

# DC Brushless Fan Motor Drivers Three-Phase Full-Wave Fan Motor Driver

# BD6326NUX

## **General description**

BD6326NUX is a three-phase sensorless fan motor driver used to cool off notebook PCs. It is controlled by a variable speed provided through the PWM input signal. Its feature is sensorless drive which doesn't require a hall device as a location detection sensor and motor downsizing can be achieved by limiting the number of external components as much as possible. Furthermore, introducing a direct PWM soft switched driving mechanism achieves silent operations and low vibrations.

#### Features

- Speed controllable by PWM input signal
- 180° Sinusoidal drive
- Power save function
- Internal RNF resistance
- Motor rotation direction select function(FR)

#### Application

■ Small fan motor notebook PCs etc.

#### Absolute maximum ratings

# Package W(Typ) x D(Typ) x H(Max) 3.00mm x 3.00mm x 0.60mm

Symbol	Limit	Unit
V <sub>CC</sub>	7	V
Pd	0.58	W
T <sub>opr</sub>	-25 to +95	°C
T <sub>stg</sub>	-55 to +150	°C
V <sub>omax</sub>	7	V
I <sub>omax</sub>	700	mA
$V_{\text{FG}}$	7	V
I <sub>FG</sub>	6	mA
T <sub>jmax</sub>	150	°C
	V <sub>CC</sub> P <sub>d</sub> T <sub>opr</sub> T <sub>stg</sub> V <sub>omax</sub> I <sub>omax</sub> V <sub>FG</sub> I <sub>FG</sub>	V <sub>CC</sub> 7   P <sub>d</sub> 0.58   T <sub>opr</sub> -25 to +95   T <sub>stg</sub> -55 to +150   V <sub>omax</sub> 7   I <sub>omax</sub> 700   V <sub>FG</sub> 7   I <sub>FG</sub> 6

(NOTE 1) Reduce by 4.64mW/°C over Ta=25°C. (On 74.2mm×74.2mm×1.6mm glass epoxy board)

(NOTE 2) This value is not to exceed Pd.

**Caution:** Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

## **Recommended operating condition**

Parameter	Symbol	Limit	Unit
Operating supply voltage range	Vcc	2.2 to 5.5	V
Input voltage range(PWM, FR terminals)	V <sub>INx</sub>	0 to $V_{CC}$	V

# **Pin Configuration**

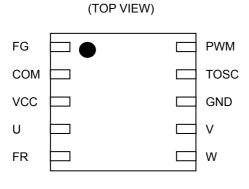


Figure 1. Pin configuration

# **Pin Description**

P/No.	T/name	Function
1	FG	FG output terminal
2	COM	Coil midpoint terminal
3	VCC	Power supply terminal
4	U	U phase output terminal
5	FR	Motor rotation direction select terminal
6	W	W phase output terminal
7	V	V phase output terminal
8	GND	GND terminal
9	TOSC	Start-up oscillation terminal
10	PWM	PWM signal input terminal

# **Block Diagram**

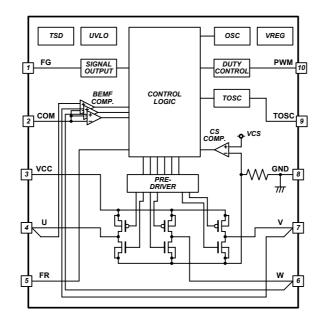


Figure 2. Block diagram

# **Electrical characteristics** (Unless otherwise specified T<sub>a</sub>=25°C, V<sub>CC</sub>=5V)

Parameter	Symbol	Limit		Unit	Conditions	
	Symbol	Min	Тур	Max	Offic	Conditions
<overall></overall>						
Circuit current STB	I <sub>CST</sub>	-	20	50	μA	
Circuit current	I <sub>CC</sub>	2.4	5.5	8.6	mA	
<pwm terminal=""></pwm>						
PWM input H level	V <sub>PH</sub>	2.5	-	Vcc	V	
PWM input L level	V <sub>PL</sub>	0	-	0.7	V	
PWM input current H	I <sub>PH</sub>	-	0	1	μA	PWM=VCC
PWM input current L	I <sub>PL</sub>	-50	-20	-	μA	PWM=GND
Input frequency	f <sub>P</sub>	20	-	50	kHz	
<fr terminal=""></fr>		1	L	1		
FR input H level	V <sub>OH</sub>	2.5	-	V <sub>CC</sub>	V	FR=H : Normal rotation
FR input L level	V <sub>OL</sub>	0	-	0.5	V	FR=L; Reverse rotation
<tosc terminal=""></tosc>		1	L	1		
TOSC frequency	f <sub>OSF</sub>	28	40	52	kHz	TOSC-GND 2200pF
TOSC charge current	l <sub>occ</sub>	-137.5	-110	-82.5	μA	TOSC=0.5V
TOSC discharge current	I <sub>ODC</sub>	75	100	125	μA	TOSC=1.0V
<fg terminal=""></fg>		1	L	1		
FG low voltage	V <sub>FGL</sub>	-	-	0.4	V	IFG=5mA
<output terminals=""></output>	I		1	1		
Output voltage	Vo	-	0.25	0.325	V	Io=250mA (H.L. total)
PWM off time	t <sub>PO</sub>	0.3	1	2	ms	
Lock protection det. time	t <sub>LDT</sub>	0.6	0.9	1.5	s	
Lock protection rel. time	t <sub>LRT</sub>	3.3	5.0	8.3	S	

About a current item, define the inflow current to IC as a positive notation, and the outflow current from IC as a negative notation.

# **Typical Performance Curves 1**

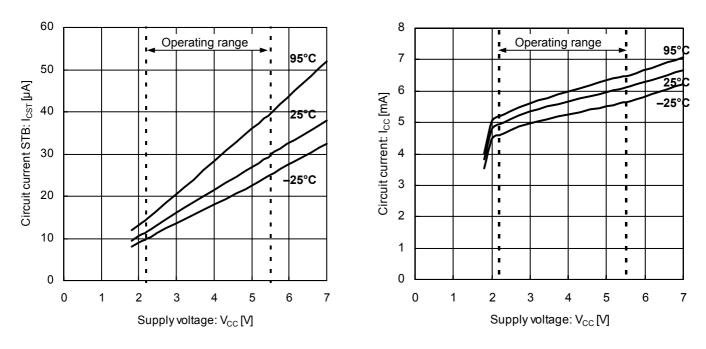


Figure 3. Circuit current STB

Figure 4. Circuit current

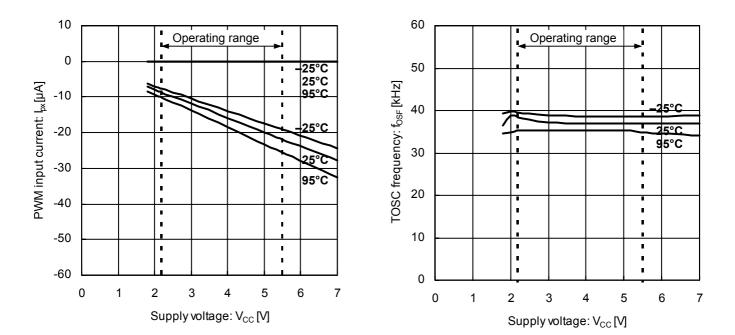


Figure 5. PWM input current

Figure 6. TOSC frequency

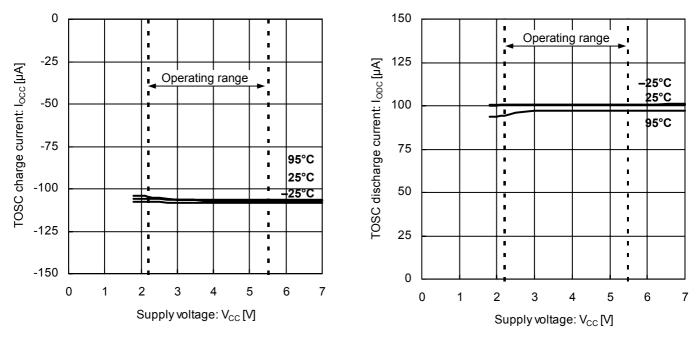


Figure 7. TOSC charge current

Figure 8. TOSC discharge current

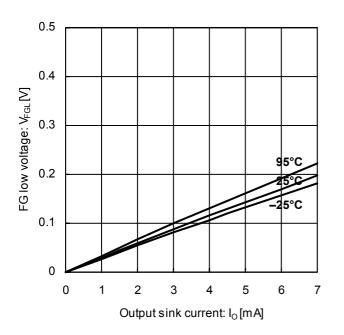


Figure 9. FG low voltage (V<sub>cc</sub>=5V)

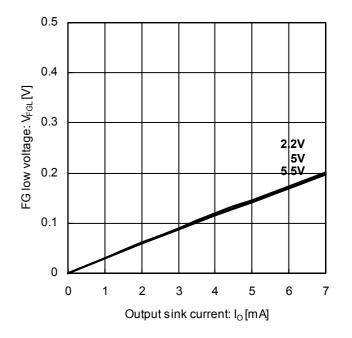


Figure 10. FG low voltage (Temp=25°C)

# **Typical Performance Curves 3**

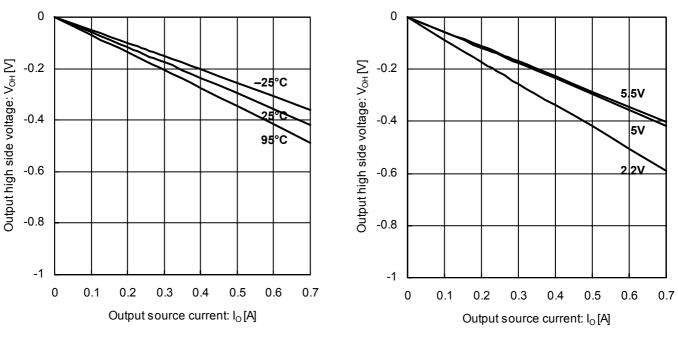
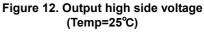


Figure 11. Output high side voltage (V<sub>cc</sub>=5V)



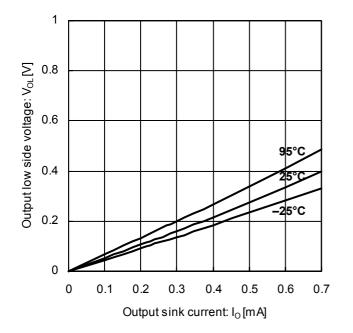
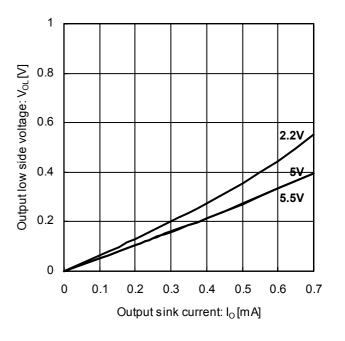
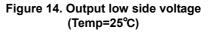


Figure 13. Output low side voltage (V<sub>cc</sub>=5V)





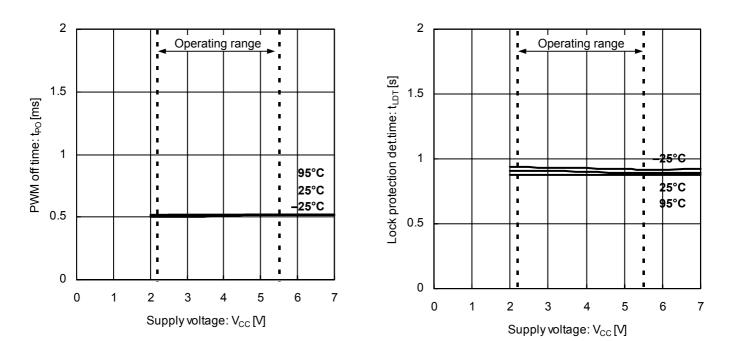


Figure 15. PWM off time

Figure 16. Lock protection det.time

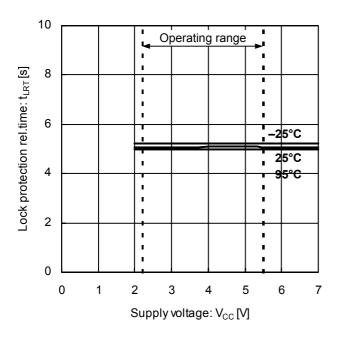


Figure 17. Lock protection rel.time

# BD6326NUX

# **Timing chart**

## 1) Sensorless Drive

BD6326NUX is a motor driver IC for driving a three-phase brushless DC motor without a hall sensor. Detecting a rotor location firstly at startup, an appropriate logic for the rotation direction is obtained using this information and given to each phase to rotate the motor. Then, the rotation of the motor induces electromotive

voltage in each phase wiring and the logic based on the induced electromotive voltage is applied to the each phase to continue rotating.

#### 1.1 BEMF detection driving mechanism (synchronized start-up mechanism)

BD6326NUX's start mechanism is synchronized start-up mechanism. BD6326NUX as BEMF detection driving starts by set output logic and monitors BEMF voltage of motor. Driving mechanism changes to BEMF detection driving after detect BEMF signal. When BEMF signal isn't detected for constant time at start-up, synchronized start-up mechanism outputs output logic forcibly by using standard synchronized signal (sync signal) and makes motor forward drive. This assistance of motor start-up as constant cycle is synchronized driving mechanism. Synchronized frequency is standard synchronized signal. Figure 18, the timing chart (outline) is shown. "Motor start-up frequency setting" generation of synchronized period is shown.

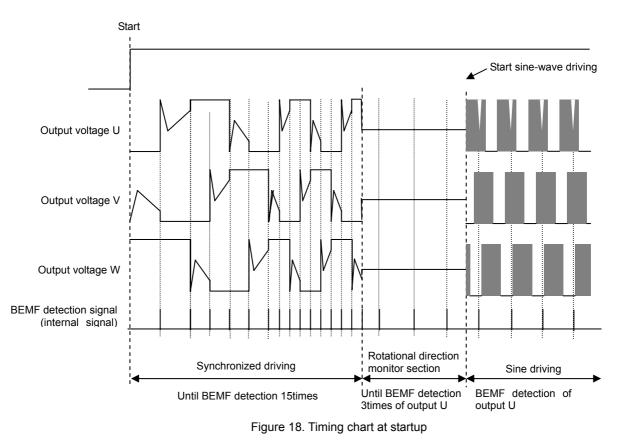


Table 1.	Setting of	electrify ang	le and outp	out duty while	e start-up

		Number of BEMF detection (from start-up)			
	Start-up	Until BEMF (output U,V,W) detection 15times successively	Until BEMF detection 3times of output U	After BEMF detection 3times of output U (after BEMF moniter section)	
Synchronized time		8000 × TOSC	Output off mode		
PWM duty	PWM = fixed 100% 150° drive		Output off mode (BEMF moniter section)	PWM control	
Electrify angle			( , ,	Sine-wave	

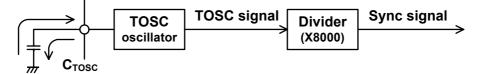
\* Disagree with above timing chart

1.2 Motor start-up frequency setting (TOSC capacitor)

The TOSC terminal starts a self-oscillation by connecting a capacitor between the TOSC terminal and GND. It becomes a start-up frequency, and synchronized time. Synchronized time can be adjusted by changing external capacitor. When the capacitor value is small, synchronized time becomes short. It is necessary to choose the best capacitor value for optimum start-up operation. For example external capacitor is 2200pF, synchronized time is 200ms (Typ).

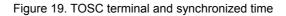
1000pF is recommended for setting value at first. Relationship between external capacitor and synchronized time is shown in below.

< Diagram of Relationship between TOSC terminal and synchronized time >



Synchronized time = 8000 x TOSC period Charge current : 100µA

discharge current : 100µA



Equation

$$Tosc = 2x \frac{C_{TOSC}V_{TSOC}}{I}$$

C<sub>TOSC</sub>: TOSC terminal capacitor value

V<sub>TOSC</sub>: TOSC terminal Hi voltage – Lo voltage= 0.57V (Typ)

I: TOSC terminal charge and discharge current

Table 2. Capacitor values and	Table 2. Capacitor values and synchronized time		
External capacitor	Synchronized time		
2200pF	200ms		
1000pF (Recommendation)	90ms		
670pF	60ms		

Example C<sub>TOSC</sub> = TOSC

 $C_{TOSC} = 2200 pF$ TOSC frequency = 40kHz (Typ) TOSC period = 25µs Synchronized time = 200ms

#### **%**Setting of Appropriate capacitor value

Appropriate value of synchronized time is differing with characteristic and parameter of motor. Appropriate value decided by start-up confirmation with various capacitor values.

<u>At first confirm start-up with 1000pF</u>, next is 1200,1500,2200pF···, and 820,680pF···etc. Appropriate capacitor value is decided after confirm maximum start-up NG value and minimum start-up NG value. For example, small BEMF voltage motor tends to small capacitor value. Set capacitor value after confirm sufficiently.

1.3 U, V, W phase and FG output signals

The timing charts of the output signals from the U, V and W phases as well as the FG terminal (at FR = Hi or no connected) is shown (Figure 20).

The detection of the BEMF voltage does with output U and detects the position of the motor rotation. The three phases are driving in the order of U, V and W phases. About FG signal output, assuming that a three-slot tetrode motor is used, two pulse outputs of FG are produced for one motor cycle.

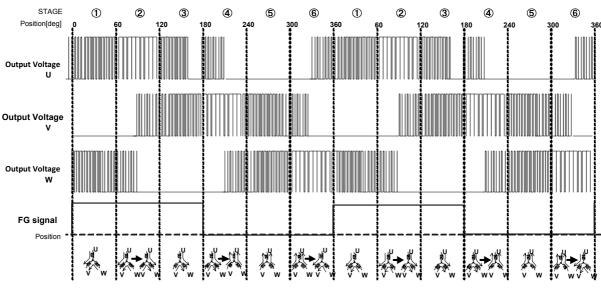


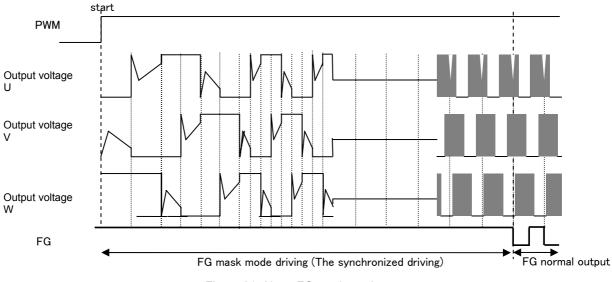
Figure 20. Timing chart of U, V, W, FG output signal (FR= Hi or no connect)

# Table 3. Truth table of normal operation

		Motor output	
Output pattern	Motor output U	Motor output V	Motor output W
1	PWM	L	PWM
2	PWM	L→PWM	PWM→L
3	$PWM \rightarrow Hi-Z (BEMF detect)$	PWM	L
4	PWM→L	PWM	L→PWM
5	L	PWM	PWM
6	L→PWM	PWM→L	PWM

\* About the output pattern, It changes in the flow of "1 $\rightarrow$ 2 $\rightarrow$ 3 ~ 6 $\rightarrow$ 1". H; High, L; Low, Hi-Z; High impedance

FG signal is masked between synchronized driving section (FG = Hi level). The FG signal is output from Rotation speed monitor section.





2) UVLO (Under voltage lock out circuit)

In the operation area under the guaranteed operating power supply voltage of 2.2 V (Typ), the transistor on the output can be turned OFF at a power supply voltage of 1.73V (Typ). A hysteresis width of 270mV is provided and a normal operation can be performed at 2.0V. This function is installed to prevent unpredictable operations, such as a large amount of current passing through the output, by means of intentionally turning OFF the output during an operation at a very low power supply voltage which may cause an abnormal function in the internal circuit.

3) Lock Protection Feature (motor stop at start-up), Automatic Recovery Circuit

To prevent passing a coll current on any phase when a motor is locked, it is provided with a function which can turn OFF the output for a certain period of time and then automatically restore itself to the normal operation. During the motor rotation, an appropriate logic based on the induced electromotive voltage can be continuously given to each phase; on the other hand, when the motor is locked at, no induced electromotive voltage is obtained. Utilizing this phenomenon to take a protective against locking, when the induced electromotive voltage is not detected for a predetermined period of time (<u>TON: 0.9sec(Typ)</u>), it is judged that the motor is locked and the output is turned OFF for a predetermined period of time (TOFF: 5.0sec(Typ)).

Moreover, if Synchronized driving doesn't change into Rotational speed monitor section between TON (0.9sec(Typ)) at start-up, it is judged that the motor is locked. The timing chart is shown (Figure 22).

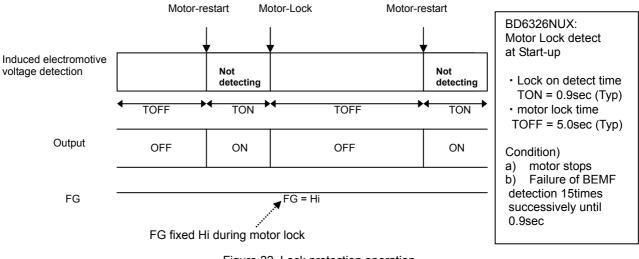


Figure 22. Lock protection operation

4) Power saving function / Speed control by PWM input

The power saving function is controlled by an input logic of the PWM terminal.

- (a) Perate mode when the PWM terminal is High.
- (b) Standby mode when the PWM terminal is Low for a time period of 1ms (Typ).

When the PWM terminal is open, High logic is set.

Input logic of the PWM terminal is set at Low and then the Standby mode becomes effective 1ms (Typ) (Figure 23). In the Standby mode, the lock protection function is deactivated and the lock protection is not effective. Therefore, this device can start up instantly even from the stop state when the input logic of the PWM terminal is set at High.

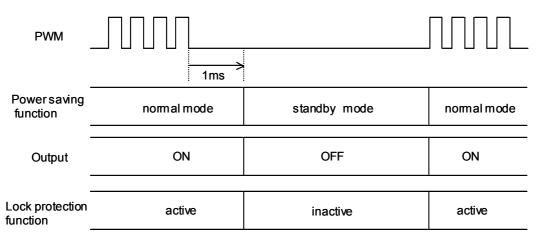


Figure 23. The power saving function

· Speed Control by PWM input

Depending on the Duty value of the input signal on the PWM terminal, the output is controlled by switching between ON and OFF. The higher Duty value results in the higher motor rotation speed, while the lower Duty values the lower speed.

5) Pole select (FG signal select terminal)

The FR terminal selects motor rotation direction (Table 4).

Table 4. FR table

FR	<b>Rotation Direction</b>
High (or open)	Normal Rotation
Low	<b>Reverse Rotation</b>

# Application circuit example (Constant values are for reference)

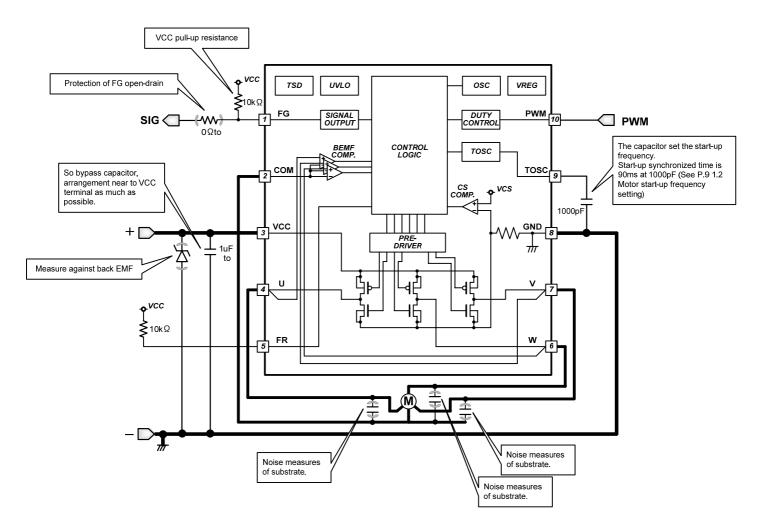


Figure 24. PWM controllable 4 wires type (FG) motor application circuit

#### Substrate design note

- a) IC power, motor outputs, and IC ground lines are made as fat as possible.
- b) IC ground (signal ground) line arranged near to (-) land.
- c) The bypass capacitor is arrangement near to VCC terminal.
- d) When substrates of outputs are noisy, add capacitor as needed.
- e) When back EMF is large, add zener diode as needed.

# **Power dissipation**

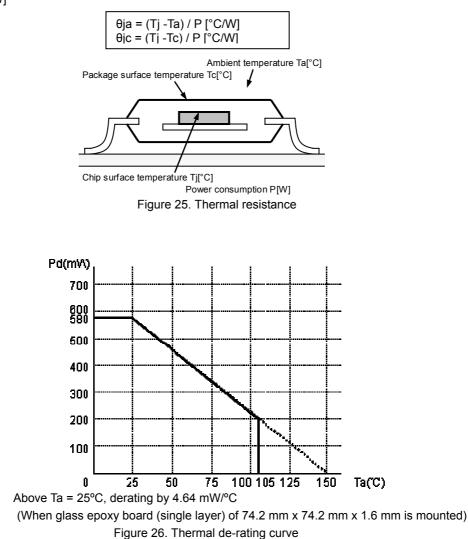
Power dissipation (total loss) indicates the power that can be consumed by IC at Ta = 25°C (normal temperature). IC is heated when it consumes power, and the temperature of IC chip becomes higher than ambient temperature. The temperature that can be accepted by IC chip depends on circuit configuration, manufacturing process, etc., and consumable power is limited. Power dissipation is determined by the temperature allowed in IC chip (maximum junction temperature) and thermal resistance of package (heat dissipation capability). The maximum junction temperature is in general equal to the maximum value in the storage temperature range.

Heat generated by consumed power of IC is radiated from the mold resin or lead frame of package. The parameter which indicates this heat dissipation capability (hardness of heat release) is called heat resistance, represented by the symbol  $\theta_{ja}$  [°C/W]. This heat resistance can estimate the temperature of IC inside the package. Figure 25 shows the model of heat resistance of the package. Heat resistance  $\theta_{ja}$ , ambient temperature Ta, junction temperature Tj, and power consumption P can be calculated by the equation below:

$$\theta$$
ja = (Tj – Ta) / P [°C/W]

Thermal de-rating curve indicates power that can be consumed by IC with reference to ambient temperature. Power that can be consumed by IC begins to attenuate at certain ambient temperature. This gradient is determined by thermal resistance  $\theta$ ja. Thermal resistance  $\theta$ ja depends on chip size, power consumption, package ambient temperature, packaging condition, wind velocity, etc., even when the same package is used. Thermal de-rating curve indicates a reference value measured at a specified condition. Figure 26 shows a thermal de-rating curve (Value when mounting FR4 glass epoxy board 74.2 [mm] x 74.2 [mm] x 1.6 [mm] (copper foil area below 3 [%])). Thermal resistance  $\theta$ jc from IC chip joint part to the package surface part of mounting the above-mentioned same substrate is shown in the following as a reference value.

θjc = 40 [°C/W]



\* 1

## Safety measure

1) Reverse connection protection diode

Reverse connection of power results in IC destruction as shown in Figure 27. When reverse connection is possible, reverse connection destruction preventive diode must be added between power supply and VCC.

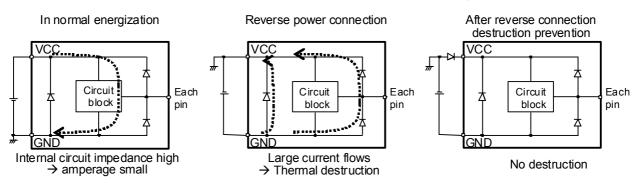


Figure 27. Flow of current when power is connected reversely

2) Measure against VCC voltage rise by back electromotive force

Back electromotive force (Back EMF) generates regenerative current to power supply. However, when reverse connection protection diode is connected. VCC voltage rises because no route is available for regenerating to power.

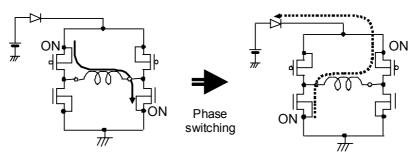


Figure 28. VCC voltage rise by back electromotive force

When the absolute maximum rated voltage may be exceeded due to voltage rise by back electromotive force, place (A) Capacitor or (B) Zener diode between VCC and GND. In addition, also take the measure (A) and (B) as shown in (C) if required.

(B) Zener diode

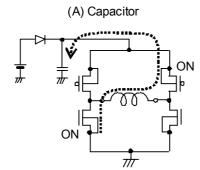
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(C) Capacitor and zener diode

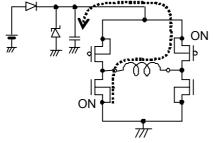


Figure 29. Measure against VCC voltage rise

3) Problem of GND line PWM switching

Do not perform PWM switching of GND line because the potential of GND terminal cannot be kept at the minimum.

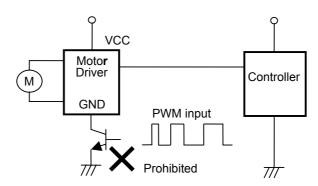


Figure 30. GND Line PWM switching prohibited

#### 4) FG output

FG output is an open drain and requires pull-up resistor.

The IC can be protected by adding resistor R1. An excess of absolute maximum rating, when FG output terminal is directly connected to power supply, could damage the IC.

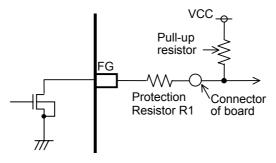


Figure 31. Protection of FG terminal

#### Location of IC

- Generally, three-phase sensorless driver is rotated motor by detecting the induced electromotive voltage. Line noise, line resistance is influenced for detecting the induced electromotive voltage. From motor to IC line should be shorted, its suggest that location of IC is on the board of Motor in below Figure 32.
- 2) In three-phase sensorless and variable speed driver, it is necessary to tuning motor and IC (each motor units).

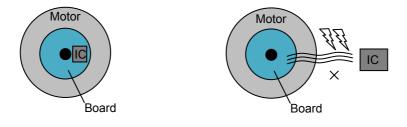
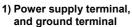
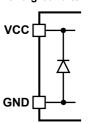
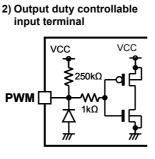


Figure 32. Location of IC

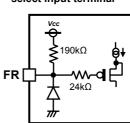
# I/O equivalent circuits (Resistance is typical.)



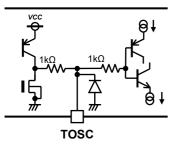


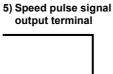


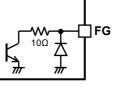
3) Motor rotation direction select input terminal



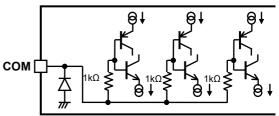
4) Start-up oscillation control terminal







6) Motor coil midpoint input terminal



7) Motor output terminal

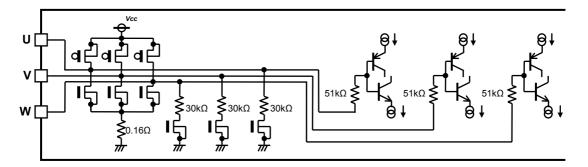


Figure 33. I/O equivalent circuits

# **Operational Notes**

### 1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition. However, pins that drive inductive loads (e.g. motor driver outputs, DC-DC converter outputs) may inevitably go below ground due to back EMF or electromotive force. In such cases, the user should make sure that such voltages going below ground will not cause the IC and the system to malfunction by examining carefully all relevant factors and conditions such as motor characteristics, supply voltage, operating frequency and PCB wiring to name a few.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

# 8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

# **Operational Notes**

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode. When GND > Pin B, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

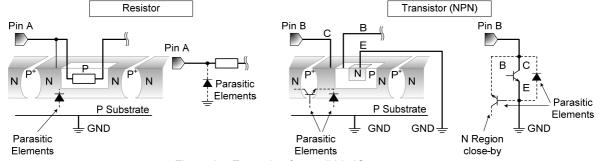


Figure 34. Example of monolithic IC structure

13. Ceramic Capacitor

When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

# 14. Area of Safe Operation (ASO)

Operate the IC such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

15. Thermal Shutdown (TSD) Circuit

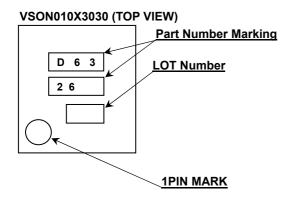
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature will rise which will activate the TSD circuit that will turn OFF all output pins. When the junction temperature falls below the TSD threshold, the circuits are automatically restored to normal operation.

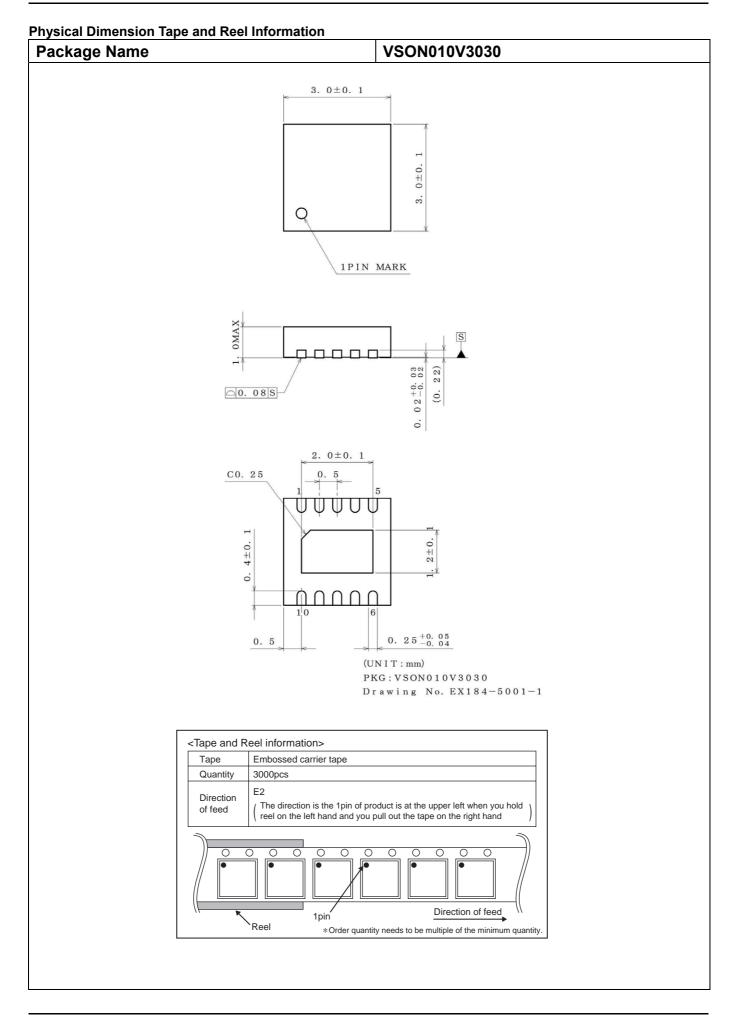
Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

# **Ordering Information**



# Marking diagram





# **Revision History**

Date	Revision	Changes
7. May. 2013	001	New Release
11.Dec.2014	002	Changed the latest format

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