

TOSHIBA BiCD Integrated Circuit Silicon Monolithic

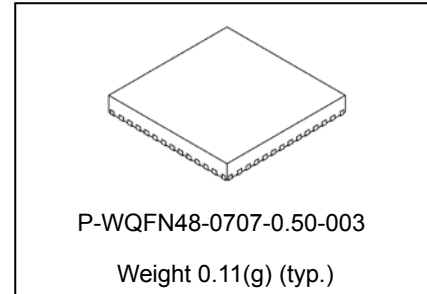
TB67S209FTG

CLOCK-in controlled Bipolar Stepping Motor Driver

The TB67S209FTG is a two-phase bipolar stepping motor driver using a PWM chopper. The clock in decoder is built in. Fabricated with the BiCD process, rating is 50 V/4.0 A.

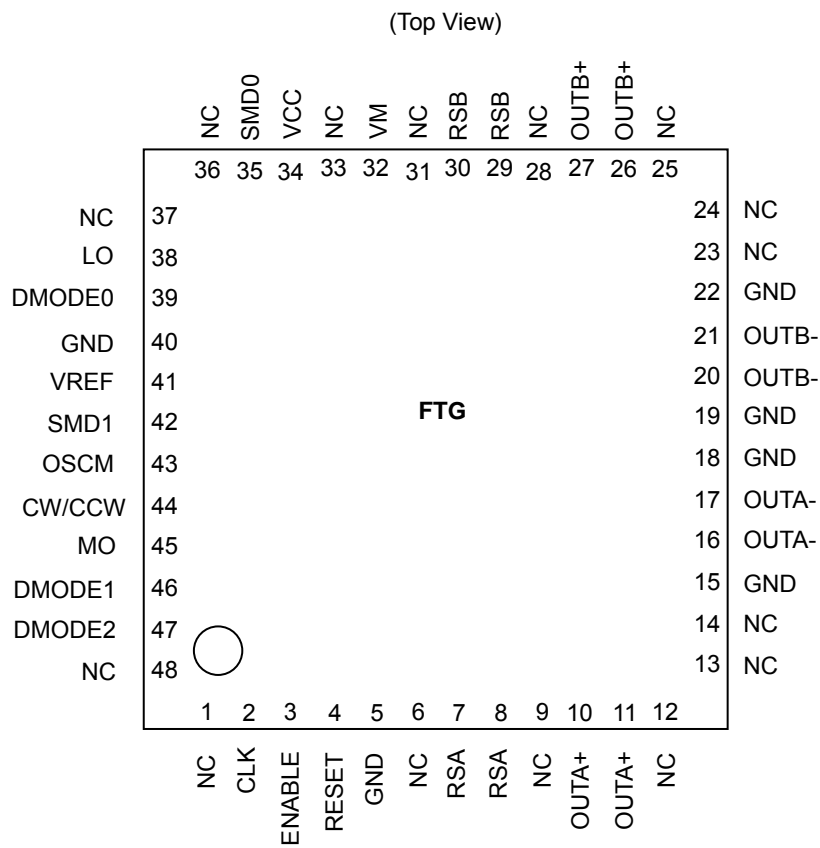
Features

- BiCD process integrated monolithic IC.
- Capable of controlling 1 bipolar stepping motor.
- PWM controlled constant-current drive.
- Allows full, half(a)(b), quarter, 1/8, 1/16, and 1/32 step operation.
- Low on-resistance (High + Low side=0.49Ω(typ.)) MOSFET output stage.
- Selectable Mixed Decay mode
- High voltage and current (For specification, please refer to absolute maximum ratings and operation ranges)
- Built-in error detection (TSD/ISD) signal output function
- Built-in error detection circuits (Thermal shutdown (TSD), over-current shutdown (ISD), and under voltage lock out (UVLO))
- Built-in VCC regulator for internal circuit use.
- Chopping frequency of a motor can be customized by external resistance and capacitor.
- TB67S209FTG: P-WQFN48-0707-0.50-003



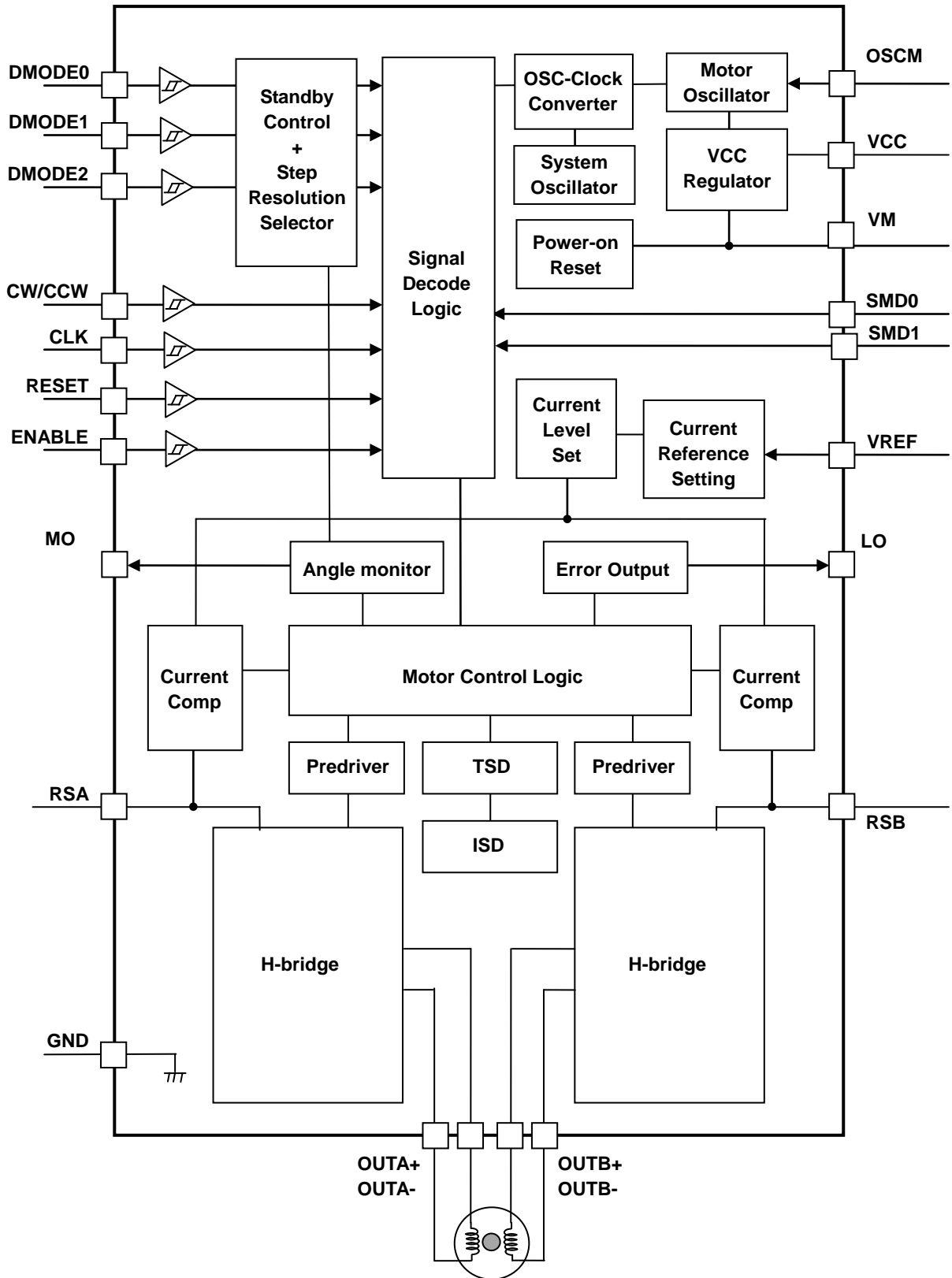
Note: Please be careful about thermal conditions during use.

Pin assignment (TB67S209FTG)



* Please mount the four corner pins of the QFN package and the exposed pad to the GND area of the PCB.

TB67S209 Block diagram



Functional blocks/circuits/constants in the block chart etc. may be omitted or simplified for explanatory purposes.

Application Notes

All the grounding wires of the TB67S209FTG must run on the solder mask on the PCB and be externally terminated at only one point. Also, a grounding method should be considered for efficient heat dissipation.

Careful attention should be paid to the layout of the output, VM and GND traces, to avoid short circuits across output pins or to the power supply or ground. If such a short circuit occurs, the device may be permanently damaged.

Also, the utmost care should be taken for pattern designing and implementation of the device since it has power supply pins (VM, RS, OUT, GND) through which a particularly large current may run. If these pins are wired incorrectly, an operation error may occur or the device may be destroyed.

The logic input pins must also be wired correctly. Otherwise, the device may be damaged owing to a current running through the IC that is larger than the specified current.

Pin explanations

TB67S209FTG (QFN48)

Pin No.1 – 28

Pin No.	Pin Name	Function
1	NC	Non-connection pin
2	CLK	CLK signal input pin
3	ENABLE	Ach/Bch output stage ON/OFF control pin
4	RESET	Electric angle reset pin
5	GND	Ground pin
6	NC	Non-connection pin
7	RSA (*)	Motor Ach current sense pin
8	RSA (*)	Motor Ach current sense pin
9	NC	Non-connection pin
10	OUTA+ (*)	Motor Ach (+) output pin
11	OUTA+ (*)	Motor Ach (+) output pin
12	NC	Non-connection pin
13	NC	Non-connection pin
14	NC	Non-connection pin
15	GND	Ground pin
16	OUTA- (*)	Motor Ach (-) output pin
17	OUTA- (*)	Motor Ach (-) output pin
18	GND	Ground pin
19	GND	Ground pin
20	OUTB- (*)	Motor Bch (-) output pin
21	OUTB- (*)	Motor Bch (-) output pin
22	GND	Ground pin
23	NC	Non-connection pin
24	NC	Non-connection pin
25	NC	Non-connection pin
26	OUTB+ (*)	Motor Bch (+) output pin
27	OUTB+ (*)	Motor Bch (+) output pin
28	NC	Non-connection pin

Pin No.29 – 48

Pin No.	Pin Name	Function
29	RSB (*)	Motor Bch current sense pin
30	RSB (*)	Motor Bch current sense pin
31	NC	Non-connection pin
32	VM	Motor power supply pin
33	NC	Non-connection pin
34	VCC	Internal VCC regulator monitor pin
35	SMD0	Selectable Mixed Decay setting pin 0
36	NC	Non-connection pin
37	NC	Non-connection pin
38	LO	Error detect signal output pin
39	DMODE0	Step resolution set pin no.0
40	GND	Ground pin
41	VREF	Motor current threshold setting pin
42	SMD1	Selectable Mixed Decay setting pin 1
43	OSCM	Oscillating circuit frequency for chopping set pin
44	CW/CCW	Motor rotation direction set pin
45	MO	Electric angle monitor pin
46	DMODE1	Step resolution set pin no.1
47	DMODE2	Step resolution set pin no.2
48	NC	Non-connection pin

Note:

- Please use the pin of NC with Open.
- (*) Two or more pins with the same name should be short-circuit near the pin.

INPUT/OUTPUT equivalent circuit (TB67S209)

Pin name	IN/OUT signal	Equivalent circuit
DMODE0 DMODE1 DMODE2 CLK ENABLE RESET CW/CCW SMD0 SMD1	Digital Input (VIH/VIL) VIH: 2.0V(min) to 5.5V(max) VIL: 0V(min) to 0.8V(max)	
LO MO	Digital Output (VOH/VOL) (Pullup resistance: 10k to 100kΩ)	
VCC VREF	VCC voltage range 4.75V(min) to 5.0V(typ.) to 5.25V(max) VREF voltage range 0V to 3.6V	
OSCM	OSCM frequency setting range 0.64MHz(min) to 1.12MHz(typ.) to 2.4MHz(max)	
OUTA+ OUTA- OUTB+ OUTB- RSA RSB	VM power supply voltage range 10V(min) to 47V(max) OUTPUT pin voltage 10V(min) to 47V(max)	

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

Function explanation (Stepping motor mode)

1. CLK Function

Each up-edge of the CLK signal will shift the motor's electrical angle per step.

CLK Input	Function
Up-edge	Shifts the electrical angle per step.
Down-edge	—(State of the electrical angle does not change.)

2. ENABLE function

The ENABLE pin controls the ON and OFF of the corresponding output stage. This pin serves to select if the motor is stopped in OFF mode (High impedance) or activated. Please set the ENABLE pin to 'L' during VM power-on and power-off sequence.

ENABLE Input	Function
H	Output stage='ON' (Normal operation mode)
L	Output stage='OFF' (High impedance mode)

3. CW/CCW function and the output pin function (Output logic at the time of a charge start)

The CW/CCW pin controls the rotation direction of the motor. When set to 'Clockwise', the current of OUTA is output first, with a phase difference of 90°. When set to 'Counter clockwise', the current of OUTB is output first with a phase difference of 90°.

CW/CCW Input	OUT (+)	OUT (-)
H: Clockwise operation(CW)	H	L
L: Counter clockwise operation(CCW)	L	H

4. Step resolution select function

DMODE0	DMODE1	DMODE2	Function
L	L	L	Standby mode (the OSCM is disabled and the output stage is set to 'OFF' status)
L	L	H	Full step resolution
L	H	L	Half step resolution(Type A)
L	H	H	Quarter step resolution
H	L	L	Half step resolution(Type B)
H	L	H	1/8 step resolution
H	H	L	1/16 step resolution
H	H	H	1/32 step resolution

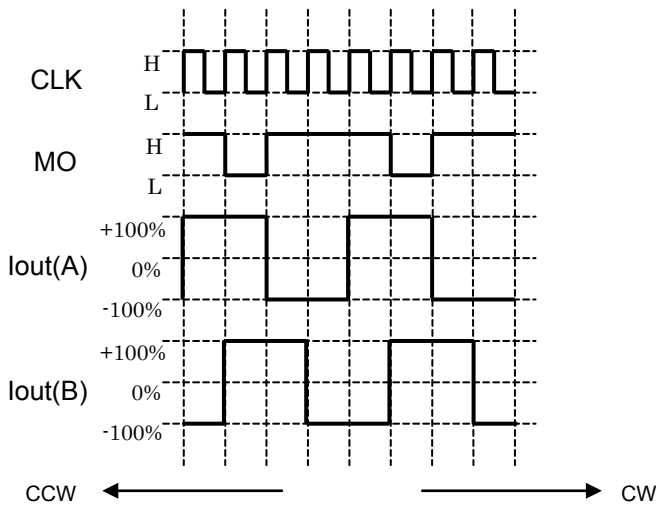
When switching the DMODE0,1,2; setting the RESET signal to Low (will set the electrical angle to the initial status), is recommended.

5. Selectable Mixed Decay function

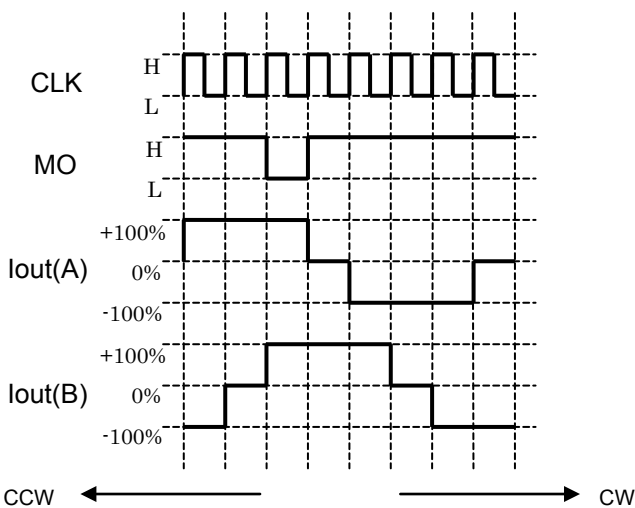
SMD0	SMD1	Function
L	L	Fast Decay: 12.5% (Fast Decay = OSCM x 2)
L	H	Fast Decay: 37.5% (Fast Decay = OSCM x 6)
H	L	Fast Decay: 50% (Fast Decay = OSCM x 8)
H	H	Fast Decay only

Step resolution setting and initial angle

[Full step resolution]



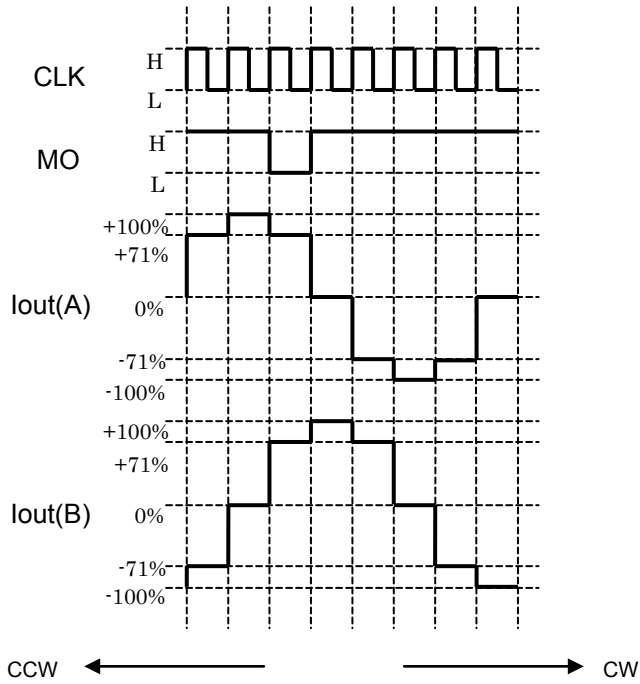
[Half step resolution (Type A)]



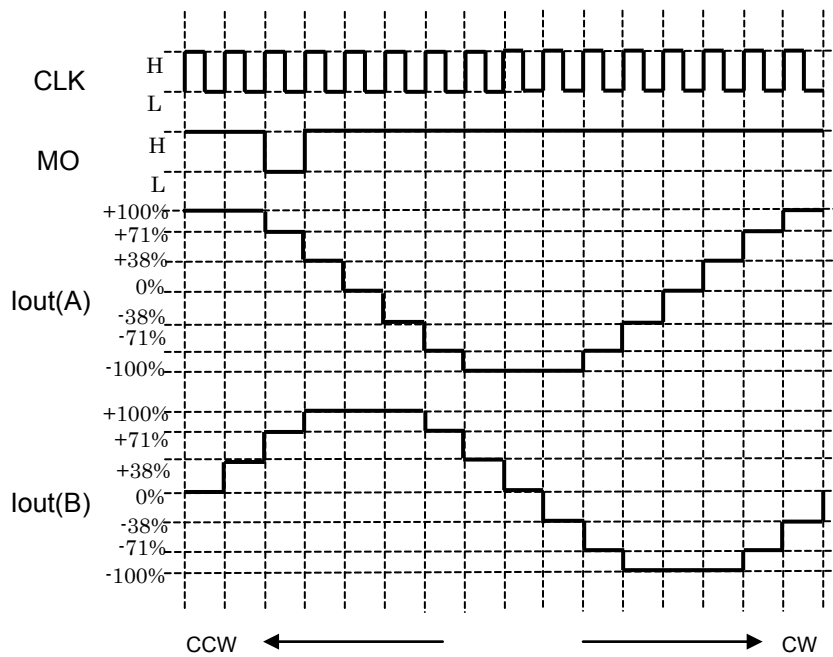
MO output shown in the timing chart is when the MO pin is pulled up.

Timing charts may be simplified for explanatory purpose.

[Half step resolution (Type B)]



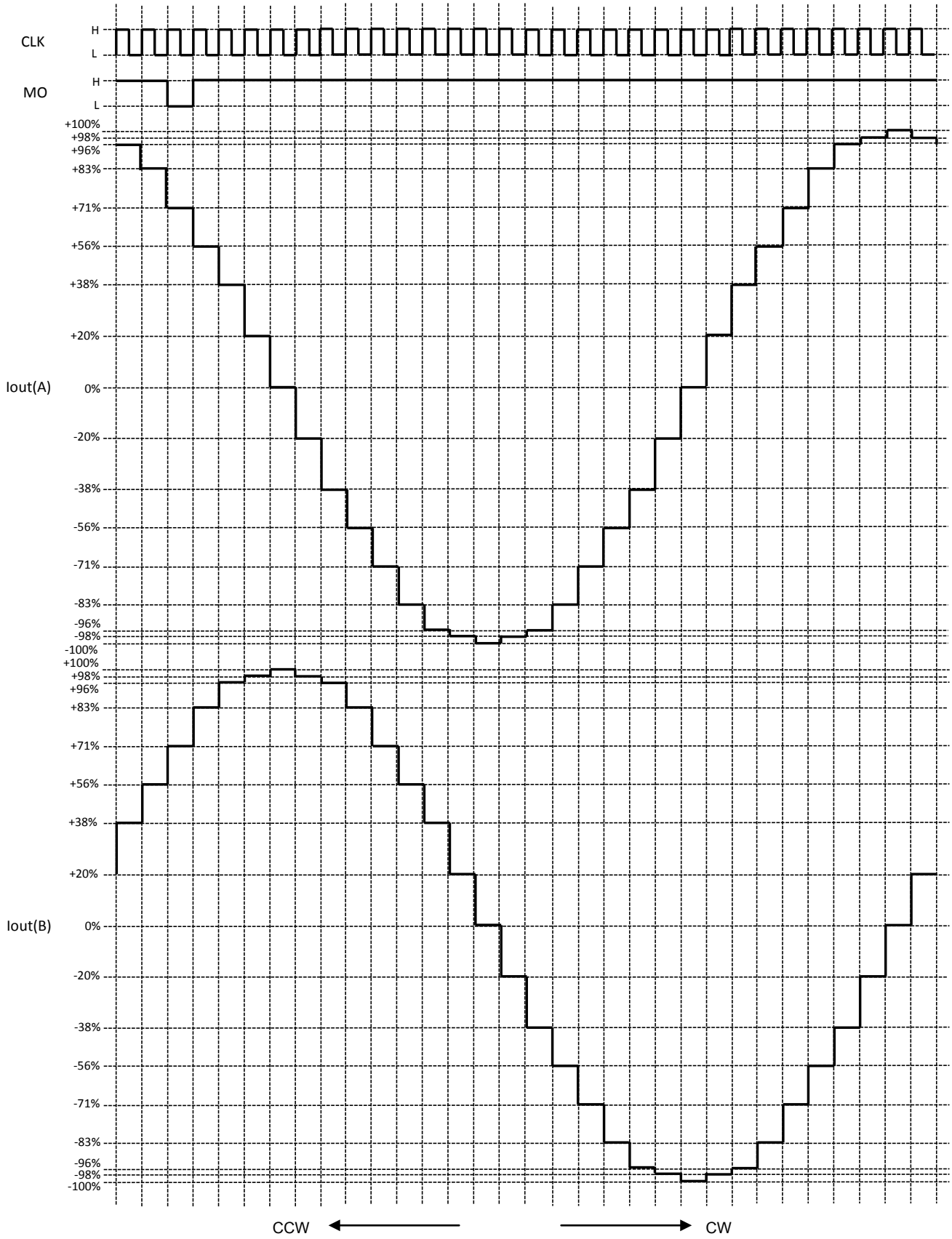
[Quarter step resolution]



MO output shown in the timing chart is when the MO pin is pulled up.

Timing charts may be simplified for explanatory purpose.

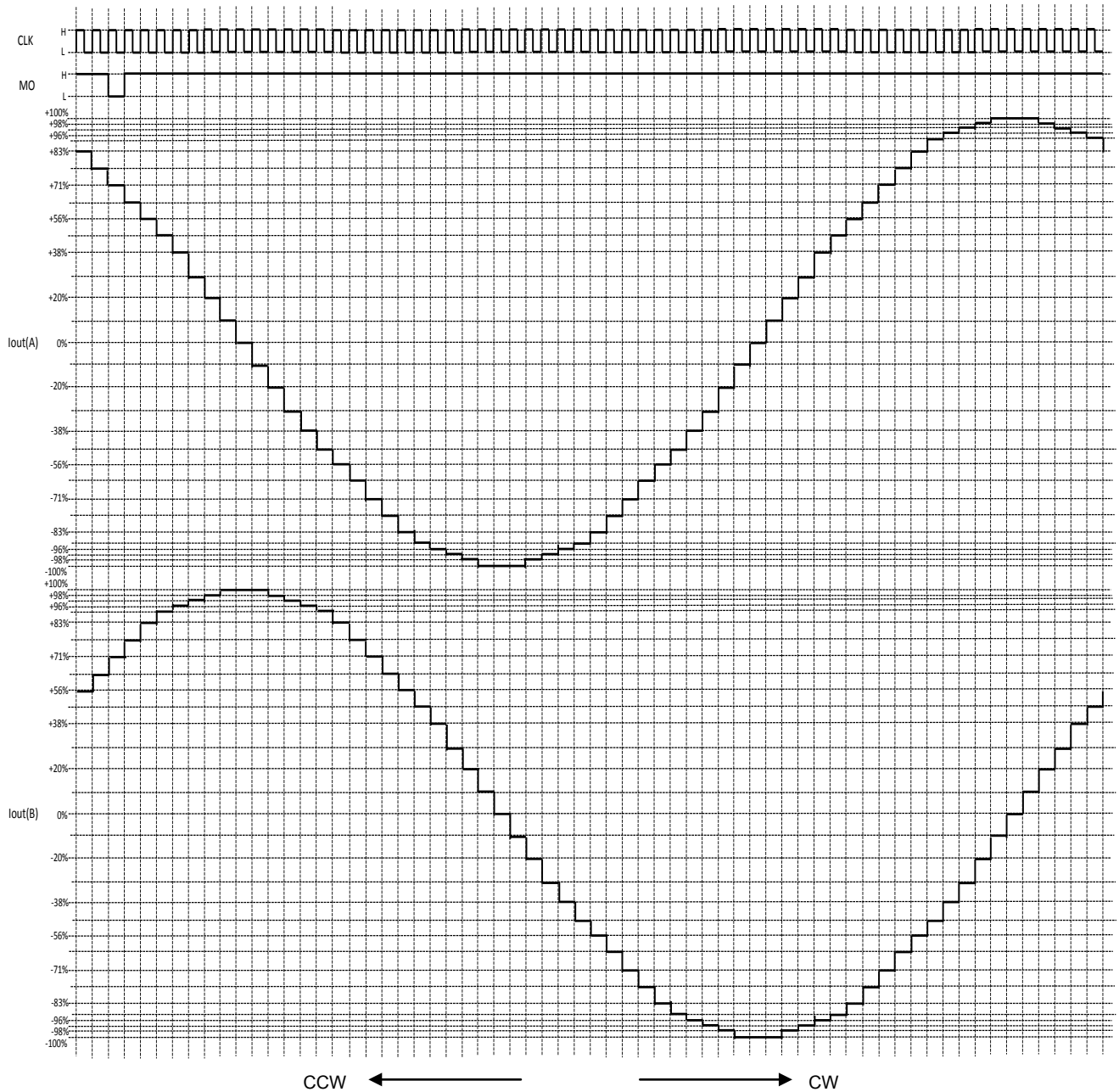
[1/8 step resolution]



MO output shown in the timing chart is when the MO pin is pulled up.

Timing charts may be simplified for explanatory purpose.

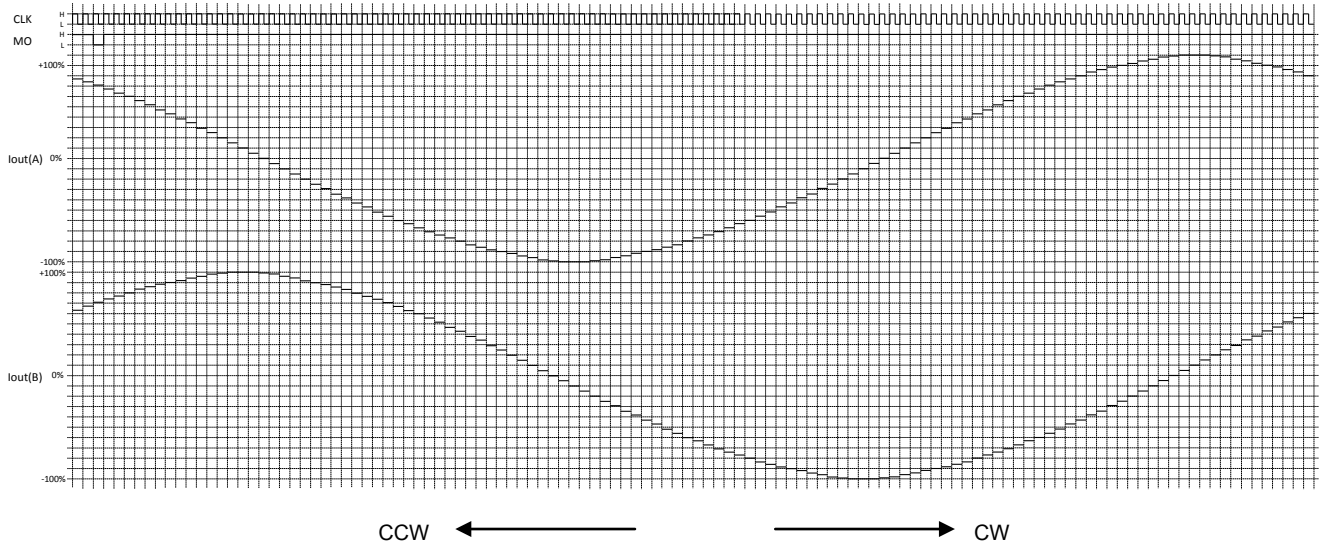
[1/16 step resolution]



MO output shown in the timing chart is when the MO pin is pulled up.

Timing charts may be simplified for explanatory purpose.

[1/32 step resolution]



MO output shown in the timing chart is when the MO pin is pulled up.

Timing charts may be simplified for explanatory purpose.

Step setting and current percentage

Current [%]	Full	Half (A)	Half (B)	Quarter	1/8	1/16	1/32
100%	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
99%						<input type="radio"/>	<input type="radio"/>
98%					<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
97%						<input type="radio"/>	<input type="radio"/>
96%					<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
94%							<input type="radio"/>
92%							<input type="radio"/>
90%						<input type="radio"/>	<input type="radio"/>
88%							<input type="radio"/>
86%							<input type="radio"/>
83%					<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
80%							<input type="radio"/>
77%						<input type="radio"/>	<input type="radio"/>
74%							<input type="radio"/>
71%			<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
67%							<input type="radio"/>
63%						<input type="radio"/>	<input type="radio"/>
60%							<input type="radio"/>
56%					<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
52%							<input type="radio"/>
47%						<input type="radio"/>	<input type="radio"/>
43%							<input type="radio"/>
38%				<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
34%							<input type="radio"/>
29%						<input type="radio"/>	<input type="radio"/>
25%							<input type="radio"/>
20%					<input type="radio"/>	<input type="radio"/>	<input type="radio"/>
15%							<input type="radio"/>
10%						<input type="radio"/>	<input type="radio"/>
5%							<input type="radio"/>
0%		<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>	<input type="radio"/>

6. RESET function

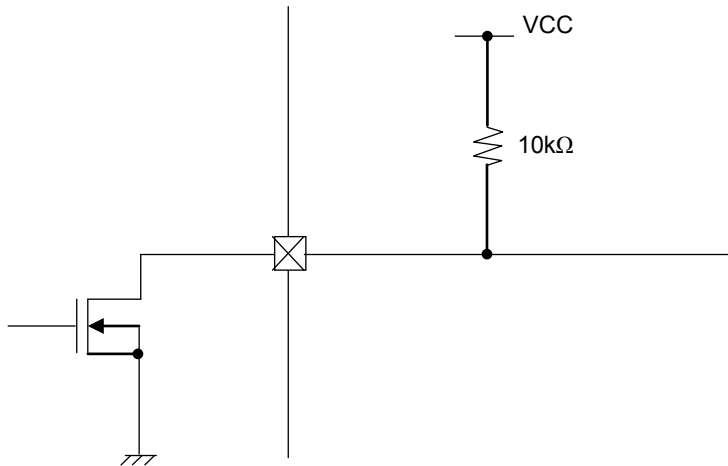
RESET Input	Function
H	Sets the electrical angle to the initial condition.
L	Normal operation mode

The current for each channel (while RESET is applied) is shown in the table below. MO will show 'L' at this time.

Step resolution setting	Ach current setting	Bch current setting	Default electrical angle
Full step	100%	100%	45°
Half step (Type A)	100%	100%	45°
Half step (Type B)	71%	71%	45°
Quarter step	71%	71%	45°
1/8 step	71%	71%	45°
1/16 step	71%	71%	45°
1/32 step	71%	71%	45°

7. LO(Error detect signal) output function

When Thermal shutdown(TSD) or Over-current shutdown(ISD) is applied, the LO voltage will be switched to Low(GND) level.

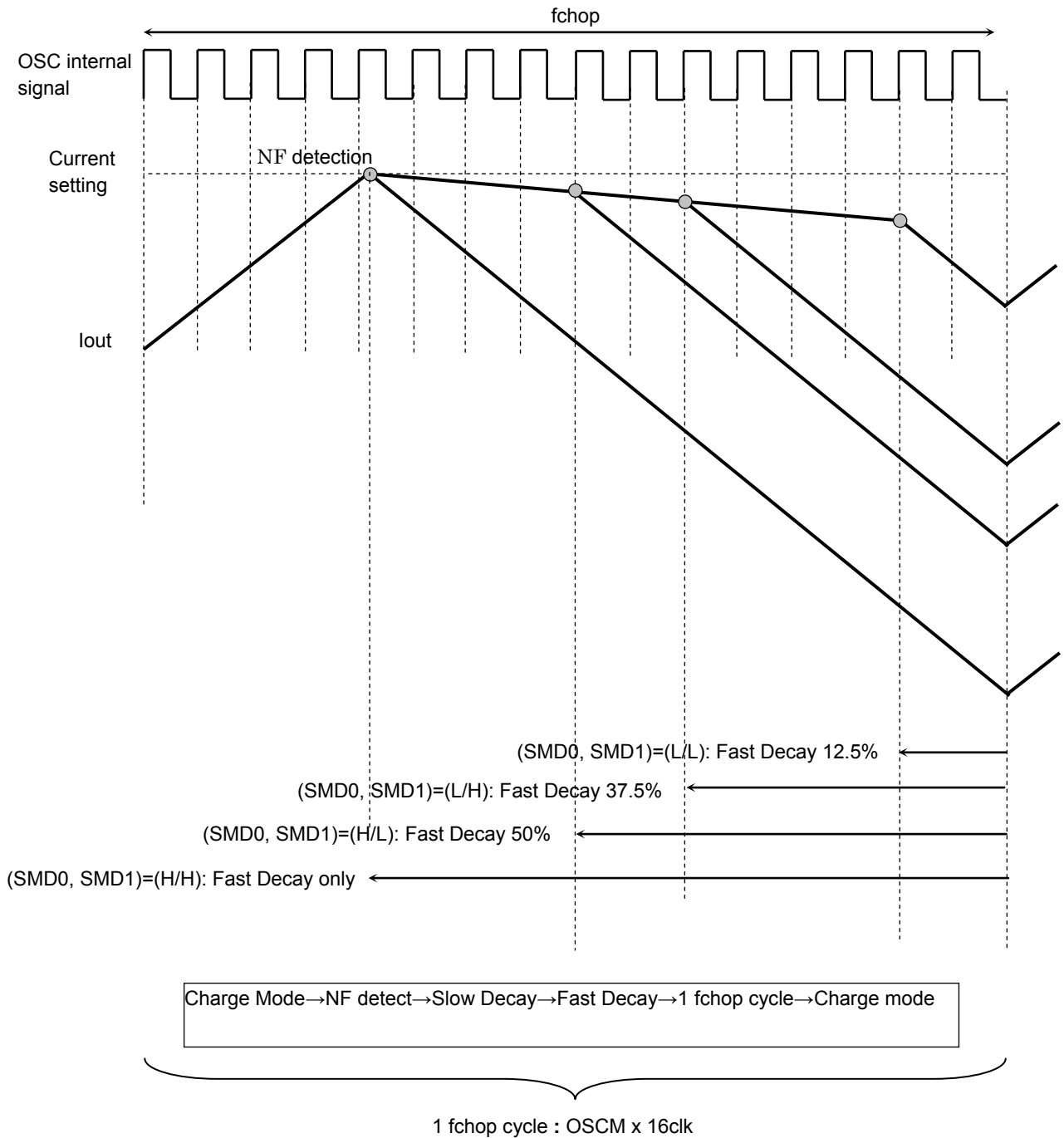


The LO is an open-drain output pin. The LO pin needs to be pulled up to 3.3V/5.0V level for proper function. During regular operation, the LO pin level will stay High (internal MOSFET=OFF). When error detection (TSD, ISD) is applied, the LO pin will show Low (internal MOSFET=ON) level.

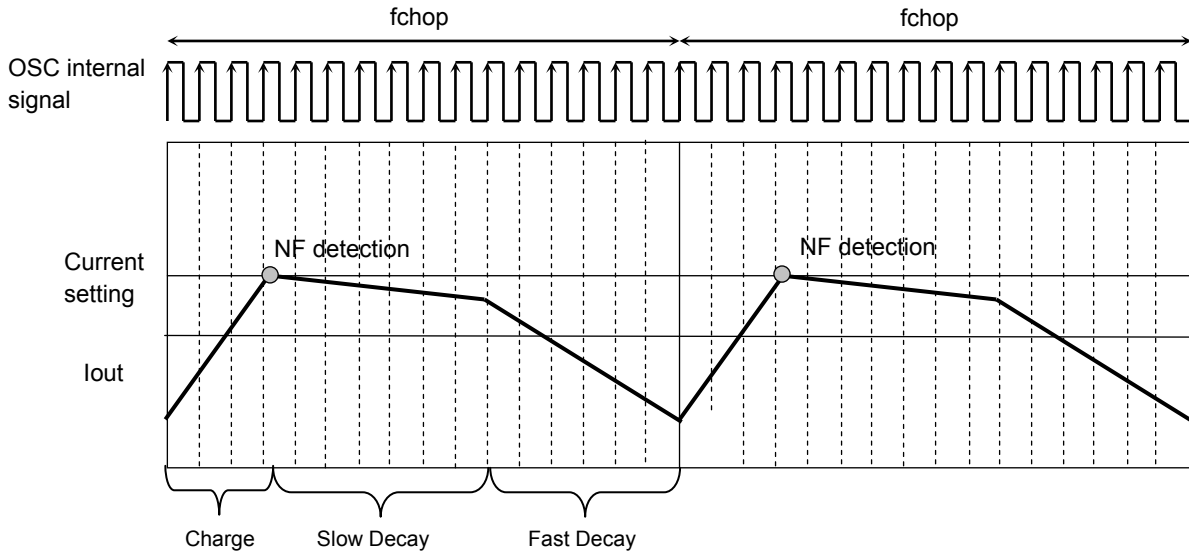
Selectable Mixed Decay function

The Selectable Mixed Decay can adjust the current regeneration amount during the period of current regeneration (Decay) using pins.

Though the Mixed Decay is determined by controlling 2 different types of decay (Fast Decay and Slow Decay), this function enables the user to select the ratio of the Mixed Decay using SMD0 and SMD1 pin. (2bit, 4 function)

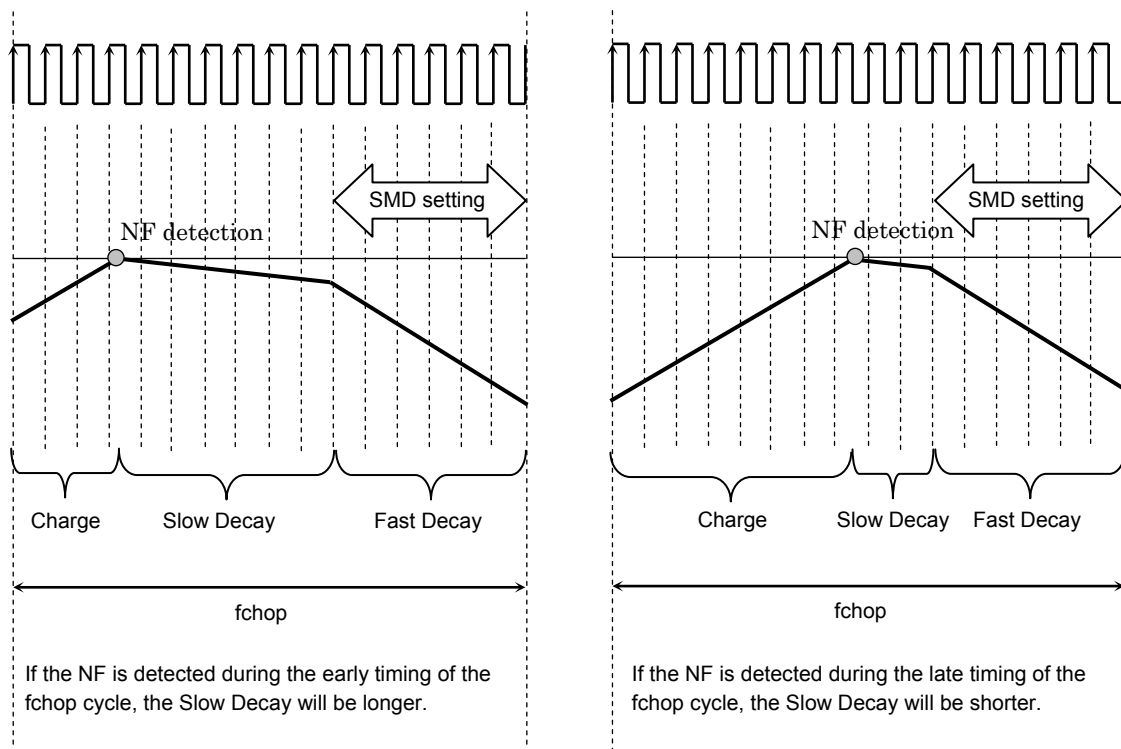


Mixed Decay waveform (current waveform)



Timing charts may be omitted for explanatory purpose.

Constant current PWM function and timings



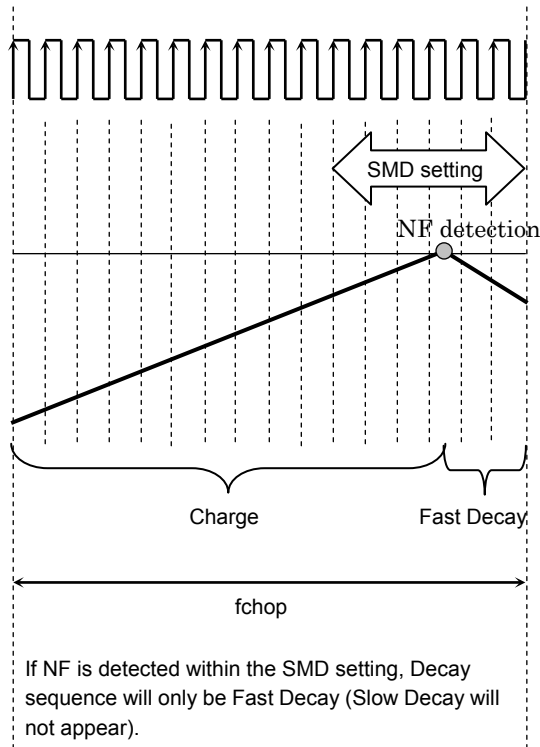
If the NF is detected during the early timing of the fchop cycle, the Slow Decay will be longer.

If the NF is detected during the late timing of the fchop cycle, the Slow Decay will be shorter.

The Charge period is determined by the operating status. Therefore the NF detect timing with in the chopping cycle will change. If NF is detected in the early period of the fchop cycle, the Slow Decay will be longer. If NF is detected in the late period of the fchop cycle, the Slow Decay will be shorter, as shown above.

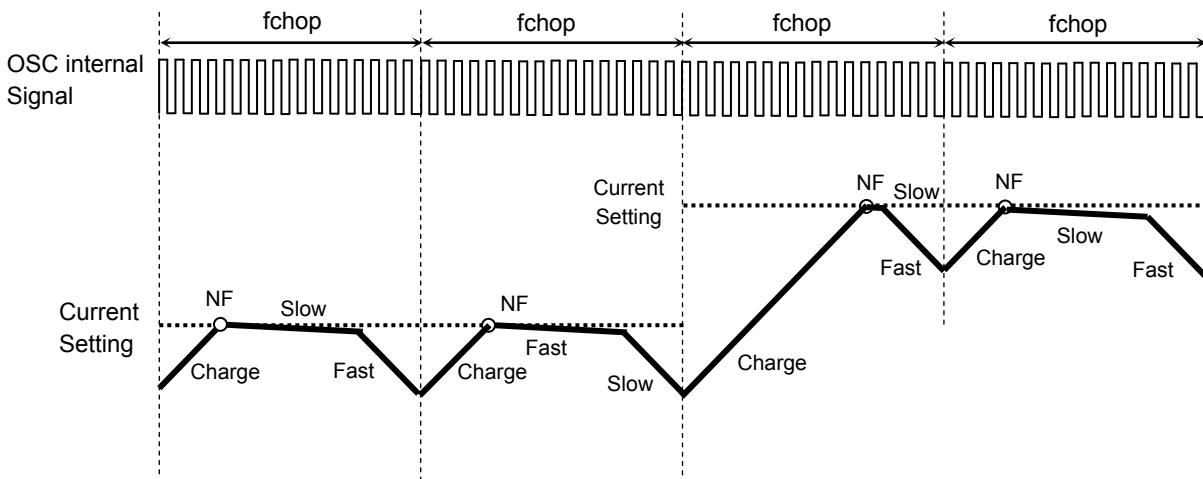
Note: The chopping cycle is determined as: $fchop - (\text{Charge} + \text{Fast decay}) = \text{Slow Decay}$
 (Fast Decay ratio can be changed by SMD0 and SMD1 setting.)

Constant current PWM function and timings

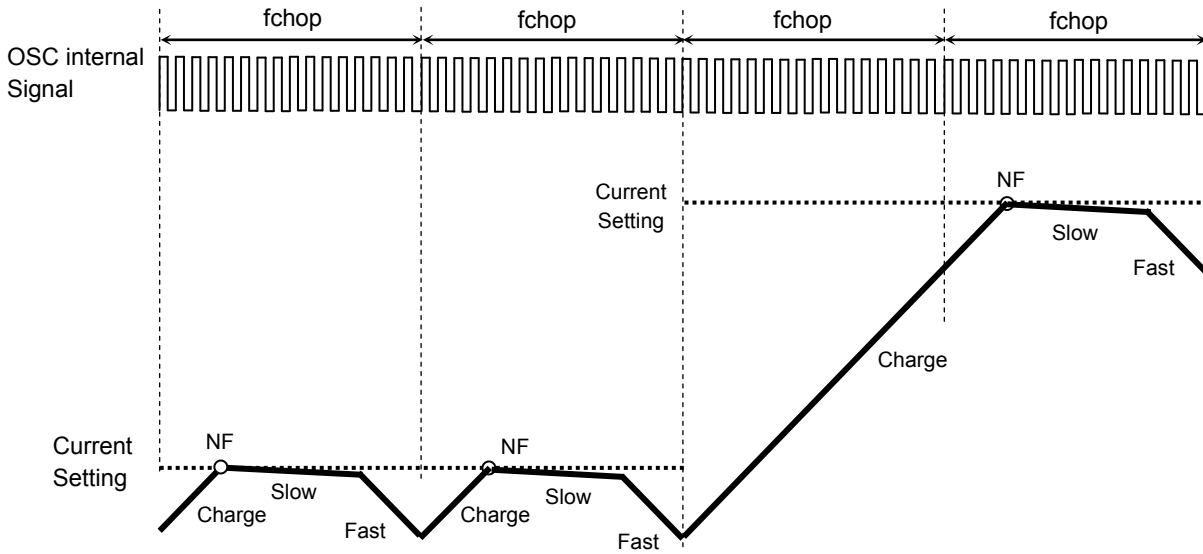


Mixed Decay current waveform

•When the next current step is higher :

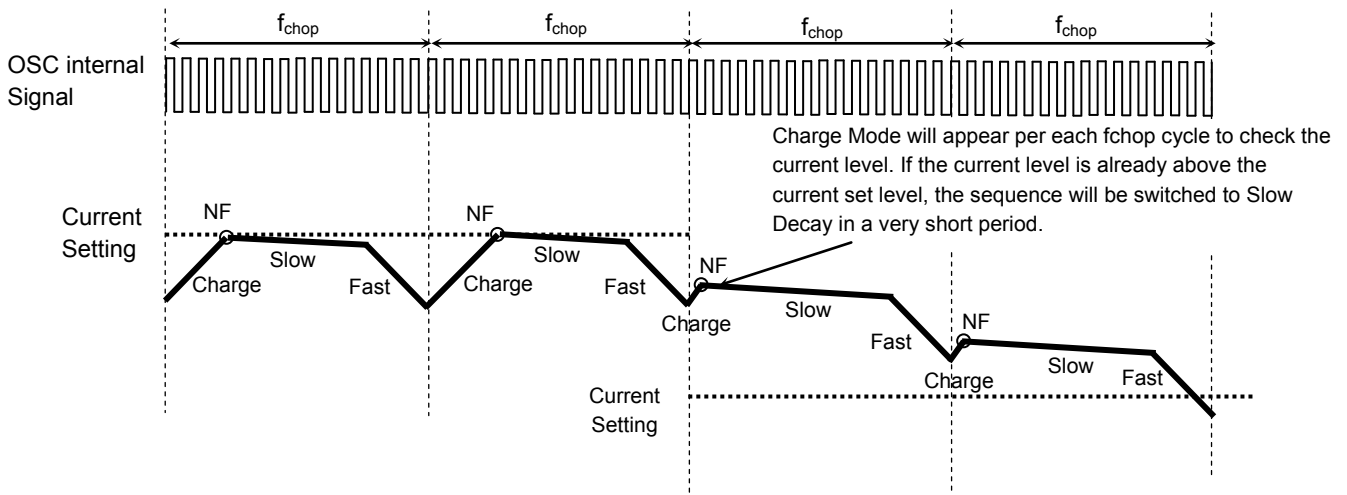


•When Charge period is more than 1 fchop cycle :

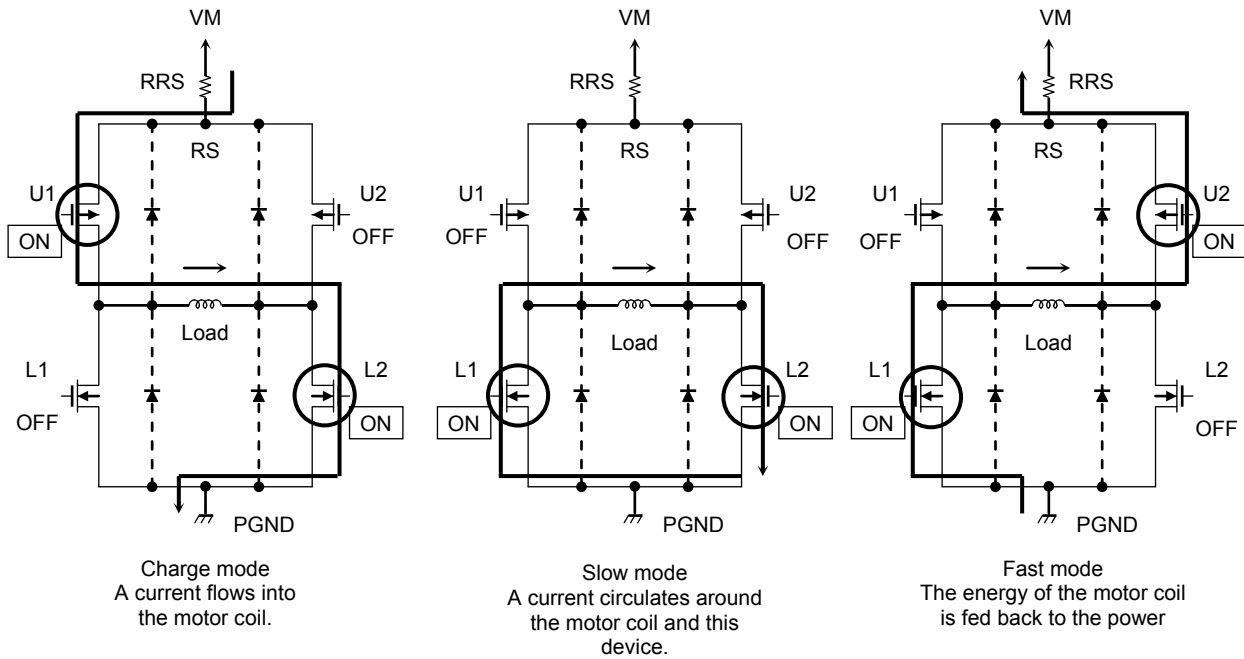


When the Charge period is longer than f_{chop} cycle, the Charge period will be extended until the motor current reaches the NF threshold. Once the current reaches the next current step, then the sequence will go on to decay mode.

•When the next current step is lower :



Output MOSFET function



Output MOSFET function

MODE	U1	U2	L1	L2
CHARGE	ON	OFF	OFF	ON
SLOW	OFF	OFF	ON	ON
FAST	OFF	ON	ON	OFF

Note: This table shows an example of when the current flows as indicated by the arrows in the figures shown above. If the current flows in the opposite direction, refer to the following table.

MODE	U1	U2	L1	L2
CHARGE	OFF	ON	ON	OFF
SLOW	OFF	OFF	ON	ON
FAST	ON	OFF	OFF	ON

This IC controls the motor current to be constant by 3 modes listed above. The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

Calculation of the Predefined Output Current

For PWM constant-current control, this IC uses a clock generated by the OSCM oscillator.

The peak output current (Setting current value) can be set via the current-sensing resistor (RS) and the reference voltage (Vref), as follows:

$$I_{out(max)} = V_{ref(gain)} \times \frac{V_{ref(V)}}{R_{RS}(\Omega)}$$

Vref(gain): the Vref decay rate is 1/ 5.0 (typ.)

For example: In the case of a 100% setup

when Vref = 3.0 V, Torque=100%, RS=0.51Ω, the motor constant current (Setting current value) will be calculated as:

$$I_{out} = 3.0V / 5.0 / 0.51\Omega = 1.18 \text{ A}$$

Calculation of the OSCM oscillation frequency (chopper reference frequency)

The OSCM oscillation frequency (fOSCM) and chopper frequency (fchop) can be calculated by the following expressions.

$$f_{OSCM} = 1 / [0.56 \times \{C \times (R1 + 500)\}]$$

.....C,R1: External components for OSCM (C=270pF , R1=5.1kΩ => About fOSCM= 1.12MHz(Typ.))

$$f_{chop} = f_{OSCM} / 16$$

.....fOSCM=1.12MHz => fchop =About 70kHz

If chopping frequency is raised, Rippl of current will become small and wave-like reproducibility will improve. However, the gate loss inside IC goes up and generation of heat becomes large.

By lowering chopping frequency, reduction in generation of heat is expectable. However, Rippl of current may become large. It is a standard about 70 kHz. A setup in the range of 50 to 100 kHz is recommended.

Absolute Maximum Ratings (Ta = 25°C)

Characteristics	Symbol	Rating	Unit	Remarks
Motor power supply	VM	50	V	—
Motor output voltage	Vout	50	V	—
Motor output current	Iout	4.0	A	(Note 1)
Internal Logic power supply	VCC	6.0	V	When externally applied.
Logic input voltage	VIN(H)	6.0	V	—
	VIN(L)	-0.4	V	—
MO output voltage	VMO	6.0	V	—
LO output voltage	VLO	6.0	V	—
MO Inflow current	IMO	30	mA	—
LO Inflow current	ILO	30	mA	—
Power dissipation	QFN48 PD	1.3	W	(Note 2)
Operating temperature	TOPR	-20 to 85	°C	—
Storage temperature	TSTR	-55 to 150	°C	—
Junction temperature	Tj(max)	150	°C	—

Note 1: Usually, the maximum current value at the time should use 70% or less of the absolute maximum ratings for a standard on thermal rating. The maximum output current may be further limited in view of thermal considerations, depending on ambient temperature and board conditions.

Note 2: Device alone (Ta =25°C) If the ambient temperature is above 25°C, derate the PD by 10.4mW/°C.

Ta: Ambient temperature

Topr: Ambient temperature while the IC is active

Tj: Junction temperature while the IC is active. Tj(max) is limited by the thermal shutdown (TSD) circuitry. Within the limit of a Tj(max) 120°C, it recommends designing in consideration of the use maximum current.

Caution) Absolute maximum ratings

The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings. Exceeding the rating (s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.

The value of even one parameter of the absolute maximum ratings should not be exceeded under any circumstances. The TB67S209FTG does not have overvoltage detection circuit. Therefore, the device is damaged if a voltage exceeding its rated maximum is applied. All voltage ratings, including supply voltages, must always be followed. The other notes and considerations described later should also be referred to.

Operation Ranges (Ta=-20 to 85°C)

Characteristics	Symbol	Min	Typ.	Max	Unit	Remarks
Motor power supply	VM	10	24	47	V	—
Motor output current	Iout	—	1.5	3.0	A	(Note 1)
Logic input voltage	VIN(H)	2.0	—	5.5	V	Logic input High Level
	VIN(L)	0	—	0.8	V	Logic input Low Level
MO output pin voltage	VMO	—	3.3	5.0	V	—
LO output pin voltage	VLO	—	3.3	5.0	V	—
Clock input frequency	fCLK	—	—	100	kHz	—
Chopper frequency	fchop(range)	40	70	150	kHz	—
Vref input voltage	Vref	GND	2.0	3.6	V	—

Note 1: Maximum current for actual usage may be limited by the operating circumstances such as operating conditions (exciting mode, operating time, and so on), ambient temperature, and heat conditions (board condition and so on). Confirm the maximum current value which can actually be used on the thermal rating under the operation environment.

Electrical Specifications 1 (Ta = 25°C, VM = 24 V, unless specified otherwise)

Characteristics		Symbol	Test condition	Min	Typ.	Max	Unit
Logic input voltage	HIGH	VIN(H)	Logic input (Note)	2.0	—	5.5	V
	LOW	VIN(L)	Logic input (Note)	0	—	0.8	V
Logic input hysteresis voltage		VIN(HYS)	Logic input (Note)	100	—	300	mV
Logic input current	HIGH	IIN(H)	VIN(H)=3.3V	—	33	—	μA
	LOW	IIN(L)	VIN(L)=0V	—	—	1	μA
MO output pin voltage	LOW	VOL(MO)	IOL=24mA output=Low	—	0.2	0.5	V
LO output pin voltage	LOW	VOL(LO)	IOL=24mA output=Low	—	0.2	0.5	V
Consumption current		IM1	Output pins=open Standby mode	—	2	3.5	mA
		IM2	Output pins=open Standby release ENABLE=Low	—	3.5	5.5	mA
		IM3	Output pins=open Full step resolution	—	5.5	7	mA
Output leakage current	High-side	IOH	VRS=VM=50V, Vout=0V	—	—	1	μA
	Low-side	IOL	VRS=VM=Vout=50V	1	—	—	μA
Motor current channel differential		ΔIout1	Vref=1.5V, R=0.2Ω (Iout=1.5A)	-5	0	5	%
Motor current setting accuracy		ΔIout2	Vref=1.5V, R=0.2Ω (Iout=1.5A)	-5	0	5	%
RS pin current		IRS	VRS=VM=24V	0	—	10	μA
Motor output ON-resistance (High-side+Low-side)		Ron(H+L)	Tj=25°C, Forward direction (High-side+Low-side)	—	0.49	0.6	Ω

Note: VIN (H) is defined as the VIN voltage that causes the outputs (OUTA, OUTB) to change when a pin under test is gradually raised from 0 V. VIN (L) is defined as the VIN voltage that causes the outputs (OUTA, OUTB) to change when the pin is then gradually lowered. The difference between VIN (H) and VIN (L) is defined as the VIN (HYS).

Note: Even if the logic input signal is input in the state of no VM voltage supply, the circuit is designed so that the EMF and leak current by a signal input are not generated. But for safe usage, please apply the logic signal after the VM power supply is asserted and the VM voltage reaches the proper operating range.

Electrical Specifications 2 (Ta =25°C, VM = 24 V, unless specified otherwise)

Characteristics	Symbol	Test condition	Min	Typ.	Max	Unit
Vref input current	Iref	Vref=2.0V	—	0	1	μA
VCC voltage	VCC	ICC=5.0mA	4.75	5.0	5.25	V
VCC current	ICC	VCC=5.0V	—	2.5	5	mA
Vref gain rate	Vref(gain)	Vref=1.5V	1/5.2	1/5.0	1/4.8	—
Thermal shutdown(TSD) threshold (Note1)	TjTSD	—	145	160	175	°C
VM recovery voltage	VMR	—	7.0	8.0	9.0	V
Over-current detection (ISD) threshold (Note2)	ISD	—	4.1	4.9	5.7	A

Note1: About TSD

When the junction temperature of the device reached the TSD threshold, the TSD circuit is triggered; the internal reset circuit then turns off the output transistors. Noise rejection blanking time is built-in to avoid misdetection. Once the TSD circuit is triggered, the device will be set to standby mode, and can be cleared by reasserting the VM power source, or setting the DMODE pins to standby mode. The TSD circuit is a backup function to detect a thermal error, therefore is not recommended to be used aggressively.

Note2: About ISD

When the output current reaches the threshold, the ISD circuit is triggered; the internal reset circuit then turns off the output transistors. The dead band time is set to avoid the incorrect operation by switching. When the ISD function is operating, the IC is set to the standby mode. After detecting overcurrent, it can be returned by re-starting the VM power supply or setting the standby mode with DMODE pins.

Back-EMF

While a motor is rotating, there is a timing at which power is fed back to the power supply. At that timing, the motor current recirculates back to the power supply due to the effect of the motor back-EMF.

If the power supply does not have enough sink capability, the power supply and output pins of the device might rise above the rated voltages. The magnitude of the motor back-EMF varies with usage conditions and motor characteristics. It must be fully verified that there is no risk that the TB67S209FTG or other components will be damaged or fail due to the motor back-EMF.

Cautions on Overcurrent Shutdown (ISD) and Thermal Shutdown (TSD)

The ISD and TSD circuits are only intended to provide temporary protection against irregular conditions such as an output short-circuit; they do not necessarily guarantee the complete IC safety.

If the device is used beyond the specified operating ranges, these circuits may not operate properly: then the device may be damaged due to an output short-circuit.

The ISD circuit is only intended to provide a temporary protection against an output short-circuit. If such a condition persists for a long time, the device may be damaged due to overstress. Overcurrent conditions must be removed immediately by external hardware.

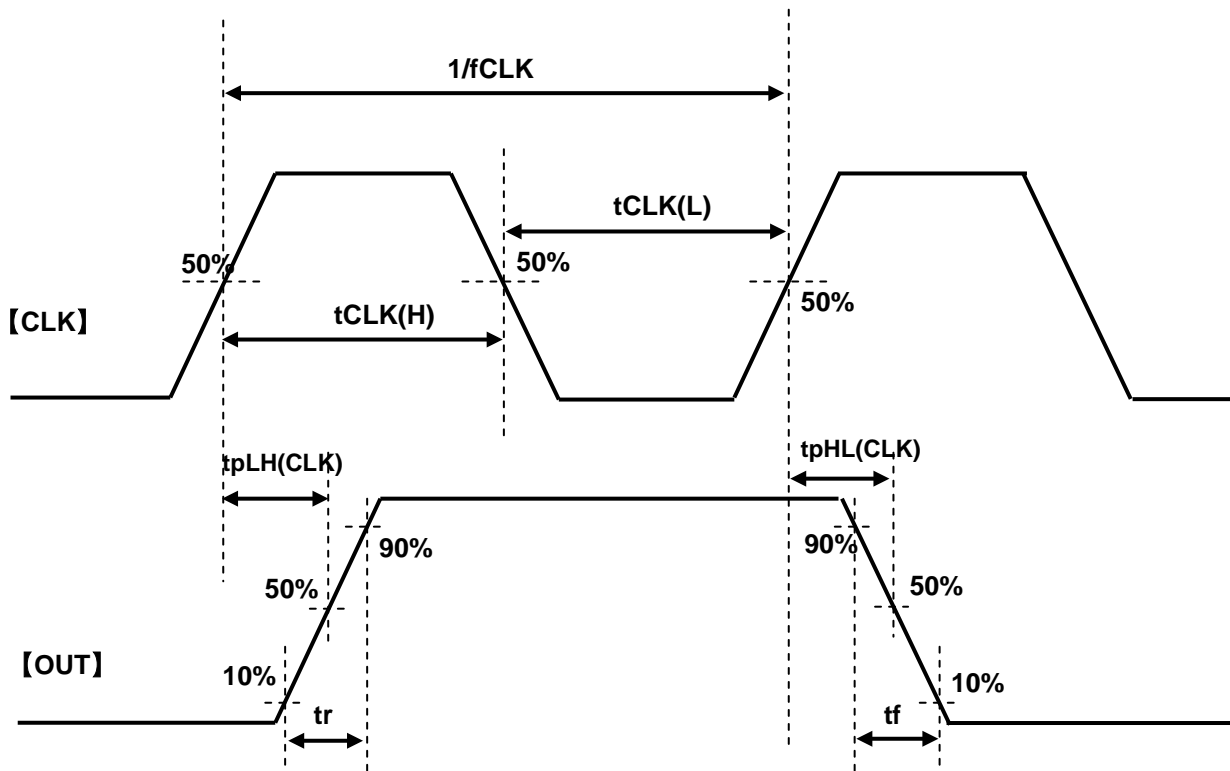
IC Mounting

Do not insert devices incorrectly or in the wrong orientation. Otherwise, it may cause breakdown, damage and/or deterioration of the device.

AC Electrical Specification (Ta = 25°C, VM = 24 V, 6.8 mH/5.7 Ω)

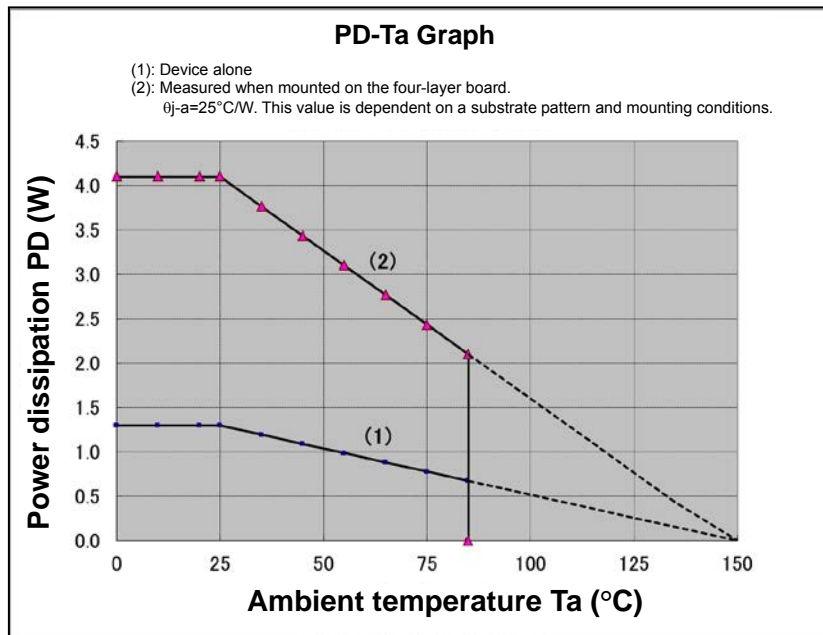
Characteristics	Symbol	Test condition	Min	Typ.	Max	Unit
Inside filter of CLK input minimum High width	tCLK(H)	The CLK(H) minimum pulse width	300	—	—	ns
Inside filter of CLK input minimum Low width	tCLK(L)	The CLK(L) minimum pulse width	250	—	—	ns
Output transistor switching specific	tr	—	30	80	130	ns
	tf	—	40	90	140	ns
	tpLH(CLK)	CLK-Output	—	1000	—	ns
	tpHL(CLK)	CLK-Output	—	1500	—	ns
Analog noise blanking time	AtBLK	VM=24V, Iout=1.5A	250	400	550	ns
Oscillator frequency accuracy	ΔfOSCM	COSC=270pF, ROOSC=5.1 kΩ	-15	—	+15	%
Oscillator reference frequency	fOSCM	COSC=270 pF, ROOSC=5.1 kΩ	952	1120	1288	kHz
Chopping frequency	fchop	Output:Active(Iout =1.5 A), fOSCM = 1120 kHz	—	70	—	kHz

AC Electrical Specification Timing chart



Timing charts may be simplified for explanatory purpose.

(Reference) About the relation between power dissipation and ambient temperature



* This value is dependent on a substrate pattern and mounting conditions. Please be careful. Moreover, power dissipation becomes small when ambient temperature is high.

Notes on Contents

1. Block Diagrams
Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.
2. Equivalent Circuits
The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.
3. Timing Charts
Timing charts may be simplified for explanatory purposes.
4. Application Circuits
The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass-production design stage.
Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.
5. Test Circuits
Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

IC Usage Considerations

Notes on handling of ICs

- (1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings. Exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
- (2) Use an appropriate power supply fuse to ensure that a large current does not continuously flow in the case of overcurrent and/or IC failure. The IC will fully break down when used under conditions that exceed its absolute maximum ratings, when the wiring is routed improperly or when an abnormal pulse noise occurs from the wiring or load, causing a large current to continuously flow and the breakdown can lead to smoke or ignition. To minimize the effects of the flow of a large current in the case of breakdown, appropriate settings, such as fuse capacity, fusing time and insertion circuit location, are required.
- (3) If your design includes an inductive load such as a motor coil, incorporate a protection circuit into the design to prevent device malfunction or breakdown caused by the current resulting from the inrush current at power ON or the negative current resulting from the back electromotive force at power OFF. IC breakdown may cause injury, smoke or ignition. Use a stable power supply with ICs with built-in protection functions. If the power supply is unstable, the protection function may not operate, causing IC breakdown. IC breakdown may cause injury, smoke or ignition.
- (4) Do not insert devices in the wrong orientation or incorrectly. Make sure that the positive and negative terminals of power supplies are connected properly.
Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
In addition, do not use any device inserted in the wrong orientation or incorrectly to which current is applied even just once.
- (5) Carefully select external components (such as inputs and negative feedback capacitors) and load components (such as speakers), for example, power amp and regulator.
If there is a large amount of leakage current such as from input or negative feedback capacitor, the IC output DC voltage will increase. If this output voltage is connected to a speaker with low input withstand voltage, overcurrent or IC failure may cause smoke or ignition. (The overcurrent may cause smoke or ignition from the IC itself.) In particular, please pay attention when using a Bridge Tied Load (BTL) connection-type IC that inputs output DC voltage to a speaker directly.

Points to remember on handling of ICs

Overcurrent detection Circuit

Overcurrent detection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the overcurrent detection circuits operate against the overcurrent, clear the overcurrent status immediately.

Depending on the method of use and usage conditions, exceeding absolute maximum ratings may cause the overcurrent detection circuit to operate improperly or IC breakdown may occur before operation. In addition, depending on the method of use and usage conditions, if overcurrent continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

Thermal Shutdown Circuit

Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over-temperature, clear the heat generation status immediately.

Depending on the method of use and usage conditions, exceeding absolute maximum ratings may cause the thermal shutdown circuit to operate improperly or IC breakdown to occur before operation.

Heat Radiation Design

When using an IC with large current flow such as power amp, regulator or driver, design the device so that heat is appropriately radiated, in order not to exceed the specified junction temperature (T_J) at any time or under any condition.

These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, when designing the device, take into consideration the effect of IC heat radiation with peripheral components.

Back-EMF

When a motor rotates in the reverse direction, stops or slows abruptly, current flows back to the motor's power supply owing to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond the absolute maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design.

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