



Features

- Integrated PLL loop filter
- ESD protection (3kV HBM/150V MM)
- High output power (5.5dBm) with low supply current (8.5mA)
- Modulation scheme ASK/FSK
 - FSK modulation is achieved by connecting an additional capacitor between the XTAL load capacitor and the open drain output of the modulating microcontroller
- Easy to design-in due to excellent isolation of the PLL from the PA and power supply
- Single Li-cell for power supply
- Supply voltage 2.0V to 4.0V in the temperature range of -40°C to $+85^{\circ}\text{C}$
- Package TSSOP8L
- Single-ended antenna output with high efficient power amplifier
- CLK output for clocking the microcontroller
- One-chip solution with minimum external circuitry

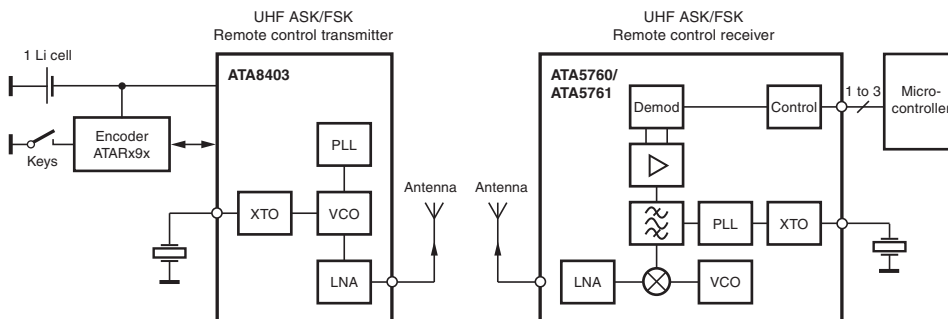
Applications

- Industrial/aftermarket remote keyless entry systems
- Alarm, telemetering, and energy metering systems
- Remote control systems for consumer and industrial markets
- Access control systems
- Home automation
- Home entertainment
- Toys

1. Description

The ATA8403 is a PLL transmitter IC, which has been developed for the demands of RF low-cost transmission systems for industrial applications at data rates up to 50kBaud ASK and 32kBaud FSK modulation scheme. The transmitting frequency range is 868MHz to 928MHz. It can be used in both FSK and ASK systems.

Figure 1-1. System Block Diagram



2. Pin Configuration

Figure 2-1. Pinning TSSOP8L

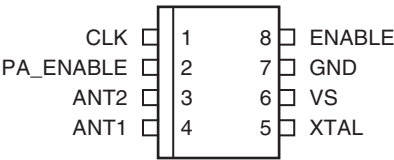


Table 2-1. Pin Description

| Pin | Symbol | Function | Configuration |
|-----|-----------|---|---------------|
| 1 | CLK | Clock output signal for microcontroller The clock output frequency is set by the crystal to $f_{XTAL}/4$ | |
| 2 | PA_ENABLE | Switches on power amplifier, which is used for ASK modulation | |
| 3 | ANT2 | Emitter of antenna output stage | |
| 4 | ANT1 | Open collector antenna output | |

Table 2-1. Pin Description (Continued)

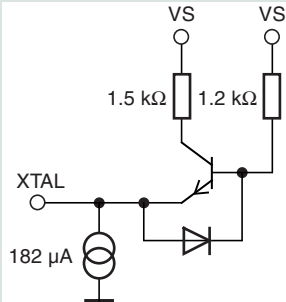
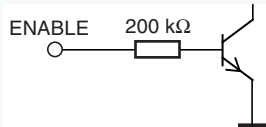
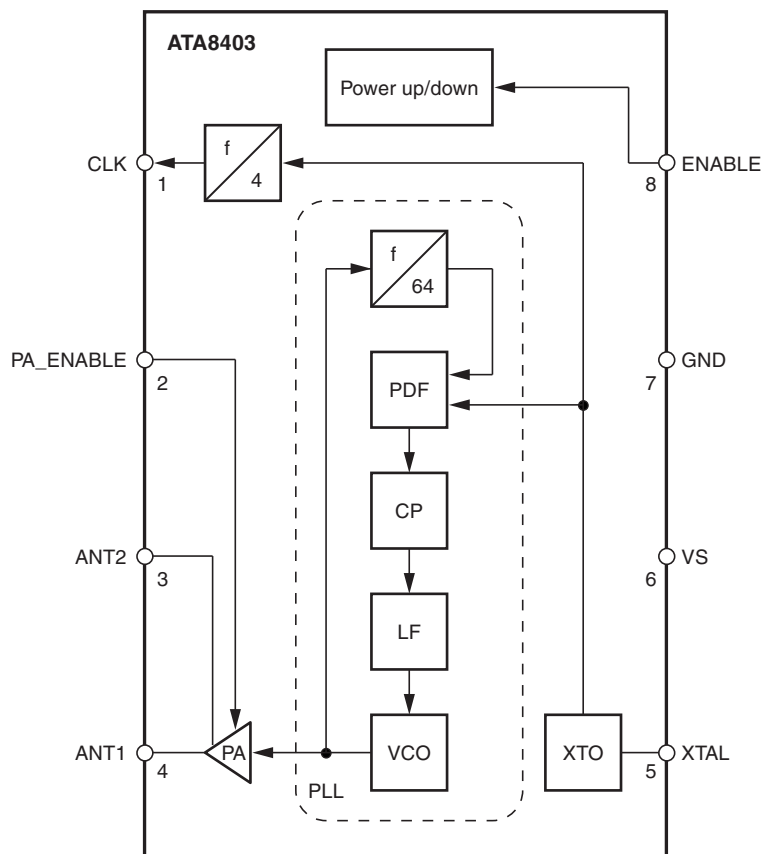
| Pin | Symbol | Function | Configuration |
|-----|--------|------------------------|---|
| 5 | XTAL | Connection for crystal |  |
| 6 | VS | Supply voltage | See ESD protection circuitry (see Figure 4-5 on page 9) |
| 7 | GND | Ground | See ESD protection circuitry (see Figure 4-5 on page 9) |
| 8 | ENABLE | Enable input |  |

Figure 2-2. Block Diagram



3. General Description

This fully integrated PLL transmitter allows particularly simple, low-cost RF miniature transmitters to be assembled. The VCO is locked to $64 \times f_{XTAL}$, and therefore a 13.5672MHz crystal is needed for a 868.3MHz transmitter and a 14.2969MHz crystal for a 915MHz transmitter. All other PLL and VCO peripheral elements are integrated.

The XTO is a series resonance oscillator so that only one capacitor together with a crystal connected in series to GND are needed as external elements.

The crystal oscillator together with the PLL typically needs $< 1\text{ms}$ until the PLL is locked and the CLK output is stable. There is a wait time of $\geq 4\text{ms}$ must be used until the CLK is used for the microcontroller and the PA is switched on.

The power amplifier is an open-collector output delivering a current pulse, which is nearly independent from the load impedance. The delivered output power is therefore controllable via the connected load impedance.

This output configuration enables a simple matching to any kind of antenna or to 50Ω . A high power efficiency of $\eta = P_{out}/(I_{S,PA} \times V_S)$ of 24% for the power amplifier at 868.3MHz results when an optimized load impedance of $Z_{Load} = (166 + j226)\Omega$ is used at 3V supply voltage.

4. Functional Description

If $ENABLE = L$ and the $PA_ENABLE = L$, the circuit is in standby mode, consuming only a very small amount of current, so that a lithium cell used as power supply can work for several years.

With $ENABLE = H$ the XTO, PLL, and the CLK driver are switched on. If PA_ENABLE remains L, only the PLL and the XTO are running and the CLK signal is delivered to the microcontroller. The VCO locks to 64 times the XTO frequency.

With $ENABLE = H$ and $PA_ENABLE = H$ the PLL, XTO, CLK driver, and the power amplifier are on. The power amplifier can be switched on and off with PA_ENABLE . This is used to perform the ASK modulation.

4.1 ASK Transmission

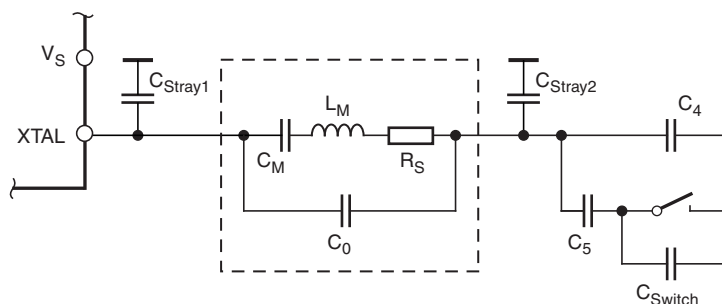
The ATA8403 is activated by $ENABLE = H$. PA_ENABLE must remain L for $t \geq 4\text{ms}$, then the CLK signal can be taken to clock the microcontroller, and the output power can be modulated by means of the PA_ENABLE pin. After transmission, PA_ENABLE is switched to L, and the microcontroller switches back to internal clocking. The ATA8403 is switched back to standby mode with $ENABLE = L$.

4.2 FSK Transmission

The ATA8403 is activated by $ENABLE = H$. PA_ENABLE must remain L for $t \geq 4\text{ms}$, then the CLK signal can be taken to clock the microcontroller and the power amplifier is switched on with $PA_ENABLE = H$. The chip is then ready for FSK modulation. The microcontroller starts to switch on and off the capacitor between the XTAL load capacitor and GND with an open-drain output port, thus changing the reference frequency of the PLL. If the switch is closed, the output frequency is lower than if the switch is open. After transmission PA_ENABLE is switched to L and the microcontroller switches back to internal clocking. The ATA8403 is switched back to standby mode with $ENABLE = L$.

The accuracy of the frequency deviation with XTAL pulling method is about $\pm 25\%$ when the following tolerances are considered.

Figure 4-1. Tolerances of Frequency Modulation



Using $C_4 = 9.2\text{pF} \pm 2\%$, $C_5 = 6.8\text{pF} \pm 5\%$, a switch port with $C_{Switch} = 3\text{pF} \pm 10\%$, stray capacitances on each side of the crystal of $C_{Stray1} = C_{Stray2} = 1\text{pF} \pm 10\%$, a parallel capacitance of the crystal of $C_0 = 3.2\text{pF} \pm 10\%$ and a crystal with $C_M = 13\text{fF} \pm 10\%$, typically results in an FSK deviation of $\pm 21.5\text{kHz}$ with worst case tolerances of $\pm 16.8\text{kHz}$ to $\pm 28.0\text{kHz}$.

4.3 CLK Output

An output CLK signal is provided for a connected microcontroller. The delivered signal is CMOS compatible if the load capacitance is lower than 10pF.

4.3.1 Clock Pulse Take-over

The clock of the crystal oscillator can be used for clocking the microcontroller. A special feature of Atmel®'s ATARx9x is that it starts with an integrated RC-oscillator to switch on the ATA8403 with ENABLE = H, and after 4ms assumes the clock signal of the transmission IC, so that the message can be sent with crystal accuracy.

4.3.2 Output Matching and Power Setting

The output power is set by the load impedance of the antenna. The maximum output power is achieved with a load impedance of $Z_{Load,opt} = (166 + j226)\Omega$ at 868.3MHz. There must be a low resistive path to V_S to deliver the DC current.

The delivered current pulse of the power amplifier is 7.7mA. The maximum output power is delivered to a resistive load of 475 Ω if the 0.53pF output capacitance of the power amplifier is compensated by the load impedance.

An optimum load impedance of:

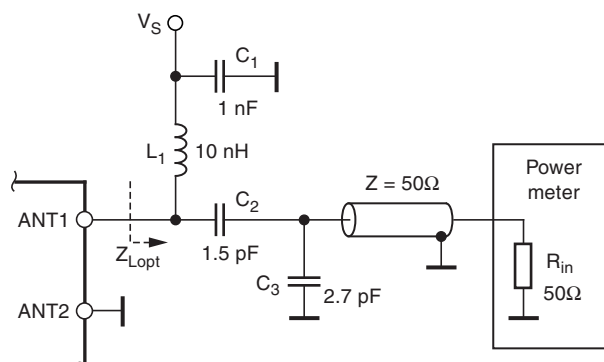
$Z_{Load} = 475\Omega \parallel j(2 \times p \times f \times 0.53pF) = (166 + j226)\Omega$ thus results in the maximum output power of 5.5dBm.

The load impedance is defined as the impedance seen from the ATA8403's ANT1, ANT2 into the matching network. Do not confuse this large signal load impedance with a small signal input impedance delivered as input characteristic of RF amplifiers and measured from the application into the IC instead of from the IC into the application for a power amplifier.

Less output power is achieved by lowering the real parallel part of 475 Ω where the parallel imaginary part should be kept constant.

Output power measurement can be done with the circuit shown in Figure 4-2. Note that the component values must be changed to compensate for the individual board parasitics until the ATA8403 has the right load impedance $Z_{Load,opt} = (166 + j226)\Omega$ at 868.3MHz. Also the damping of the cable used to measure the output power must be calibrated out.

Figure 4-2. Output Power Measurement



4.4 Application Circuit

A value of 68nF/X7R is recommended for the supply-voltage blocking capacitor C_3 (see Figure 4-3 on page 7 and Figure 4-4 on page 8). C_1 and C_2 are used to match the loop antenna to the power amplifier where C_1 typically is 3.9pF/NP0 and C_2 is 1pF/NP0. For C_2 , two capacitors in series should be used to achieve a better tolerance value and to have the possibility of realizing the $Z_{Load,opt}$ using standard valued capacitors.

C_1 , together with the pins of ATA8403 and the PCB board wires, forms a series resonance loop that suppresses the 1st harmonic. Therefore, the position of C_1 on the PCB is important. Normally the best suppression is achieved when C_1 is placed as close as possible to the pins ANT1 and ANT2.

The loop antenna should not exceed a width of 1.5mm, otherwise the Q-factor of the loop antenna is too high.

L_1 ($\approx 50nH$ to $100nH$) can be printed on PCB. C_4 should be selected so that the XTO runs on the load resonance frequency of the crystal. Normally, a 15pF load-capacitance crystal results in a value of 12pF.

Figure 4-3. ASK Application Circuit

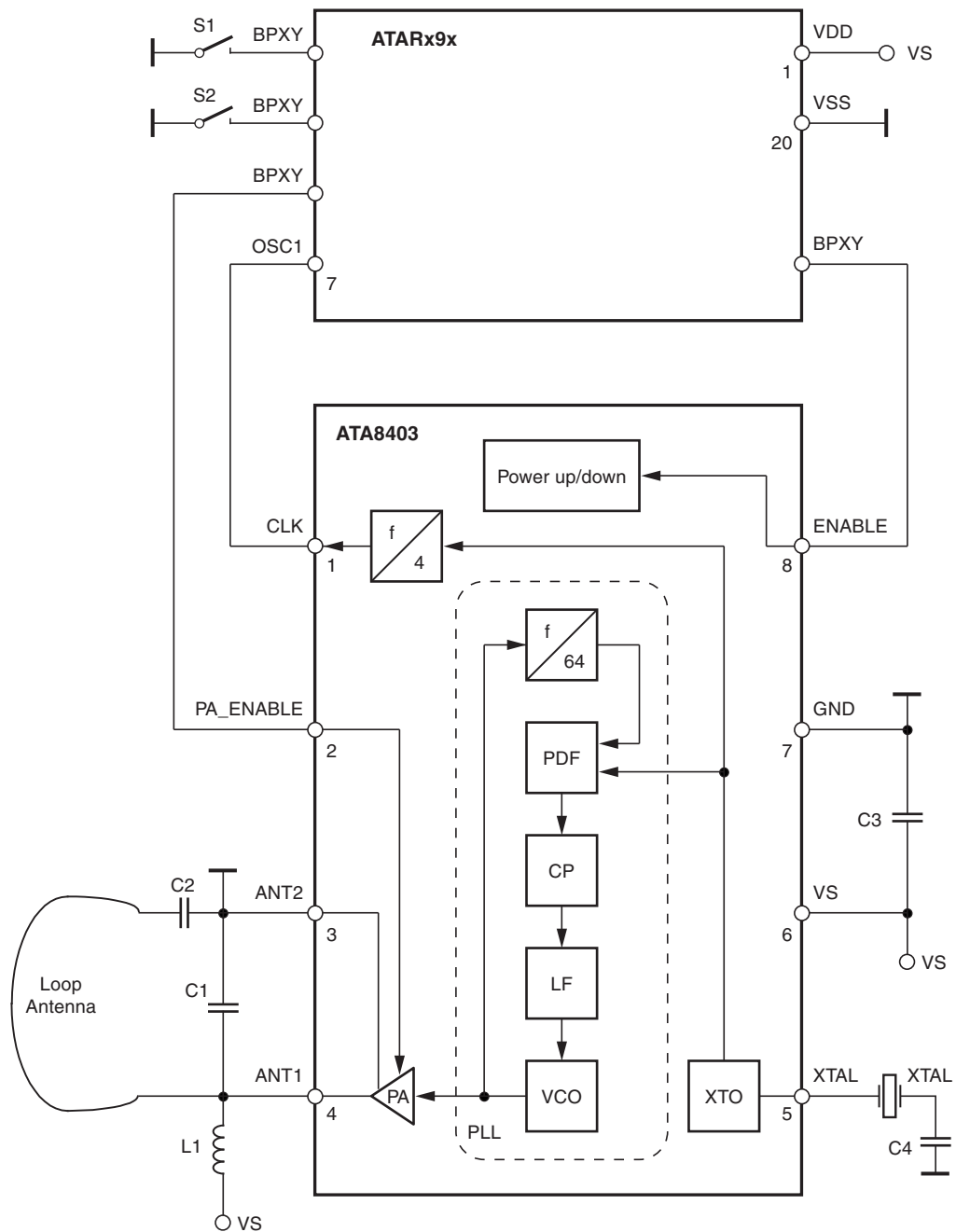


Figure 4-4. FSK Application Circuit

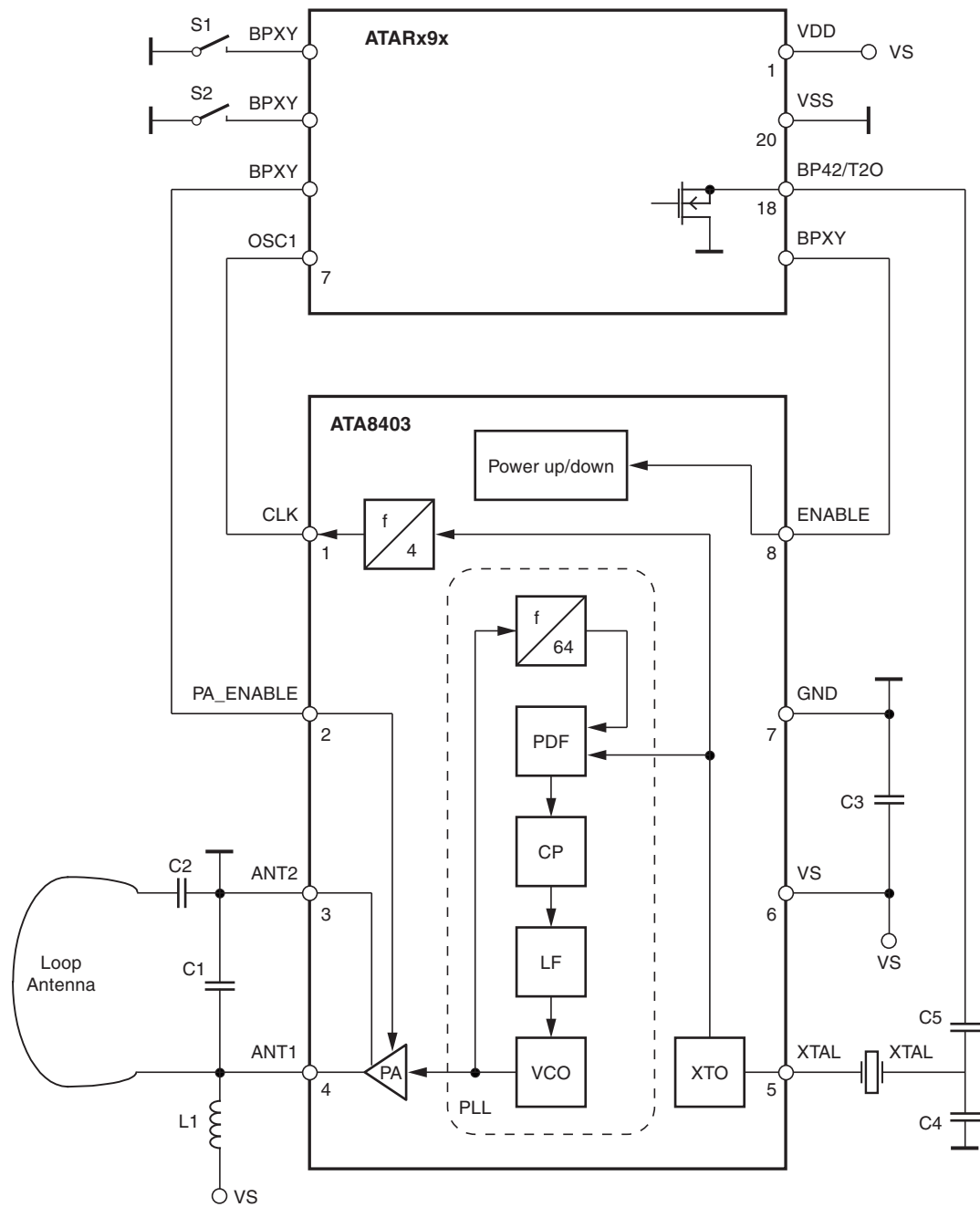
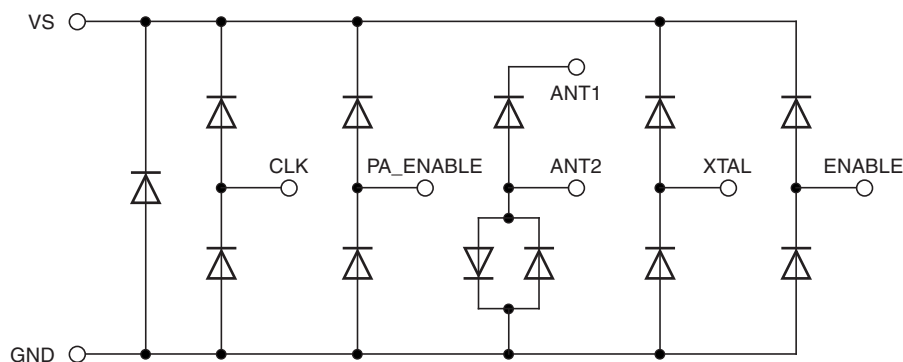


Figure 4-5. ESD Protection Circuit



5. Absolute Maximum Ratings

Stresses beyond those listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

| Parameters | Symbol | Minimum | Maximum | Unit |
|----------------------|---------------------|---------|---------------------|------|
| Supply voltage | V_S | | 5 | V |
| Power dissipation | P_{tot} | | 100 | mW |
| Junction temperature | T_J | | 150 | °C |
| Storage temperature | T_{stg} | -55 | +85 | °C |
| Ambient temperature | T_{amb} | -55 | +85 | °C |
| Input voltage | V_{maxPA_ENABLE} | -0.3 | $(V_S + 0.3)^{(1)}$ | V |

Note: 1. If $V_S + 0.3$ is higher than 3.7V, the maximum voltage will be reduced to 3.7V.

6. Thermal Resistance

| Parameters | Symbol | Value | Unit |
|------------------|------------|-------|------|
| Junction ambient | R_{thJA} | 170 | K/W |

7. Electrical Characteristics

$V_S = 2.0V$ to $4.0V$, $T_{amb} = 25^\circ C$ unless otherwise specified.

Typical values are given at $V_S = 3.0V$ and $T_{amb} = 25^\circ C$. All parameters are referred to GND (pin 7).

| Parameters | Test Conditions | Symbol | Min. | Typ. | Max. | Unit |
|----------------|---|-------------------|------|------|------|----------|
| Supply current | Power down, $V_{ENABLE} < 0.25V$, $-40^\circ C$ to $85^\circ C$ $V_{PA_ENABLE} < 0.25V$, $25^\circ C$ (100% correlation tested) | I_{S_off} | | < 10 | 350 | nA nA |
| Supply current | Power up, PA off, $V_S = 3V$, $V_{ENABLE} > 1.7V$, $V_{PA_ENABLE} < 0.25V$ | I_S | | 3.6 | 4.6 | mA |
| | Power up, $V_S = 3.0$, $V_{ENABLE} > 1.7V$, $V_{PA_ENABLE} > 1.7V$ | $I_{S_Transmit}$ | | 8.5 | 11 | mA |

Note: 1. If V_S is higher than 3.6V, the maximum voltage will be reduced to 3.6V.

7. Electrical Characteristics (Continued)

$V_S = 2.0V$ to $4.0V$, $T_{amb} = 25^\circ C$ unless otherwise specified.

Typical values are given at $V_S = 3.0V$ and $T_{amb} = 25^\circ C$. All parameters are referred to GND (pin 7).

| Parameters | Test Conditions | Symbol | Min. | Typ. | Max. | Unit |
|--|--|--------------------------------------|------------------|-------------|--------------------------|-------------------|
| Output power | $V_S = 3.0V$, $T_{amb} = 25^\circ C$, $f = 868.3MHz$, $Z_{Load} = (166 + j226)\Omega$ | P_{Ref} | 3.5 | 5.5 | 8 | dBm |
| Output power variation for the full temperature range | $T_{amb} = 25^\circ C$, $V_S = 3.0V$ $V_S = 2.0V$ | ΔP_{Ref} ΔP_{Ref} | | | -1.5 -4.0 | dB dB |
| Output power variation for the full temperature range | $T_{amb} = 25^\circ C$, $V_S = 3.0V$ $V_S = 2.0V$, $P_{Out} = P_{Ref} + \Delta P_{Ref}$ | ΔP_{Ref} ΔP_{Ref} | | | -2.0 -4.5 | dB dB |
| Achievable output-power range | Selectable by load impedance | P_{Out_typ} | -3 | | +5.5 | dBm |
| Spurious emission | $f_{CLK} = f_0/128$ Load capacitance at pin CLK = 10pF $f_0 \pm 1 \times f_{CLK}$ $f_0 \pm 4 \times f_{CLK}$ Other spurious are lower | | | -52 -52 | | dBc dBc |
| Oscillator frequency XTO (= phase comparator frequency) | $f_{XTO} = f_0/64$ f_{XTAL} = resonant frequency of the XTAL, $C_M \leq 10fF$, load capacitance selected accordingly $T_{amb} = 25^\circ C$ | f_{XTO} | | f_{XTAL} | | ppm |
| PLL loop bandwidth | | | | 250 | | kHz |
| Phase noise of phase comparator | Referred to $f_{PC} = f_{XTO}$, 25kHz distance to carrier | | | -116 | -110 | dBc/Hz |
| In-loop phase noise PLL | 25kHz distance to carrier | | | -80 | -74 | dBc/Hz |
| Phase noise VCO | At 1MHz At 36MHz | | | -89 -120 | -86 -117 | dBc/Hz dBc/Hz |
| Frequency range of VCO | | f_{VCO} | 868 | | 928 | MHz |
| Clock output frequency (CMOS microcontroller compatible) | | | | $f_0/256$ | | MHz |
| Voltage swing at pin CLK | $C_{Load} \leq 10pF$ | V_{Oh} V_{Ol} | $V_S \times 0.8$ | | $V_S \times 0.2$ | V V |
| Series resonance R of the crystal | | R_s | | | 110 | Ω |
| Capacitive load at pin XTO | | | | | 7 | pF |
| FSK modulation frequency rate | Duty cycle of the modulation signal = 50% | | 0 | | 32 | kHz |
| ASK modulation frequency rate | Duty cycle of the modulation signal = 50% | | 0 | | 50 | kHz |
| ENABLE input | Low level input voltage High level input voltage Input current high | V_{Il} V_{Ih} I_{In} | 1.7 | | 0.25 20 | V V μA |
| PA_ENABLE input | Low level input voltage High level input voltage Input current high | V_{Il} V_{Ih} I_{In} | 1.7 | | 0.25 $V_S^{(1)}$ 5 | V V μA |

Note: 1. If V_S is higher than 3.6V, the maximum voltage will be reduced to 3.6V.

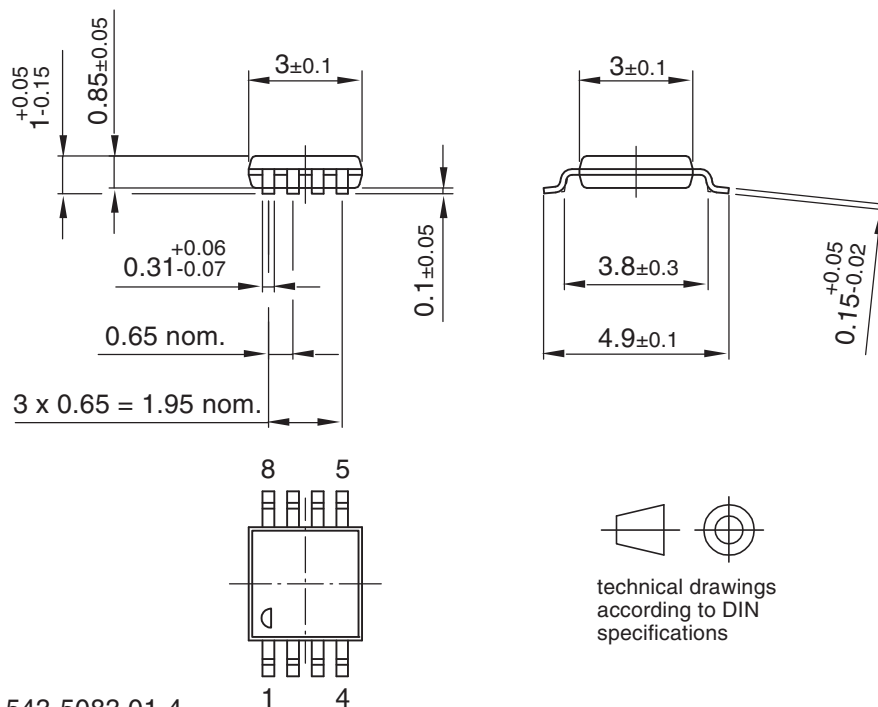
8. Ordering Information

| Extended Type Number | Package | MOQ | Remarks |
|----------------------|---------|----------|---------------------------|
| ATA8403C-6AQY | TSSOP8L | 5000 pcs | Taped and reeled, Pb-free |

9. Package Information

Package: TSSOP 8L

Dimensions in mm



Drawing-No.: 6.543-5083.01-4

Issue: 2; 15.03.04

10. Revision History

Please note that the following page numbers referred to in this section refer to the specific revision mentioned, not to this document.

| Revision No. | History |
|-------------------|---|
| 4983D-INDCO-08/12 | <ul style="list-style-type: none">• Features on page 1 changed |
| 4983C-INDCO-03/12 | <ul style="list-style-type: none">• Features on page 1 changed• Section 8 “Ordering Information” on page 11 changed |
| 4983B-INDCO-10/08 | <ul style="list-style-type: none">• Put datasheet in the newestest template• Section 7 “Electrical Characteristics” on page 10 changed |



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