

LM26 SOT-23, ±3°C Accurate, Factory Preset Thermostat

General Description

The LM26 is a precision, single digital-output, low-power thermostat comprised of an internal reference. DAC, temperature sensor and comparator. Utilizing factory programming, it can be manufactured with different trip points as well as different digital output functionality. The trip point (T_{OS}) can be preset at the factory to any temperature in the range of -55°C to +110°C in 1°C increments. The LM26 has one digital output (OS/OS/US/US), one digital input (HYST) and one analog output (V_{TEMP}). The digital output stage can be preset as either open-drain or push-pull. In addition, it can be factory programmed to be active HIGH or LOW. The digital output can be factory programmed to indicate an over temperature shutdown event (OS or OS) or an under temperature shutdown event (US or US). When preset as an overtemperature shutdown (OS) it will go LOW to indicate that the die temperature is over the internally preset T_{OS} and go HIGH when the temperature goes below (T_{OS}-T_{HYST}). Similarly, when preprogrammed as an undertemperature shutdown (US) it will go HIGH to indicate that the temperature is below T_{US} and go LOW when the temperature is above $(T_{US}+T_{HYST})$. The typical hysteresis, T_{HYST}, can be set to 2°C or 10°C and is controlled by the state of the HYST pin. A V_{TEMP} analog output provides a voltage that is proportional to temperature and has a -10.82mV/°C output slope.

Available parts are detailed in the ordering information. For other part options, contact a National Semiconductor Distributor or Sales Representative for information on minimum order qualification. The LM26 is currently available in a 5-lead SOT-23 package.

Applications

- Microprocessor Thermal Management
- Appliances
- Portable Battery Powered Systems

Fan Control

- Industrial Process Control
- HVAC Systems
- Remote Temperature Sensing
- Electronic System Protection

Features

- Internal comparator with pin programmable 2°C or 10°C hysteresis
- No external components required
- Open Drain or push-pull digital output; supports CMOS logic levels
- Internal temperature sensor with V_{TEMP} output pin
- V_{TEMP} output allows after-assembly system testing
- Internal voltage reference and DAC for trip-point setting
- Currently available in 5-pin SOT-23 plastic package
- Excellent power supply noise rejection
- UL Recognized Component R

Key Specifications

- Power Supply Voltage
- Power Supply Current 40µA(max)
 - 20µÀ(typ)

2.7V to 5.5V

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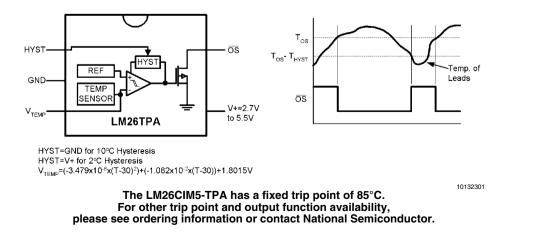
Hysteresis Temperature

2°C or 10°C(typ)

Temperature Trip Point Accuracy

Temperature Range	LM26CIM
–55°C to +110°C	±3°C (max)
+120°C	±4°C (max)

LM26CIM5-TPA Simplified Block Diagram and Connection Diagram

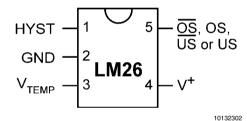


Ordering Information

For more detailed information on the suffix meaning see the part number template at the end of the Electrical Characteristics Section. Contact National Semiconductor for other set points and output options.

Order Number			NS Package	Trip Point Setting	Output Function
Bulk Rail	3000 Units in Tape &		Number		
	Reel	Top Mark			
LM26CIM5-HHD	LM26CIM5X-HHD	THHD	MA05B	0°C	Open Drain US
LM26CIM5-NPA	LM26CIM5X-NPA	TNPA	MA05B	45°C	Open Drain OS
LM26CIM5-PHA	LM26CIM5X-PHA	TPHA	MA05B	50°C	Open Drain OS
LM26CIM5-RPA	LM26CIM5X-RPA	TRPA	MA05B	65°C	Open Drain OS
LM26CIM5-SHA	LM26CIM5X-SHA	TSHA	MA05B	70°C	Open Drain OS
LM26CIM5-SPA	LM26CIM5X-SPA	TSPA	MA05B	75°C	Open Drain OS
LM26CIM5-TPA	LM26CIM5X-TPA	TTPA	MA05B	85°C	Open Drain OS
LM26CIM5-VHA	LM26CIM5X-VHA	TVHA	MA05B	90°C	Open Drain OS
LM26CIM5-VPA	LM26CIM5X-VPA	TVPA	MA05B	95°C	Open Drain OS
LM26CIM5-XHA	LM26CIM5X-XHA	TXHA	MA05B	100°C	Open Drain OS
LM26CIM5-XPA	LM26CIM5X-XPA	TXPA	MA05B	105°C	Open Drain OS
LM26CIM5-YHA	LM26CIM5X-YHA	TYHA	MA05B	110°C	Open Drain OS
LM26CIM5-YPA	LM26CIM5X-YPA	TYPA	MA05B	115°C	Open Drain OS
LM26CIM5-ZHA	LM26CIM5X-ZHA	TZHA	MA05B	120°C	Open Drain OS

Connection Diagram



Pin Descriptions

Pin	Pin	Function	Connection
Number	Name		
1	HYST	Hysteresis control, digital input	GND for 10°C or V+ for 2°C
2	GND	Ground, connected to the back side of the die through lead frame.	System GND
3	V_{TEMP}	Analog output voltage proportional to temperature	Leave floating or connect to a high impedance node.
4	V+	Supply input	2.7V to 5.5V with a 0.1µF bypass capacitor. For PSRR information see <i>Section Titled NOISE CONSIDERATIONS</i> .
5	ŌS	Overtemperature Shutdown open-drain active low thermostat digital output	Controller interrupt, system or power supply shutdown; pull-up resistor $\geq 10 k\Omega$
	OS	Overtemperature Shutdown push-pull active high thermostat digital output	Controller interrupt, system or power supply shutdown
	US	Undertemperature Shutdown open-drain active low thermostat digital output	System or power supply shutdown; pull-up resistor ≥ 10 k Ω
	US	Undertemperature Shutdown push-pull active high thermostat digital output	System or power supply shutdown

Note: pin 5 functionality and trip point setting are programmed during LM26 manufacture.

Absolute Maximum Ratin	GS (Note 1)	Storage Temperature	-65°C to + 150°C
Input Voltage Input Current at any pin (Note 2)	6.0V 5mA	ESD Susceptibility (Note 4) Human Body Model Machine Model	2500V 250V
Package Input Current(Note 2) Package Dissipation at $T_A = 25^{\circ}$ C	20mA	Operating Ratings (N	ote 1)
(Note 3) Soldering Information SOT23 Package	500mW	Specified Temperature Range LM26CIM	T_{MIN} ≤ T_A ≤ T_{MAX} –55°C ≤ T _A ≤ +125°C
Vapor Phase (60 seconds) Infrared (15 seconds)	215°C 220°C	Positive Supply Voltage (V+) Maximum V _{OUT}	+2.7V to +5.5V +5.5V

LM26 Electrical Characteristics

The following specifications apply for $V_{+} = 2.7V_{DC}$ to $5.5V_{DC}$, and V_{TEMP} load current = 0µA unless otherwise specified. Boldface limits apply for $T_A = T_J = T_{MIN}$ to T_{MAX} ; all other limits $T_A = T_J = 25^{\circ}C$ unless otherwise specified.

Symbol	Parameter	Conditions	Typical (Note 6)	LM26CIM Limits (Note 7)	Units (Limits)
emperature		•			•
	Trip Point Accuracy (Includes V _{REF} , DAC,	-55°C ≤ T _A ≤ +110°C		±3	°C (max)
	Comparator Offset, and Temperature Sensitivity errors)	+120°C		±4	°C (max)
	Trip Point Hysteresis	HYST = GND	11		°C
		HYST = V+	2		°C
	V _{TEMP} Output Temperature Sensitivity		-10.82		mV/°C
	V _{TEMP} Temperature Sensitivity Error to Equation:	–30°C ≤ T _A ≤ 120°C, 2.7V ≤ V+ ≤ 5.5V		±3	°C (max)
	$V_{O} = (-3.479 \times 10^{-6} \times (T-30)^{2}) + (-1.082 \times 10^{-2} \times (T-30)) + 1.8015 V$	-55° C ≤ T _A ≤ 120°C, 4.5V ≤ V+ ≤ 5.5V		±3	°C (max)
		$T_A = 30^{\circ}C$		±2.5	°C (max)
	V _{TEMP} Load Regulation	Source ≤ 1 µA	0.070		mV
		Sink ≤ 40 µA		0.7	mV (max)
	V _{TEMP} Line Regulation	$+2.7V \le V^+ \le +5.5V$,	-0.2		mV/V
		–30°C ≤ T _A ≤ +120°C			
۱ _S	Supply Current		16	20 40	μΑ (max) μΑ (max)
igital Outp	ut and Input				•
I _{OUT("1")}	Logical "1" Output Leakage Current (Note 9)	V+ = +5.0V	0.001	1	µA (max)
V _{OUT("0")}	Logical "0" Output Voltage	I _{OUT} = +1.2mA and V+≥2.7V; I _{OUT} = +3.2mA and V+≥4.5V; (Note 8)		0.4	V (max)
V _{OUT("1")}	Logical "1" Push-Pull Output Voltage	I _{SOURCE} = 500µA, V+ ≥ 2.7V		0.8 × V+	V (min)
		I _{SOURCE} = 800µA, V +≥4.5V		V+ – 1.5	V (min)
V _{IH}	HYST Input Logical "1" Threshold Voltage			0.8 × V+	V (min)
V _{IL}	HYST Input Logical "0" Threshold Voltage			0.2 × V+	V (max)

LM26

3

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

Note 2: When the input voltage (V_i) at any pin exceeds the power supply $(V_i < GND \text{ or } V_i > V^+)$, the current at that pin should be limited to 5mA. The 20mA maximum package input current rating limits the number of pins that can safely exceed the power supplies with an input current of 5mA to four. Under normal operating conditions the maximum current that pins 2, 4 or 5 can handle is limited to 5mA each.

Note 3: The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{Jmax} (maximum junction temperature), θ_{JA} (junction to ambient thermal resistance) and T_A (ambient temperature). The maximum allowable power dissipation at any temperature is $P_D = (T_{Jmax} - T_A)/\theta_{JA}$ or the number given in the Absolute Maximum Ratings, whichever is lower. For this device, $T_{Jmax} = 150^{\circ}$ C. For this device the typical thermal resistance (θ_{JA}) of the different package types when board mounted follow:

Package Type	θ _{JA}	
SOT23-5, MA05B	250°C/W	

Note 4: The human body model is a 100pF capacitor discharge through a 1.5kΩ resistor into each pin. The machine model is a 200pF capacitor discharged directly into each pin.

Note 5: See the URL "http://www.national.com/packaging/" for other recommendations and methods of soldering surface mount devices.

Note 6: Typicals are at $T_J = T_A = 25^{\circ}C$ and represent most likely parametric norm.

Note 7: Limits are guaranteed to National's AOQL (Average Outgoing Quality Level).

Note 8: Care should be taken to include the effects of self heating when setting the maximum output load current. The power dissipation of the LM26 would increase by 1.28mW when I_{OUT} =3.2mA and V_{OUT} =0.4V. With a thermal resistance of 250°C/W, this power dissipation would cause an increase in the die temperature of about 0.32°C due to self heating. Self heating is not included in the trip point accuracy specification.

Note 9: The 1µA limit is based on a testing limitation and does not reflect the actual performance of the part. Expect to see a doubling of the current for every 15°C increase in temperature. For example, the 1nA typical current at 25°C would increase to 16nA at 85°C.

Part Number Template

The series of digits labeled xyz in the part number LM26CIM-xyz, describe the set point value and the function of the output as follows:

The place holders xy describe the set point temperature as shown in the following table.

x (10x)	y (1x)	Temperature (°C)
A	-	-5
В	-	-4
С	-	-3
D	-	-2
E	-	-1
F	-	-0
Н	Н	0
J	J	1
К	К	2
L	L	3

x (10x)	y (1x)	Temperature (°C)
N	N	4
Р	Р	5
R	R	6
S	S	7
Т	Т	8
V	V	9
Х	-	10
Y	-	11
Z	-	12

The value of z describes the assignment/function of the output as shown in the following table:

Active-Low/High	Open-Drain/ Push- Pull	os /us	Value of z	Digital Output Function
0	0	0	А	Active-Low, Open-Drain, OS output
0	0	1	В	Active-Low, Open-Drain, US output
1	1	0	С	Active-High, Push-Pull, OS output
1	1	1	D	Active-High, Push-Pull, US output

For example:

- the part number LM26CIM5-TPA has T_{OS} = 85°C, and programmed as an active-low open-drain overtemperature shutdown output.
- the part number LM26CIM5-FPD has T_{US} = -5°C, and programmed as an active-high, push-pull undertemperature shutdown output.

Active-high open-drain and active-low push-pull options are available, please contact National Semiconductor for more information.

Functional Description

LM26 OPTIONS

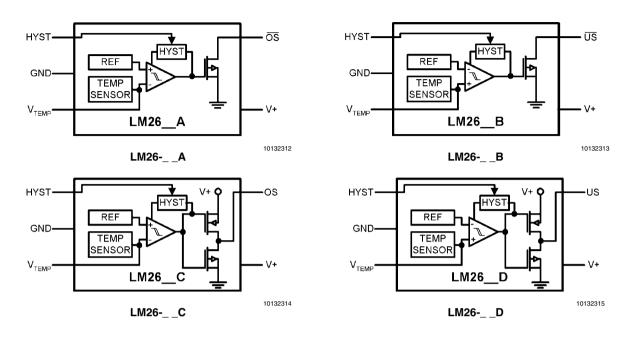


FIGURE 1. Output Pin Options Block Diagrams

The LM26 can be factory programmed to have a trip point anywhere in the range of -55° C to $+110^{\circ}$ C.

Applications Hints

AFTER-ASSEMBLY PCB TESTING

The LM26's V_{TEMP} output allows after-assembly PCB testing by following a simple test procedure. Simply measuring the V_{TEMP} output voltage will verify that the LM26 has been assembled properly and that its temperature sensing circuitry is functional. The V_{TEMP} output has very weak drive capability that can be overdriven by 1.5mA. Therefore, one can simply force the V_{TEMP} voltage to cause the digital output to change state, thereby verifying that the comparator and output circuitry function after assembly. Here is a sample test procedure that can be used to test the LM26CIM5-TPA which has an 85°C trip point.

1. Turn on V⁺ and measure V_{TEMP}. Then calculate the temperature reading of the LM26 using the equation:

$$V_{O} = (-3.479 \times 10^{-6} \times (T - 30)^{2}) + (-1.082 \times 10^{-2} \times (T - 30)) + 1.8015 V$$
(1)

or

$$\Gamma = -1525.04 + \sqrt{2.4182 \times 10^6 + \frac{1.8015 - V_{\text{TEMP}}}{3.479 \times 10^{-6}}}$$
(2)

 Verify that the temperature measured in step one is within (±3°C + error of reference temperature sensor) of the ambient/board temperature. The ambient/board temperature (reference temperature) should be measured using an extremely accurate calibrated temperature sensor.

- 3. 1. Observe that \overline{OS} is high.
 - 2. Drive V_{TEMP} to ground.
 - 3. Observe that \overline{OS} is now low.
 - 4. Release the V_{TEMP} pin.
 - 5. Observe that \overline{OS} is now high.
- 4. 1. Observe that OS is high.
 - 2. Drive V_{TEMP} voltage down gradually.
 - 3. When \overline{OS} goes low, note the V_{TEMP} voltage.
 - 4. V_{TEMP} Trig = V_{TEMP} at $\overline{\text{OS}}$ trigger (HIGH->LOW)
 - 5. Calculate Ttrig using Equation 2.
- 5. 1. Gradually raise V_{TEMP} until \overline{OS} goes HIGH. Note V_{TEMP} .
 - 2. Calculate T_{HYST} using Equation 2.

V_{TEMP} LOADING

The V_{TEMP} output has very weak drive capability (1 µA source, 40 µA sink). So care should be taken when attaching circuitry to this pin. Capacitive loading may cause the V_{TEMP} output to oscillate. Simply adding a resistor in series as shown in *Figure 2* will prevent oscillations from occurring. To determine the value of the resistor follow the guidelines given in *Table 1*. The same value resistor will work for either placement of the resistor. If an additional capacitive load is placed directly on the LM26 output, rather than across C_{LOAD}, it should be at least a factor of 10 smaller than C_{LOAD}.

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TABLE 1. Resistive compensation for capacitive loading of V_{TEMP} CLOAD R (Ω) 0 ≤100pF 1nF 8200 10nF 3000 100nF 1000 430 ≥1µF Heavy Capacitive os/os/ Load, Cable/Wiring HYST US/US ^{GND} LM26 VTEMP ٧ C_{LOAD} 0.1μF R 10132317 a) R in series with capacitor Heavy Capacitive OS/OS/ Load, Cable/ HYST US/US Wiring GNDLM26 TEME V 0.1µf C_{LOAD} 10132318 b) R in series with signal path

FIGURE 2. Resistor placement for capacitive loading compensation of V_{TEMP}

NOISE CONSIDERATIONS

The LM26 has excellent power supply noise rejection. Listed below is a variety of signals used to test the LM26 power supply rejection. False triggering of the output was not observed when these signals where coupled into the V+ pin of the LM26.

- square wave 400kHz, 1Vp-p
- square wave 2kHz, 200mVp-p
- sine wave 100Hz to 1MHz, 200mVp-p

Testing was done while maintaining the temperature of the LM26 one degree centigrade way from the trip point with the output not activated.

MOUNTING CONSIDERATIONS

The LM26 can be applied easily in the same way as other integrated-circuit temperature sensors. It can be glued or ce-

mented to a surface. The temperature that the LM26 is sensing will be within about $+0.06^{\circ}$ C of the surface temperature to which the LM26's leads are attached to.

This presumes that the ambient air temperature is almost the same as the surface temperature; if the air temperature were much higher or lower than the surface temperature, the actual temperature measured would be at an intermediate temperature between the surface temperature and the air temperature.

To ensure good thermal conductivity, the backside of the LM26 die is directly attached to the GND pin (pin 2). The temperatures of the lands and traces to the other leads of the LM26 will also affect the temperature that is being sensed.

Alternatively, the LM26 can be mounted inside a sealed-end metal tube, and can then be dipped into a bath or screwed into a threaded hole in a tank. As with any IC, the LM26 and accompanying wiring and circuits must be kept insulated and dry, to avoid leakage and corrosion. This is especially true if the circuit may operate at cold temperatures where condensation can occur. Printed-circuit coatings and varnishes such as Humiseal and epoxy paints or dips are often used to ensure that moisture cannot corrode the LM26 or its connections.

The junction to ambient thermal resistance (θ_{JA}) is the parameter used to calculate the rise of a part's junction temperature due to its power dissipation. For the LM26 the equation used to calculate the rise in the die junction temperature is as follows:

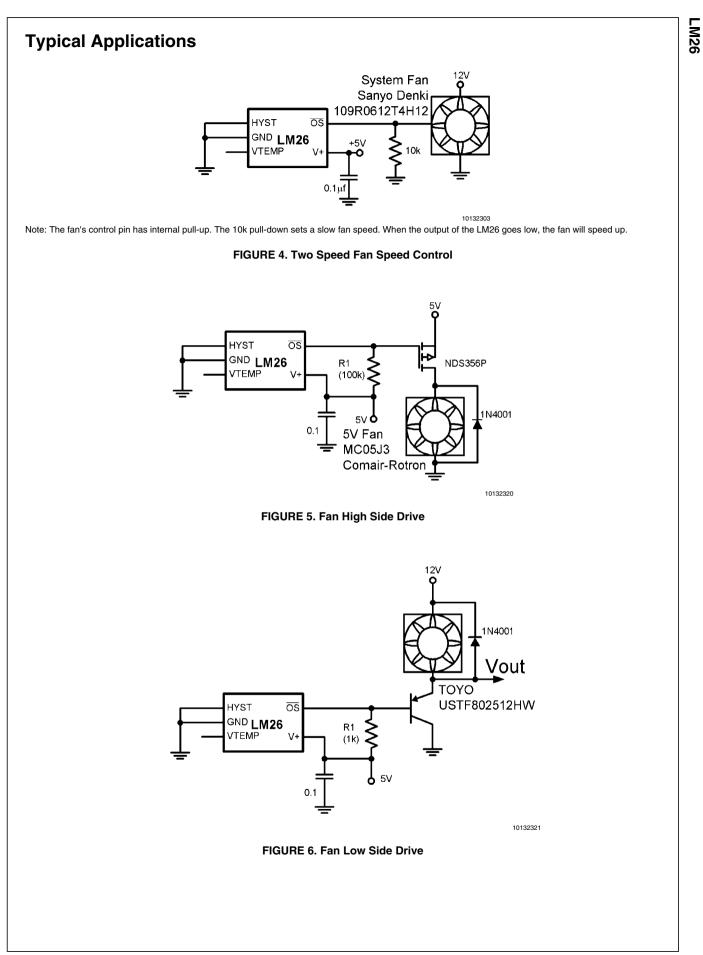
$$\Gamma_{J} = T_{A} + \Theta_{JA} (V^{+}|_{Q} + (V^{+} - V_{TEMP})|_{L_{TEMP}} + V_{DO}|_{DO})$$
(3)

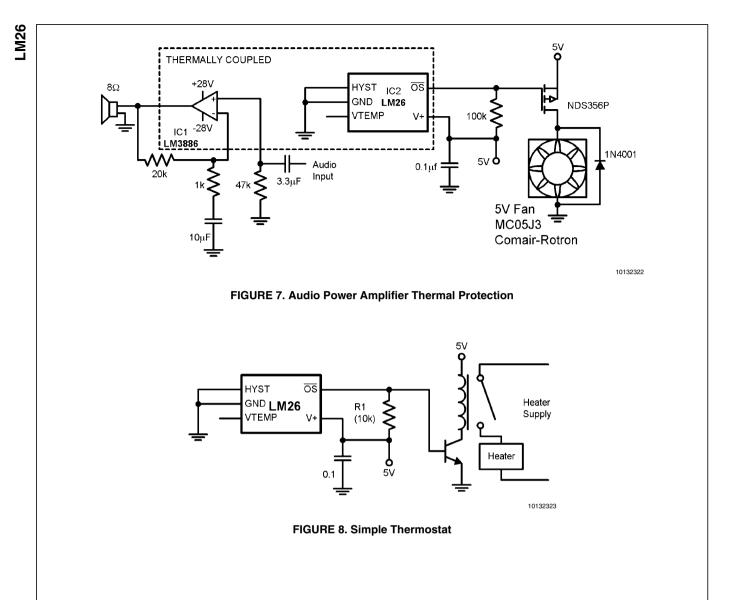
where T_A is the ambient temperature, V⁺ is the power supply voltage, I_Q is the quiescent current, I_{L_TEMP} is the load current on the V_{TEMP} output, V_{DO} is the voltage on the digital output, and I_{DO} is the load current on the digital output. Since the LM26's junction temperature is the actual temperature being measured, care should be taken to minimize the load current that the LM26 is required to drive.

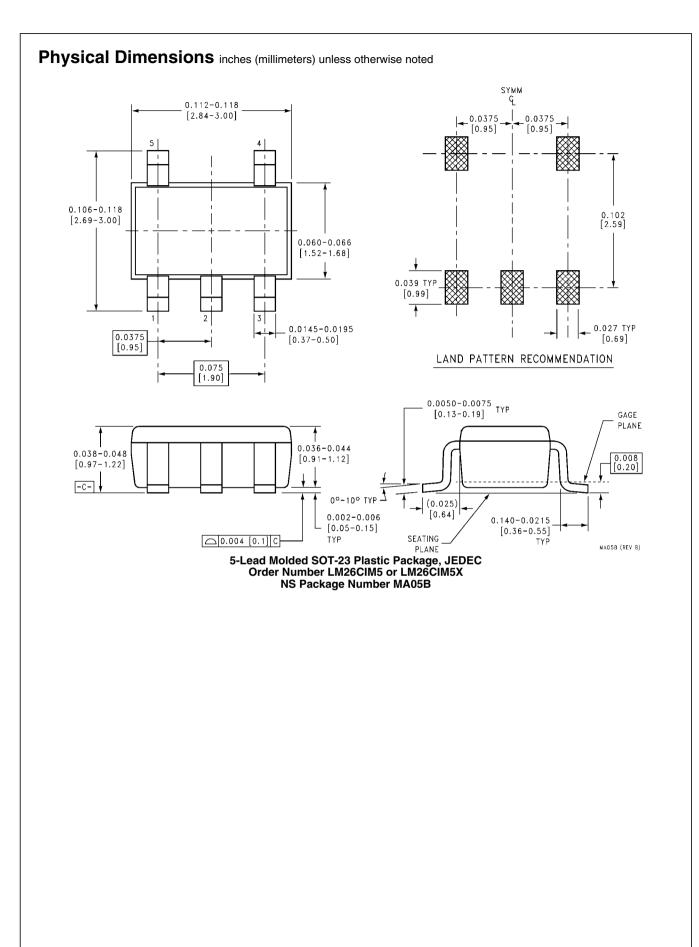
The tables shown in *Figure 3* summarize the thermal resistance for different conditions and the rise in die temperature of the LM26 without any loading on V_{TEMP} and a 10k pull-up resistor on an open-drain digital output with a 5.5V power supply.

	SOT: no hea		SOT23-5 small heat sink		
	θ _{JA} (°C/W)	T _J −T _A (°C)	θ _{JA} (°C/W)	T _J −T _A (°C)	
Still Air	250	0.11	TBD	TBD	
Moving Air	TBD	TBD	TBD	TBD	

FIGURE 3. Thermal resistance (θ_{JA}) and temperature rise due to self heating $(T_J - T_A)$







LM26

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