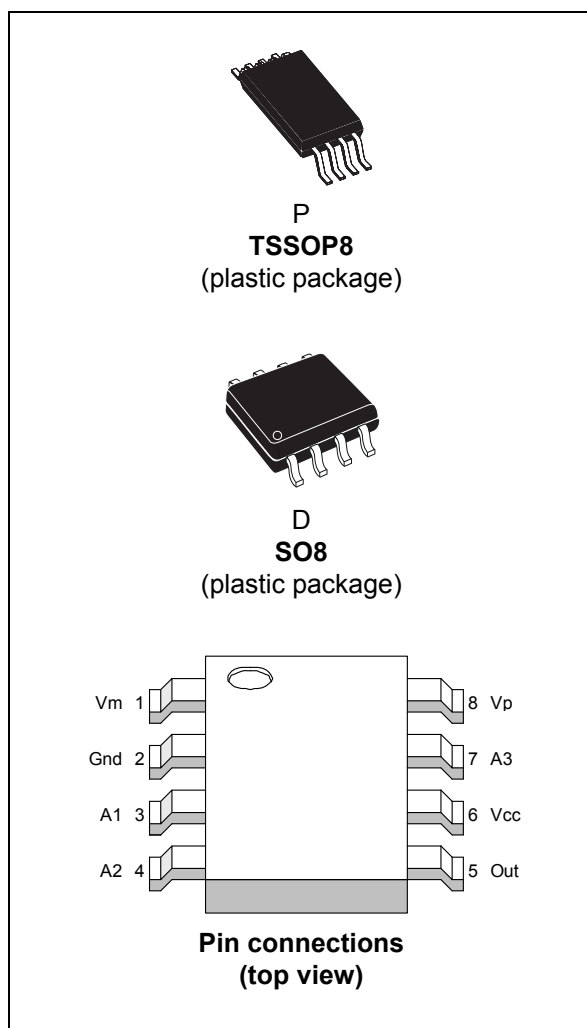


## High-side current sense amplifier plus signal conditioning amplifier

Datasheet - production data



- Output amplifier for tailor-made signal conditioning
- -40 °C to 125 °C operating temperature range
- 4 kV ESD protection

### Applications

- Battery chargers
- Automotive current monitoring
- Notebook computers
- DC motor control
- Photo-voltaic systems
- Precision current sources
- Uninterruptible power supplies
- High-end power supplies

### Description

The TSC102 measures a small differential voltage on a high-side shunt resistor and translates it into a ground-referenced output voltage.

The device's wide input common-mode voltage range, low quiescent current and tiny TSSOP8 packaging enable use in a wide variety of applications (also available in SO8 package).

The input common-mode and power supply voltages are independent. The common-mode voltage can range from 2.8 to 30 V in operating conditions.

The TSC102 is rugged against abnormal conditions on the input pins: Vp and Vm can withstand up to 60 V in case of voltage spikes, as little as -16 V in case of reversed battery, and up to 4 kV in case of electrostatic discharge.

In addition to the current sensing amplifier, the TSC102 offers a fully accessible amplifier for output signal conditioning. The device's overall current consumption is lower than 420  $\mu$ A.

### Features

- Independent supply and input common-mode voltages
- Wide common-mode operating range: 2.8 to 30 V
- Wide common-mode surviving range: -16 to 60 V (reversed battery and load-dump conditions)
- Low current consumption:  $I_{CC}$  max = 420  $\mu$ A

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# 1 Application schematic and pin description

The TSC102 high-side current sense amplifier features a 2.8 V to 30 V input common-mode range that is independent of the supply voltage. The main advantage of this feature is that it allows high-side current sensing at voltages much greater than the supply voltage ( $V_{CC}$ ).

**Figure 1. Application schematics**

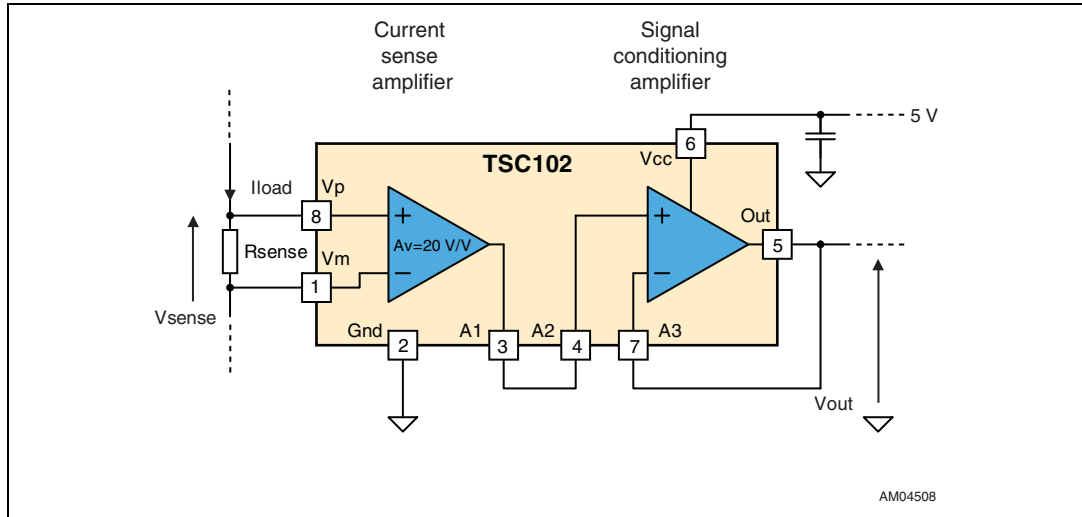


Table 1 describes the function of each pin. Their position is shown in the illustration on the cover page and in Figure 1 above.

**Table 1. Pin description**

Symbol	Type	Function
Out	Analog output	Out voltage is proportional to the magnitude of the sense voltage $V_p - V_m$
Gnd	Power supply	Ground line
$V_{CC}$		Positive power supply line
$V_p$	Analog input	Connection for the external sense resistor. The measured current enters the shunt on the $V_p$ side
$V_m$		Connection for the external sense resistor. The measured current exits the shunt on the $V_m$ side
A1		Connection to current sensing amplifier output
A2		Connection to signal conditioning amplifier non-inverting input
A3		Connection to signal conditioning amplifier inverting input

## 2 Absolute maximum ratings and operating conditions

**Table 2. Absolute maximum ratings**

Symbol	Parameter	Value	Unit
$V_{id}$	Input pins differential voltage ( $V_p-V_m$ )	$\pm 20$	V
$V_i$	Current sensing input pin voltages ( $V_p$ and $V_m$ ) <sup>(1)</sup>	-16 to 60	
$V_1$	Voltage for pins A1, A2, A3, Out, $V_{cc}$ <sup>(1)</sup>	-0.3 to 7	
$T_{stg}$	Storage temperature	-55 to 150	°C
$T_j$	Maximum junction temperature	150	
$R_{thja}$	TSSOP8 thermal resistance junction to ambient	120	°C/W
	SO8 thermal resistance junction to ambient	125	
ESD	HBM: human body model for $V_m$ and $V_p$ pins <sup>(2)</sup>	4	kV
	HBM: human body model <sup>(3)</sup>	2.5	
	MM: machine model <sup>(4)</sup>	200	V
	CDM: charged device model <sup>(5)</sup>	1.5	kV

1. These voltage values are measured with respect to the GND pin.
2. Human body model for  $V_m$  and  $V_p$ : a 100 pF capacitor is charged to the specified voltage, then discharged through a 1.5 kΩ resistor between the  $V_p$  or  $V_m$  pin and Gnd while the other pins are floating.
3. Human body model: a 100 pF capacitor is charged to the specified voltage, then discharged through a 1.5 kΩ resistor between two pins of the device. This is done for all couples of connected pin combinations while the other pins are floating.
4. Machine model: a 200 pF capacitor is charged to the specified voltage, then discharged directly between two pins of the device with no external series resistor (internal resistor < 5 Ω). This is done for all couples of connected pin combinations while the other pins are floating.
5. Charged device model: all pins plus package are charged together to the specified voltage and then discharged directly to ground.

**Table 3. Operating conditions**

Symbol	Parameter	Value	Unit
$V_{CC}$	DC supply voltage from $T_{min}$ to $T_{max}$	3.5 to 5.5	V
$T_{oper}$	Operational temperature range ( $T_{min}$ to $T_{max}$ )	-40 to 125	°C
$V_{icm}$	Common-mode voltage range ( $V_m$ pin voltage)	2.8 to 30	V

### 3 Electrical characteristics

Unless otherwise specified, the electrical characteristics given in the following tables have been measured under the following test conditions.

- $T_{amb} = 25\text{ °C}$ ,  $V_{CC} = 5\text{ V}$ ,  $V_{sense} = V_p - V_m = 50\text{ mV}$ ,  $V_m = 12\text{ V}$ .
- No load on Out pin.
- Signal conditioning amplifier used as a buffer (pin A3 connected to pin Out and pin A1 connected to pin A2).

**Table 4. Supply**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$I_{CC}$	Total supply current	$V_{sense} = 0\text{ V}$ , pin A1 open, pin A2 shorted to Gnd $T_{min} < T_{amb} < T_{max}$	-	240	420	$\mu\text{A}$
$I_{CC1}$		$V_{sense} = 50\text{ mV}$ , pin A1 connected to pin A2 $T_{min} < T_{amb} < T_{max}$		420	700	

**Table 5. Current sensing amplifier input stage**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
DC CMR1	DC common-mode rejection Variation of $V_{a1}$ versus $V_{icm}$ referred to input <sup>(1)</sup>	$2.8\text{ V} < V_m < 30\text{ V}$ $-40\text{ °C} < T_{amb} < 150\text{ °C}$	90	100		dB
AC CMR1	AC common-mode rejection Variation of $V_{a1}$ versus $V_{icm}$ referred to input (peak-to-peak voltage variation)	$2.8\text{ V} < V_m < 30\text{ V}$ 1 kHz sine wave		75		
		$2.8\text{ V} < V_m < 30\text{ V}$ 10 kHz sine wave		60		
SVR1	Supply voltage rejection Variation of $V_{a1}$ versus $V_{CC}$ <sup>(2)</sup>	$3.5\text{ V} < V_{CC} < 5.5\text{ V}$ $-40\text{ °C} < T_{amb} < 125\text{ °C}$	85	90		
$V_{os}$	Input offset voltage <sup>(3)</sup>	$T_{amb} = 25\text{ °C}$ $-40\text{ °C} < T_{amb} < 125\text{ °C}$			$\pm 1.5$ $\pm 2.3$	mV
$dV_{os}/dT$	Input offset drift versus T	$-40\text{ °C} < T_{amb} < 125\text{ °C}$		$\pm 3$	$\pm 8$	$\mu\text{V}/\text{°C}$
$I_{lk}$	Input leakage current	$V_{CC} = 0\text{ V}$ $T_{min} < T_{amb} < T_{max}$			1	$\mu\text{A}$
$I_{ib}$	Input bias current	$V_{sense} = 0\text{ V}$ $T_{min} < T_{amb} < T_{max}$		5	7	

1. See [Section 6: Parameter definitions](#) for the definition of CMR

2. See [Section 6](#) for the definition of SVR

3. See [Section 6](#) for the definition of  $V_{os}$

**Table 6. Current sensing amplifier output stage**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$A_v$	Gain (variation of $V_{a1}$ versus $V_{sense}$ )			20		V/V
$V_{oh1}$	A1 node high-level saturation voltage $V_{oh1} = V_{cc} - V_{a1}$	$V_{sense} = 1\text{ V}$ $I_{a1} = 1\text{ mA}$ $-40\text{ °C} < T_{amb} < 125\text{ °C}$		85	185	mV
$V_{ol1}$	A1 node low-level saturation voltage	$V_{sense} = -1\text{ V}$ $I_{a1} = 1\text{ mA}$ $-40\text{ °C} < T_{amb} < 125\text{ °C}$		75	165	
$I_{sc1}$	Short-circuit current	A1 connected to $V_{CC}$ or Gnd	10	30		mA
$\Delta V_{a1}/\Delta T$	Output voltage drift versus $T^{(1)}$	$T_{min} < T_{amb} < T_{max}$			$\pm 400$	ppm/°C
$\Delta V_{a1}/\Delta I_{a1}$	Output stage load regulation	$-5\text{ mA} < I_{a1} < +5\text{ mA}$ $I_{a1}$ sink or source current		0.4	$\pm 2$	mV/mA
$\Delta V_{a1}$	Total output voltage accuracy <sup>(2)</sup>	$V_{sense} = 50\text{ mV}$ $T_{amb} = 25\text{ °C}$ $T_{min} < T_{amb} < T_{max}$			$\pm 2.5$ $\pm 4$	%
$\Delta V_{a1}$	Total output voltage accuracy <sup>(2)</sup>	$V_{sense} = 100\text{ mV}$ $T_{amb} = 25\text{ °C}$ $T_{min} < T_{amb} < T_{max}$			$\pm 2.5$ $\pm 4$	
$\Delta V_{a1}$	Total output voltage accuracy <sup>(2)</sup>	$V_{sense} = 20\text{ mV}$ $T_{amb} = 25\text{ °C}$ $T_{min} < T_{amb} < T_{max}$			$\pm 8$ $\pm 10$	
$\Delta V_{a1}$	Total output voltage accuracy <sup>(2)</sup>	$V_{sense} = 10\text{ mV}$ $T_{amb} = 25\text{ °C}$ $T_{min} < T_{amb} < T_{max}$			$\pm 13$ $\pm 16$	

1. See [Section 6: Parameter definitions](#) for the definition of output voltage drift versus temperature.
2. Output voltage accuracy is the difference with the expected theoretical output voltage  $V_{a1-th} = A_v * V_{sense}$ . See [Section 6](#) for a more detailed definition.

**Table 7. Current sensing amplifier frequency response**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_s$	$V_{a1}$ settling to 1% final value	$V_{sense} = 10\text{ mV to }100\text{ mV}$ , $C_{load} = 47\text{ pF}$	-	7	-	$\mu\text{s}$
SR	Slew rate	$V_{sense} = 10\text{ mV to }100\text{ mV}$	0.2	0.4	-	V/ $\mu\text{s}$
BW	3 dB bandwidth	$C_{load} = 47\text{ pF}$	-	800	-	kHz

**Table 8. Current sensing amplifier noise**

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$e_N$	Equivalent input noise voltage	$f = 1\text{ kHz}$	-	50	-	nV/ $\sqrt{\text{Hz}}$

Table 9. Signal conditioning amplifier

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$V_{icm}$	Common-mode voltage range	$T_{min} < T_{amb} < T_{max}$	0		$V_{CC}$	
$V_{IO}$	Input offset voltage	$V_{a2} = 1\text{ V}$ $T_{amb} = 25\text{ }^\circ\text{C}$ $-40\text{ }^\circ\text{C} < T_{amb} < 150\text{ }^\circ\text{C}$			$\pm 3.5$ $\pm 4.5$	mV
$\Delta V_{IO}$	Input offset voltage drift	$T_{min} < T_{amb} < T_{max}$		5		$\mu\text{V}/^\circ\text{C}$
$I_{ib}$	Input bias current	$V_{a2} = V_{a3} = V_{CC}/2$		10		pA
$V_{oh2}$	Output high-level saturation voltage ( $V_{oh2} = V_{CC} - V_{out}$ )	$V_{a2} = 1\text{ V}$ $V_{a3} = 0\text{ V}$ $I_{out} = 1\text{ mA}$ $-40\text{ }^\circ\text{C} < T_{amb} < 125\text{ }^\circ\text{C}$		85	185	mV
$V_{ol2}$	Output low-level saturation voltage	$V_{a2} = 0\text{ V}$ $V_{a3} = 1\text{ V}$ $I_{out} = 1\text{ mA}$ $-40\text{ }^\circ\text{C} < T_{amb} < 125\text{ }^\circ\text{C}$		75	165	
$I_{sc2}$	Short-circuit current	Out connected to $V_{CC}$ or Gnd	12	30		mA
$\Delta V_{out}/\Delta I_{out}$	Output stage load regulation	$-10\text{ mA} < I_{out} < +10\text{ mA}$ $V_{a2} = 1\text{ V}$ $I_{out}$ sink or source current			300	$\mu\text{V}/\text{mA}$
CMR2	DC common-mode rejection Variation of $V_{IO}$ versus $V_{icm}$	$T_{min} < T_{amb} < T_{max}$ $0\text{ V} < V_{a2} < 3\text{ V}$ $0\text{ V} < V_{a2} < 5\text{ V}$	70 60	95 80		dB
SVR2	Supply voltage rejection Variation of $V_{IO}$ versus $V_{CC}$	$3.5\text{ V} < V_{CC} < 5.5\text{ V}$ $V_{a2} = 1\text{ V}$ $-40\text{ }^\circ\text{C} < T_{amb} < 125\text{ }^\circ\text{C}$	85	105		
GBP	Gain bandwidth product	$R_L = 10\text{ k}\Omega$ , $C_{load} = 100\text{ pF}$ , $f = 100\text{ kHz}$		1		MHz
PM	Phase margin	$R_L = 10\text{ k}\Omega$ , $C_{load} = 100\text{ pF}$		65		deg
SR	Slew rate	$R_L = 10\text{ k}\Omega$ , $C_{load} = 100\text{ pF}$ $V_{a2} = 0.5\text{ V}$ to $4.5\text{ V}$ A3 connected to OUT (follower configuration) Slew rate measured from 10% to 90% of $V_{out}$ step	0.2	0.4		$\text{V}/\mu\text{s}$

## 4 Electrical characteristics curves: current sense amplifier

Unless otherwise specified, the test conditions for the following curves are:

- $T_{amb} = 25\text{ }^{\circ}\text{C}$ ,  $V_{CC} = 5\text{ V}$ ,  $V_{sense} = V_p - V_m = 50\text{ mV}$ ,  $V_m = 12\text{ V}$ .
- no load on Out pin.
- signal conditioning amplifier used as a buffer (pin A3 connected to pin Out and pin A1 connected to pin A2).

Figure 2. Output voltage vs. Vsense

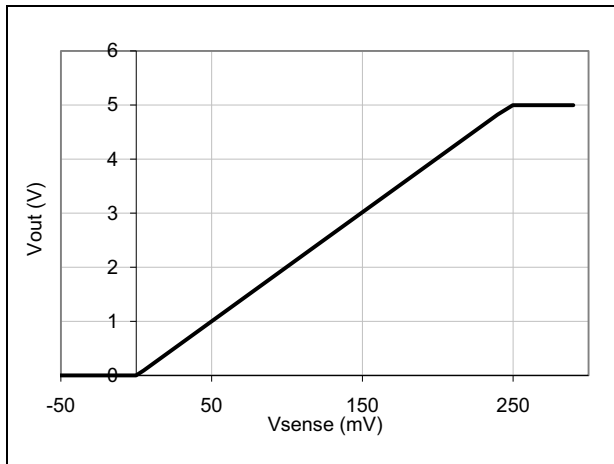


Figure 3. A1 pin voltage accuracy vs. Vsense

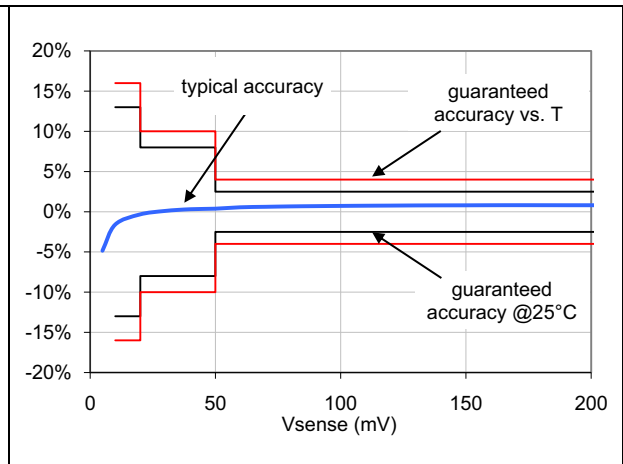


Figure 4. Supply current vs. supply voltage

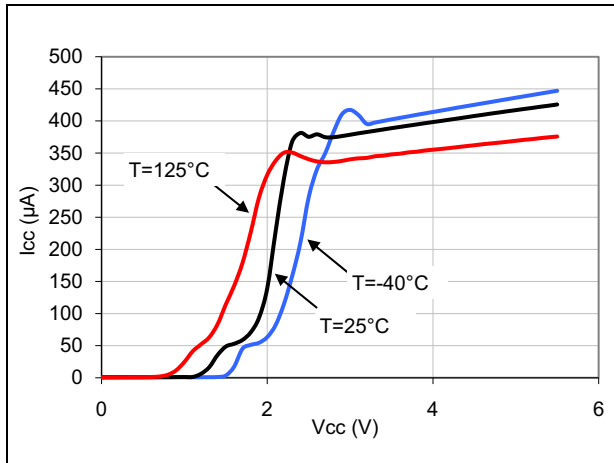


Figure 5. Supply current vs. Vsense

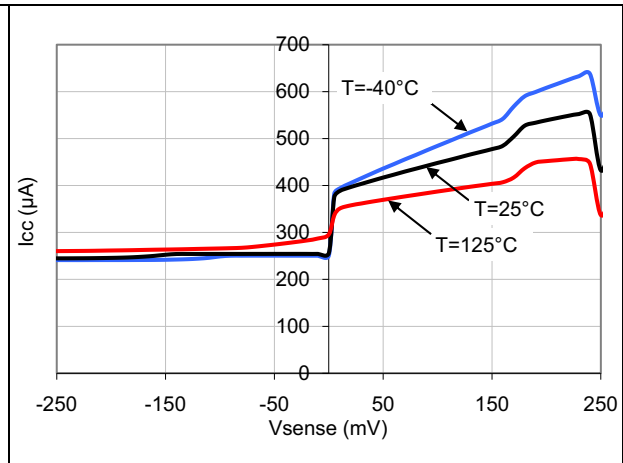




Figure 6. Vp pin input bias current vs. Vsense

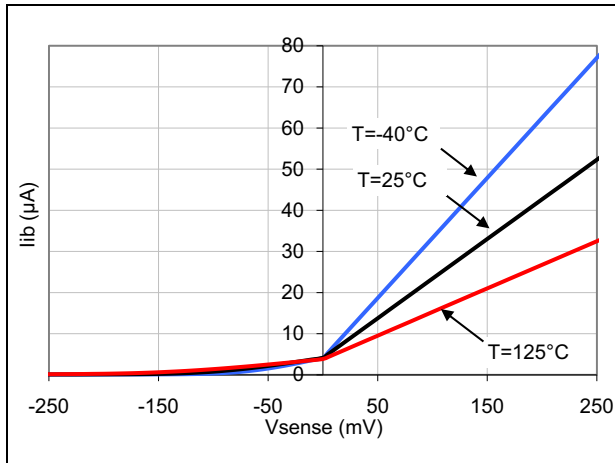


Figure 7. Vm pin input bias current vs. Vsense

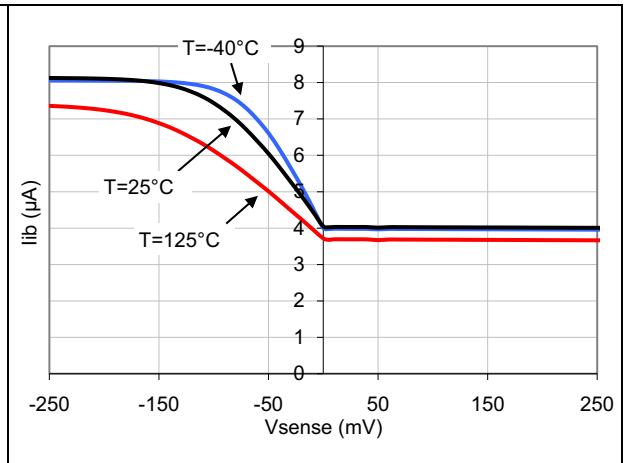


Figure 8. Output stage low-state saturation voltage versus output current (Vsense = -1 V)

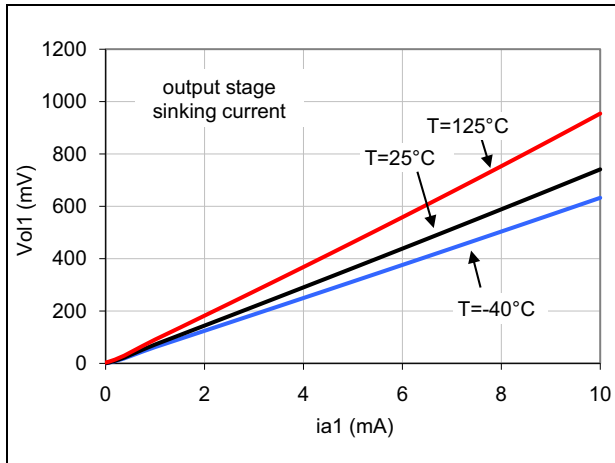


Figure 9. Output stage high-state saturation voltage versus output current (Vsense = +1 V)

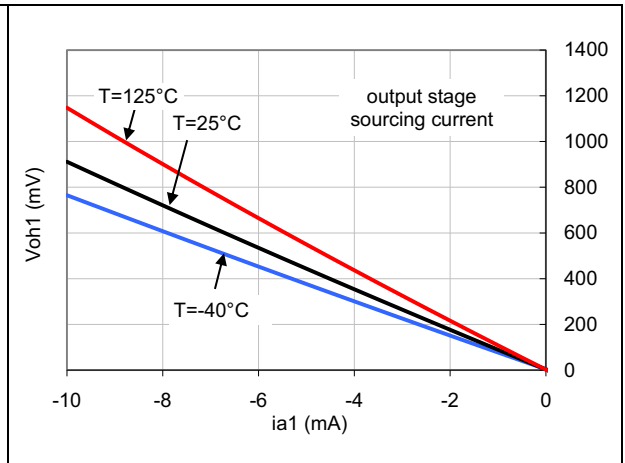


Figure 10. Output stage load regulation

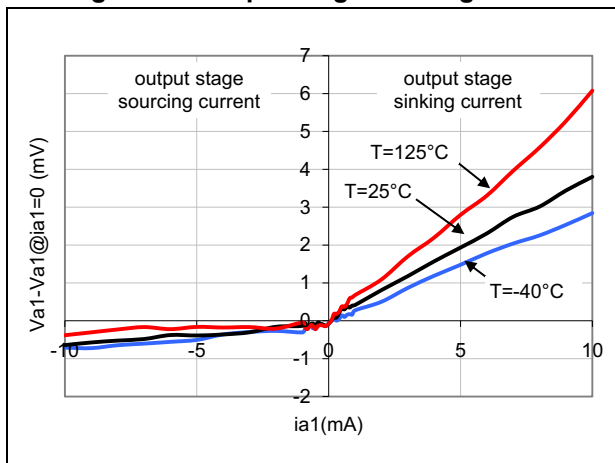


Figure 11. Step response

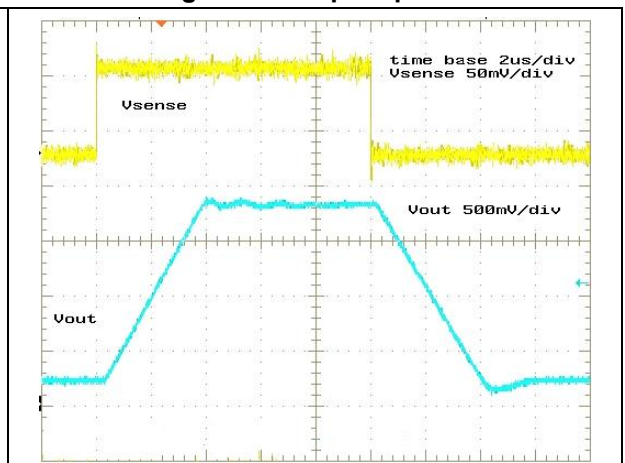


Figure 12. Bode diagram

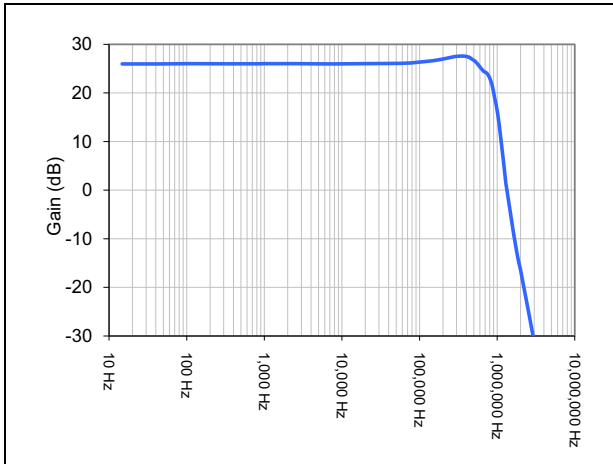
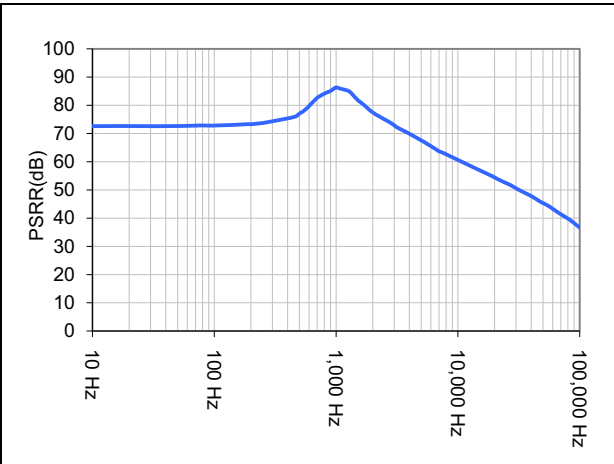


Figure 13. Power supply rejection ratio

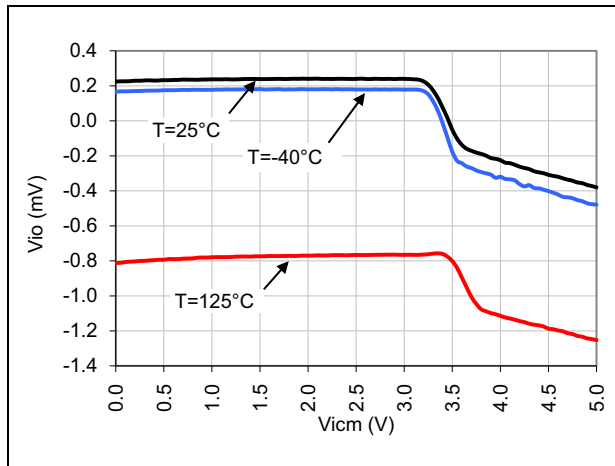


## 5 Electrical characteristics curves: signal conditioning amplifier

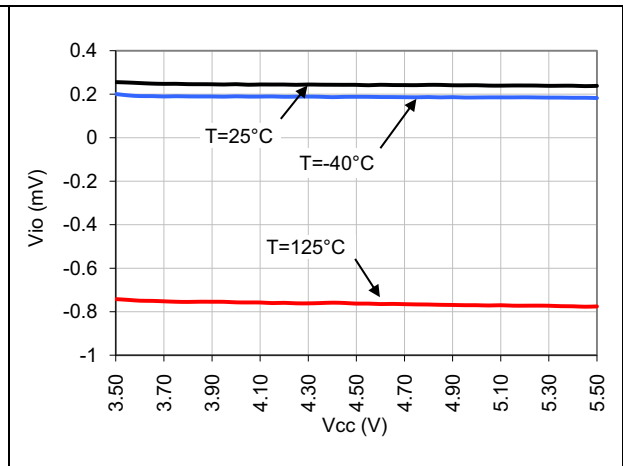
Unless otherwise specified, the test conditions for the following curves are:

- $T_{amb} = 25\text{ }^{\circ}\text{C}$ ,  $V_{CC} = 5\text{ V}$
- no load on Out.
- signal conditioning amplifier tested as standalone amplifier.

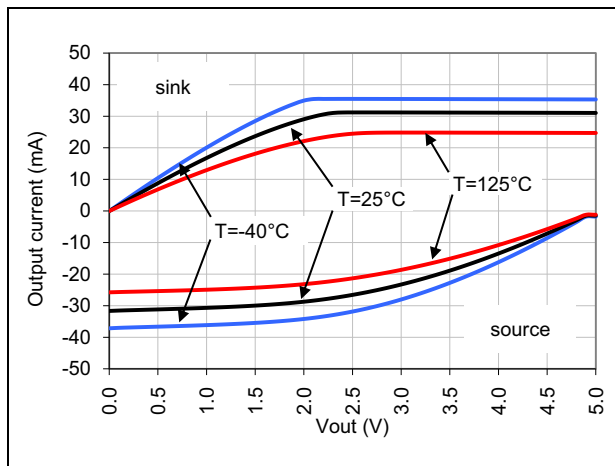
**Figure 14. Input offset voltage versus input common-mode voltage**



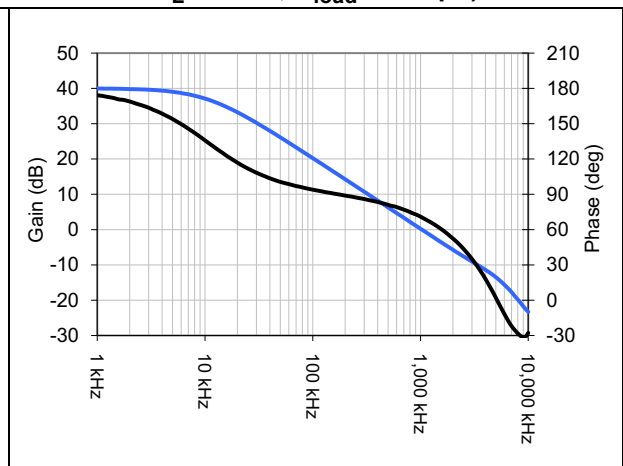
**Figure 15. Input offset voltage versus supply voltage ( $V_{icm} = V_{CC}/2$ )**



**Figure 16. Output current versus output voltage**



**Figure 17. Bode diagram ( $V_{out} = V_{CC}/2$ ,  $R_L = 10\text{ k}\Omega$ ,  $C_{load} = 100\text{ pF}$ )**



## 6 Parameter definitions

### 6.1 Common-mode rejection ratio (CMR)

The common-mode rejection ratio (CMR) measures the ability of the current sensing amplifier to reject any DC voltage applied on both inputs  $V_p$  and  $V_m$ . The CMR is referred back to the input so that its effect can be compared with the applied differential signal. The CMR is defined by the formula:

$$\text{CMR} = -20 \cdot \log \frac{\Delta V_{a1}}{\Delta V_{icm} \cdot A_v}$$

### 6.2 Supply voltage rejection ratio (SVR)

The supply voltage rejection ratio (SVR) measures the ability of the current sensing amplifier to reject any variation of the supply voltage  $V_{CC}$ . The SVR is referred back to the input so that its effect can be compared with the applied differential signal. The SVR is defined by the formula:

$$\text{SVR} = -20 \cdot \log \frac{\Delta V_{a1}}{\Delta V_{cc} \cdot A_v}$$

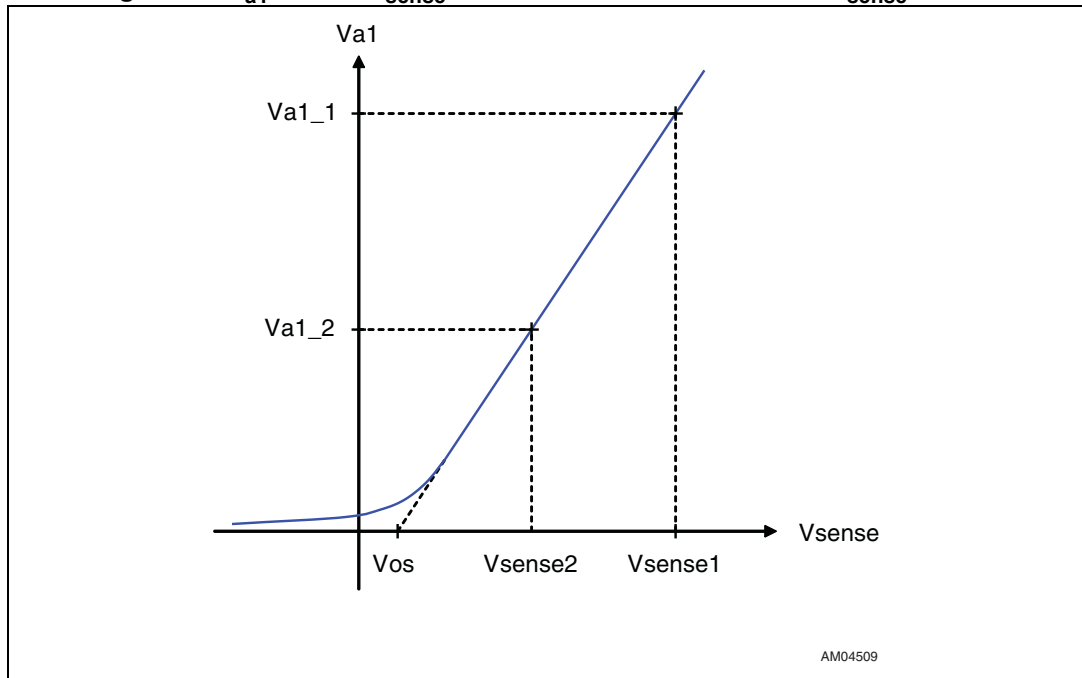
### 6.3 Gain ( $A_v$ ) and input offset voltage ( $V_{os}$ )

The input offset voltage is defined as the intersection between the linear regression of the  $V_{a1}$  versus  $V_{sense}$  curve with the X-axis (see [Figure 18](#)). If  $V_{a11}$  is the output voltage with  $V_{sense} = V_{sense1} = 50$  mV and  $V_{a12}$  is the output voltage with  $V_{sense} = V_{sense2} = 5$  mV, then  $V_{os}$  can be calculated with the formula:

$$V_{os} = V_{sense1} - \left( \frac{V_{sense1} - V_{sense2}}{V_{a11} - V_{a12}} \cdot V_{out1} \right)$$

The amplification gain  $A_v$  is defined as the ratio between the output voltage and the input differential voltage.

$$A_v = \frac{V_{out}}{V_{sense}}$$

Figure 18.  $V_{a1}$  versus  $V_{sense}$  characteristics: detail for low  $V_{sense}$  values

## 6.4 Output voltage drift versus temperature

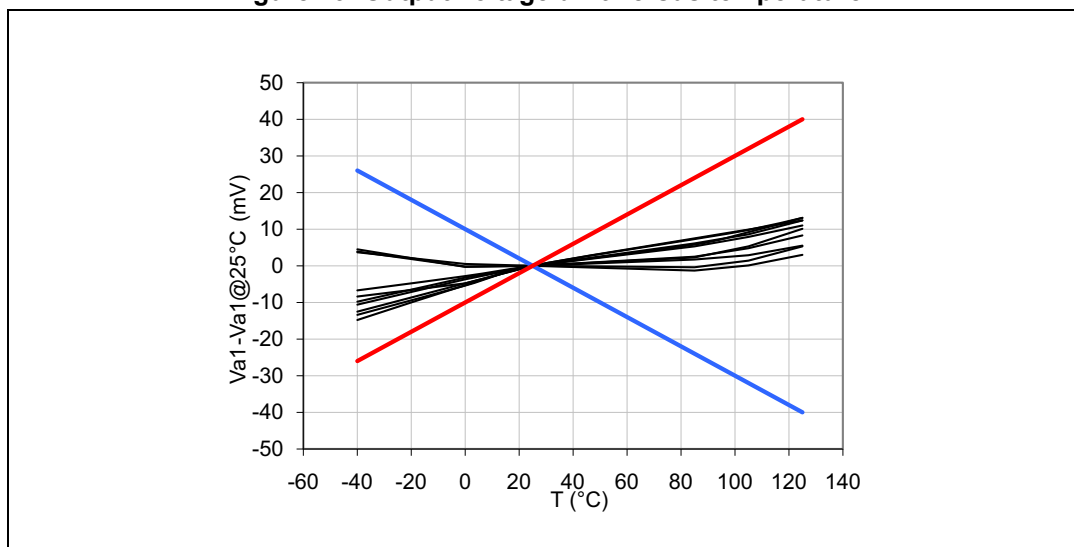
The output voltage drift versus temperature is defined as the maximum variation of  $V_{a1}$  with respect to its value at 25 °C, over the temperature range. It is calculated as follows:

$$\frac{\Delta V_{a1}}{\Delta T} = \max \frac{V_{a1}(T_{amb}) - V_{a1}(25^\circ \text{C})}{T_{amb} - 25^\circ \text{C}}$$

with  $T_{min} < T_{amb} < T_{max}$ .

[Figure 19](#) provides a graphical definition of the output voltage drift versus temperature. On this chart,  $V_{a1}$  is always within the area defined by the maximum and minimum variation of  $V_{a1}$  versus  $T$ , and  $T = 25^\circ \text{C}$  is considered to be the reference.

Figure 19. Output voltage drift versus temperature



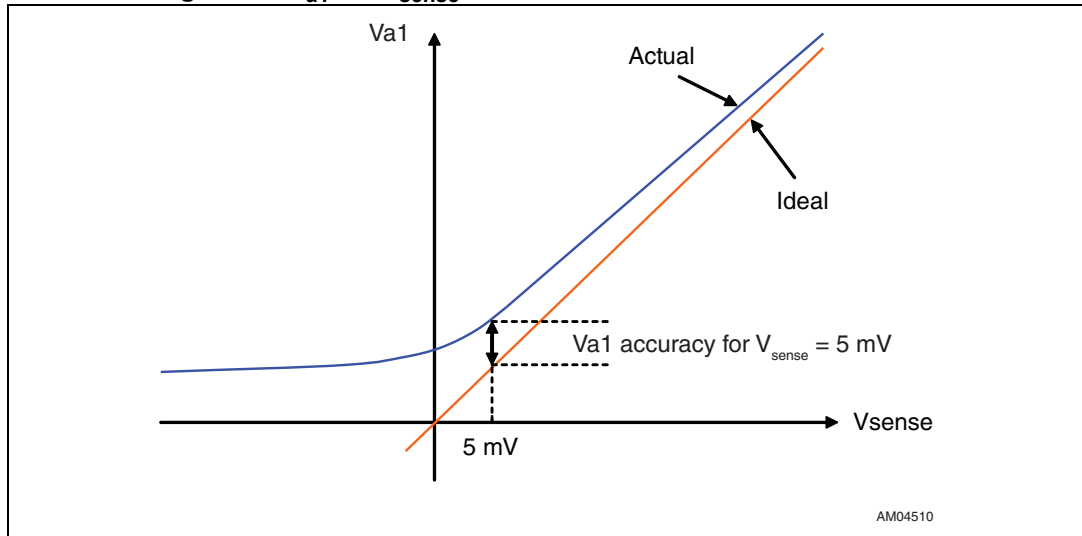
## 6.5 Output voltage accuracy

The output voltage accuracy is the difference between the actual output voltage and the theoretical output voltage. Ideally, the current sensing output voltage should be equal to the input differential voltage multiplied by the theoretical gain, as in the following formula.

$$V_{a1-th} = Av \cdot V_{sense}$$

The actual value is very slightly different, mainly due to the effects of the input offset voltage  $V_{OS}$  and the non-linearity.

**Figure 20.  $V_{a1}$  vs.  $V_{sense}$  theoretical and actual characteristics**



The output voltage accuracy, expressed as a percentage, can be calculated with the following formula:

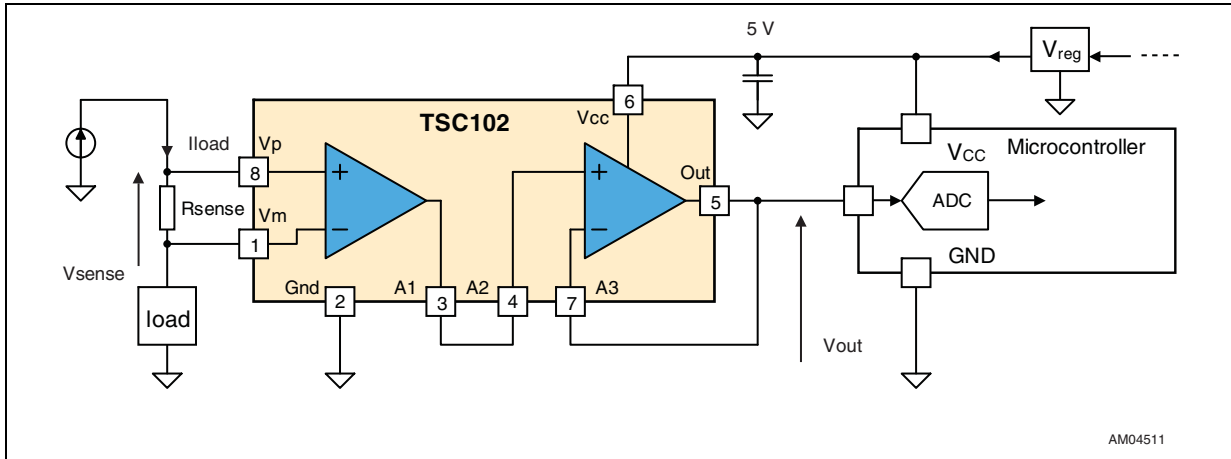
$$\Delta V_{a1} = \frac{\text{abs}(V_{a1} - (Av \cdot V_{sense}))}{Av \cdot V_{sense}}$$

with  $Av = 20 \text{ V/V}$ .

# 7 Application information

The TSC102 can be used to measure current and feed back the information to a microcontroller, as shown in [Figure 21](#).

**Figure 21. Typical application schematic**



This fully-accessible output amplifier offers wide schematic possibilities, as shown in the following examples.

**Figure 22. Gain higher than 20**

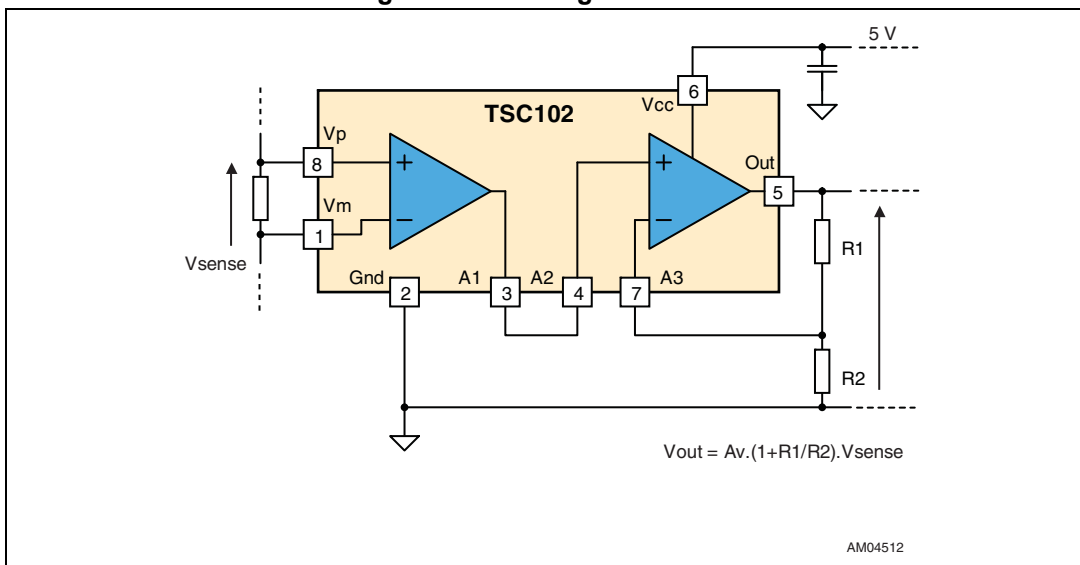




Figure 23. Gain lower than 20

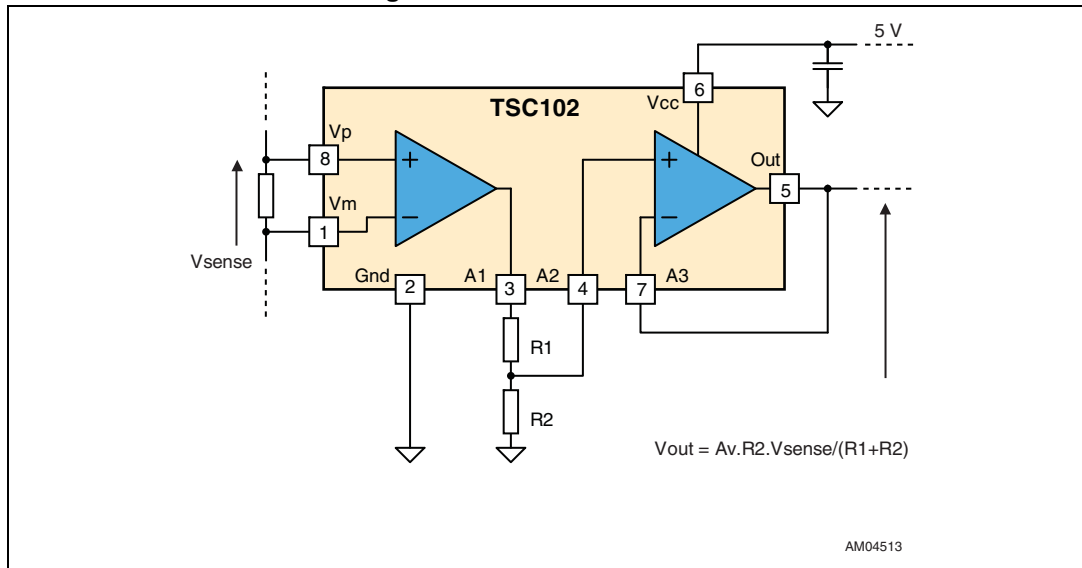


Figure 24. Overcurrent protection

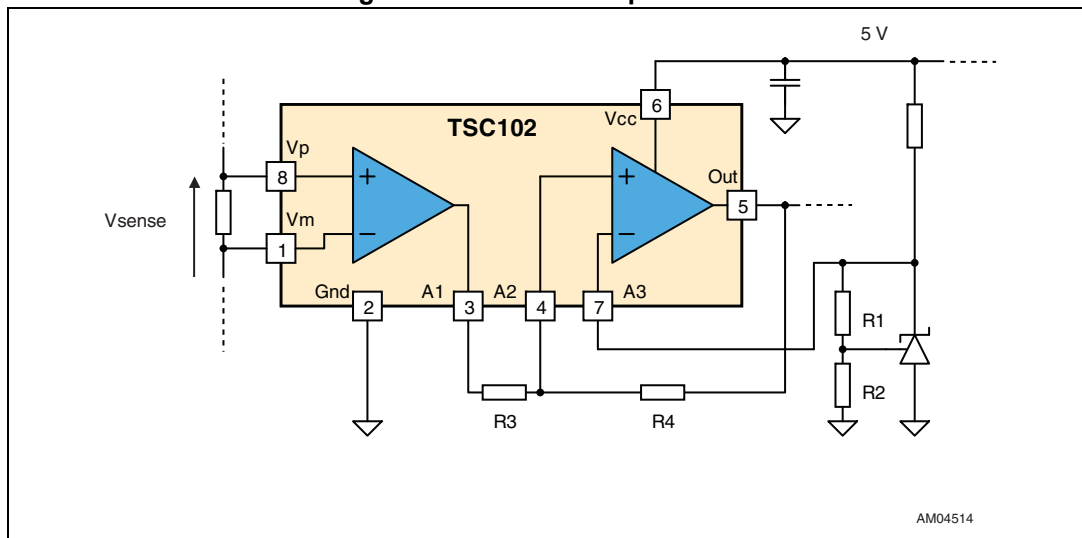
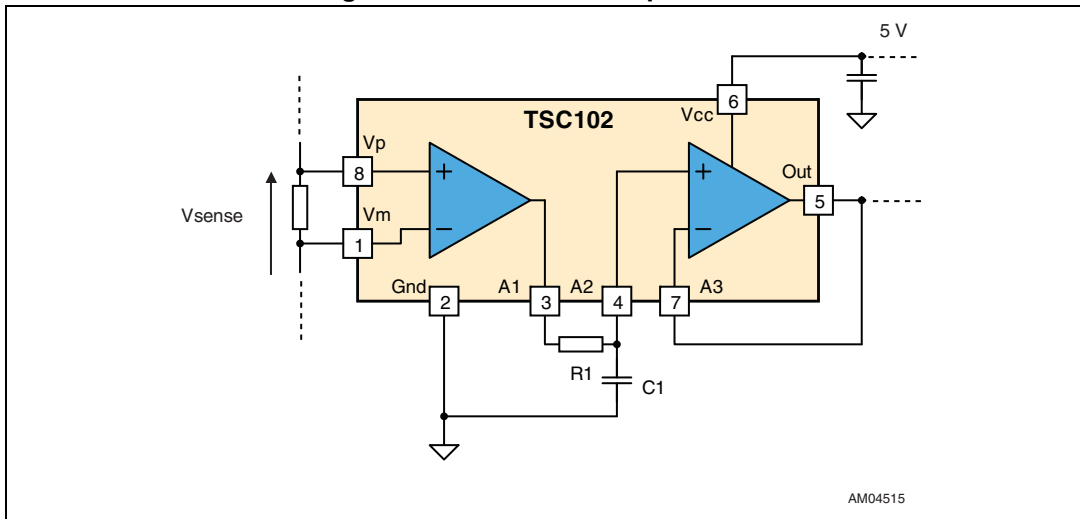
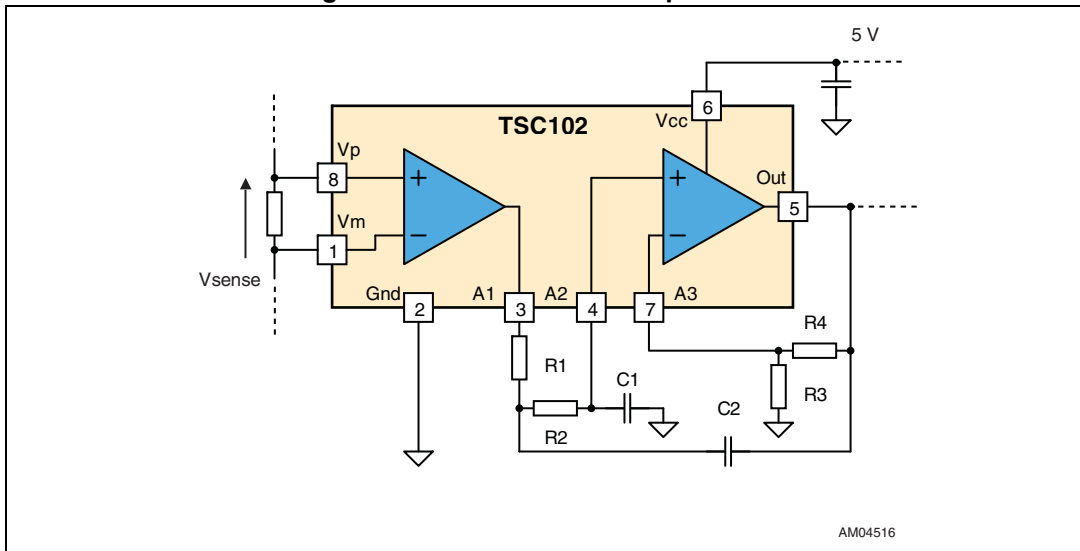


Figure 25. First-order low-pass filter



AM04515

Figure 26. Second-order low-pass filter



AM04516

## 8 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK<sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK<sup>®</sup> specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK<sup>®</sup> is an ST trademark.

### 8.1 TSSOP8 package information

Figure 27. TSSOP8 package mechanical drawing

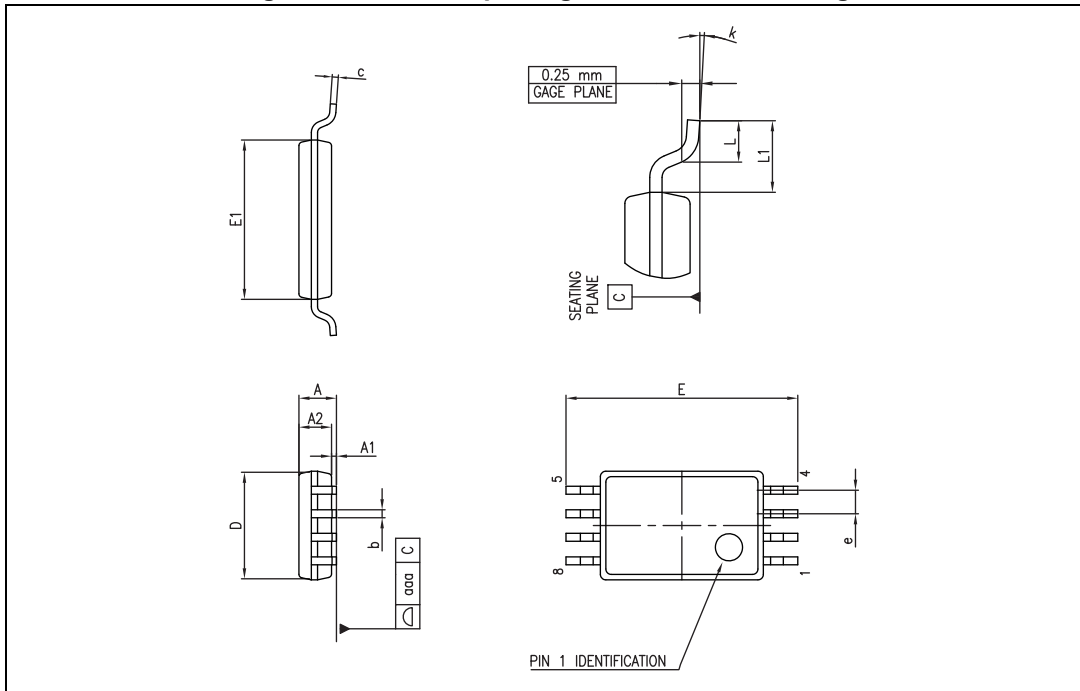


Table 10. TSSOP8 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.20			0.047
A1	0.05		0.15	0.002		0.006
A2	0.80	1.00	1.05	0.031	0.039	0.041
b	0.19		0.30	0.007		0.012
c	0.09		0.20	0.004		0.008
D	2.90	3.00	3.10	0.114	0.118	0.122
E	6.20	6.40	6.60	0.244	0.252	0.260
E1	4.30	4.40	4.50	0.169	0.173	0.177
e		0.65			0.0256	
k	0°		8°	0°		8°
L	0.45	0.60	0.75	0.018	0.024	0.030
L1		1			0.039	
aaa			0.10			0.004

## 8.2 SO8 package information

Figure 28. SO8 package mechanical drawing

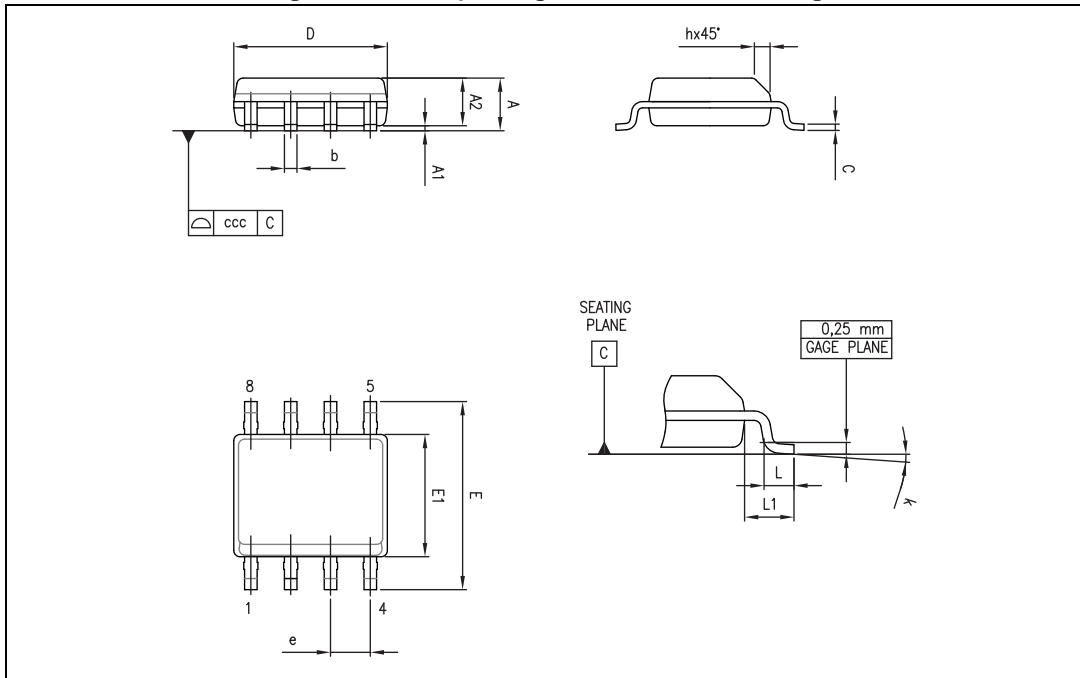


Table 11. SO8 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.75			0.069
A1	0.10		0.25	0.004		0.010
A2	1.25			0.049		
b	0.28		0.48	0.011		0.019
c	0.17		0.23	0.007		0.010
D	4.80	4.90	5.00	0.189	0.193	0.197
E	5.80	6.00	6.20	0.228	0.236	0.244
E1	3.80	3.90	4.00	0.150	0.154	0.157
e		1.27			0.050	
h	0.25		0.50	0.010		0.020
L	0.40		1.27	0.016		0.050
L1		1.04			0.040	
k	0		8°	1°		8°
ccc			0.10			0.004

## 9 Ordering information

**Table 12. Order codes**

Part number	Temperature range	Package	Packing	Marking
TSC102IPT	-40 °C, +125 °C	TSSOP8	Tape and reel	102I
TSC102IDT		SO8		TSC102I
TSC102IYPT	-40 °C, +125 °C automotive grade	TSSOP8 <sup>(1)</sup>		102Y
TSC102IYDT		SO8 <sup>(1)</sup>		TSC102IY

1. Qualification and characterization according to AEC Q100 and Q003 or equivalent, advanced screening according to AEC Q001 and Q 002 or equivalent.

## 10 Revision history

**Table 13. Document revision history**

Date	Revision	Changes
09-Nov-2009	1	Initial release.
03-Mar-2011	2	Added automotive grade qualification for SO8 package (note 2. under <a href="#">Table 12</a> ).
31-Jan-2014	3	<a href="#">Table 12</a> : Updated automotive-grade footnotes.

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