

MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

General Description

The MAX5703/MAX5704/MAX5705 single-channel, low-power, 8-/10-/12-bit, voltage-output digital-to-analog converters (DACs) include output buffers and an internal reference that is selectable to be 2.048V, 2.500V, or 4.096V. The MAX5703/MAX5704/MAX5705 accept a wide supply voltage range of 2.7V to 5.5V with extremely low power (< 1mW) consumption to accommodate most low-voltage applications. A precision external reference input allows rail-to-rail operation and presents a 100k Ω (typ) load to an external reference.

The MAX5703/MAX5704/MAX5705 have a 50MHz, 3-wire SPI/QSPI™/MICROWIRE®/DSP-compatible serial interface. The DAC output is buffered and has a low supply current of 155 μ A (typical at 3V) and a low offset error of ± 0.5 mV (typical). On power-up, the MAX5703/MAX5704/MAX5705 reset the DAC outputs to zero, providing additional safety for applications that drive valves or other transducers which need to be off on power-up. The internal reference is initially powered down to allow use of an external reference.

The MAX5703/MAX5704/MAX5705 include a user-configurable active-low asynchronous input, \overline{AUX} for additional flexibility. This input can be programmed to asynchronously clear (\overline{CLR}) or temporarily gate (\overline{GATE}) the DAC output to a user-programmable value. A dedicated active-low asynchronous \overline{LDAC} input is also included. This allows simultaneous output updates of multiple devices.

The MAX5703/MAX5704/MAX5705 are available in 10-pin TDFN/ μ MAX® packages and are specified over the -40°C to +125°C temperature range.

Applications

- Programmable Voltage and Current Sources
- Gain and Offset Adjustment
- Automatic Tuning and Optical Control
- Power Amplifier Control and Biasing
- Process Control and Servo Loops
- Portable Instrumentation
- Data Acquisition

QSPI is a trademark of Motorola, Inc.

MICROWIRE is a registered trademark of National Semiconductor Corp.

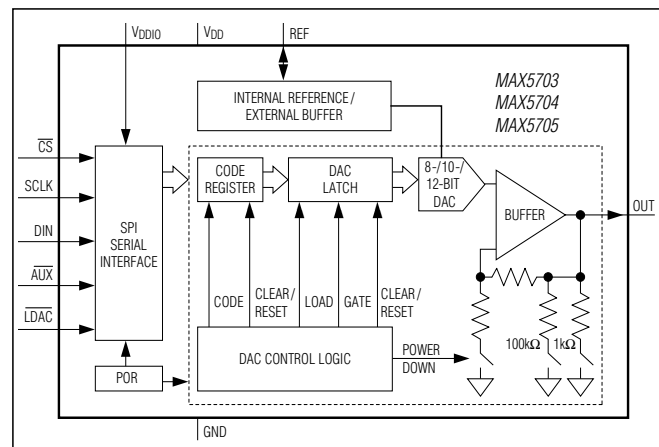
μ MAX is a registered trademark of Maxim Integrated Products, Inc.

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maximintegrated.com.

Benefits and Features

- ◆ **Single High-Accuracy DAC Channel**
 - ◇ 12-Bit Accuracy Without Adjustments
 - ◇ ± 1 LSB INL Buffered Voltage Output
 - ◇ Monotonic Over All Operating Conditions
- ◆ **Three Precision Selectable Internal References**
 - ◇ 2.048V, 2.500V, or 4.096V
- ◆ **Internal Output Buffer**
 - ◇ Rail-to-Rail Operation with External Reference
 - ◇ 6.3 μ s Settling Time
 - ◇ Output Directly Drives 2k Ω Loads
- ◆ **Small, 10-Pin, 2mm x 3mm TDFN and 3mm x 5mm μ MAX Packages**
- ◆ **Wide 2.7V to 5.5V Supply Range**
- ◆ **Flexible 1.8V to 5.5V V_{DDIO}**
- ◆ **50MHz, 3-Wire, SPI/QSPI/MICROWIRE/DSP-Compatible Serial Interface**
- ◆ **Power-On-Reset to Zero-Scale DAC Output**
- ◆ **User-Configurable Asynchronous I/O Functions: \overline{CLR} , \overline{LDAC} , \overline{GATE}**
- ◆ **Three Software-Selectable Power-Down Output Impedances: 1k Ω , 100k Ω , or High Impedance**
- ◆ **Low 155 μ A DAC Supply Current at 3V**

Functional Diagram



Ordering Information appears at end of data sheet.

For related parts and recommended products to use with this part, refer to: www.maximintegrated.com/MAX5703.related

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ABSOLUTE MAXIMUM RATINGS

V_{DD} to GND.....	-0.3V to +6V	Maximum Continuous Current into Any Pin	± 50 mA
V_{DDIO} to GND	-0.3V to +6V	Operating Temperature Range	-40°C to +125°C
OUT, REF to GND	-0.3V to lower of ($V_{DD} + 0.3$ V) and +6V	Storage Temperature Range.....	-65°C to +150°C
\overline{CS} , SCLK, DIN, AUX, LDAC to GND.....	-0.3V to +6V	Lead Temperature (soldering, 10s)	+300°C
Continuous Power Dissipation ($T_A = +70^\circ\text{C}$)		Soldering Temperature (reflow)	+260°C
TDFN (derate 14.9mW/°C above +70°C).....	1188.7mW		
μ MAX (derate 8.8mW/°C above +70°C)	707.3mW		

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

PACKAGE THERMAL CHARACTERISTICS (Note 1)

TDFN	
Junction-to-Ambient Thermal Resistance (θ_{JA})	67.3°C/W
μ MAX	
Junction-to-Ambient Thermal Resistance (θ_{JA})	113.1°C/W
Junction-to-Ambient Thermal Resistance (θ_{JC}).....	36°C/W

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maximintegrated.com/thermal-tutorial.

ELECTRICAL CHARACTERISTICS

($V_{DD} = 2.7$ V to 5.5V, $V_{DDIO} = 1.8$ V to 5.5V, $V_{GND} = 0$ V, $C_L = 200$ pF, $R_L = 2$ k Ω , $T_A = -40^\circ\text{C}$ to +125°C, unless otherwise noted.) (Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
DC PERFORMANCE (Note 3)						
Resolution and Monotonicity	N	MAX5703	8			Bits
		MAX5704	10			
		MAX5705	12			
Integral Nonlinearity (Note 4)	INL	MAX5703, 8 bits	-0.25	± 0.05	+0.25	LSB
		MAX5704, 10 bits	-0.5	± 0.2	+0.5	
		MAX5705, 12 bits	-1	± 0.5	+1	
Differential Nonlinearity (Note 4)	DNL	MAX5703, 8 bits	-0.25	± 0.05	+0.25	LSB
		MAX5704, 10 bits	-0.5	± 0.1	+0.5	
		MAX5705, 12 bits	-1	± 0.2	+1	
Offset Error (Note 5)	OE		-5	± 0.5	+5	mV
Offset Error Drift				± 10		$\mu\text{V}/^\circ\text{C}$
Gain Error (Note 5)	GE		-1.0	± 0.1	+1.0	%FS
Gain Temperature Coefficient		With respect to V_{REF}		± 2.5		ppm of FS/°C
Zero-Scale Error			0		+10	mV
Full-Scale Error		With respect to V_{REF}	-0.5		+0.5	%FS

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = 2.7V$ to $5.5V$, $V_{DDIO} = 1.8V$ to $5.5V$, $V_{GND} = 0V$, $C_L = 200pF$, $R_L = 2k\Omega$, $T_A = -40^\circ C$ to $+125^\circ C$, unless otherwise noted.)
(Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
DAC OUTPUT CHARACTERISTICS							
Output Voltage Range (Note 6)		No load		0		V_{DD}	V
		2k Ω load to GND		0		$V_{DD} - 0.2$	
		2k Ω load to V_{DD}		0.2		V_{DD}	
Load Regulation		$V_{OUT} = V_{FS}/2$	$V_{DD} = 3V \pm 10\%$, $ I_{OUT} \leq 5mA$		300		$\mu V/mA$
			$V_{DD} = 5V \pm 10\%$, $ I_{OUT} \leq 10mA$		300		
DC Output Impedance		$V_{OUT} = V_{FS}/2$	$V_{DD} = 3V \pm 10\%$, $ I_{OUT} \leq 5mA$		0.3		Ω
			$V_{DD} = 5V \pm 10\%$, $ I_{OUT} \leq 10mA$		0.3		
Capacitive Load Handling	C_L				500		pF
Resistive Load Handling	R_L			2			k Ω
Short-Circuit Output Current		$V_{DD} = 5.5V$	Sourcing (output short to GND)		30		mA
			Sinking (output shorted to V_{DD})		40		
DYNAMIC PERFORMANCE							
Voltage-Output Slew Rate	SR	Positive and negative			2.0		V/ μs
Voltage-Output Settling Time		$1/4$ scale to $3/4$ scale, to ≤ 1 LSB, MAX5703			2.8		μs
		$1/4$ scale to $3/4$ scale, to ≤ 1 LSB, MAX5704			5.2		
		$1/4$ scale to $3/4$ scale, to ≤ 1 LSB, MAX5705			6.3		
DAC Glitch Impulse		Major code transition			5.0		nV·s
Digital Feedthrough		Code = 0, all digital inputs from 0V to V_{DDIO}			0.5		nV·s
Power-Up Time		Startup calibration time (Note 7)			200		μs
		From power-down mode			60		μs
DC Power-Supply Rejection		$V_{DD} = 3V \pm 10\%$ or $5V \pm 10\%$			100		$\mu V/V$

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = 2.7V$ to $5.5V$, $V_{DDIO} = 1.8V$ to $5.5V$, $V_{GND} = 0V$, $C_L = 200pF$, $R_L = 2k\Omega$, $T_A = -40^\circ C$ to $+125^\circ C$, unless otherwise noted.)
(Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Output Voltage-Noise Density (DAC Output at Midscale)		External reference	f = 1kHz		88		nV/ \sqrt{Hz}
			f = 10kHz		79		
		2.048V internal reference	f = 1kHz		108		
			f = 10kHz		98		
		2.5V internal reference	f = 1kHz		117		
			f = 10kHz		110		
4.096V internal reference	f = 1kHz		152				
	f = 10kHz		145				
Integrated Output Noise (DAC Output at Midscale)		External reference	f = 0.1Hz to 10Hz		10		μV_{P-P}
			f = 0.1Hz to 10kHz		72		
			f = 0.1Hz to 300kHz		298		
		2.048V internal reference	f = 0.1Hz to 10Hz		11		
			f = 0.1Hz to 10kHz		89		
			f = 0.1Hz to 300kHz		370		
		2.5V internal reference	f = 0.1Hz to 10Hz		12		
			f = 0.1Hz to 10kHz		99		
			f = 0.1Hz to 300kHz		355		
		4.096V internal reference	f = 0.1Hz to 10Hz		13		
			f = 0.1Hz to 10kHz		128		
			f = 0.1Hz to 300kHz		400		
Output Voltage-Noise Density (DAC Output at Full Scale)		External reference	f = 1kHz		113		nV/ \sqrt{Hz}
			f = 10kHz		100		
		2.048V internal reference	f = 1kHz		172		
			f = 10kHz		157		
		2.5V internal reference	f = 1kHz		195		
			f = 10kHz		180		
4.096V internal reference	f = 1kHz		279				
	f = 10kHz		258				
Integrated Output Noise (DAC Output at Full Scale)		External reference	f = 0.1Hz to 10Hz		12		μV_{P-P}
			f = 0.1Hz to 10kHz		88		
			f = 0.1Hz to 300kHz		280		
		2.048V internal reference	f = 0.1Hz to 10Hz		14		
			f = 0.1Hz to 10kHz		135		
			f = 0.1Hz to 300kHz		530		
		2.5V internal reference	f = 0.1Hz to 10Hz		15		
			f = 0.1Hz to 10kHz		160		
			f = 0.1Hz to 300kHz		550		
		4.096V internal reference	f = 0.1Hz to 10Hz		23		
			f = 0.1Hz to 10kHz		220		
			f = 0.1Hz to 300kHz		610		

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = 2.7V$ to $5.5V$, $V_{DDIO} = 1.8V$ to $5.5V$, $V_{GND} = 0V$, $C_L = 200pF$, $R_L = 2k\Omega$, $T_A = -40^\circ C$ to $+125^\circ C$, unless otherwise noted.)
(Note 2)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
REFERENCE INPUT						
Reference Input Range	V_{REF}		1.24		V_{DD}	V
Reference Input Current	I_{REF}	$V_{REF} = V_{DD} = 5.5V$		55	75	μA
Reference Input Impedance	R_{REF}		75	100		$k\Omega$
REFERENCE OUTPUT						
Reference Output Voltage	V_{REF}	$V_{REF} = 2.048V$, $T_A = +25^\circ C$	2.043	2.048	2.053	V
		$V_{REF} = 2.5V$, $T_A = +25^\circ C$	2.494	2.500	2.506	
		$V_{REF} = 4.096V$, $T_A = +25^\circ C$	4.086	4.096	4.106	
Reference Output Noise Density	V_{REF}	$V_{REF} = 2.048V$	$f = 1kHz$	129		nV/ \sqrt{Hz}
			$f = 10kHz$	122		
		$V_{REF} = 2.500V$	$f = 1kHz$	158		
			$f = 10kHz$	151		
		$V_{REF} = 4.096V$	$f = 1kHz$	254		
			$f = 10kHz$	237		
Integrated Reference Output Noise	V_{REF}	$V_{REF} = 2.048V$	$f = 0.1Hz$ to $10Hz$	12		μV_{P-P}
			$f = 0.1Hz$ to $10kHz$	110		
			$f = 0.1Hz$ to $300kHz$	390		
		$V_{REF} = 2.500V$	$f = 0.1Hz$ to $10Hz$	15		
			$f = 0.1Hz$ to $10kHz$	129		
			$f = 0.1Hz$ to $300kHz$	430		
		$V_{REF} = 4.096V$	$f = 0.1Hz$ to $10Hz$	20		
			$f = 0.1Hz$ to $10kHz$	205		
			$f = 0.1Hz$ to $300kHz$	525		
Reference Temperature Coefficient (Note 8)		MAX5705A	± 4	± 12	ppm/ $^\circ C$	
		MAX5703/MAX5704/MAX5705B	± 10	± 25		
Reference Drive Capacity		External load		25		$k\Omega$
Reference Capacitive Load Handling				200		pF
Reference Load Regulation		$I_{SOURCE} = 0$ to $500\mu A$		1.0		mV/mA
Reference Line Regulation				0.1		mV/V

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = 2.7V$ to $5.5V$, $V_{DDIO} = 1.8V$ to $5.5V$, $V_{GND} = 0V$, $C_L = 200pF$, $R_L = 2k\Omega$, $T_A = -40^\circ C$ to $+125^\circ C$, unless otherwise noted.)
(Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
POWER REQUIREMENTS							
Supply Voltage	V_{DD}	$V_{REF} = 4.096V$		4.5		5.5	V
		All other options		2.7		5.5	
I/O Supply Voltage	V_{DDIO}			1.8		5.5	V
Supply Current (DAC Output at Midscale) (Note 9)	I_{DD}	External reference	$V_{REF} = 3V$		135	190	μA
			$V_{REF} = 5V$		165	225	
		Internal reference, reference pin undriven	$V_{REF} = 2.048V$		190	265	
			$V_{REF} = 2.5V$		205	280	
			$V_{REF} = 4.096V$		250	340	
		Internal reference, reference pin driven	$V_{REF} = 2.048V$		215	300	
			$V_{REF} = 2.5V$		225	315	
			$V_{REF} = 4.096V$		275	375	
Supply Current (DAC Output at Full Scale) (Note 9)	I_{DD}	External reference	$V_{REF} = 3V$		155	210	μA
			$V_{REF} = 5V$		200	265	
		Internal reference, reference pin undriven	$V_{REF} = 2.048V$		205	280	
			$V_{REF} = 2.5V$		220	300	
			$V_{REF} = 4.096V$		275	375	
		Internal reference, reference pin driven	$V_{REF} = 2.048V$		225	310	
			$V_{REF} = 2.5V$		240	330	
			$V_{REF} = 4.096V$		300	410	
Power-Down Mode Supply Current (DAC Powered Down, Reference Remains Active) (Note 9)	I_{DD}	Internal reference, reference pin driven	$V_{REF} = 2.048V$		90	135	μA
			$V_{REF} = 2.5V$		93	135	
			$V_{REF} = 4.096V$		100	150	
Power-Down Mode Supply Current (Note 9)	I_{PD}	External reference, $V_{DD} = V_{REF}$			0.4	2	μA
Digital Supply Current (Note 9)	I_{DDIO}					1.0	μA
DIGITAL INPUT CHARACTERISTICS (\overline{CS}, SCLK, DIN, LDAC, \overline{AUX})							
Input High Voltage	V_{IH}	$2.2V < V_{DDIO} < 5.5V$			$0.7 \times V_{DDIO}$		V
		$1.8V < V_{DDIO} < 2.2V$			$0.8 \times V_{DDIO}$		
Input Low Voltage	V_{IL}	$2.2V < V_{DDIO} < 5.5V$			$0.3 \times V_{DDIO}$		V
		$1.8V < V_{DDIO} < 2.2V$			$0.2 \times V_{DDIO}$		

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = 2.7V$ to $5.5V$, $V_{DDIO} = 1.8V$ to $5.5V$, $V_{GND} = 0V$, $C_L = 200pF$, $R_L = 2k\Omega$, $T_A = -40^\circ C$ to $+125^\circ C$, unless otherwise noted.)
(Note 2)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Hysteresis Voltage	V_H				0.15		V
Input Leakage Current (Note 9)	I_{IN}				± 0.1	± 1	μA
Input Capacitance	C_{IN}				3		pF
SPI TIMING CHARACTERISTICS (\overline{CS}, SCLK, DIN, LDAC, AUX) (Note 10)							
SCLK Frequency		$2.7V \leq V_{DDIO} \leq 5.5V$		0		50	MHz
		$1.8V \leq V_{DDIO} < 2.7V$		0		33	
SCLK Period	t_{SCLK}	$2.7V \leq V_{DDIO} \leq 5.5V$		20			ns
		$1.8V \leq V_{DDIO} < 2.7V$		30			
SCLK Pulse Width High	t_{CH}			8			ns
SCLK Pulse Width Low	t_{CL}			8			ns
\overline{CS} Fall to SCLK Fall Setup Time	t_{CSS0}	To first SCLK falling edge	$2.7V \leq V_{DDIO} \leq 5.5V$	8			ns
			$1.8V \leq V_{DDIO} < 2.7V$	12			
\overline{CS} Fall to SCLK Fall Hold Time	t_{CSH0}	Applies to inactive SCLK falling edge preceding the first SCLK falling edge		0			ns
\overline{CS} Rise to SCLK Fall Hold Time	t_{CSH1}	Applies to the 24th SCLK falling edge		0			ns
\overline{CS} Rise to SCLK Fall	t_{CSA}	Applies to the 24th SCLK falling edge, aborted sequence		12			ns
SCLK Fall to \overline{CS} Fall	t_{CSF}	Applies to 24th SCLK falling edge		100			ns
\overline{CS} Pulse Width High	t_{CSPW}			20			ns
DIN to SCLK Fall Setup Time	t_{DS}			5			ns
DIN to SCLK Fall Hold Time	t_{DH}			4.5			ns
CLR Pulse Width Low	t_{CLPW}			20			ns
CLR Rise to \overline{CS} Fall	t_{CSC}	Required for command to be executed		20			ns
LDAC Pulse Width Low	t_{LDPW}			20			ns
LDAC Fall to SCLK Fall Hold	t_{LDH}	Applies to 24th SCLK falling edge		20			ns

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ELECTRICAL CHARACTERISTICS (continued)

($V_{DD} = 2.7V$ to $5.5V$, $V_{DDIO} = 1.8V$ to $5.5V$, $V_{GND} = 0V$, $C_L = 200pF$, $R_L = 2k\Omega$, $T_A = -40^\circ C$ to $+125^\circ C$, unless otherwise noted.)
(Note 2)

- Note 2:** Electrical specifications are production tested at $T_A = +25^\circ C$. Specifications over the entire operating temperature range are guaranteed by design and characterization. Typical specifications are at $T_A = +25^\circ C$.
- Note 3:** DC Performance is tested without load.
- Note 4:** Linearity is tested with unloaded outputs to within 20mV of GND and V_{DD} .
- Note 5:** Gain and offset calculated from measurements made with $V_{REF} = V_{DD}$ at code 30 and 4065 for MAX5705, code 8 and 1016 for MAX5704, and code 2 and 254 for MAX5703.
- Note 6:** Subject to zero and full-scale error limits and V_{REF} settings.
- Note 7:** On power-up, the device initiates an internal 200 μs (typ) calibration sequence. All commands issued during this time will be ignored.
- Note 8:** Specification is guaranteed by design and characterization.
- Note 9:** Static logic inputs with $V_{IL} = V_{GND}$ and $V_{IH} = V_{DDIO}$.
- Note 10:** All timing is tested with $V_{IL} = V_{GND}$ and $V_{IH} = V_{DDIO}$.

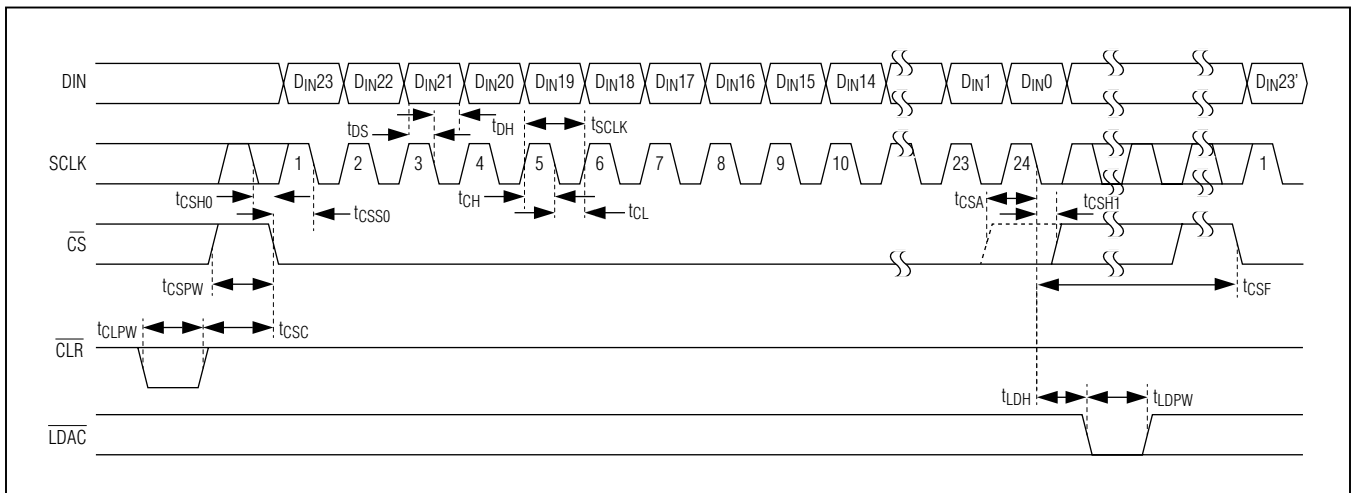


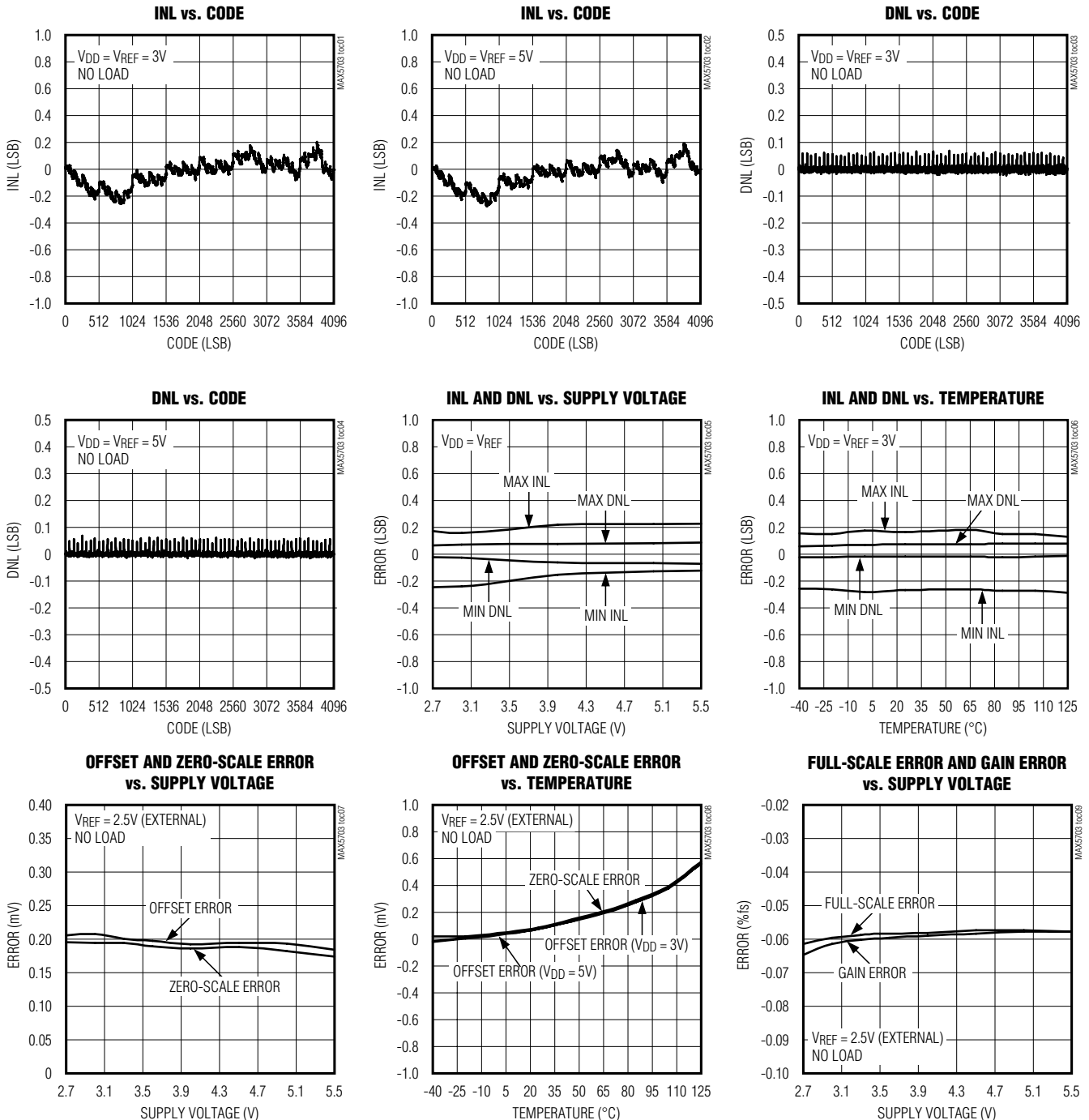
Figure 1. SPI Serial Interface Timing Diagram

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Typical Operating Characteristics

(MAX5705, 12-bit performance, $T_A = +25^\circ\text{C}$, unless otherwise noted.)



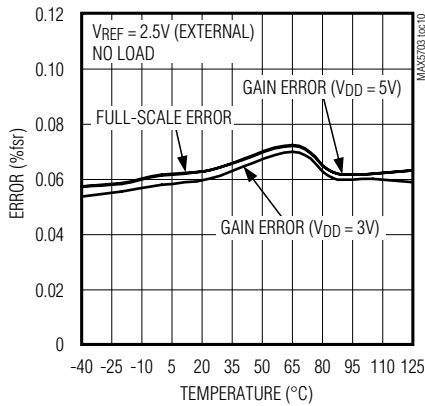
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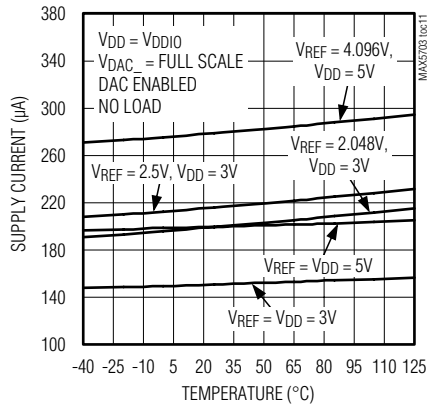
Typical Operating Characteristics (continued)

(MAX5705, 12-bit performance, $T_A = +25^\circ\text{C}$, unless otherwise noted.)

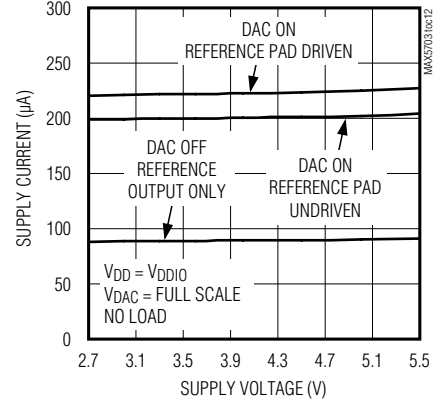
FULL-SCALE ERROR AND GAIN ERROR vs. TEMPERATURE



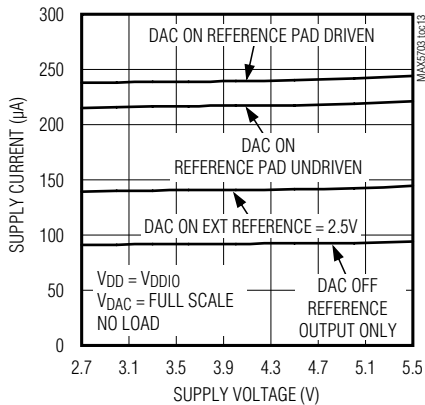
SUPPLY CURRENT vs. TEMPERATURE (PIN UNDRIVEN FOR INTERNAL REF MODES)



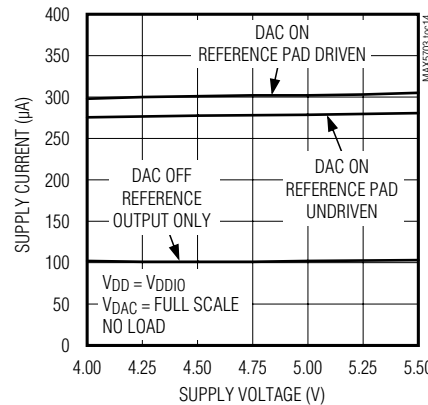
SUPPLY CURRENT vs. SUPPLY VOLTAGE (2.048V INTERNAL REFERENCE)



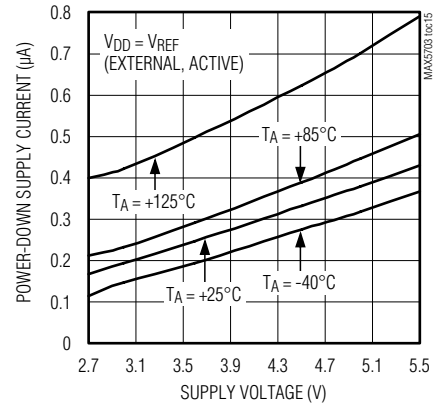
SUPPLY CURRENT vs. SUPPLY VOLTAGE (2.500V INTERNAL REFERENCE)



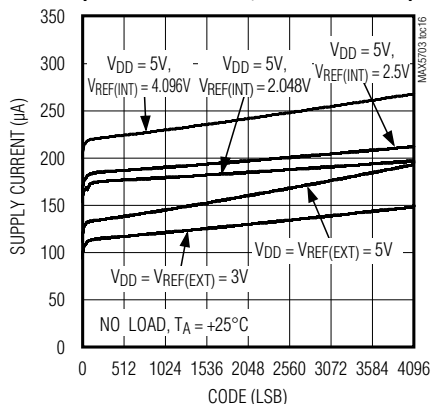
SUPPLY CURRENT vs. SUPPLY VOLTAGE (4.096V INTERNAL REFERENCE)



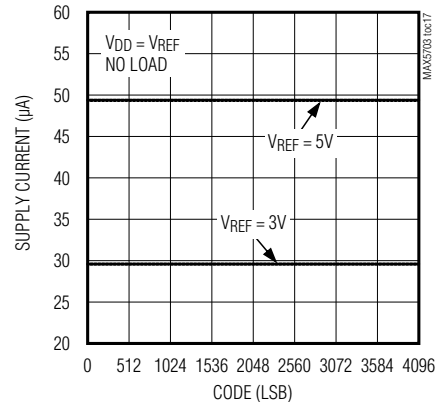
POWER-DOWN MODE CURRENT vs. SUPPLY VOLTAGE



SUPPLY CURRENT vs. CODE (FOR INTERNAL REF, PIN IS UNDRIVEN)



IREF (EXTERNAL) vs. CODE

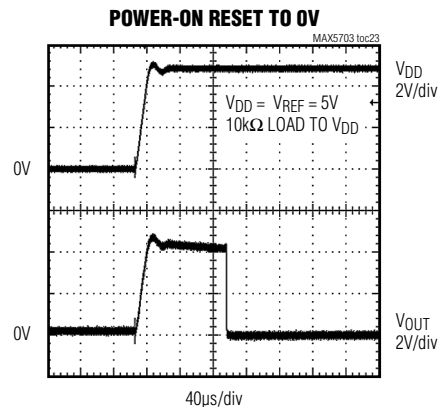
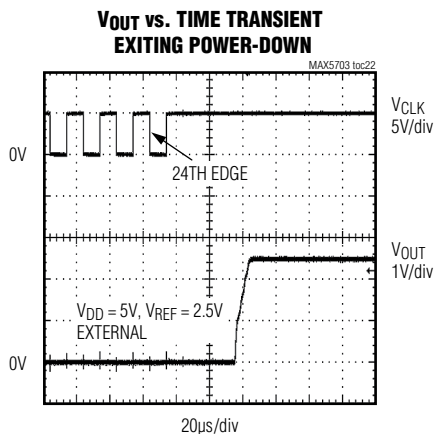
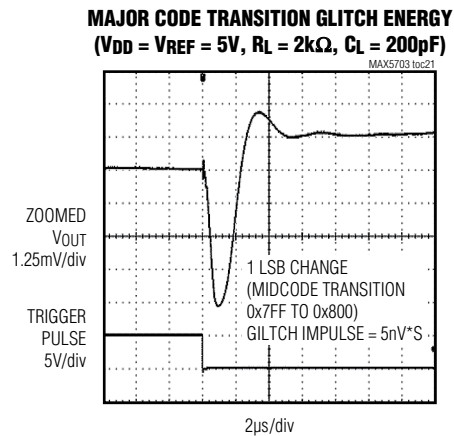
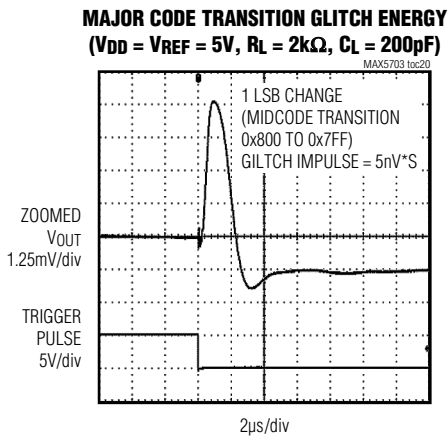
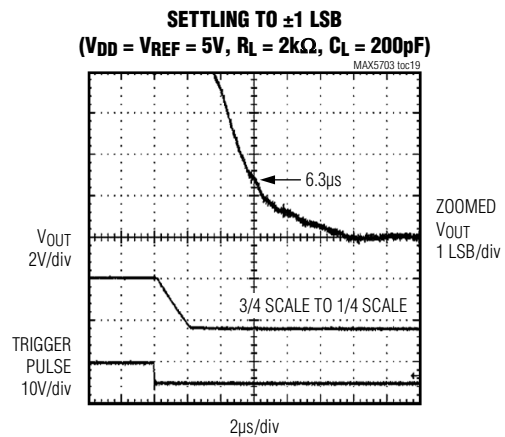
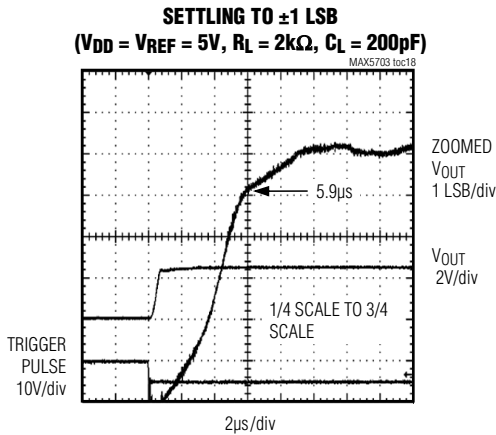


MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

Typical Operating Characteristics (continued)

(MAX5705, 12-bit performance, $T_A = +25^\circ\text{C}$, unless otherwise noted.)



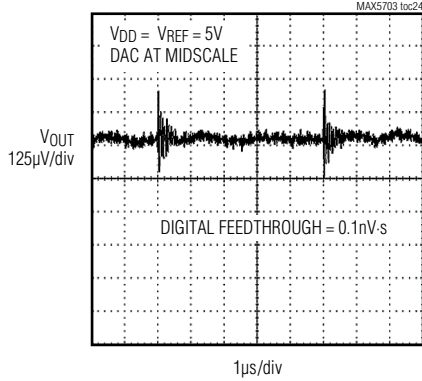
MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

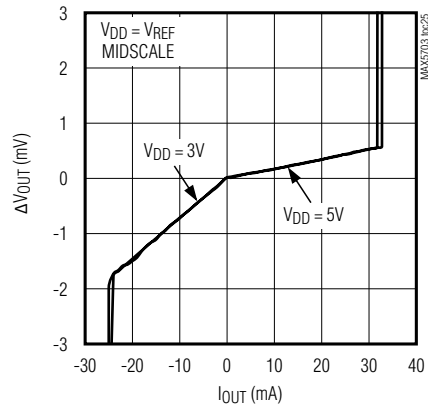
Typical Operating Characteristics (continued)

(MAX5705, 12-bit performance, $T_A = +25^\circ\text{C}$, unless otherwise noted.)

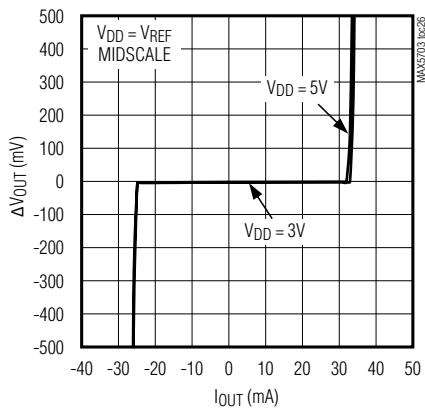
DIGITAL FEEDTHROUGH
($V_{DD} = V_{REF} = 5\text{V}$, $R_L = 2\text{k}\Omega$, $C_L = 200\text{pF}$)



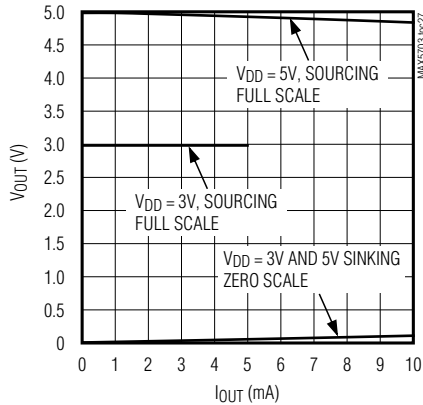
OUTPUT LOAD REGULATION



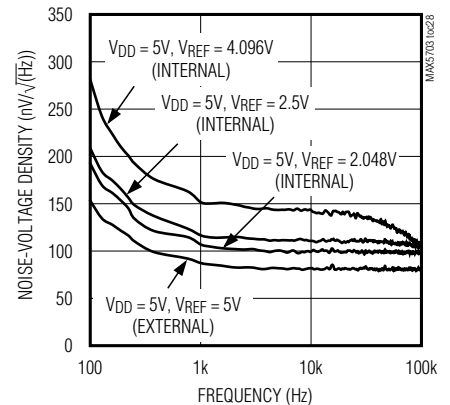
OUTPUT CURRENT LIMITING



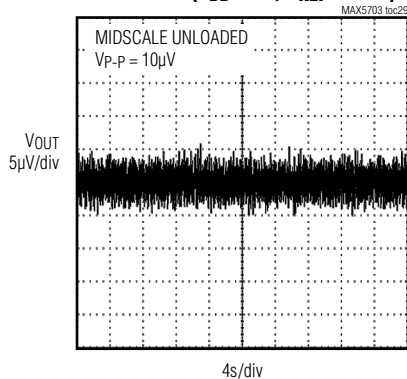
HEADROOM AT RAILS vs. OUTPUT CURRENT ($V_{DD} = V_{REF}$)



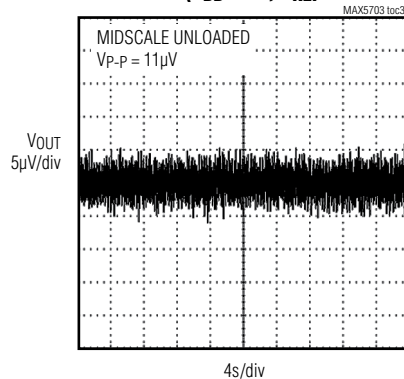
NOISE-VOLTAGE DENSITY vs. FREQUENCY (DAC AT MIDSCALE)



0.1Hz TO 10Hz OUTPUT NOISE, EXTERNAL REFERENCE ($V_{DD} = 5\text{V}$, $V_{REF} = 4.5\text{V}$)



0.1Hz TO 10Hz OUTPUT NOISE, INTERNAL REFERENCE ($V_{DD} = 5\text{V}$, $V_{REF} = 2.048\text{V}$)

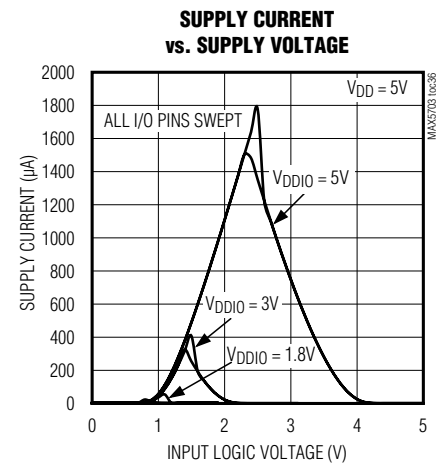
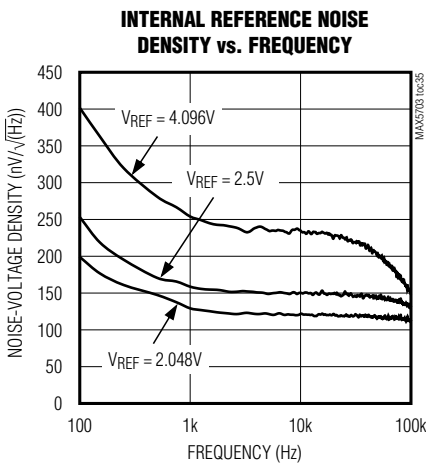
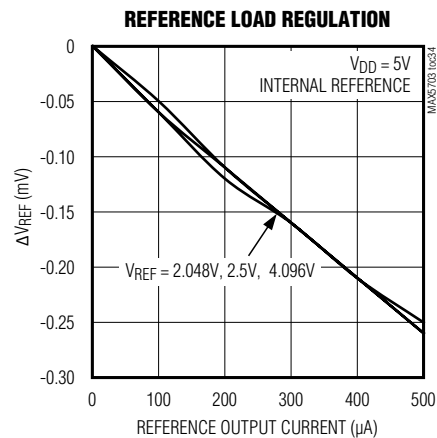
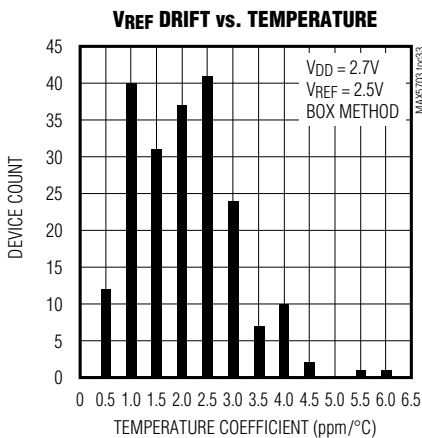
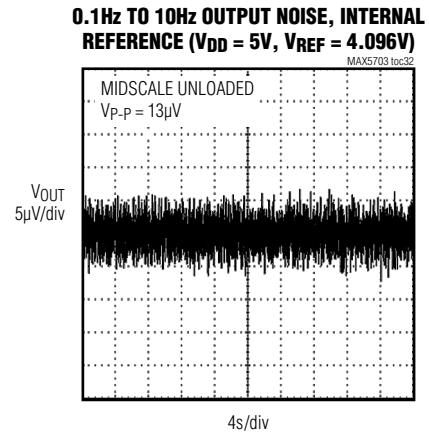
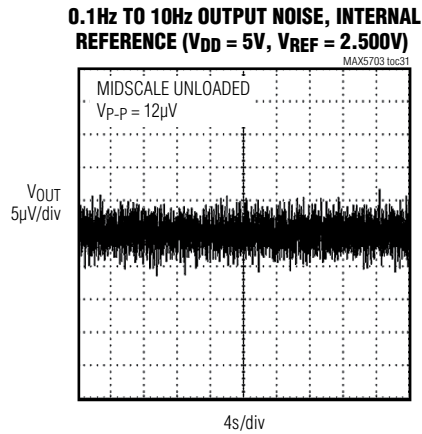


MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

Typical Operating Characteristics (continued)

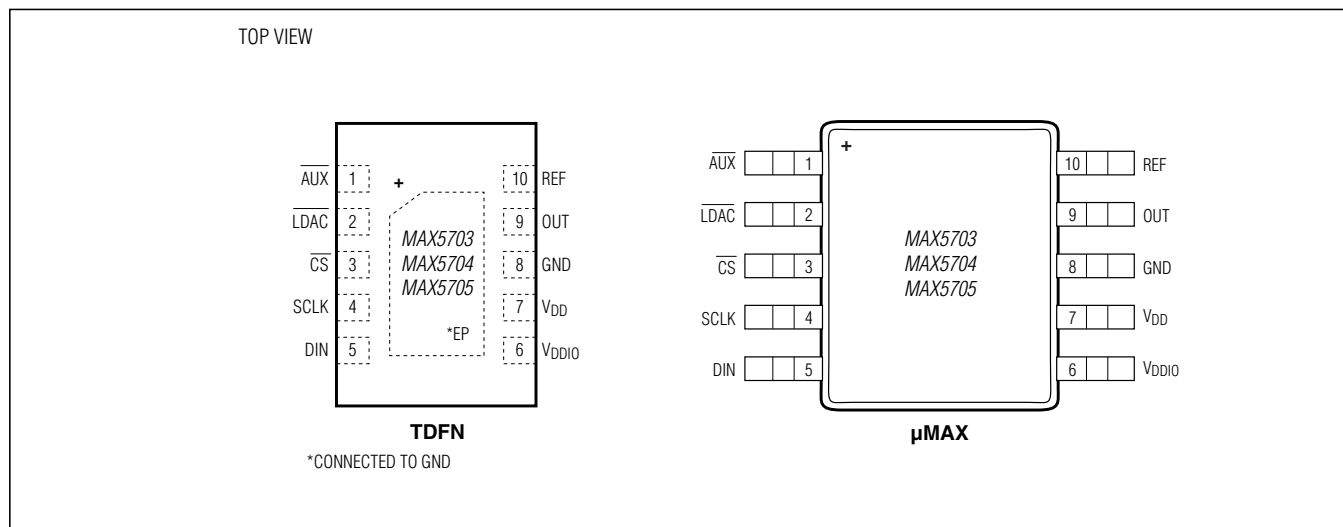
(MAX5705, 12-bit performance, $T_A = +25^\circ\text{C}$, unless otherwise noted.)



MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

Pin Configurations



Pin Description

PIN	NAME	FUNCTION
1	$\overline{\text{AUX}}$	Active-Low Auxilliary Asynchronous Input. User configurable, see Table 7. If not using the $\overline{\text{AUX}}$ functions, connect this input to V_{DDIO} .
2	$\overline{\text{LDAC}}$	Dedicated Active-Low Asynchronous Load DAC
3	$\overline{\text{CS}}$	SPI Chip-Select Input
4	SCLK	SPI Interface Clock Input
5	DIN	SPI Interface Data Input
6	V_{DDIO}	Digital Interface Power-Supply Input
7	V_{DD}	Supply Voltage Input. Bypass V_{DD} with a 0.1 μF capacitor to GND.
8	GND	Ground
9	OUT	Buffered DAC Output
10	REF	Reference Voltage Input/Output
—	EP	Exposed Pad (TDFN Only). Connect to ground.

MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

Detailed Description

The MAX5703/MAX5704/MAX5705 are single-channel, low-power, 8-/10-/12-bit voltage-output digital-to-analog converters (DACs) with an internal output buffer. The wide supply voltage range of 2.7V to 5.5V and low power consumption accommodate low-power and low-voltage applications. The devices present a 100k Ω (typ) load to the external reference. The internal output buffer allows rail-to-rail operation. An internal voltage reference is available with software selectable options of 2.048V, 2.500V, or 4.096V. The devices feature a 50MHz, 3-wire SPI/QSPI/MICROWIRE/DSP-compatible serial interface to save board space and reduce complexity in isolated applications. The MAX5703/MAX5704/MAX5705 include a serial-in/parallel-out shift register, internal CODE and DAC registers, a power-on-reset (POR) circuit to initialize the DAC output to code zero, and control logic. A user-configurable $\overline{\text{AUX}}$ pin is available to asynchronously clear or gate the device output independent of the serial interface.

DAC Output (OUT)

The MAX5703/MAX5704/MAX5705 include an internal buffer on the DAC output. The internal output buffer provides improved load regulation for the DAC output. The output buffer slews at 1V/ μs (typ) and drives up to 2k Ω in parallel with 500pF. The analog supply voltage (V_{DD}) determines the maximum output voltage range of the devices as V_{DD} powers the output buffer. Under no-load conditions, the output buffer drives from GND to V_{DD} , subject to offset and gain errors. With a 2k Ω load to GND, the output buffer drives from GND to within and 200mV of V_{DD} . With a 2k Ω load to V_{DD} , the output buffer drives from V_{DD} to within 200mV of GND.

The DAC ideal output voltage is defined by:

$$V_{\text{OUT}} = V_{\text{REF}} \times \frac{D}{2^N}$$

Where D = code loaded into the DAC register, V_{REF} = reference voltage, N = resolution.

Internal Register Structure

The user interface is separated from the DAC logic to minimize digital feedthrough. Within the serial interface is an input shift register, the contents of which can be routed to control registers or the DAC itself, as determined by the user command.

Within the device there is a CODE register followed by a DAC Latch register (see the [Functional Diagram](#)). The contents of the CODE register hold pending DAC output settings which can later be loaded into the DAC registers. The CODE register can be updated using both CODE and CODE_LOAD user commands. The contents of the DAC register hold the current DAC output settings. The DAC register can be updated directly from the serial interface using the CODE_LOAD commands or can upload the current contents of the CODE register using LOAD commands or the $\overline{\text{LDAC}}$ input.

The contents of both CODE and DAC registers are maintained during all software power-down states, so that when the DAC is returned to a normal operating mode, it returns to its previously stored output settings. Any CODE or LOAD commands issued during software power-down states continue to update the register contents. The SW_CLEAR command clears the contents of the CODE and DAC registers to the user-programmable default values. The SW_RESET command resets all configuration registers to their power-on default states, while resetting the CODE and DAC registers to zero scale.

Internal Reference

The MAX5703/MAX5704/MAX5705 include an internal precision voltage reference that is software selectable to be 2.048V, 2.500V, or 4.096V. When an internal reference is selected, that voltage is available on the REF pin for other external circuitry (see the [Typical Operating Circuits](#)) and can drive a 25k Ω load.

External Reference

The external reference input features a typical input impedance of 100k Ω and accepts an input voltage from +1.24V to V_{DD} . Connect an external voltage supply between REF and GND to apply an external reference. The MAX5703/4/5 power up and reset to external reference mode. Visit www.maximintegrated.com/products/references for a list of available external voltage-reference devices.

$\overline{\text{AUX}}$ Input

The MAX5703/MAX5704/MAX5705 provide an asynchronous $\overline{\text{AUX}}$ (active-low) input. Use the CONFIG command to program the device to use the input in one of the following modes: CLR (default), GATE, or disabled. If not using the $\overline{\text{AUX}}$ functions, connect this input to V_{DDIO} .

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$\overline{\text{CLR}}$ Mode

In $\overline{\text{CLR}}$ mode, the $\overline{\text{AUX}}$ input performs an asynchronous level sensitive CLEAR operation when pulled low. If $\overline{\text{CLR}}$ is configured and asserted, all CODE and DAC data registers are cleared to their default/return values as defined by the configuration settings. Other user-configuration settings are not affected.

Some SPI interface commands are gated by $\overline{\text{CLR}}$ activity during the transfer sequence. If $\overline{\text{CLR}}$ is issued during a command write sequence, any gated commands within the sequence are ignored. Any non-gated commands appearing in the transfer sequence are executed. For the gating condition to be removed, drive $\overline{\text{CLR}}$ high, satisfying the t_{CSC} requirements.

$\overline{\text{GATE}}$ Mode

Use of the $\overline{\text{GATE}}$ mode provides a means of momentarily holding the DAC in a user-selectable default/return state, returning the DAC to the last programmed state upon removal. The MAX5703/MAX5704/MAX5705 also feature a software-accessible GATE command. While asserted in $\overline{\text{GATE}}$ mode, the $\overline{\text{AUX}}$ pin does not interfere with RETURN, CODE, or DAC register updates and related load activity.

$\overline{\text{LDAC}}$ Input

The MAX5703/MAX5704/MAX5705 provide a dedicated asynchronous $\overline{\text{LDAC}}$ (active-low) input. The $\overline{\text{LDAC}}$ input performs an asynchronous level sensitive LOAD operation when pulled low. Use of the $\overline{\text{LDAC}}$ input mode provides a means of updating multiple devices together as a group. Users wishing to control the DAC update instance independently of the I/O instruction should hold $\overline{\text{LDAC}}$ high during programming cycles. Once programming is complete, $\overline{\text{LDAC}}$ may be strobed and the new CODE register content is loaded into the DAC latch output. Users wishing to load new DAC data in direct response to I/O CODE register activity should connect $\overline{\text{LDAC}}$ permanently low; in this configuration, the MAX5703/MAX5704/MAX5705 DAC output updates in response to each completed I/O CODE instruction update edge. A software LOAD command is also provided.

The $\overline{\text{LDAC}}$ operation does not interact with the user interface directly. However, in order to achieve the best possible glitch performance, timing with respect to the interface update edge should follow t_{LDH} specifications when issuing CODE commands.

V_{DDIO} Input

The MAX5703/MAX5704/MAX5705 feature a separate supply pin (V_{DDIO}) for the digital interface (1.8V to 5.5V). Connect V_{DDIO} to the I/O supply of the host processor.

SPI Serial Interface

The MAX5703/MAX5704/MAX5705 3-wire serial interface is compatible with MICROWIRE/SPI/QSPI and DSPs. The interface provides three inputs: SCLK, $\overline{\text{CS}}$, and DIN. The chip-select input ($\overline{\text{CS}}$, active-low) frames the data loaded through the serial data input (DIN). Following a $\overline{\text{CS}}$ input high-to-low transition, the data is shifted in synchronously and latched into the input register on each falling edge of the serial clock input (SCLK). Each serial operation word is 24-bits long. The DAC data is left justified as shown in Table 1. The serial input register transfers its contents to the destination registers after loading 24 bits of data on the 24th SCLK falling edge. To initiate a new SPI operation, drive $\overline{\text{CS}}$ high and then low to begin the next operation sequence, being sure to meet all relevant timing requirements. During $\overline{\text{CS}}$ high periods, SCLK is ignored, allowing communication to other devices on the same bus. SPI operations consisting of more than 24 SCLK cycles are executed on the 24th SCLK falling edge, using the first three bytes of data available. SPI operations consisting of less than 24 SCLK cycles will not be executed. The content of the SPI operation consists of a command byte followed by a two byte data word.

[Figure 1](#) shows the timing diagram for the complete 3-wire serial interface transmission. The DAC code settings (D) for the MAX5703/MAX5704/MAX5705 are accepted in an offset binary format (see [Table 1](#)). Otherwise, the expected data format for each command is listed in [Table 2](#).

SPI User-Command Register Map

This section lists the user-accessible commands and registers for the MAX5703/MAX5704/MAX5705.

[Table 2](#) provides detailed information about the SPI Command Registers.

MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

CODE Command

The CODE command (B[23:20] = 1000) updates the CODE register content for the DAC. Changes to the CODE register content based on this command will not affect the DAC output directly unless the $\overline{\text{LDAC}}$ input is in a low state. Otherwise, a subsequent hardware or software LOAD operation will be required to move this content to the active DAC latch. This command is gated when $\overline{\text{CLR}}$ is asserted, updates to this register are ignored while the register is being cleared. See [Table 1](#) and [Table 2](#).

LOAD Command

The LOAD command (B[23:20] = 1001) updates the DAC latch register content by uploading the current contents of the CODE register. This command is gated when $\overline{\text{CLR}}$ is asserted, updates to this register are ignored while the register is being cleared. See [Table 2](#).

CODE_LOAD Command

The CODE_LOAD command (B[23:20] = 1010 and 1011) updates the CODE register contents as well as the DAC register content of the DAC. This command is gated when $\overline{\text{CLR}}$ is asserted, updates to these registers are ignored while the register is being cleared. See [Table 1](#) and [Table 2](#).

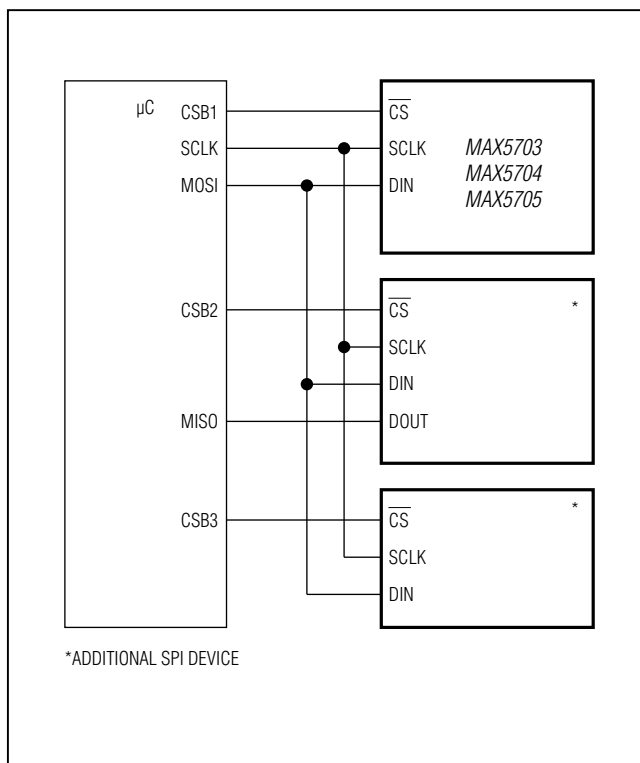


Figure 2. Typical SPI Application Circuit

Table 1. DAC Data Bit Positions

PART	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
MAX5703	D7	D6	D5	D4	D3	D2	D1	D0	X	X	X	X	X	X	X	X
MAX5704	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	X	X	X	X	X	X
MAX5705	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	X	X	X	X

MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

Table 2. SPI Commands Summary

COMMAND	B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0	DESCRIPTION
DAC COMMANDS																									
CODE	1	0	0	0	X	X	X	X											CODE REGISTER DATA[3:0]		X	X	X	X	Writes data to the CODE register
LOAD	1	0	0	1	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	Transfers data from the CODE registers to the DAC register
CODE_LOAD	1	0	1	0	X	X	X	X											CODE AND DAC REGISTER DATA[3:0]		X	X	X	X	Simultaneously writes data to the CODE register while updating DAC register
CODE_LOAD	1	0	1	1	X	X	X	X											CODE AND DAC REGISTER DATA[3:0]		X	X	X	X	Simultaneously writes data to the CODE register while updating DAC register
RETURN	0	1	1	1	X	X	X	X											RETURN REGISTER DATA[3:0]		X	X	X	X	Updates the RETURN register contents for the DAC

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Table 2. SPI Commands Summary (continued)

COMMAND	B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0	DESCRIPTION
CONFIGURATION COMMANDS																									
REF	0	0	1	0		0 = No Drive 1 = Drive Pn	0 = Default 1 = Always ON	00 = EXT 01 = 2.5V 10 = 2.0V 11 = 4.1V	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	Sets the reference operating mode.
	0	0	1	1	X		Type: 000 = END 001 = GATE 100 = CLR 101 = RST Other = No Effect	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	Executes a software operation of the type chosen
POWER	0	1	0	0	X			X	X	X	X	X	X	X	X	X									Sets the Power mode
	0	1	0	1	X			X	X	X	X	X	X	X	X	X									Updates the function of the AUX input
DEFAULT	0	1	1	0	X			X	X	X	X	X	X	X	X										Sets the default value for the DAC
	0	0	0	0	X			X	X	X	X	X	X	X	X										Sets the default value for the DAC
NO OPERATION COMMANDS																									
No Operation	0	0	0	0	X			X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	These commands will have no effect on the part.
	1	1	1	1	X			X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	
Reserved Commands: Any commands not specifically listed above are reserved for Maxim internal use only.																									

MAX5703/MAX5704/MAX5705

Ultra-Small, Single-Channel, 8-/10-/12-Bit Buffered Output Voltage DACs with Internal Reference and SPI Interface

REF Command

The REF (B[23:20] = 0010) command updates the global reference setting used for the DAC. Set B[17:16] = 00 to use an external reference for the DAC or set B[17:16] to 01, 10, or 11 to select either the 2.5V, 2.048V, or 4.096V internal reference, respectively.

If RF3 (B19) is set to zero (default) in the REF command, the REF I/O will not be driven by the internal reference circuit, saving current. If RF3 is set to one, the REF I/O will be driven by the internal reference circuit, consuming an additional 25µA (typ) of current when the reference is powered; when the reference is powered down, the REF I/O will be high-impedance.

If RF2 (B18) is set to zero (default) in the REF command, the reference will be powered down any time the DAC is powered down (in STANDBY mode). If RF2 (B18) is set to one, the reference will remain powered even if the DAC is powered down, allowing continued operation of external circuitry. In this mode, the 1µA shutdown state is not available. See [Table 3](#).

SOFTWARE Commands

The SOFTWARE (B[23:20] = 0011) commands provide a means of issuing several flexible software actions. See [Table 4](#).

The SOFTWARE Command Action Mode is selected by B[18:16]:

- END (000): Used to end any active gate operation, returning to normal operation (default).
- GATE (001): DAC contents will be gated to their DEFAULT selected values until the gate condition is removed.
- CLEAR (100): All CODE and DAC contents will be cleared to their DEFAULT selected values.
- RESET (101): All CODE, DAC, RETURN, and configuration registers reset to their power-up defaults (including REF, POWER, and CONFIG settings), simulating a power cycle reset.
- OTHER: No effect.

Table 3. REF (0010) Command Format

B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
0	0	1	0	RF3	RF2	RF1	RF0	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X
REF COMMAND				0 = REF Not driven 1 = REF Driven	0 = Off in Standby 1 = On in Standby	Ref Mode: 00 = EXT 01 = 2.5V 10 = 2.0V 11 = 4.0V			Don't Care							Don't Care							
DEFAULT VALUES				0	0	0	0	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X
COMMAND BYTE								DATA HIGH BYTE								DATA LOW BYTE							

Table 4. SOFTWARE (0011) Command Format

B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0
0	0	1	1	X	SW2	SW1	SW0	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X
SOFTWARE COMMANDS				Don't Care	Mode: 000: END 001: GATE 100: CLR 101: RST Other: No Effect			Don't Care							Don't Care								
DEFAULT VALUES				X	0	0	0	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	X
COMMAND BYTE								DATA HIGH BYTE								DATA LOW BYTE							

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POWER Command

The MAX5703/MAX5704/MAX5705 feature a software-controlled POWER mode command (B[23:20] = 0100).

In power-down, the DAC output is disconnected from the buffer and is grounded with either one of the two selectable internal resistors or set to high impedance. See [Table 5](#) and [Table 6](#) for the selectable internal resistor values in power-down mode. In power-down

mode, the DAC register retains its value so that the output is restored when the device powers up. The serial interface remains active in power-down mode with all registers accessible.

In power-down mode, the internal reference can be powered down or it can be set to remain powered-on for external use. Also, in power-down mode, parts using the external reference do not load the REF pin. See [Table 5](#).

Table 5. POWER (0100) Command Format

B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0	
0	1	0	0	X	X	X	X	X	X	X	X	X	X	X	X	PD1	PD0	X	X	X	X	X	X	
POWER COMMAND				Don't Care				Don't Care								Power Mode: 00 = Normal 01 = 1kΩ 10 = 100kΩ 11 = Hi-Z		Don't Care						
DEFAULT VALUES				X	X	X	X	X	X	X	X	X	X	X	X	X	0	0	X	X	X	X	X	X
COMMAND BYTE								DATA HIGH BYTE								DATA LOW BYTE								

Table 6. Selectable DAC Output Impedance in Power-Down Mode

PD1 (B7)	PD0 (B6)	OPERATING MODE
0	0	Normal operation
0	1	Power-down with internal 1kΩ pulldown resistor to GND.
1	0	Power-down with internal 100kΩ pulldown resistor to GND.
1	1	Power-down with high-impedance output.

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CONFIG Command

The CONFIG command (B[23:20] = 0101) updates the function of the $\overline{\text{AUX}}$ input enabling its gate, load, or clear (default) operation mode. See [Table 7](#).

$\overline{\text{AUX}}$ Config settings are written by B[5:3]:

GATE (011): $\overline{\text{AUX}}$ functions as a $\overline{\text{GATE}}$. DAC code is gated to DEFAULT value input when pin is low.

CLEAR (110): $\overline{\text{AUX}}$ functions as a $\overline{\text{CLR}}$ input (default). CODE and DAC content is cleared to DEFAULT value if pin is low.

NONE (111): $\overline{\text{AUX}}$ functions are disabled.

OTHER: $\overline{\text{AUX}}$ function is not altered.

Note: CONFIG should not be programmed with the $\overline{\text{AUX}}$ pin asserted (low) or unexpected behavior could result.

DEFAULT Command

DEFAULT (0110): The DEFAULT command selects the default value for the DAC. These default values are used for all future clear and gate operations. The new default setting is determined by bits DF[2:0]. See [Table 8](#).

Available default values are:

POR (000): DAC defaults to power-on reset value (default).

ZERO (001): DAC defaults to zero scale.

MID (010): DAC defaults to midscale.

FULL (011): DAC defaults to full scale.

RETURN (100): DAC defaults to value specified by the RETURN register

OTHER: No effect, the default setting remains unchanged.

Note: The selected default values do not apply to resets initiated by SW_RESET commands or supply cycling, both of which return the DACs to the power-on reset state (zero scale).

Table 7. CONFIG (0101) Command Format

B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0	
0	1	0	1	X	X	X	X	X	X	X	X	X	X	X	X	X	X	AB2	AB1	AB0	X	X	X	
CONFIG COMMAND				Don't Care				Don't Care								Don't Care		AUXB Mode: 011 = GATE 110 = CLEAR 111 = NONE Other = No Effect			Don't Care			
DEFAULT VALUES				X	X	X	X	X	X	X	X	X	X	X	X	X	X	X	1	1	0	X	X	X
COMMAND BYTE								DATA HIGH BYTE								DATA LOW BYTE								

Table 8. DEFAULT (0110) Command Format

B23	B22	B21	B20	B19	B18	B17	B16	B15	B14	B13	B12	B11	B10	B9	B8	B7	B6	B5	B4	B3	B2	B1	B0	
0	1	1	0	X	X	X	X	X	X	X	X	X	X	X	X	DF2	DF1	DF0	X	X	X	X	X	
DEFAULT COMMAND				Don't Care				Don't Care								Default Values: 000: POR 001: ZERO 010: MID 011: FULL 100: RETURN Other: No Effect			Don't Care					
DEFAULT VALUES				X	X	X	X	X	X	X	X	X	X	X	X	X	0	0	0	X	X	X	X	X
COMMAND BYTE								DATA HIGH BYTE								DATA LOW BYTE								

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RETURN Command

The RETURN command (B[23:20] = 0111) updates the RETURN register content for the DAC. If the DEFAULT configuration register is set to RETURN mode, the DAC will be cleared or gated to the RETURN register value in the event of a SW or HW CLEAR or GATE condition. It is not necessary to program this register if the DEFAULT = RETURN mode will not be used. The data format for the RETURN register is identical to that used for CODE and LOAD operations. See [Table 1](#) and [Table 2](#).

Applications Information

Power-On Reset (POR)

When power is applied to V_{DD} , the DAC output is set to zero scale. To optimize DAC linearity, wait until the supplies have settled and the internal setup and calibration sequence completes (200 μ s, typ).

Power Supplies and Bypassing Considerations

Bypass V_{DD} with high-quality ceramic capacitors to a low-impedance ground as close as possible to the device. Minimize lead lengths to reduce lead inductance. Connect GND to the analog ground plane.

Layout Considerations

Digital and AC transient signals on GND can create noise at the output. Connect GND to form the star ground for the DAC system. Refer remote DAC loads to this system ground for the best possible performance. Use proper grounding techniques, such as a multilayer board with a low-inductance ground plane, or star connect all ground return paths back to the MAX5703/MAX5704/MAX5705 GND. Carefully layout the traces between channels to reduce AC cross-coupling. Do not use wire-wrapped boards and sockets. Use shielding to maximize noise immunity. Do not run analog and digital signals parallel to one another, especially clock signals. Avoid routing digital lines underneath the MAX5703/MAX5704/MAX5705 package.

Definitions

Integral Nonlinearity (INL)

INL is the deviation of the measured transfer function from a straight line drawn between two codes once offset and gain errors have been nullified.

Differential Nonlinearity (DNL)

DNL is the difference between an actual step height and the ideal value of 1 LSB. If the magnitude of the $DNL \leq 1$ LSB, the DAC guarantees no missing codes and is monotonic. If the magnitude of the $DNL \geq 1$ LSB, the DAC output may still be monotonic.

Offset Error

Offset error indicates how well the actual transfer function matches the ideal transfer function. The offset error is calculated from two measurements near zero code and near maximum code.

Gain Error

Gain error is the difference between the ideal and the actual full-scale output voltage on the transfer curve, after nullifying the offset error. This error alters the slope of the transfer function and corresponds to the same percentage error in each step.

Zero-Scale Error

Zero-scale error is the difference between the DAC output voltage when set to code zero and ground. This includes offset and other die level nonidealities.

Full-Scale Error

Full-scale error is the difference between the DAC output voltage when set to full scale and the reference voltage. This includes offset, gain error, and other die level nonidealities.

Settling Time

The settling time is the amount of time required from the start of a transition, until the DAC output settles to the new output value within the converter's specified accuracy.

Digital Feedthrough

Digital feedthrough is the amount of noise that appears on the DAC output when the DAC digital control lines are toggled.

Digital-to-Analog Glitch Impulse

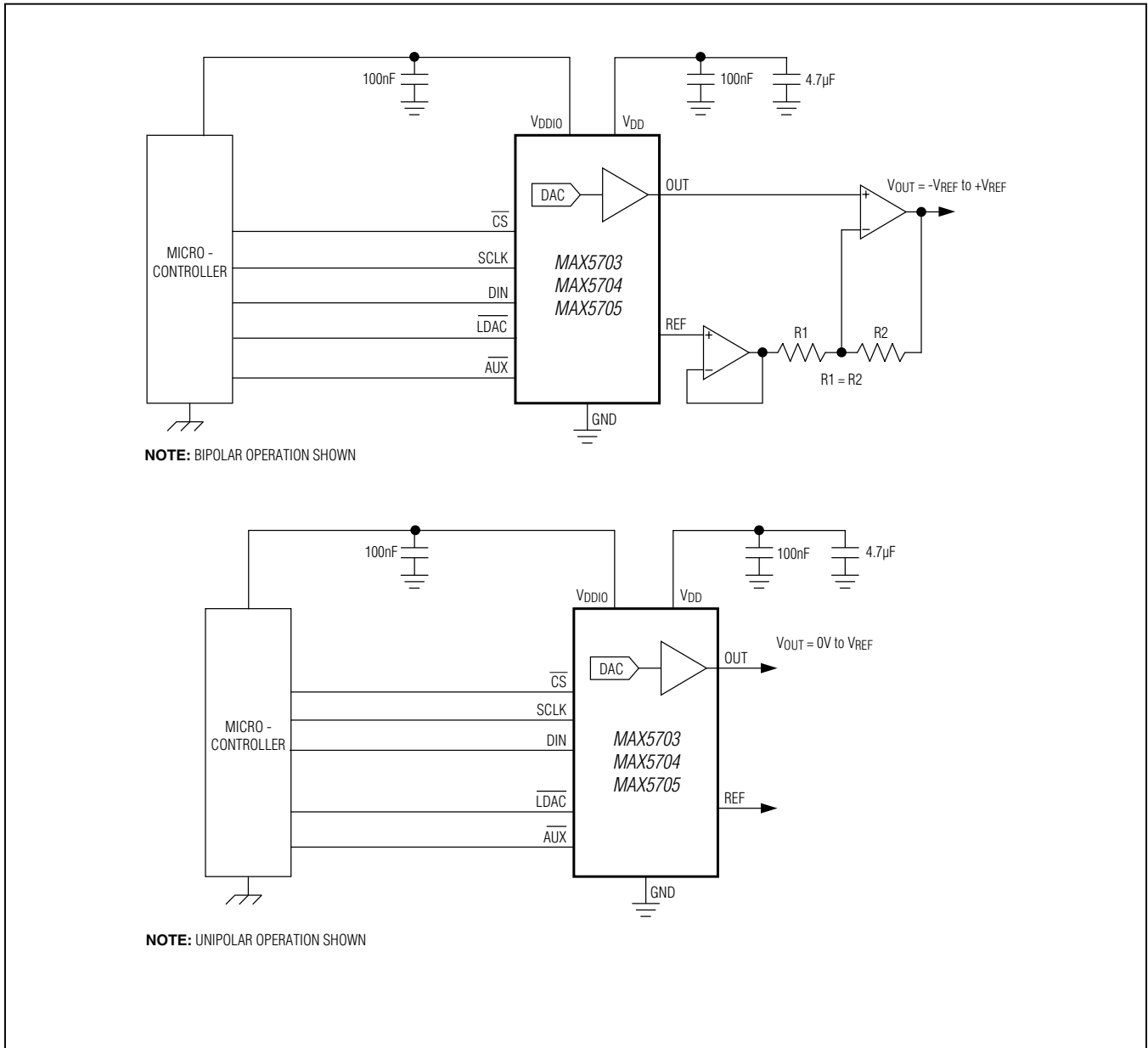
A major carry transition occurs at the midscale point where the MSB changes from low to high and all other bits change from high to low, or where the MSB changes from high to low and all other bits change from low to high. The duration of the magnitude of the switching glitch during a major carry transition is referred to as the digital-to-analog glitch impulse.

The digital-to-analog power-up glitch is the duration of the magnitude of the switching glitch that occurs as the device exits power-down mode.

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Typical Operating Circuits



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Ordering Information

PART	PIN-PACKAGE	RESOLUTION (BIT)	INTERNAL REFERENCE TEMPCO (ppm/°C)
MAX5703 ATB+T	10 TDFN-EP*	8	10 (typ), 25 (max)
MAX5703AUB+	10 μ MAX	8	10 (typ), 25 (max)
MAX5704 ATB+T	10 TDFN-EP*	10	10 (typ), 25 (max)
MAX5704AUB+	10 μ MAX	10	10 (typ), 25 (max)
MAX5705 AAUB+	10 μ MAX	12	4 (typ), 12 (max)
MAX5705BATB+T	10 TDFN-EP*	12	10 (typ), 25 (max)
MAX5705BAUB+	10 μ MAX	12	10 (typ), 25 (max)

Note: All devices are specified over the -40°C to +125°C temperature range.

+Denotes a lead(Pb)-free/RoHS-compliant package.

T = Tape and reel.

*EP = Exposed pad.

Chip Information

PROCESS: BiCMOS

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
10 TDFN-EP	T1032N+1	21-0429	90-0082
10 μ MAX	U10+2	21-0061	90-0330

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	11/12	Initial release	—
1	2/13	Released MAX5703/MAX5704. Updated the <i>Electrical Characteristics</i> .	2–8, 25
2	6/13	Released the MAX5703/MAX5704/MAX5705 TDFN packages.	25
3	11/14	Added details to $\overline{\text{AUX}}$ input description	14, 15, 22



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- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
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- Формирование склада под заказчика.
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- Тестирование поставляемой продукции.
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- Входной контроль качества.
- Наличие сертификата ISO.

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- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



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