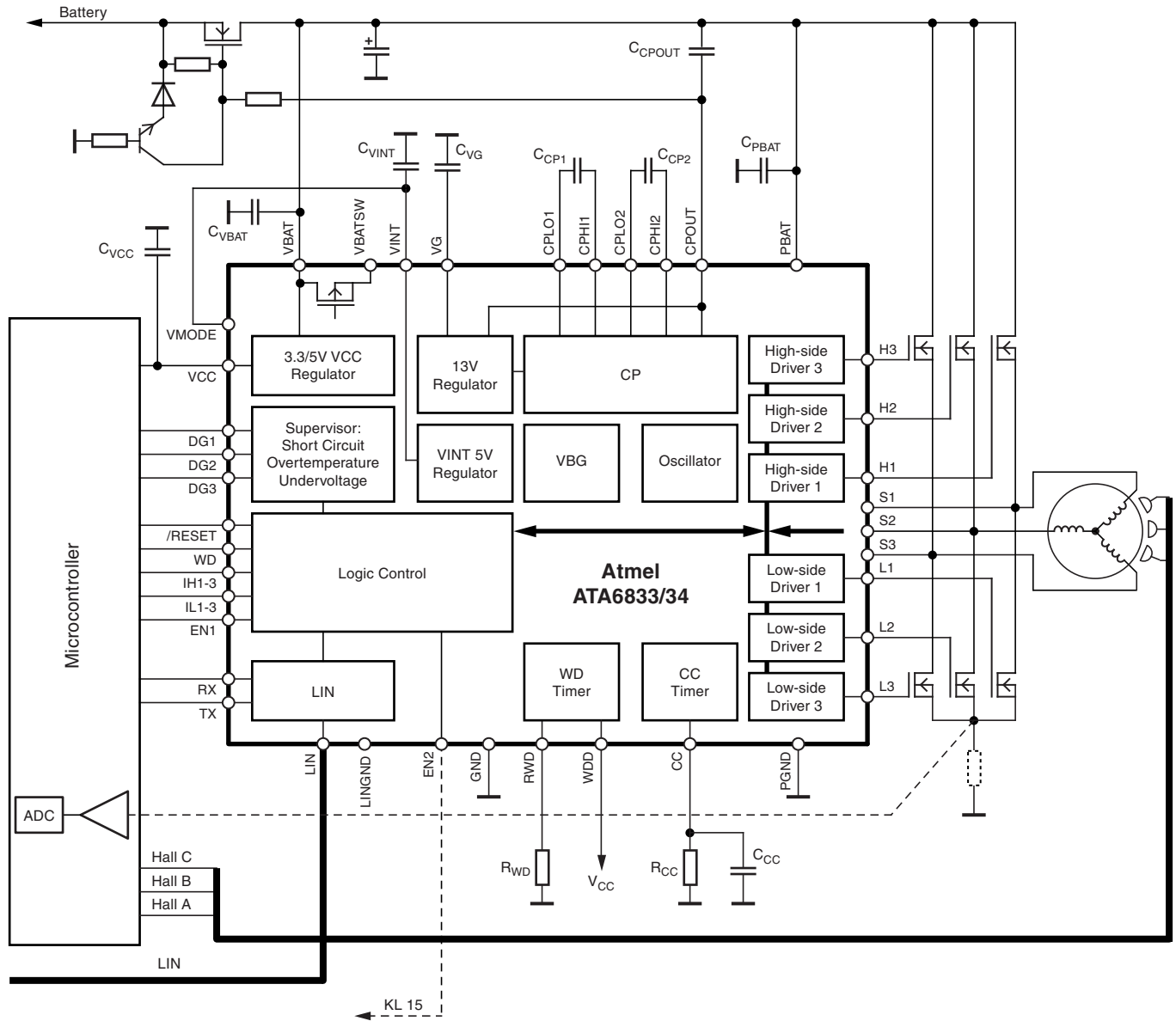
No.	Parameters	Test Conditions	Pin	Symbol	Min.	Typ.	Max.	Unit	Type*
	Cross Conduction Timer								
8.27	Cross conduction time constant			$K_{CC}$	0.345	0.405	0.465		B
<b>9</b>	<b>Input EN2</b>								
9.1	Input low level threshold			$V_{IL}$	2.3		3.6	V	A
9.2	Input high level threshold			$V_{IH}$	2.8		4.0	V	A
9.3	Hysteresis			HYS		0.47		V	C
9.4	Pull-down resistor			$R_{PD}$	50	100	200	k $\Omega$	A
9.5	Debounce time			$t_{db}$	10	20	25	$\mu s$	A
<b>10</b>	<b>Diagnostic Outputs DG1, DG2, DG3</b>								
10.1	Low level output current	$V_{DG} = 0.4V$		IL	2			mA	A
10.2	High level output current	$V_{DG} = VCC - 0.4V$		IH			-2	mA	A

\*) Type means: A = 100% tested, B = 100% correlation tested, C = Characterized on samples, D = Design parameter

## 9. Application

This section describes the principal application for which the ATA6833/ATA6834 was designed.

Figure 9-1. Typical Application



**Table 9-1. Typical External Components**

Component	Function	Min.	Typical	Max.
CVINT	Blocking capacitor at VINT	100nF	220nF/10V	470nF
C <sub>VCC</sub>	Blocking capacitor at VCC	1.5μF		10μF
ESL (C <sub>VCC</sub> )	Serial inductance to C <sub>VCC</sub> including PCB	1nH		20nH
ESR (C <sub>VCC</sub> )	Serial resistance to C <sub>VCC</sub> including PCB	2Ω		15Ω
CVG	Blocking capacitor at VG	220nF	470nF, 25V	1μF
CCP1	Charge pump shuffle capacitor	47nF	220nF/25V	470nF
CCP2	Charge pump shuffle capacitor	47nF	220nF/25V	470nF
CCPOUT	Charge pump reservoir capacitor	220nF	470nF, 25V	1μF
RRWD	Resistor defining internal bias currents for watchdog oscillator	10kΩ	33kΩ	91kΩ
RCC	Cross conduction time definition resistor	5kΩ	10kΩ	
CCC	Cross conduction time definition capacitor		330pF	5nF
C <sub>VBAT</sub>	Blocking capacitor VBAT		100nF	
C <sub>PBAT</sub>	Blocking capacitor PBAT		100nF	

## 10. Ordering Information

Extended Type Number	Package	Remarks
ATA6833-PLQW	QFN48	
ATA6834-PLQW	QFN48	

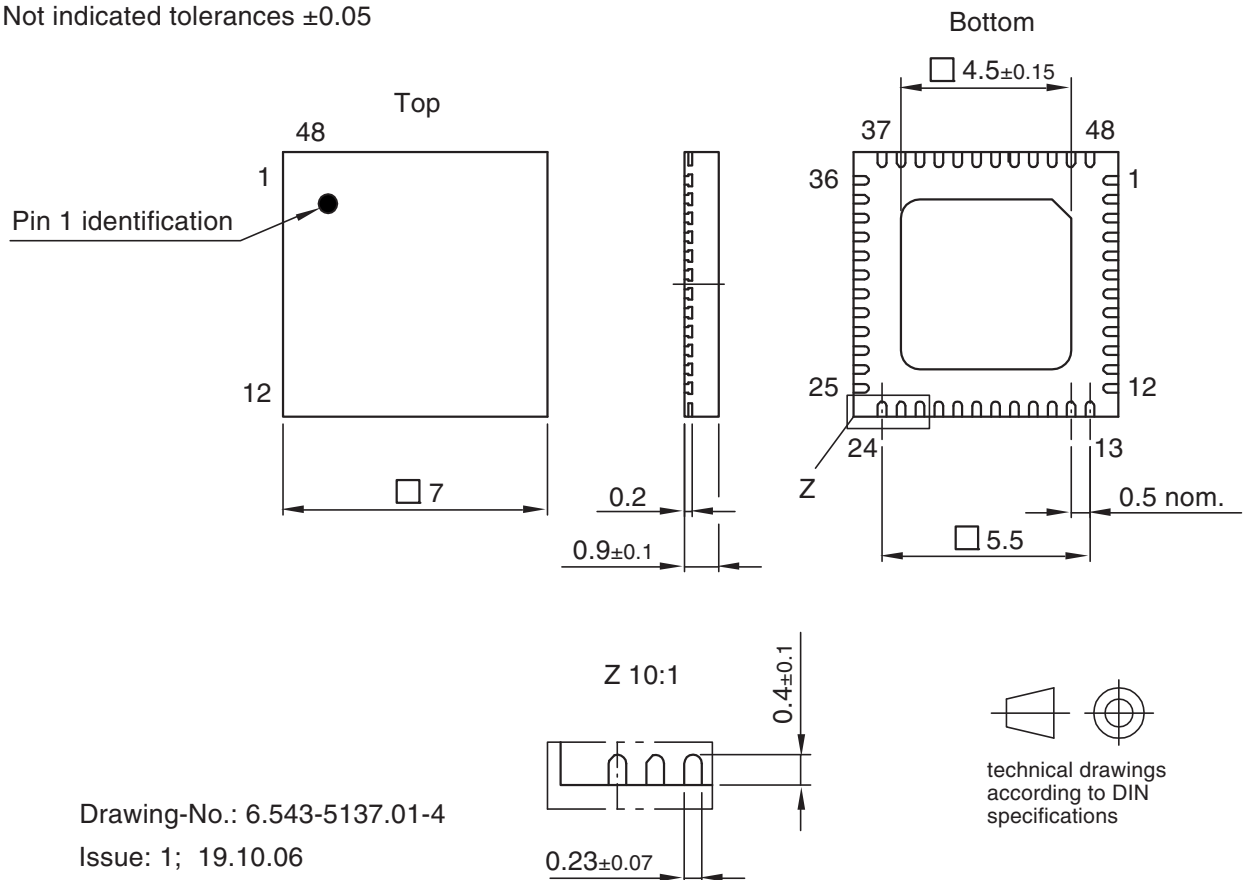
## 11. Package Information

Package: VQFN\_7 x 7\_48L

Exposed pad 4.5 x 4.5

Dimensions in mm

Not indicated tolerances  $\pm 0.05$



Drawing-No.: 6.543-5137.01-4

Issue: 1; 19.10.06



## 12. Revision History

Please note that the following page numbers referred to in this section refer to the specific revision mentioned, not to this document.

Revision No.	History
9122I-AUTO-03/12	<ul style="list-style-type: none"><li>• Section 4 “Absolute Maximum Ratings” on page 13 changed</li></ul>
9122H-AUTO-08/11	<ul style="list-style-type: none"><li>• Page 9: Example test changed and text under figure 3-5 added</li><li>• El. Characteristics table, page 17, row 3.8 changed</li><li>• El. Characteristics table, page 22, row 8.27 changed</li><li>• Page 23: Typ. Application drawing changed</li><li>• Page 24: Table 9-1 rows <math>C_{VBAT}</math> and <math>C_{PBAT}</math> added</li></ul>
9122G-AUTO-10/10	<ul style="list-style-type: none"><li>• Section 8 “Electrical Characteristics” numbers 8.23 and 8.24 on page 21 changed</li></ul>
9122E-AUTO-02/10	<ul style="list-style-type: none"><li>• Page 12: Note added</li><li>• Page 21: El. Char. Table: rows 8.11 and 8.12 changed</li></ul>
9122D-AUTO-09/09	<ul style="list-style-type: none"><li>• Section 4 “Absolute Maximum Ratings” on page 14 changed</li><li>• Section 8 “Electrical Characteristics” numbers 2.7, 2.8, 3.1, 3.2, 4.20, 4.21, 6.4, 8.19, 8.20, 8.23, 8.24 and 8.27 on pages 16 to 22 changed</li></ul>
9122C-AUTO-04/09	<ul style="list-style-type: none"><li>• Features on page 1 changed</li><li>• Section 7 “Noise and Surge Immunity, ESD and Latch-up” on page 15 changed</li></ul>
9122B-AUTO-10/08	<ul style="list-style-type: none"><li>• Put datasheet in the latest template</li><li>• Section 8 “Electrical Characteristics” on pages 15 to 21 changed</li></ul>



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