

Infrared Remote Control System-on-Chip

General Description

The MAXQ617 is a low-power, 16-bit MAXQ® microcontroller designed for low-power applications including universal remote controls, consumer electronics, and white goods. The device combines a powerful 16-bit RISC microcontroller and integrated peripherals including a universal synchronous/asynchronous receiver-transmitter (USART), an SPI master/slave and two I²C communications ports, along with an IR module with carrier frequency generation and flexible port I/O capable of multiplexed keypad control. An internal amplifier eliminates the need for external circuitry to drive the IR receiver pin.

The device includes 80KB of flash memory and 4KB of data SRAM.

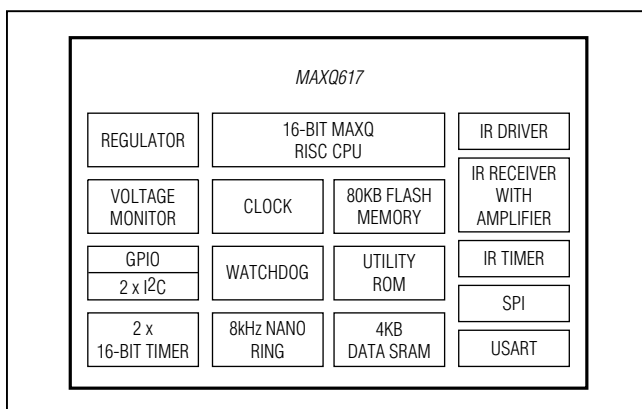
For the ultimate in low-power battery-operated performance, the device includes an ultra-low-power stop mode (0.2µA typ). In this mode, the minimum amount of circuitry is powered. Wake-up sources include external interrupts, the power-fail interrupt, and a timer interrupt. The microcontroller runs from a wide 1.67V to 3.6V operating voltage.

Applications

Universal Remote Controls for Tablets

Universal Remote Controls for Smartphones

Block Diagram



MAXQ is a registered trademark of Maxim Integrated Products, Inc.

Features

- ◆ High-Performance, Low-Power, 16-Bit RISC Core
- ◆ Internal 12MHz Oscillator Requires no External Components
- ◆ 1.67V to 3.6V Operating Voltage
- ◆ 33 Total Instructions for Simplified Programming
- ◆ Three Independent Data Pointers Accelerate Data Movement with Automatic Increment/Decrement
- ◆ Dedicated Pointer for Direct Read from Code Space
- ◆ 16-Bit Instruction Word, 16-Bit Data Bus
- ◆ 16 x 16-Bit General-Purpose Working Registers
- ◆ Memory Features
 - ◇ 80KB Flash Memory
 - ◇ 4KB Data SRAM
- ◆ Additional Peripherals
 - ◇ Power-Fail Warning
 - ◇ Power-On Reset (POR)/Brownout Reset
 - ◇ Automatic IR Carrier Frequency Generation and Modulation
 - ◇ IR Learning Amplifier
 - ◇ IR Transmit Driver with 200mA (min) Sink Current at 1.8V
 - ◇ Two 16-Bit Programmable Timers/Counters with Prescaler and Capture/Compare
 - ◇ One SPI, Two I²C, and One USART Port
 - ◇ Programmable Watchdog Timer
 - ◇ 8kHz Nanopower Ring Oscillator Wake-Up Timer
 - ◇ Up to 10 General-Purpose I/Os
- ◆ Low Power Consumption
 - ◇ 0.2µA (typ), 2.0µA (max) in Stop Mode, T_A = +25°C, Power-Fail Monitor Disabled
 - ◇ 2.0mA (typ) at 12MHz in Active Mode

[Ordering Information](#) appears at end of data sheet.

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ABSOLUTE MAXIMUM RATINGS

(All voltages with respect to GND.)

Voltage Range on V_{DD}-0.3V to +3.6V
 Voltage Range on Any Lead Except V_{DD} ..-0.3V to ($V_{DD} + 0.5V$)
 Continuous Power Dissipation ($T_A = +70^\circ C$)
 WLP (multilayer board)
 (derate 17.20mW/ $^\circ C$ above $+70^\circ C$).....1600mW

Operating Temperature Range.....-20 $^\circ C$ to +70 $^\circ C$
 Storage Temperature Range.....-65 $^\circ C$ to +150 $^\circ C$
 Soldering Temperature (reflow)+260 $^\circ C$
 Continuous Output Current
 Any Single I/O Pin.....32mA
 All I/O Pins Combined.....32mA

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

PACKAGE THERMAL CHARACTERISTICS

WLP

Junction-to-Ambient Thermal Resistance (θ_{JA})58 $^\circ C/W$

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to www.maxim-ic.com/thermal-tutorial.

ELECTRICAL CHARACTERISTICS

(Limits are 100% tested at $T_A = +25^\circ C$ and $T_A = +85^\circ C$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked GBD are guaranteed by design and not production tested.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|-----------------|--|-----------|------|------|---------|
| Supply Voltage | V_{DD} | | V_{RST} | | 3.6 | V |
| 1.8V Internal Regulator | V_{REG18} | | 1.59 | 1.8 | 1.98 | V |
| Power-Fail Warning Voltage | V_{PFW} | PFWARNCN[1:0] = 00 (default) | 1.64 | 1.67 | 1.70 | V |
| | | GBD, all other values of PFWARNCN[1:0] as shown in Table 3 | -3% | | +3% | |
| Power-Fail Reset Voltage | V_{RST} | | 1.61 | | 1.67 | V |
| Power-Fail Warning/Reset Offset | $V_{PFWOFFSET}$ | PFWARNCN[1:0] = 00, $V_{PFW} > V_{RST}$ | | 30 | | mV |
| Power-On Reset Voltage | V_{POR} | Monitors V_{DD} | | 1.2 | | V |
| RAM Data Retention Voltage | V_{DRV} | | | 1.0 | | V |
| Active Current | I_{DD_1} | $f_{SYSCLK} = 12MHz$, executing code from flash memory, all inputs connected to GND/ V_{DD} , outputs do not source or sink current | | 2 | 3.5 | mA |
| Stop Mode Current | I_{S1} | $T_A = +25^\circ C$ (power-fail off) | | 0.15 | 2.0 | μA |
| | | $T_A = 0^\circ C$ to $+70^\circ C$ (power-fail off) | | 0.15 | 8 | |
| | I_{S2} | $T_A = +25^\circ C$ (power-fail on) | | 22 | 31 | |
| | | $T_A = 0^\circ C$ to $+70^\circ C$ (power-fail on) | | 27.6 | 38 | |
| Power Consumption During Power-On Reset | I_{POR} | During POR while $V_{DD} < V_{POR}$ | | 100 | | nA |

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ELECTRICAL CHARACTERISTICS (continued)

(Limits are 100% tested at $T_A = +25^\circ\text{C}$ and $T_A = +85^\circ\text{C}$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked GBD are guaranteed by design and not production tested.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|----------------|---|------------------------|-----|-----------------|---------------|
| Stop Mode Resume Time | t_{ON} | | 375 + (8192 t_{CK}) | | | μs |
| Input Low Voltage for IRRX and All Port Pins | V_{IL} | | V_{GND} | | 0.3 V_{DD} | V |
| Input High Voltage for IRRX and All Port Pins | V_{IH} | | 0.7 V_{DD} | | V_{DD} | V |
| Input Hysteresis (Schmitt) | V_{IHYS} | $V_{DD} = 3.3\text{V}$, $T_A = +25^\circ\text{C}$ | 300 | | | mV |
| IRRX Input Filter Pulse-Width Reject | t_{IRRX_R} | | 50 | | | ns |
| IRRX Input Filter Pulse-Width Accept | t_{IRRX_A} | | 300 | | | ns |
| IRTX Sink Current | I_{IRTX} | $V_{IRTX} \geq 0.25\text{V}$ | 200 | | | mA |
| Output Low Voltage for All Port Pins | V_{OL} | $V_{DD} = 3.6\text{V}$, $I_{OL} = 11\text{mA}$ | 0.4 | | 0.5 | V |
| | | $V_{DD} = 2.35\text{V}$, $I_{OL} = 8\text{mA}$ | 0.4 | | 0.5 | |
| | | $V_{DD} = 1.8\text{V}$, $I_{OL} = 4.5\text{mA}$ | 0.4 | | 0.5 | |
| Output High Voltage All Port Pins | V_{OH} | $I_{OH} = -2\text{mA}$ | $V_{DD} - 0.5$ | | V_{DD} | V |
| Input/Output Pin Capacitance for All Port Pins | C_{IO} | | 15 | | | pF |
| Input Leakage Current for All Pins | I_L | Internal pullup disabled | -100 | | +100 | nA |
| Input Pullup Resistor for $\overline{\text{RESET}}$, IRRX, and All Port Pins | R_{PU} | $V_{DD} = 3.0\text{V}$, $V_{OL} = 0.4\text{V}$ | 16 | 28 | 39 | k Ω |
| | | $V_{DD} = 1.8\text{V}$, $V_{OL} = 0.4\text{V}$ | 18 | 31 | 43 | k Ω |
| LEARNING AMPLIFIER | | | | | | |
| IRRX Amplifier Input Low Detection | I_{DL} | $V_{DD} = 1.8\text{V}$, $I_{RRCVEN} = 1$ | 0.2 | | | μA |
| IRRX Amplifier Input High Detection | I_{DH} | $V_{DD} = 1.8\text{V}$, $I_{RRCVEN} = 1$ | 1.25 | | | μA |
| CLOCK | | | | | | |
| Internal Oscillator Frequency | f_{OSC} | | 12 | | | MHz |
| Internal Oscillator Variability | f_{OSC_VAR} | $T_A = -20^\circ\text{C}$ to $+70^\circ\text{C}$ | $\pm 1\%$ | | | MHz |
| | | $T_A = +25^\circ\text{C}$, $V_{DD} = 1.8\text{V}$ | $\pm 0.5\%$ | | | |
| | | $T_A = +15^\circ\text{C}$ to $+40^\circ\text{C}$, $V_{DD} = 1.8\text{V} \pm 5\%$ | $\pm 0.5\%$ | | | |

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ELECTRICAL CHARACTERISTICS (continued)

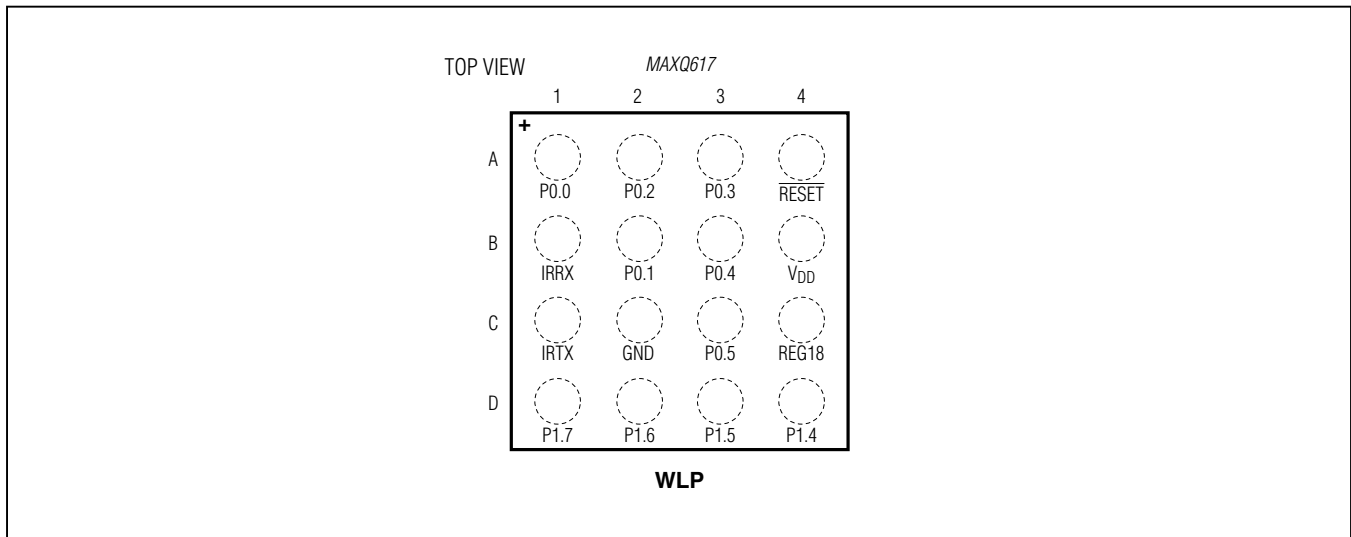
(Limits are 100% tested at $T_A = +25^\circ\text{C}$ and $T_A = +85^\circ\text{C}$. Limits over the operating temperature range and relevant supply voltage range are guaranteed by design and characterization. Specifications marked GBD are guaranteed by design and not production tested.)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|----------------|---|--------------|-------------|-----------------------|---------------|
| System Clock Period | t_{CK} | | | $1/f_{OSC}$ | | ns |
| System Clock Frequency | f_{CK} | | | $1/t_{CK}$ | | MHz |
| NANOPOWER RING | | | | | | |
| Nanopower Ring Frequency | f_{NANO} | $T_A = +25^\circ\text{C}$ | 3.0 | 8.0 | 20.0 | kHz |
| | | $T_A = +25^\circ\text{C}$, $V_{DD} = \text{POR voltage}$ | 1.7 | 2.4 | | |
| WAKE-UP TIMER | | | | | | |
| Wakeup Timer Interval | t_{WAKEUP} | | $1/f_{NANO}$ | | 65,535/ f_{NANO} | s |
| FLASH MEMORY | | | | | | |
| System Clock During Flash Programming/Erase | $f_{FPSYSCLK}$ | $f_{FPSYSCLK}/(\text{FCKDIV}[3:0]+1)$ must equal 1MHz, verify PFI = 0 before calling utility ROM. | | f_{OSC} | | MHz |
| Flash Erase Time | t_{ME} | Mass erase | | 40 | | ms |
| | t_{ERASE} | Page erase | | 40 | | ms |
| Flash Programming Time per Word | t_{PROG} | Excluding utility ROM overhead | | 40 | | μs |
| Write/Erase Cycles | | | 20,000 | | | Cycles |
| Data Retention | | $T_A = +25^\circ\text{C}$ | 100 | | | Years |

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Ball Configuration



Ball Description

| BALL | NAME | FUNCTION |
|--------------------------|-----------------|---|
| POWER BALLS | | |
| B4 | V _{DD} | Supply Voltage |
| C2 | GND | Ground. Connect directly to the ground plane. |
| C4 | REG18 | 1.8V Regulator Output. This pin must be connected to ground through a 1.0μF external ceramic-chip capacitor. The capacitor must be placed as close to this pin as possible. No devices other than the capacitor should be connected to this pin. |
| RESET BALLS | | |
| A4 | RESET | Digital, Active-Low Reset Input/Output. The device remains in reset as long as this pin is low and begins executing from the utility ROM at address 8000h when this pin returns to a high state. The pin includes pullup current source; if this pin is driven by an external device, it should be driven by an open-drain source capable of sinking in excess of 4mA. This pin can be left unconnected if there is no need to place the device in a reset state using an external signal. This pin is driven low as an output when an internal reset condition occurs. |
| IR FUNCTION BALLS | | |
| C1 | IRTX | IR Transmit Output. This pin defaults to a high-impedance input with the weak pullup disabled during all forms of reset. Software must configure this pin after release from reset to remove the high-impedance input condition. |
| B1 | IRRX | IR Receive Input. This pin defaults to a high-impedance input with the weak pullup disabled during all forms of reset. Software must configure this pin after release from reset to remove the high-impedance input condition and to enable the IR amplifier if desired. |

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Ball Description (continued)

| BALL | NAME | FUNCTION | |
|---|------------------------------|---|---|
| GENERAL-PURPOSE I/O AND SPECIAL FUNCTION BALLS | | | |
| | | Port 0 General-Purpose, Digital I/O Pins. These port pins function as general-purpose I/O pins with their input and output states controlled by the PD0, PO0, and PI0 registers. All port pins default to high-impedance mode after a reset. Software must configure these pins after release from reset to remove the high-impedance condition. All special functions must be enabled from software before they can be used. | |
| | | GPIO PORT PIN | SPECIAL FUNCTION |
| A1 | P0.0 RX0 MOSI INT0 | P0.0 | RX0: USART receive MOSI: SPI master out slave in INT0: External interrupt 0 |
| B2 | P0.1 TX0 MISO INT1 | P0.1 | TX0: USART transmit MISO: SPI master in slave out INT1: External interrupt 1 |
| A2 | P0.2 SCL0 SCLK INT2 | P0.2 | SCL0: I ² C clock SCLK: SPI clock INT2: External interrupt 2 |
| A3 | P0.3 SDA0 SSEL INT3 | P0.3 | SDA0: I ² C data SSEL: SPI slave select INT3: External interrupt 3 |
| B3 | P0.4 TBA0 INT4 | P0.4 | TBA0: Timer B A0 INT4: External interrupt 4 |
| C3 | P0.5 TBB0 INT5 | P0.5 | TBB0: Timer B B0 INT5: External interrupt 5 |
| | | Port 1 General-Purpose, Digital I/O Pins. These port pins function as general-purpose I/O pins with their input and output states controlled by the PD1, PO1, and PI1 registers. The JTAG pins default to their JTAG function with weak pullups enabled after a reset. The JTAG function can be disabled using the TAP bit in the SC register. | |
| | | GPIO PORT PIN | SPECIAL FUNCTION |
| D4 | P1.4/TCK/SCL1 | P1.4 | SCL1: I ² C clock, TCK: JTAG test clock |
| D3 | P1.5/TDI/SDA1 | P1.5 | SDA1: I ² C data, TDI: JTAG data input |
| D2 | P1.6/TMS | P1.6 | TMS: JTAG test mode select |
| D1 | P1.7/TDO | P1.7 | TDO: JTAG Data Output. TDO functions as the test-data output on reset and defaults to an input with a weak pullup. The output function of the test data is only enabled during the TAP's Shift_IR or Shift_DR states. |

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Detailed Description

The MAXQ617 provides integrated, low-cost solutions that simplify the design of IR communications equipment such as universal remote controls. Standard features include the highly optimized, single-cycle, MAXQ, 16-bit RISC core; 80KB flash memory; 4KB data RAM; soft stack; 16 general-purpose registers; and three data pointers. The MAXQ core has the industry's best MIPS/mA rating, allowing developers to achieve the same performance as competing microcontrollers at substantially lower clock rates. Lower active-mode current combined with the even lower stop-mode current (0.2 μ A typ) results in increased battery life. Application-specific peripherals include flexible timers for generating IR carrier frequencies and modulation. A high-current IR drive pin operates with an internal receiver amplifier without external components. It also includes general-purpose I/O pins ideal for keypad matrix input, and a power-fail-detection circuit to notify the application when the supply voltage is nearing the microcontroller's minimum operating voltage.

The internal 12MHz oscillator requires no external components and executes instructions. Operating from DC to 12MHz, almost all instructions execute in a single clock cycle (83.3ns at 12MHz), enabling nearly 12MIPS true-code operation. When active device operation is not required, an ultra-low-power stop mode can be invoked from software, resulting in quiescent current consumption of less than 0.2 μ A (typ) and 2.0 μ A (max). The combination of high-performance instructions and ultra-low stop-mode current increases battery life over competing microcontrollers. An integrated POR circuit with brownout support resets the device to a known condition following a power-up cycle or brownout condition. Additionally, a power-fail warning flag is set, and a power-fail interrupt can be generated when the system voltage falls below the power-fail warning voltage, V_{PFW} . The configurable power-fail warning feature allows the application to notify the user that the system supply is low and appropriate action should be taken.

Microprocessor

The device is based on Maxim Integrated's low-power, 16-bit MAXQ20S. The core supports the Harvard memory architecture with separate 16-bit program and data address buses. A fixed 16-bit instruction word is standard, but data can be arranged in 8 or 16 bits. The MAXQ core in the device is implemented as a pipe-

lined processor with performance approaching 1MIPS per MHz. The 16-bit data path is implemented around register modules, and each register module contributes specific functions to the core. The accumulator module consists of sixteen 16-bit registers and is tightly coupled with the arithmetic logic unit (ALU). A configurable soft stack supports program flow.

Execution of instructions is triggered by data transfer between functional register modules or between a functional register module and memory. Because data movement involves only source and destination modules, circuit switching activities are limited to active modules only. For power-conscious applications, this approach localizes power dissipation and minimizes switching noise. The modular architecture also provides a maximum of flexibility and reusability that are important for a microprocessor used in embedded applications.

The MAXQ instruction set is highly orthogonal. All arithmetical and logical operations can use any register in conjunction with the accumulator. Data movement is supported from any register to any other register. Memory is accessed through specific data-pointer registers with autoincrement/decrement support.

Memory

The microcontroller incorporates several memory types:

- 80KB flash memory
- 4KB SRAM data memory
- Dedicated utility ROM
- Soft stack

Stack Memory

The device provides a soft stack that can be used to store program return addresses (for subroutine calls and interrupt handling) and other general-purpose data. This soft stack is located in the SRAM data memory, which means that the SRAM data memory must be shared between the soft stack and general-purpose application data storage. However, the location and size of the soft stack is determined by the user, providing maximum flexibility when allocating resources for a particular application. The stack is used automatically by the processor when the CALL, RET, and RETI instructions are executed and when an interrupt is serviced. An application can also store and retrieve values explicitly using the stack by means of the PUSH, POP, and POPI instructions.

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The SP pointer indicates the current top of the stack, which initializes by default to the top of the SRAM data memory. As values are pushed onto the stack, the SP pointer decrements, which means that the stack grows downward towards the bottom (lowest address) of the data memory. Popping values off the stack causes the SP pointer value to increase. Refer to the *User's Guide* for more details.

Utility ROM

The utility ROM is located in program space beginning at address 8000h. This ROM includes the following routines:

- Production test routines (internal memory tests, memory loader, etc.), which are used for internal testing only, and are generally of no use to the end-application developer
- User-callable routines for buffer copying and fast table lookup (more information on these routines can be found in the *User's Guide*)

Following any reset, execution begins in the utility ROM at address 8000h. At this point, unless test mode has been invoked (which requires special programming through the JTAG interface), the utility ROM in the device always automatically jumps to location 0000h, which is the beginning of user application code.

Watchdog Timer

The internal watchdog timer greatly increases system reliability. The timer resets the device if software execution is disturbed. The watchdog timer is a free-running counter designed to be periodically reset by the application software. If software is operating correctly, the counter is

periodically reset and never reaches its maximum count. However, if software operation is interrupted, the timer does not reset, triggering a system reset and optionally a watchdog timer interrupt. This protects the system against electrical noise or electrostatic discharge (ESD) upsets that could cause uncontrolled processor operation. The internal watchdog timer is an upgrade to older designs with external watchdog devices, reducing system cost and simultaneously increasing reliability.

The watchdog timer functions as the source of both the watchdog timer timeout and the watchdog timer reset. The timeout period can be programmed in a range of 2^{15} to 2^{24} system clock cycles. An interrupt is generated when the timeout period expires if the interrupt is enabled. All watchdog timer resets follow the programmed interrupt timeouts by 512 system clock cycles. If the watchdog timer is not restarted for another full interval in this time period, a system reset occurs when the reset timeout expires. See [Table 1](#).

IR Carrier Generation and Modulation Timer

The dedicated IR timer/counter module simplifies low-speed infrared (IR) communication. The IR timer implements two pins (IRTX and IRRX) for supporting IR transmit and receive, respectively. The IRTX pin has no corresponding port pin designation, so the standard PD, PO, and PI port control status bits are not present. However, the IRTX pin output can be manipulated high or low using the PWCN.IRTXOUT and PWCN.IRTXOE bits when the IR timer is not enabled (i.e., IREN = 0).

Table 1. Watchdog Interrupt Timeout (Sysclk = 12MHz, CD[1:0] = 00)

| WD[1:0] | WATCHDOG CLOCK | WATCHDOG INTERRUPT TIMEOUT | WATCHDOG RESET AFTER WATCHDOG INTERRUPT (μs) |
|---------|------------------------|----------------------------|--|
| 00 | $\text{Sysclk}/2^{15}$ | 2.7ms | 42.7 |
| 01 | $\text{Sysclk}/2^{18}$ | 21.9ms | 42.7 |
| 10 | $\text{Sysclk}/2^{21}$ | 174.7ms | 42.7 |
| 11 | $\text{Sysclk}/2^{24}$ | 1.4s | 42.7 |

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The IR timer is composed of a carrier generator and a carrier modulator. The carrier generation module uses the 16-bit IR carrier register (IRCA) to define the high and low time of the carrier through the IR carrier high byte (IRCAH) and IR carrier low byte (IRCAL). The carrier modulator uses the IR data bit (IRDATA) and IR modulator time register (IRMT) to determine whether the carrier or the idle condition is present on IRTX.

The IR timer is enabled when the IR enable bit (IREN) is set to 1. The IR Value register (IRV) defines the beginning value for the carrier modulator. During transmission, the IRV register is initially loaded with the IRMT value and begins down counting towards 0000h, whereas in receive mode it counts upward from the initial IRV register value. During the receive operation, the IRV register can be configured to reload with 0000h when capture occurs on detection of selected edges or can be allowed to continue free-running throughout the receive operation. An overflow occurs when the IR timer value rolls over from 0FFFFh to 0000h. The IR overflow flag (IROV) is set to 1 and an interrupt is generated if enabled (IRIE = 1).

Carrier Generation Module

The IRCAH byte defines the carrier high time in terms of the number of IR input clocks, whereas the IRCAL byte defines the carrier low time.

- IR Input Clock (f_{IRCLK}) = $f_{SYS}/2^{IRDIV[2:0]}$
- Carrier Frequency ($f_{CARRIER}$) = $f_{IRCLK}/(IRCAH + IRCAL + 2)$
- Carrier High Time = $IRCAH + 1$
- Carrier Low Time = $IRCAL + 1$
- Carrier Duty Cycle = $(IRCAH + 1)/(IRCAH + IRCAL + 2)$

During transmission, the IRCA register is latched for each IRV down-count interval, and is sampled along with the IRTXPOL and IRDATA bits at the beginning of each new IRV down-count interval so that duty-cycle variation and frequency shifting is possible from one interval to the next, which is illustrated in [Figure 1](#).

[Figure 2](#) illustrates the basic carrier generation and its path to the IRTX output pin. The IR transmit polarity bit (IRTXPOL) defines the starting/idle state and the carrier polarity of the IRTX pin when the IR timer is enabled.

IR Transmission

During IR transmission (IRMODE = 1), the carrier generator creates the appropriate carrier waveform, while the carrier modulator performs the modulation. The carrier modulation can be performed as a function of carrier cycles or IRCLK cycles dependent on the setting of the IRCFME bit. When IRCFME = 0, the IRV down counter is clocked by the carrier frequency and thus the modulation is a function of carrier cycles ([Figure 3](#)). When IRCFME = 1, the IRV down counter is clocked by IRCLK, allowing carrier modulation timing with IRCLK resolution.

The IRTXPOL bit defines the starting/idle state as well as the carrier polarity for the IRTX pin. If IRTXPOL = 1, the IRTX pin is set to a logic-high when the IR timer module is enabled. If IRTXPOL = 0, the IRTX pin is set to a logic-low when the IR timer is enabled.

A separate register bit, IR data (IRDATA), is used to determine whether the carrier generator output is output to the IRTX pin for the next IRMT carrier cycles. When IRDATA = 1, the carrier waveform (or inversion of this waveform if IRTXPOL = 1) is output on the IRTX pin during the next IRMT cycles. When IRDATA = 0, the idle condition, as defined by IRTXPOL, is output on the IRTX pin during the next IRMT cycles.

The IR timer acts as a down counter in transmit mode. An IR transmission starts when the IREN bit is set to 1 when IRMODE = 1; when the IRMODE bit is set to 1 when IREN = 1; or when IREN and IRMODE are both set to 1 in the same instruction. The IRMT and IRCA registers, along with the IRDATA and IRTXPOL bits, are sampled at the beginning of the transmit process and every time the IR timer value reloads its value. When the IRV reaches 0000h value, on the next carrier clock, it does the following:

- 1) Reloads IRV with IRMT.
- 2) Samples IRCA, IRDATA, and IRTXPOL.
- 3) Generates IRTX accordingly.
- 4) Sets IRIF to 1.
- 5) Generates an interrupt to the CPU if enabled (IRIE = 1).

To terminate the current transmission, the user can switch to receive mode (IRMODE = 0) or clear IREN to 0.

Carrier Modulation Time = IRMT + 1 carrier cycles

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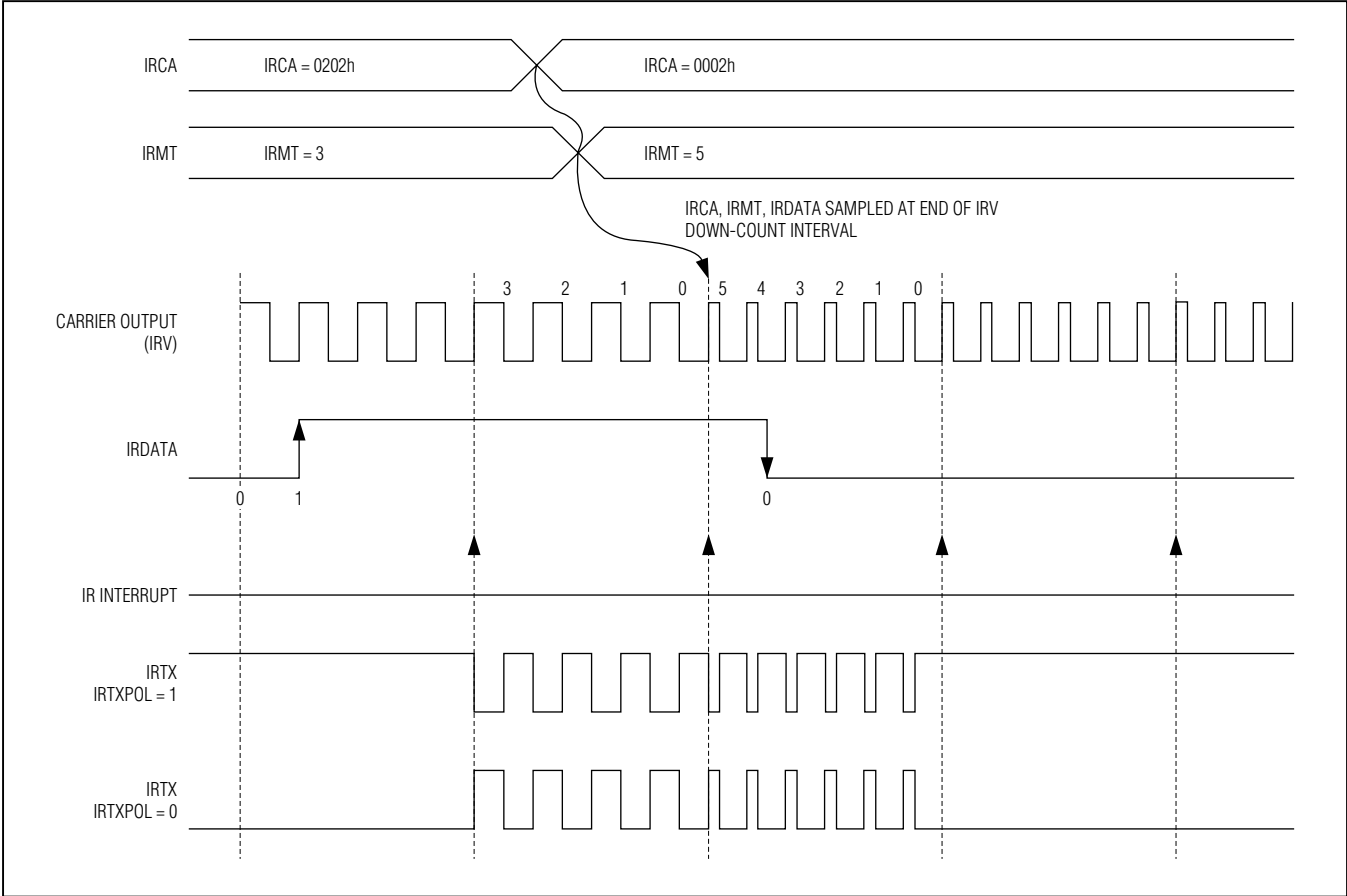


Figure 1. IR Transmit Frequency Shifting Example (IRCFME = 0)

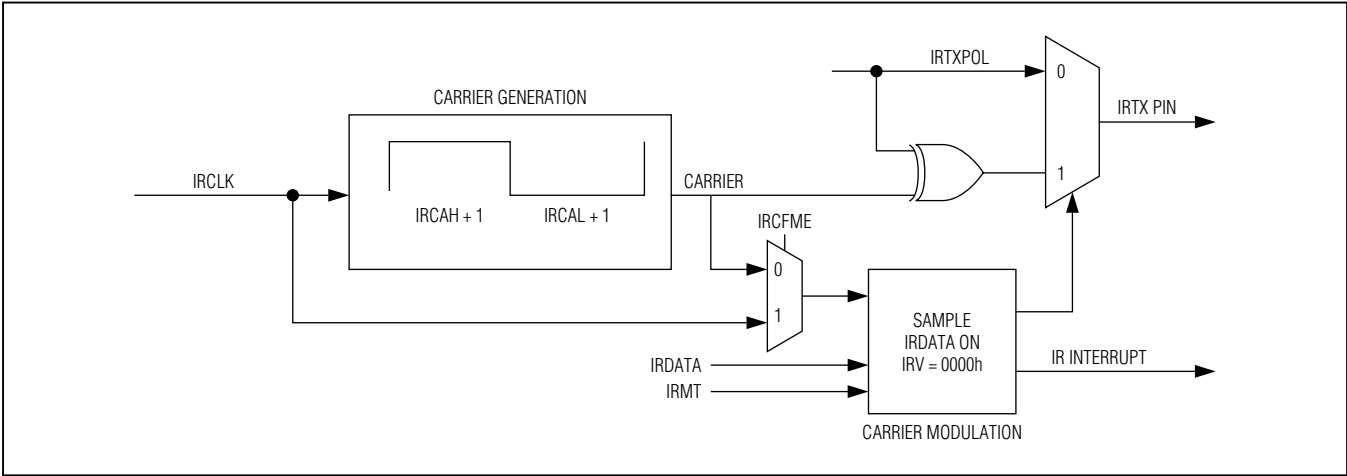


Figure 2. IR Transmit Carrier Generation and Carrier Modulator Control

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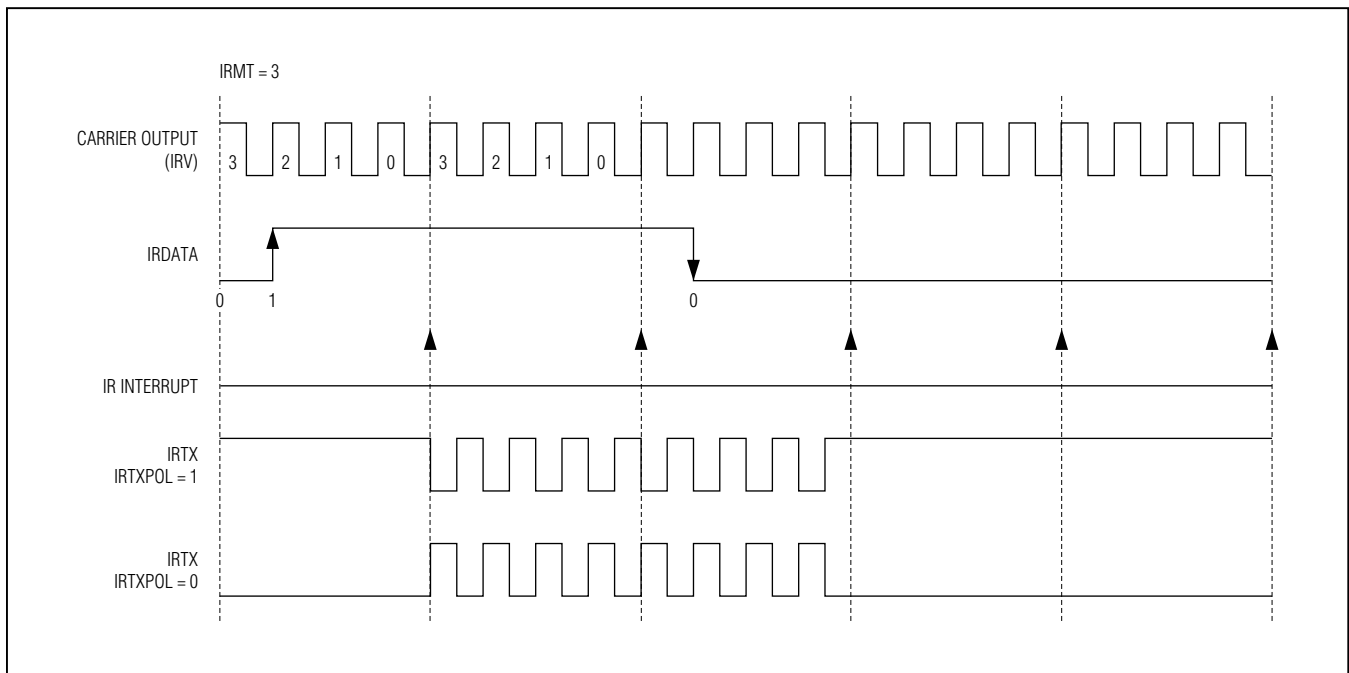


Figure 3. IR Transmission Waveform (IRCFME = 0)

IR Receive

When configured in receive mode (IRMODE = 0), the IR hardware supports the IRRX capture function. The IRRXSEL[1:0] bits define which edge(s) of the IRRX pin should trigger the IR timer capture function.

The IR module starts operating in the receive mode when IRMODE = 0 and IREN = 1. Once started, the IR timer (IRV) starts up counting from 0000h when a qualified capture event as defined by IRRXSEL happens. The IRV register is, by default, counting carrier cycles as defined by the IRCA register. However, the IR carrier frequency detect (IRCFME) bit can be set to 1 to allow clocking of the IRV register directly with the IRCLK for finer resolution. When IRCFME = 0, the IRCA defined carrier is counted by IRV. When IRCFME = 1, the IRCLK clocks the IRV register (Figure 4).

On the next qualified event, the IR module does the following:

- 1) Captures the IRRX pin state and transfers its value to IRDATA. If a falling edge occurs, IRDATA = 0. If a rising edge occurs, IRDATA = 1.
- 2) Transfers its current IRV value to the IRMT.
- 3) Resets IRV content to 0000h (if IRXRL = 1).
- 4) Continues counting again until the next qualified event.

If the IR timer value rolls over from 0FFFFh to 0000h before a qualified event happens, the IR timer overflow (IROV) flag is set to 1 and an interrupt is generated, if enabled. The IR module continues to operate in receive mode until it is stopped by switching into transmit mode (IRMODE = 1) or clearing IREN = 0.

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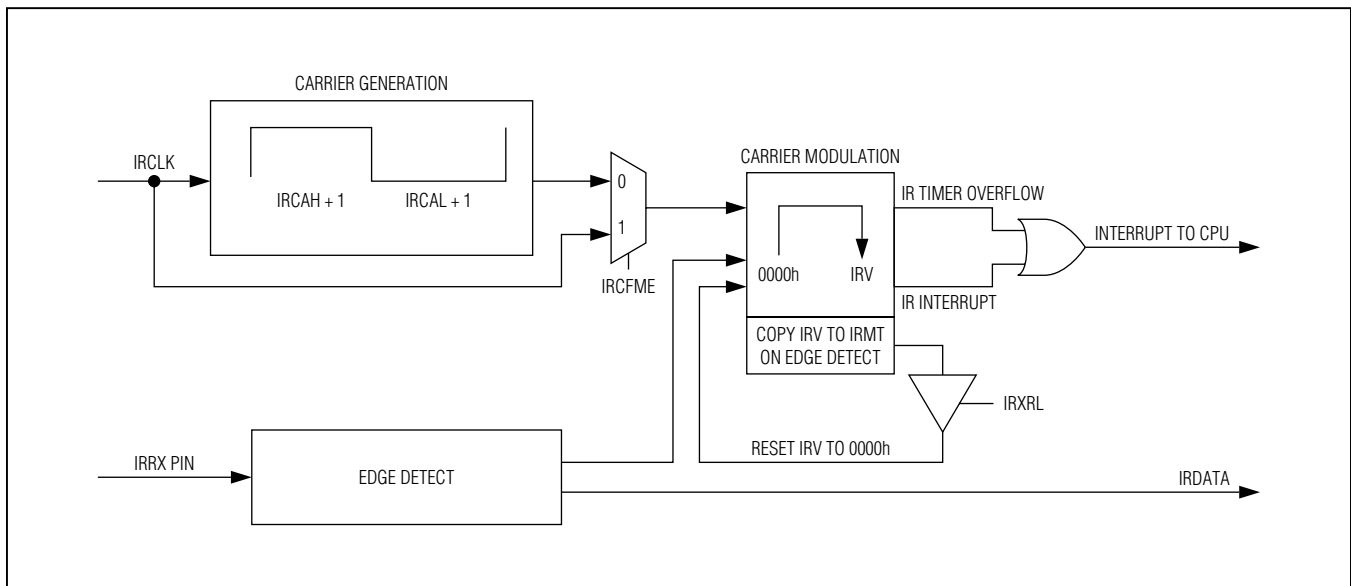


Figure 4. IR Capture

Carrier Burst-Count Mode

A special mode reduces the CPU processing burden when performing IR learning functions. Typically, when operating in an IR learning capacity, some number of carrier cycles are examined for frequency determination. Once the frequency has been determined, the IR receive function can be reduced to counting the number of carrier pulses in the burst and the duration of the combined mark-space time within the burst. To simplify this process, the receive burst-count mode (as enabled by the **RXBCNT** bit) can be used. When **RXBCNT** = 0, the standard IR receive capture functionality is in place. When **RXBCNT** = 1, the **IRV** capture operation is disabled and the interrupt flag associated with the capture no longer denotes a capture. In the carrier burst-count mode, the **IRMT** register only counts qualified edges. The **IRIF** interrupt flag (normally used to signal a capture when **RXBCNT** = 0) now becomes set if two **IRCA** cycles elapse without getting a qualified edge. The **IRIF** interrupt flag thus denotes absence of the carrier and the beginning of a space in the receive signal. When the **RXBCNT** bit is changed from 0 to 1, the **IRMT** register is set to 0001h. The **IRCFME** bit is still used to define

whether the **IRV** register is counting system **IRCLK** clocks or **IRCA**-defined carrier cycles. The **IRXRL** bit defines whether the **IRV** register is reloaded with 0000h on detection of a qualified edge (per the **IRRXSEL[1:0]** bits). Figure 5 and the descriptive sequence embedded in the figure illustrate the expected usage of the receive burst-count mode.

16-Bit Timers/Counters

The microcontroller provides two timers/counters that support the following functions:

- 16-bit timer/counter
- 16-bit up/down autoreload
- Counter function of external pulse
- 16-bit timer with capture
- 16-bit timer with compare
- Input/output enhancements for pulse-width modulation
- Set/reset/toggle output state on comparator match
- Prescaler with 2^n divider (for $n = 0, 2, 4, 6, 8, 10$)

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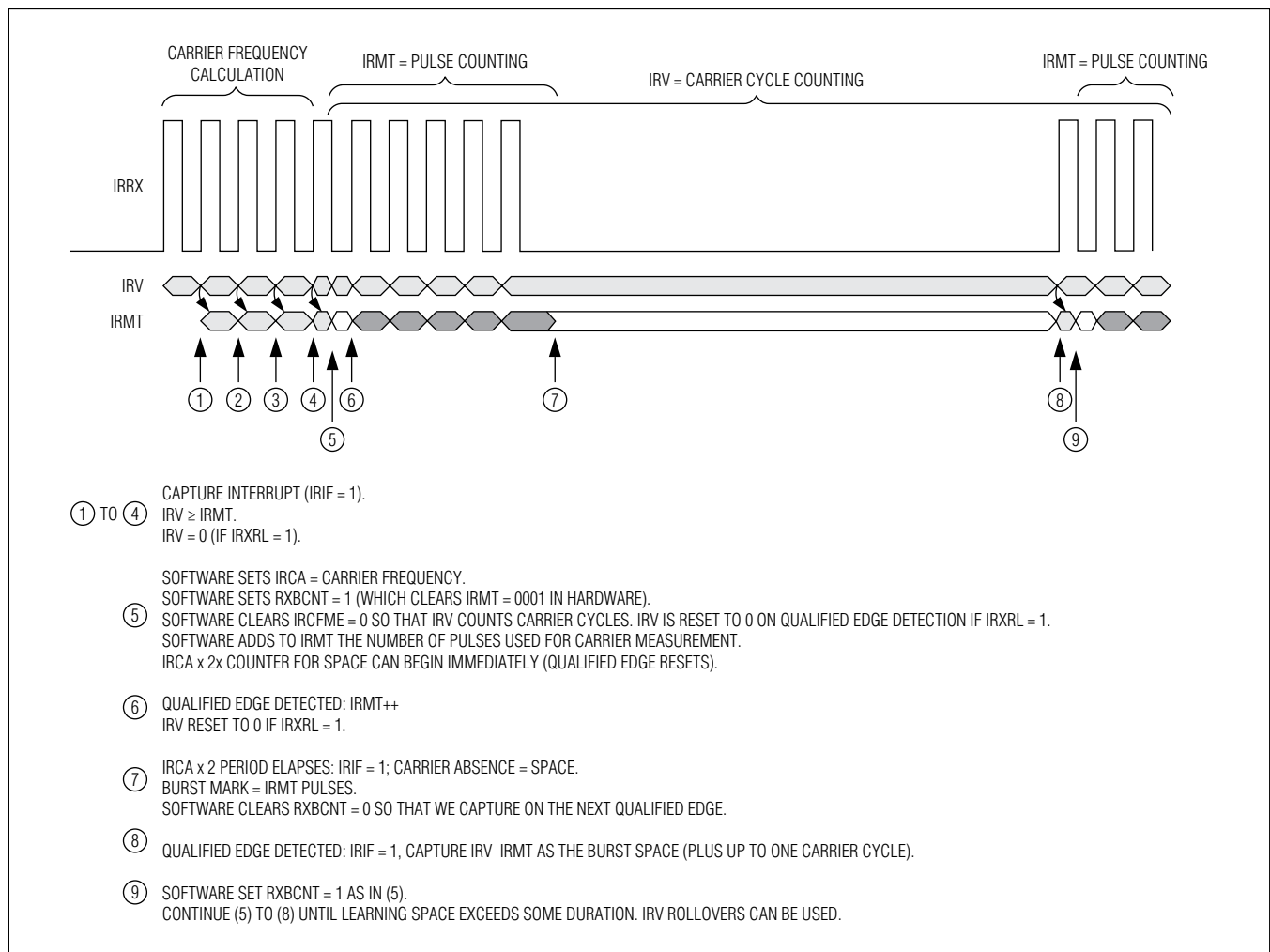


Figure 5. Receive Burst-Count Example

Serial Peripherals

Serial Peripheral Interface (SPI)

The device provides two SPI ports. The SPI is an inter-device bus protocol that provides fast, synchronous, full-duplex communications between devices. The integrated SPI interface acts as either an SPI master or slave device. The master drives the synchronous clock and selects which of several slaves is being addressed. Every SPI peripheral consists of a single shift register and control circuitry so that an addressed serial peripheral interface SPI peripheral is simultaneously transmitting and receiving. The maximum SPI master transfer rate is Sysclk/2.

When operating as an SPI slave, the device can support up to Sysclk/4 SPI transfer rate. Data can be transferred as an 8-bit or 16-bit value, MSB first. In addition, the SPI module supports configuration of the active SSEL state through the slave active-select pin.

Four signals are used in SPI communication:

- **SCLK:** The synchronous clock used by all devices. The master drives this clock and the slaves receive the clock. Note that SCLK can be gated and need not be driven between SPI transactions.
- **MOSI:** Master out-slave in. This is the main data line driven by the master to all slaves on the SPI bus. Only the selected slave clocks data from MOSI.

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- **MISO:** Master in-slave out. This is the main data line driven by the selected slave to the master. Only the selected slave may drive this circuit. In fact, it is the only circuit in the SPI bus arrangement that a slave is ever permitted to drive.
- **SSEL:** This signal is unique to each slave. When active (generally low), the selected slave must drive MISO.

I²C Bus

The microcontroller provides two internal I²C bus master/slave peripherals for communication with a wide variety of other I²C-enabled devices. The I²C bus is a 2-wire, bidirectional bus using two bus lines—the serial data line (SDA) and the serial clock line (SCL)—and a ground line. Both the SDA and SCL lines must be driven as open-collector/drain outputs. External resistors are required to pull the lines to a logic-high state.

The device supports both the master and slave protocols. In the master mode, the device has ownership of the I²C bus, drives the clock, and generates the START and STOP signals. This allows it to send data to a slave or receive data from a slave as required. In slave mode, the device relies on an externally generated clock to drive SCL and responds to data and commands only when requested by the I²C master device.

USART

The device provides two USART peripherals with operation modes described in [Table 2](#). The USART provides the following features:

- 2-wire interface
- Full-duplex operation for asynchronous data transfers
- Half-duplex operation for synchronous data transfers
- Programmable interrupt when transmit or receive data operation completes
- Independent programmable baud-rate generator
- Optional 9th bit parity support
- Start/stop bit support

Table 2. USART Mode Details

| MODE | TYPE | START BITS | DATA BITS | STOP BITS |
|--------|--------------|------------|-----------|-----------|
| Mode 0 | Synchronous | N/A | 8 | N/A |
| Mode 1 | Asynchronous | 1 | 8 | 1 |
| Mode 2 | Asynchronous | 1 | 8 + 1 | 1 |
| Mode 3 | Asynchronous | 1 | 8 + 1 | 1 |

General-Purpose I/O

The microcontroller provides port pins for general-purpose I/O that have the following features:

- CMOS output drivers
- Schmitt trigger inputs
- Optional weak pullup to V_{DD} when operating in input mode

While the microcontroller is in a reset state, all port pins become high impedance with both weak pullups and input buffers disabled, unless otherwise noted.

From a software perspective, each port appears as a group of peripheral registers with unique addresses. Special function pins can also be used as general-purpose I/O pins when the special functions are disabled. For a detailed description of the special functions available for each pin, refer to the *User's Guide*.

On-Chip Oscillator

The device provides an internal 12MHz oscillator that requires no external components, thereby reducing system cost, PCB area, and radiated EMI.

Operating Modes

The lowest power mode of operation is stop mode. In this mode, CPU state and memories are preserved, but the CPU is not actively running. Wake-up sources include external I/O interrupts, the power-fail warning interrupt, wake-up timer, or a power-fail reset. Any time the microcontroller is in a state where code does not need to be executed, the user software can put the device into stop mode. The nanopower ring oscillator is an internal ultra-low-power (400nA) 8kHz ring oscillator that can be used to drive a wake-up timer that exits stop mode. The wake-up timer is programmable by software in steps of 125μs up to approximately 8s.

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The power-fail monitor is always on during normal operation. However, it can be selectively disabled during stop mode to minimize power consumption. This feature is enabled using the power-fail monitor disable (PFD) bit in the PWCN register. The reset default state for the PFD bit is 1, which disables the power-fail monitor function during stop mode. If power-fail monitoring is disabled (PFD = 1) during stop mode, the circuitry responsible for generating a power-fail warning or reset is shut down and neither condition is detected. Thus, the $V_{DD} < V_{RST}$ condition does not invoke a reset state.

Power-Fail Detection

Figure 6, Figure 7, Figure 8, Table 4, Table 5, and Table 6 show the power-fail detection and response during normal and stop-mode operation. If a reset is caused by a power-fail, the power-fail monitor can be set to one of the following intervals:

- Always on—continuous monitoring
- 2^{11} nanopower ring oscillator clocks (~256ms)
- 2^{12} nanopower ring oscillator clocks (~512ms)
- 2^{13} nanopower ring oscillator clocks (~1.024s)

In the case where the power-fail circuitry is periodically turned on, the power-fail detection is turned on for two

nanopower ring-oscillator cycles. If $V_{DD} > V_{RST}$ during detection, V_{DD} is monitored for an additional nanopower ring-oscillator period. If V_{DD} remains above V_{RST} for the third nanopower ring period, the CPU exits the reset state and resumes normal operation from utility ROM at 8000h after satisfying the crystal warmup period.

The voltage (V_{PFW}) below which a power-fail warning is generated is user-configurable through the power-fail warning trip point control (PFWARNCN) bits. Table 3 shows the supported V_{PFW} values.

If a reset is generated by any other event, such as the RESET pin being driven low externally or the watchdog timer, the power-fail, internal regulator, and crystal remain on during the CPU reset. In these cases, the CPU exits the reset state in less than 20 crystal cycles after the reset source is removed.

Table 3. Power-Fail Warning Voltage Selection

| PFWARNCN | NOMINAL VOLTAGE |
|----------|-----------------|
| 00 | 1.67 |
| 01 | 1.9 |
| 10 | 2.5 |
| 11 | 2.7 |

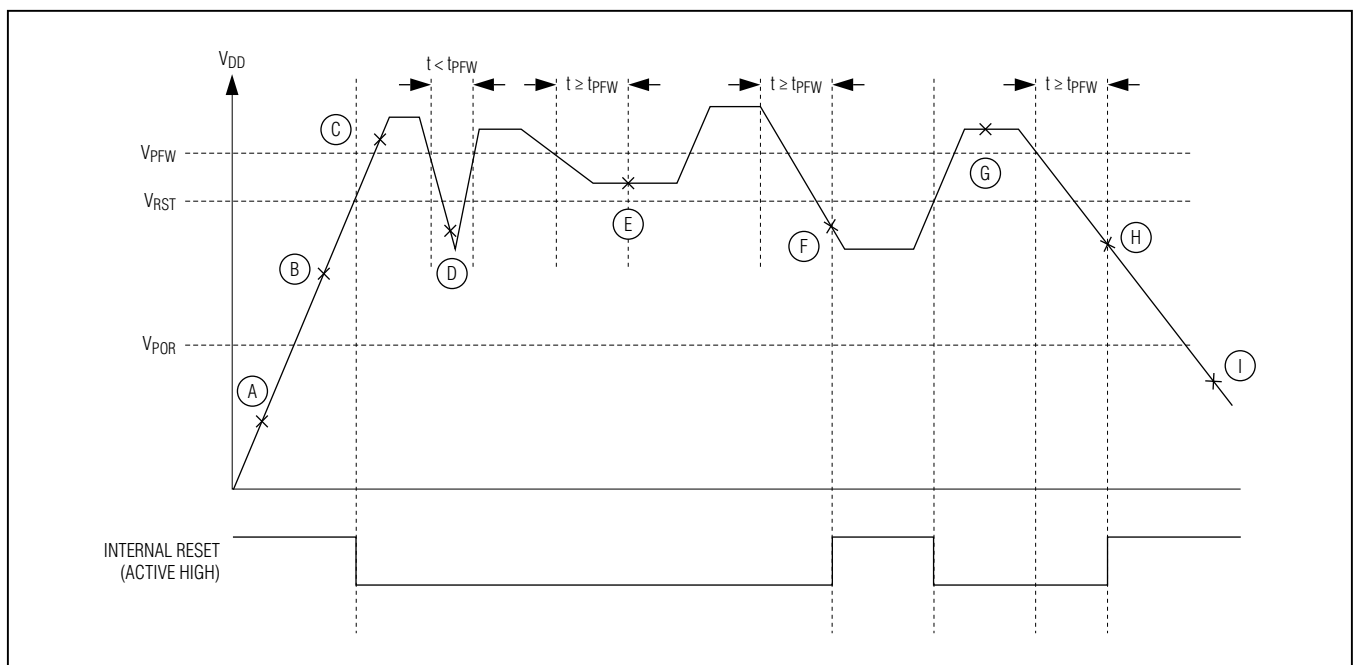


Figure 6. Power-Fail Detection During Normal Operation

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Table 4. Power-Fail Detection States During Normal Operation

| STATE | POWER-FAIL | INTERNAL REGULATOR | CRYSTAL OSCILLATOR | SRAM RETENTION | COMMENTS |
|-------|----------------------|--------------------|--------------------|----------------|---|
| A | On | Off | Off | — | $V_{DD} < V_{POR}$. |
| B | On | On | On | — | $V_{POR} < V_{DD} < V_{RST}$. Crystal warmup time, t_{XTAL_RDY} . CPU held in reset. |
| C | On | On | On | — | $V_{DD} > V_{RST}$. CPU normal operation. |
| D | On | On | On | — | Power drop too short. Power-fail not detected. |
| E | On | On | On | — | $V_{RST} < V_{DD} < V_{PFW}$. PFI is set when $V_{RST} < V_{DD} < V_{PFW}$ and maintains this state for at least t_{PFW} , at which time a power-fail interrupt is generated (if enabled). CPU continues normal operation. |
| F | On (Periodically) | Off | Off | Yes | $V_{POR} < V_{DD} < V_{RST}$. Power-fail detected. CPU goes into reset. Power-fail monitor turns on periodically. |
| G | On | On | On | — | $V_{DD} > V_{RST}$. Crystal warmup time, t_{XTAL_RDY} . CPU resumes normal operation from 8000h. |
| H | On (Periodically) | Off | Off | Yes | $V_{POR} < V_{DD} < V_{RST}$. Power-fail detected. CPU goes into reset. Power-fail monitor turns on periodically. |
| I | Off | Off | Off | — | $V_{DD} < V_{POR}$. Device held in reset. No operation allowed. |

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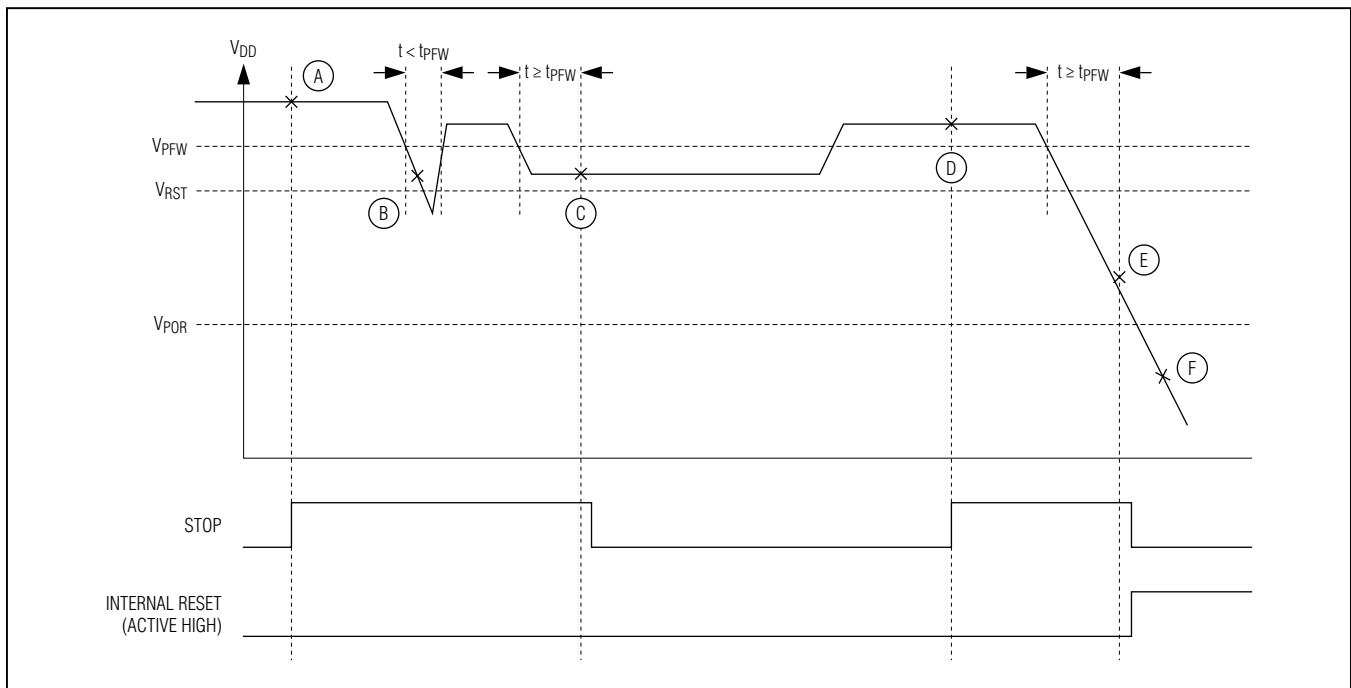


Figure 7. Stop Mode Power-Fail Detection States with Power-Fail Monitor Enabled

Table 5. Stop Mode Power-Fail Detection States with Power-Fail Monitor Enabled

| STATE | POWER-FAIL | INTERNAL REGULATOR | CRYSTAL OSCILLATOR | SRAM RETENTION | COMMENTS |
|-------|----------------------|--------------------|--------------------|----------------|---|
| A | On | Off | Off | Yes | Application enters stop mode. $V_{DD} > V_{RST}$. CPU in stop mode. |
| B | On | Off | Off | Yes | Power drop too short. Power-fail not detected. |
| C | On | On | On | Yes | $V_{RST} < V_{DD} < V_{PFW}$. Power-fail warning detected. Turn on regulator and crystal. Crystal warmup time, t_{XTAL_RDY} . Exit stop mode. |
| D | On | Off | Off | Yes | Application enters stop mode. $V_{DD} > V_{RST}$. CPU in stop mode. |
| E | On (Periodically) | Off | Off | Yes | $V_{POR} < V_{DD} < V_{RST}$. Power-fail detected. CPU goes into reset. Power-fail monitor turns on periodically. |
| F | Off | Off | Off | — | $V_{DD} < V_{POR}$. Device held in reset. No operation allowed. |

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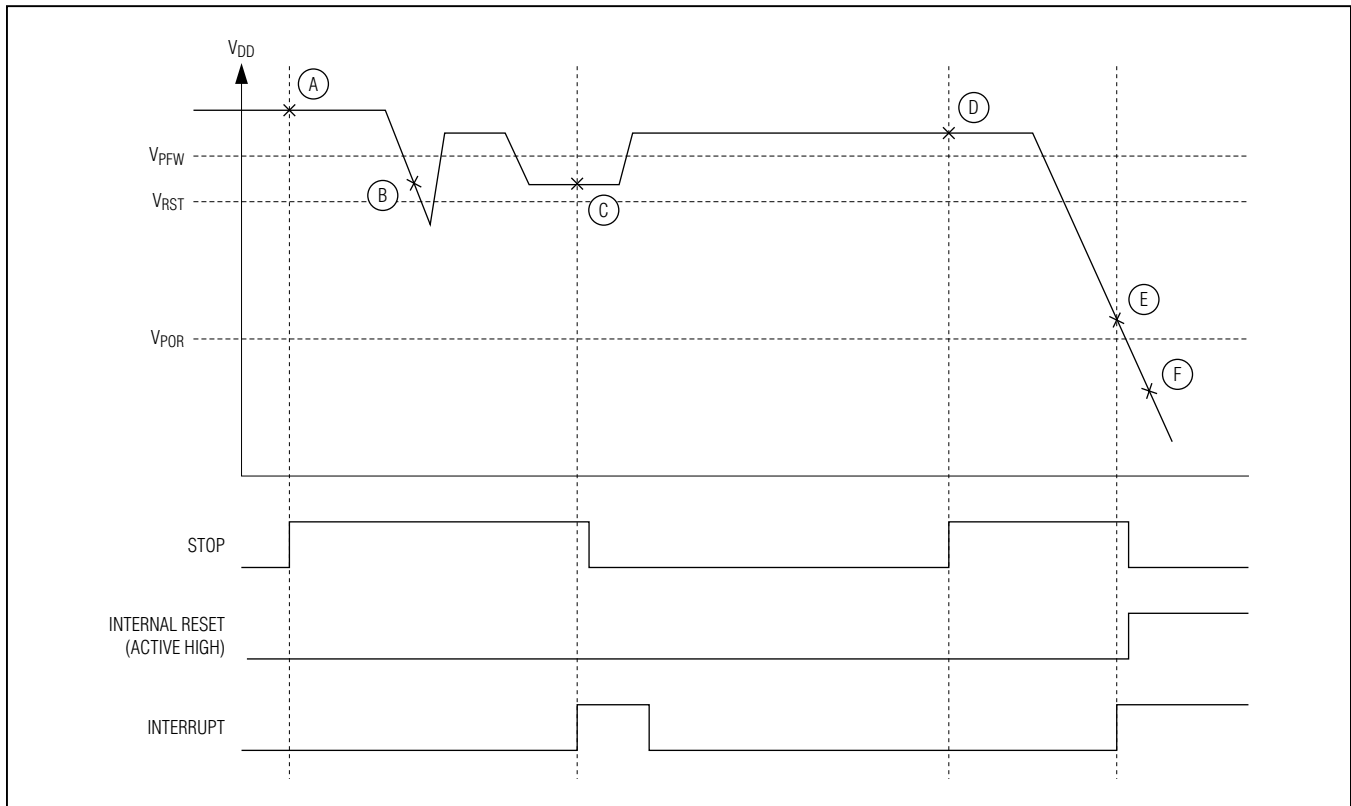


Figure 8. Stop Mode Power-Fail Detection with Power-Fail Monitor Disabled

Table 6. Stop Mode Power-Fail Detection States with Power-Fail Monitor Disabled

| STATE | POWER-FAIL | INTERNAL REGULATOR | CRYSTAL OSCILLATOR | SRAM RETENTION | COMMENTS |
|-------|------------|--------------------|--------------------|----------------|---|
| A | Off | Off | Off | Yes | Application enters stop mode. $V_{DD} > V_{RST}$. CPU in stop mode. |
| B | Off | Off | Off | Yes | $V_{DD} < V_{PFW}$. Power-fail not detected because power-fail monitor is disabled. |
| C | On | On | On | Yes | $V_{RST} < V_{DD} < V_{PFW}$. An interrupt occurs that causes the CPU to exit stop mode. Power-fail monitor is turned on, detects a power-fail warning, and sets the power-fail interrupt flag. Turn on regulator and crystal. Crystal warmup time, t_{XTAL_RDY} . On stop mode exit, CPU vectors to the higher priority of power-fail and the interrupt that causes stop mode exit. |

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Table 6. Stop Mode Power-Fail Detection States with Power-Fail Monitor Disabled (continued)

| STATE | POWER-FAIL | INTERNAL REGULATOR | CRYSTAL OSCILLATOR | SRAM RETENTION | COMMENTS |
|-------|----------------------|--------------------|--------------------|----------------|--|
| D | Off | Off | Off | Yes | Application enters stop mode. $V_{DD} > V_{RST}$. CPU in stop mode. |
| E | On (Periodically) | Off | Off | Yes | $V_{POR} < V_{DD} < V_{RST}$. An interrupt occurs that causes the CPU to exit stop mode. Power-fail monitor is turned on, detects a power-fail, and puts CPU in reset. Power-fail monitor is turned on periodically. |
| F | Off | Off | Off | — | $V_{DD} < V_{POR}$. Device held in reset. No operation allowed. |

Applications Information

The low-power, high-performance RISC architecture of this device makes it an excellent fit for many portable or battery-powered applications. It is ideally suited for applications such as universal remote controls that require the cost-effective integration of IR transmit/receive capability.

Grounds and Bypassing

Careful PCB layout significantly minimizes system-level digital noise that could interact with the microcontroller or peripheral components. The use of multilayer boards is essential to allow the use of dedicated power planes. The area under any digital components should be a continuous ground plane if possible. Keep bypass capacitor leads short for best noise rejection and place the capacitors as close to the leads of the devices as possible.

CMOS design guidelines for any semiconductor require that no pin be taken above V_{DD} or below GND. Violation of this guideline can result in a hard failure (damage to the silicon inside the device) or a soft failure (unintentional modification of memory contents). Voltage spikes above or below the device's absolute maximum ratings can potentially cause a devastating IC latchup.

Microcontrollers commonly experience negative voltage spikes through either their power pins or general-purpose I/O pins. Negative voltage spikes on power pins are especially problematic as they directly couple to the internal power buses. Devices such as keypads can conduct electrostatic discharges directly into the microcontroller and seriously damage the device. System designers must protect components against these transients that can corrupt system memory.

MAXQ617

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Additional Documentation

Engineers must have the following documents to fully use this device:

- This data sheet, containing pin descriptions, feature overviews, and electrical specifications.
- The device-appropriate user guide, containing detailed information and programming guidelines for core features and peripherals.
- Errata sheets for specific revisions noting deviations from published specifications.

For information regarding these documents, visit Technical Support at support.maximintegrated.com/micro.

Development and Technical Support

Contact technical support for information about highly versatile, affordable development tools, available from Maxim Integrated and third-party vendors.

- Evaluation kits
- Compilers
- Integrated development environments (IDEs)
- USB interface modules for programming and debugging

For technical support, go to support.maximintegrated.com/micro.

Ordering Information/Selector Guide

| PART | TEMP RANGE | OPERATING VOLTAGE (V) | PROGRAM MEMORY (KB) | DATA MEMORY (KB) | GPIO | PIN-PACKAGE |
|-----------------|----------------|-----------------------|---------------------|------------------|------|-------------|
| MAXQ617V-XXXX+T | -20°C to +70°C | 1.67 to 3.6 | 80 Flash | 4 | 10 | 16 WLP |

Note: The 4-digit suffix “-XXXX” indicates a device preprogrammed at Maxim Integrated with proprietary customer-supplied software. For more information on factory preprogramming of this device, contact Maxim Integrated at support.maximintegrated.com/micro.

+Denotes a lead(Pb)-free/RoHS-compliant package.

T = Tape and reel.

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a “+”, “#”, or “-” in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | OUTLINE NO. | LAND PATTERN NO. |
|--------------|--------------|-------------------------|--|
| 16 WLP | W162K2+1 | 21-0491 | Refer to Application Note 1891 |

Infrared Remote Control System-on-Chip

Appendix A

I²C SERIAL PERIPHERAL SPECIFICATIONS

(Figure 9 and Figure 10)

| PARAMETER | SYMBOL | CONDITIONS | STANDARD MODE | | FAST MODE | | UNITS |
|--|-----------------|--|---------------------|---------------------|----------------------|---------------------|-------|
| | | | MIN | MAX | MIN | MAX | |
| Input Low Voltage | V_{IL_I2C} | Supply voltages that mismatch I ² C bus levels must relate input levels to the R_P pullup voltage | -0.5 | $0.3 \times V_{DD}$ | -0.5 | $0.3 \times V_{DD}$ | V |
| Input High Voltage | V_{IH_I2C} | Supply voltages that mismatch I ² C bus levels must relate input levels to the R_P pullup voltage | $0.7 \times V_{DD}$ | | $0.7 \times V_{DD}$ | $V_{DD} + 0.5V$ | V |
| Input Hysteresis (Schmitt) | V_{IHYS_I2C} | $V_{DD} > 2V$ | | | $0.05 \times V_{DD}$ | | V |
| Output Logic-Low (Open Drain or Open Collector) | V_{OL_I2C} | $V_{DD} > 2V$, 3mA sink current | 0 | 0.4 | 0 | 0.4 | V |
| Capacitive Load for Each Bus Line | C_B | | 400 | | 400 | | pF |
| Output Fall Time from V_{IH_MIN} to V_{IL_MAX} with Bus Capacitance from 10pF to 400pF | t_{OF_I2C} | t_{RF_I2C} exceeds t_{OF_I2C} , which permits RS to be connected as shown in figure | 250 | | $20 + 0.1C_B$ | 250 | ns |
| Pulse Width of Spike Filtering That Must Be Suppressed by Input Filter | t_{SP_I2C} | | | | 0 | 50 | ns |
| Input Current on I/O | I_{IN_I2C} | Input voltage from $0.1 \times V_{DD}$ to $0.9 \times V_{DD}$ | -10 | +10 | -10 | +10 | μA |
| I/O Capacitance | C_{IO_I2C} | | 10 | | 10 | | pF |
| I ² C Bus Operating Frequency | f_{I2C} | | 0 | 100 | 0 | 400 | kHz |
| System Frequency | f_{SYS} | | 0.90 | | 3.60 | | MHz |
| I ² C Bit Rate | f_{I2C} | | $f_{SYS}/8$ | | $f_{SYS}/8$ | | Hz |
| Hold Time After (Repeated) START | $t_{HD:STA}$ | | 4.0 | | 0.6 | | μs |
| Clock Low Period | t_{LOW_I2C} | | 4.7 | | 1.3 | | μs |
| Clock High Period | t_{HIGH_I2C} | | 4.0 | | 0.6 | | μs |
| Setup Time for Repeated START | $t_{SU:STA}$ | | 4.7 | | 0.6 | | μs |

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I²C SERIAL PERIPHERAL SPECIFICATIONS (continued)

(Figure 9 and Figure 10)

| PARAMETER | SYMBOL | CONDITIONS | STANDARD MODE | | FAST MODE | | UNITS |
|---|---------------|--|---------------------|------|---------------------|-----|---------|
| | | | MIN | MAX | MIN | MAX | |
| Hold Time for Data | $t_{HD:DAT}$ | A device must internally provide a hold time of at least 300ns for $V_{IH_I2C(MIN)}$ to bridge the undefined region of the falling edge of SCL. The maximum $t_{HD:DAT}$ needs to be met only if the device does not stretch the SCL low period | 0 | 3.45 | 0 | 0.9 | μ s |
| Setup Time for Data | $t_{SU:DAT}$ | A fast-mode I ² C bus device can be used in a standard-mode I ² C bus system; if such a device does not stretch the low period of the SCL signal, it must output the next data bit to the SDA line $t_{R_I2C(MAX)} + t_{SU:DAT} = 1000 + 250 = 1250$ ns (according to the standard-mode I ² C specification) before the SCL line is released | 250 | | 100 | | ns |
| SDA/SCL Fall Time | t_{F_I2C} | | | 300 | 20 + $0.1C_B$ | 300 | ns |
| SDA/SCL Rise Time | t_{R_I2C} | | | 1000 | 20 + $0.1C_B$ | 300 | ns |
| Setup Time for STOP | $t_{SU:STO}$ | | 4.0 | | 0.6 | | μ s |
| Bus Free Time Between STOP and START | t_{BUF} | | 4.7 | | 1.3 | | μ s |
| Noise Margin at the Low Level for Each Connected Device (Including Hysteresis) | V_{nL_I2C} | | $0.1 \times V_{DD}$ | | $0.1 \times V_{DD}$ | | V |
| Noise Margin at the High Level for Each Connected Device (Including Hysteresis) | V_{nH_I2C} | | $0.2 \times V_{DD}$ | | $0.2 \times V_{DD}$ | | V |

Infrared Remote Control System-on-Chip

I²C Serial Diagrams

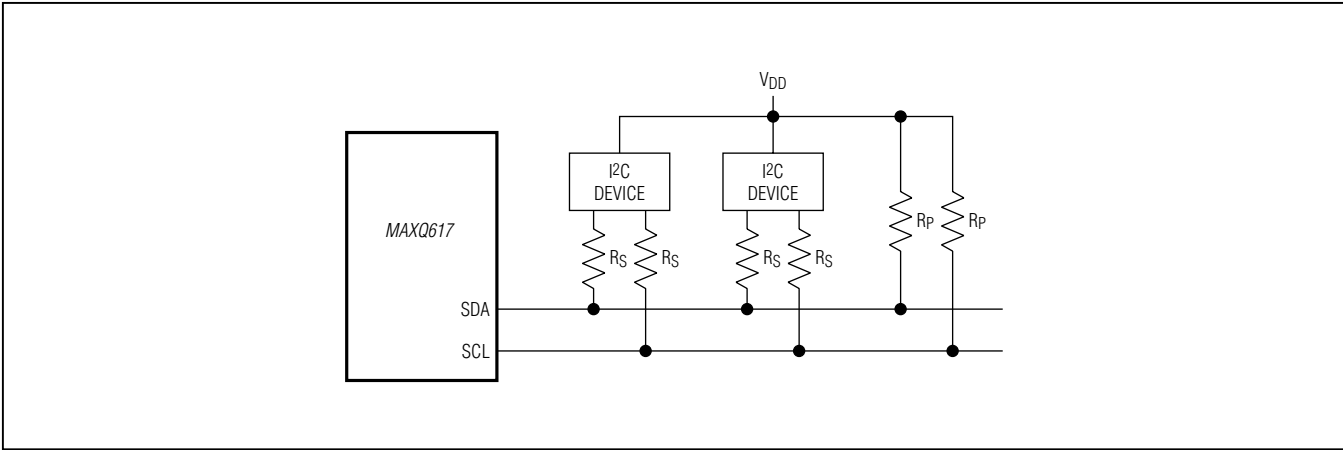


Figure 9. Series Resistors (RS) for Protecting Against High-Voltage Spikes

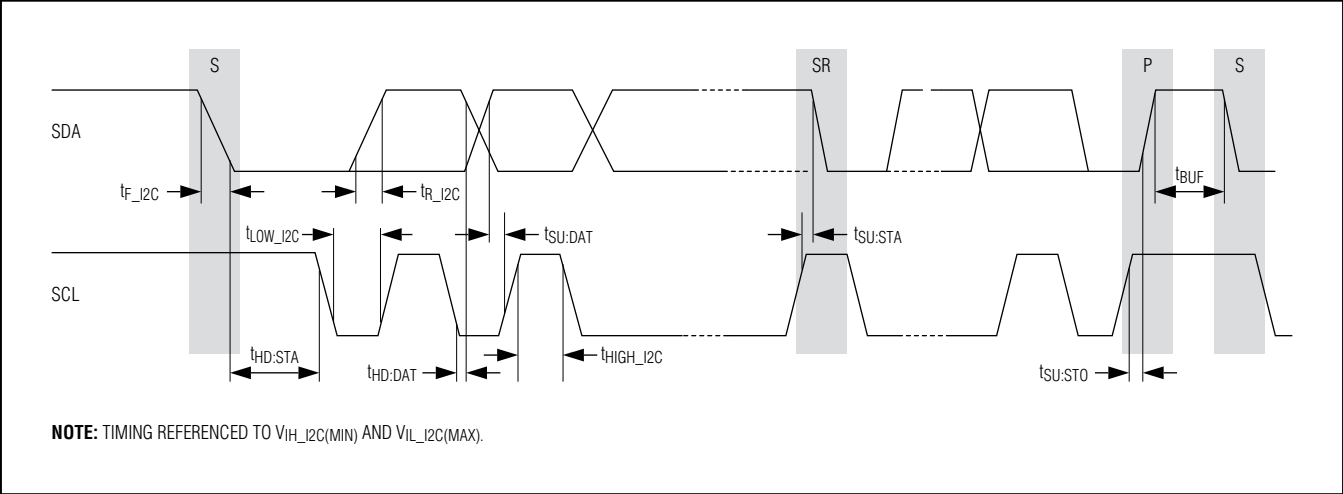


Figure 10. I²C Bus Controller Timing Diagram

Infrared Remote Control System-on-Chip**SERIAL PERIPHERAL INTERFACE (SPI) SPECIFICATIONS**

(Figure 11 and Figure 12)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|--|--------------------------|------------|---------------------|-------------|------------|-------|
| SPI Master Operating Frequency | $1/t_{MCK}$ | | | | $f_{CK}/2$ | MHz |
| SPI Slave Operating Frequency | $1/t_{SCK}$ | | | | $f_{CK}/4$ | MHz |
| SCLK Output Pulse-Width High/Low | t_{MCH} , t_{MCL} | | $t_{MCK}/2$ - 35 | | | ns |
| MOSI Output Hold Time After SCLK Sample Edge | t_{MOH} | | $t_{MCK}/2$ - 35 | | | ns |
| MOSI Output Valid to Sample Edge | t_{MOV} | | $t_{MCK}/2$ - 35 | | | ns |
| MISO Input Valid to SCLK Sample Edge Rise/Fall Setup | t_{MIS} | | 35 | | | ns |
| MISO Input to SCLK Sample Edge Rise/Fall Hold | t_{MIH} | | 0 | | | ns |
| SCLK Input Pulse-Width High/Low | t_{SCH} , t_{SCL} | | | $t_{SCK}/2$ | | ns |
| SSEL Active to First Shift Edge | t_{SSE} | | | 50 | | ns |
| MOSI Input to SCLK Sample Edge Rise/Fall Setup | t_{SIS} | | 35 | | | ns |
| MOSI Input from SCLK Sample Edge Transition Hold | t_{SIH} | | 35 | | | ns |
| MISO Output Valid After SCLK Shift Edge Transition | t_{SOV} | | | | 70 | ns |
| SCLK Inactive to SSEL Rising | t_{SD} | | 35 | | | ns |

Infrared Remote Control System-on-Chip

SPI Timing Diagrams

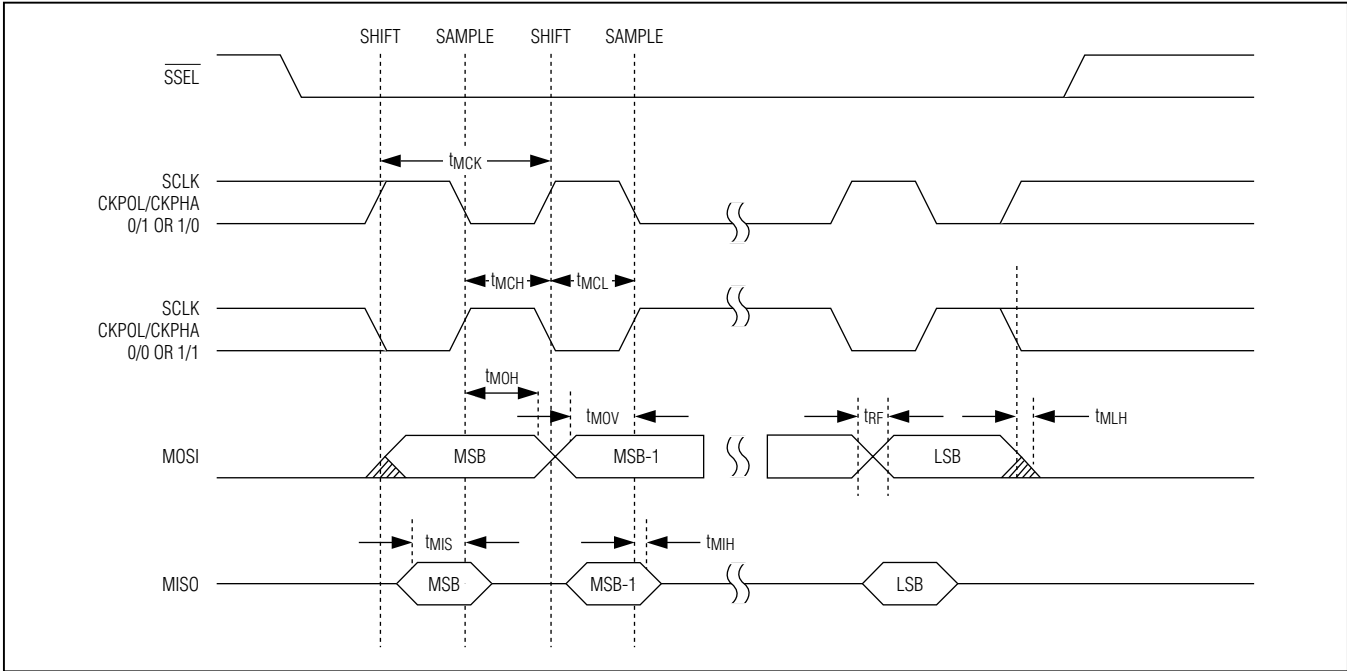


Figure 11. SPI Master Communication Timing

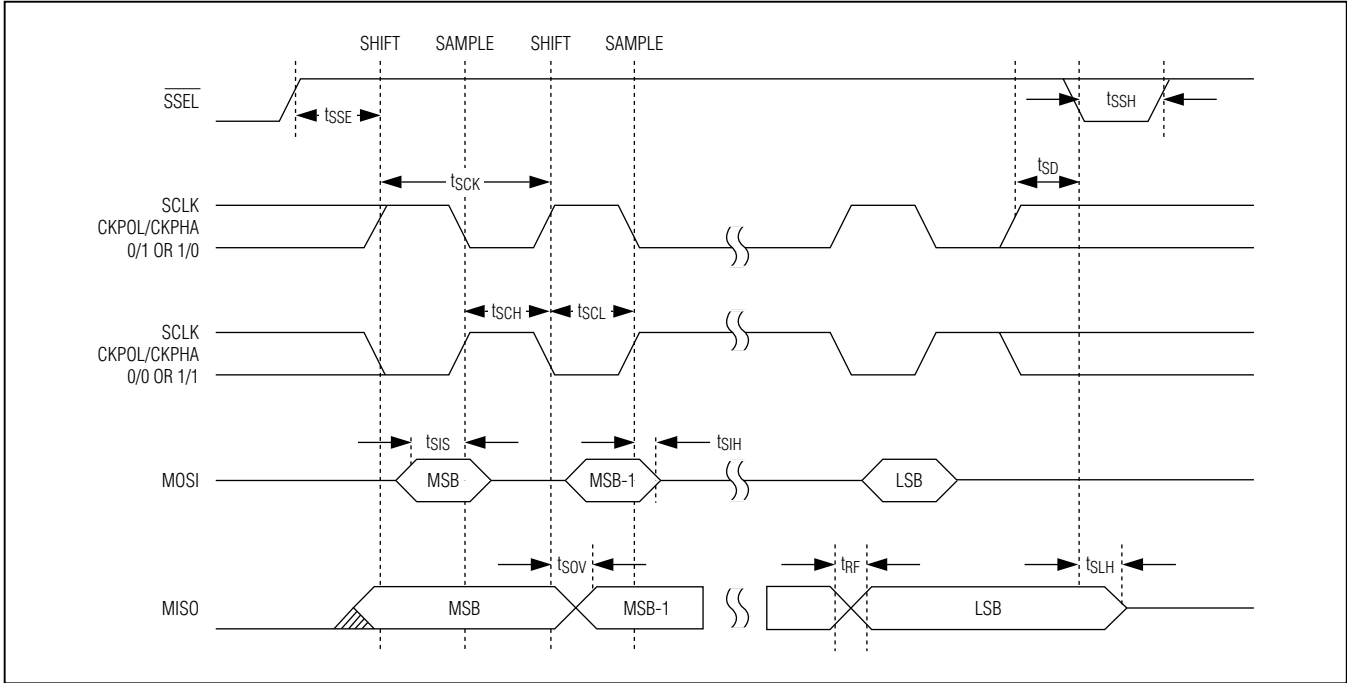


Figure 12. SPI Slave Communication Timing

Infrared Remote Control System-on-Chip

USART MODE 0 SPECIFICATIONS

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|------------|------------|-----|--------------|-----|-------|
| TXD Clock Period | t_{XLXL} | SM2 = 0 | | $12t_{CLCL}$ | | ns |
| | | SM2 = 1 | | $4t_{CLCL}$ | | ns |
| TXD Clock High Time | t_{XHXL} | SM2 = 0 | | $3t_{CLCL}$ | | ns |
| | | SM2 = 1 | | $2t_{CLCL}$ | | ns |
| RXD Output Data Valid to TXD Clock Rising Edge | t_{QVXH} | SM2 = 0 | | $10t_{CLCL}$ | | ns |
| | | SM2 = 1 | | $3t_{CLCL}$ | | ns |
| RXD Output Data Hold from TXD Clock Rising Edge | t_{XHQH} | SM2 = 0 | | $2t_{CLCL}$ | | ns |
| | | SM2 = 1 | | t_{CLCL} | | ns |
| RXD Input Data Valid to TXD Clock Rising Edge | t_{DVXH} | SM2 = 0 | | t_{CLCL} | | ns |
| | | SM2 = 1 | | t_{CLCL} | | ns |
| RXD Input Data Hold after TXD Clock Rising Edge | t_{XHDH} | SM2 = 0 | | t_{CLCL} | | ns |
| | | SM2 = 1 | | t_{CLCL} | | ns |

USART Timing

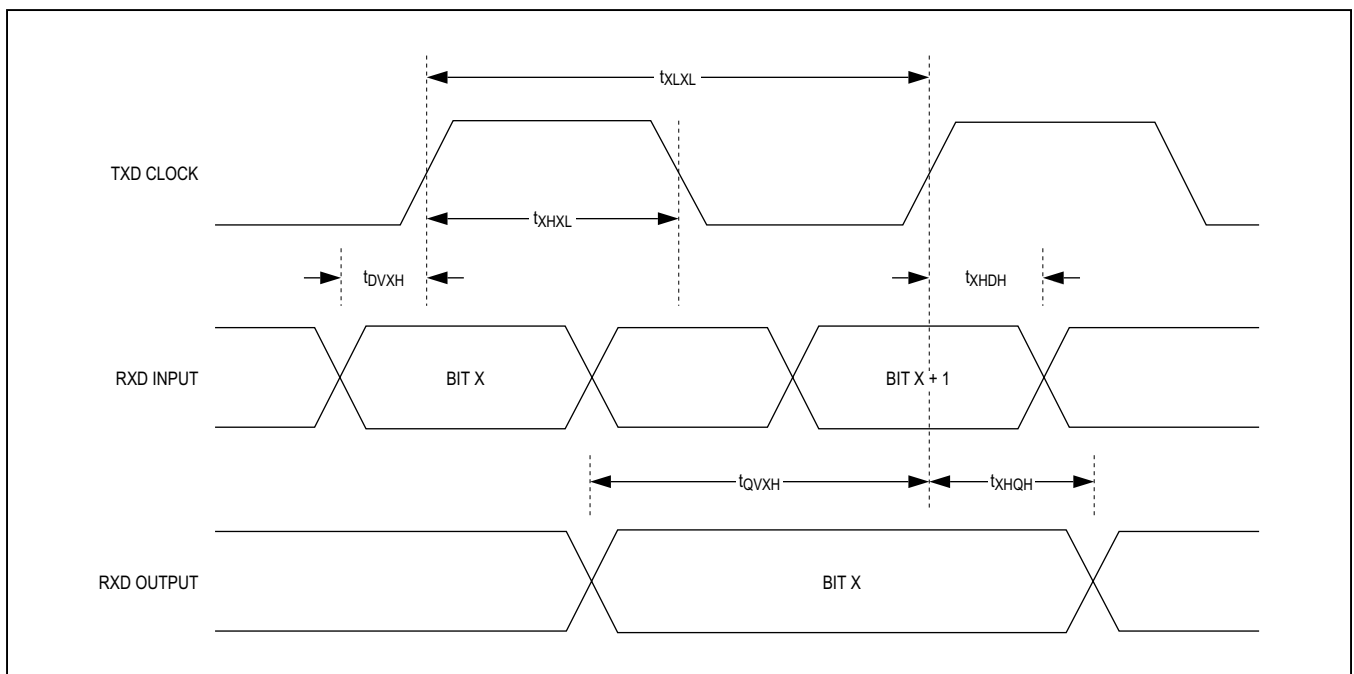


Figure 13. USART Timing Diagram

MAXQ617

Infrared Remote Control System-on-Chip

Revision History

| REVISION NUMBER | REVISION DATE | DESCRIPTION | PAGES CHANGED |
|-----------------|---------------|-----------------|---------------|
| 0 | 12/13 | Initial release | — |



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Мы предлагаем:

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- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
- Поставку компонентов, требующих военную и космическую приемку.
- Входной контроль качества.
- Наличие сертификата ISO.

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- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



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