

Power Supply IC Series for TFT-LCD Panels

Multi-channel

System Power Supply IC + Gamma Buffer

BD8162AEKV

General Description

The BD8162AEKV is a system power supply IC that provides control 6 power supply channels and 4 gamma output channels + VCOM required for TFT-LCD panels on a single chip. All channels have built-in control input and Power-Good output functions, enabling free sequence control setting just by changing channels. In addition, the BD8162AEKV is a user-friendly IC incorporating input switch, short-circuit protection, and protection detection output circuits.

Features

- Step-up DC/DC Converter with Built-in 3A FET
- Step-down DC/DC Converter with Built-in 2A FET
- Synchronous Rectification Step-down DC/DC Converter with Built-in 2A FET
- 3ch LDO Regulator (500mA, 200mA, 20mA)
- Positive/Negative Charge Pumps
- 4ch Gamma Buffer Amplifier + VCOM
- Protection Circuits:
 - Under-Voltage Lockout Protection Circuit
 - Thermal Shutdown Circuit
 - Timer Latch Type Short-Circuit Protection Circuit
- Controllable Startup Sequence

Key Specifications

- Power Supply Voltage 1 Range: 4.2V to 14V
- Oscillating Frequency: 200kHz to 800kHz(Variable)
- Operating Temperature Range: -40°C to +105°C

Package

W(Typ) x D(Typ) x H(Max)



HTQFP64V
12.00mm x 12.00mm x 1.00mm

Applications

LCD TV power supplies

Typical Application Circuit (1)

1. Application used to input Vcc12V:

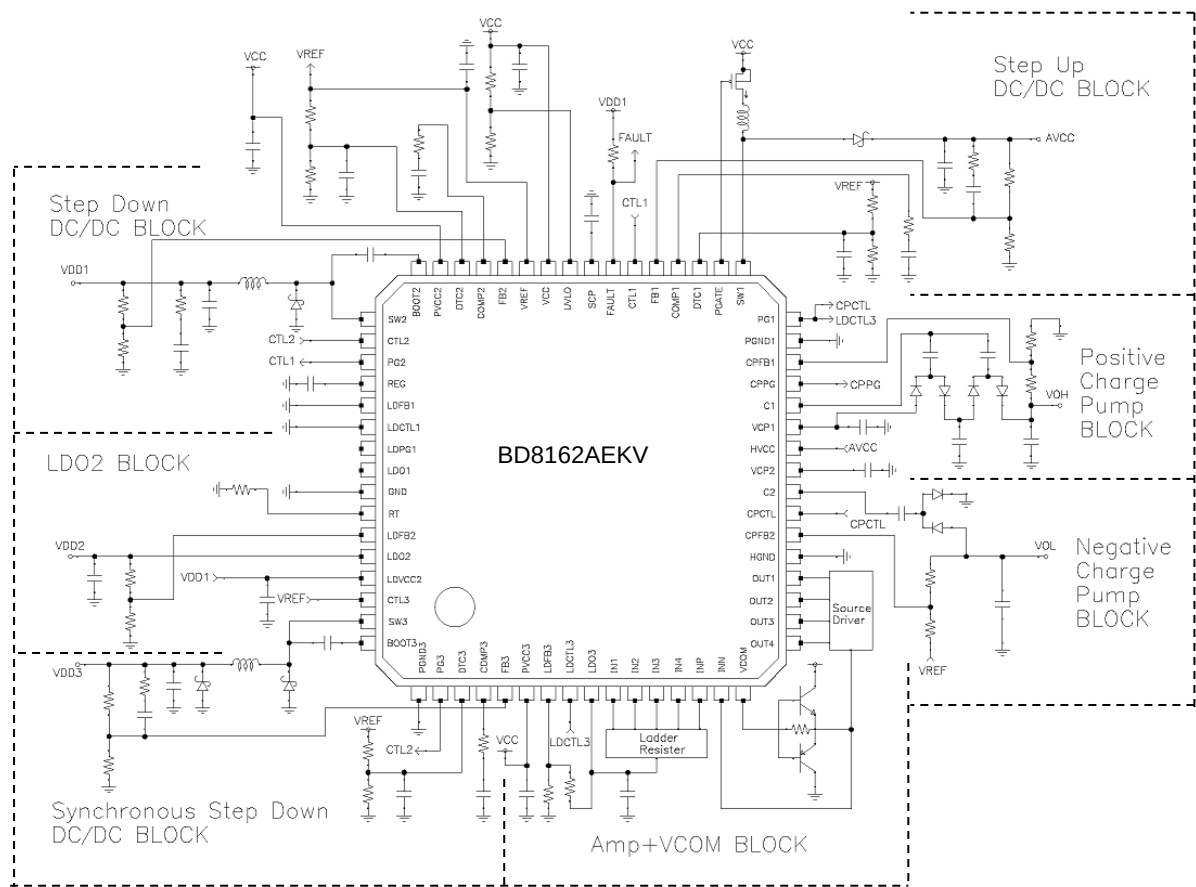
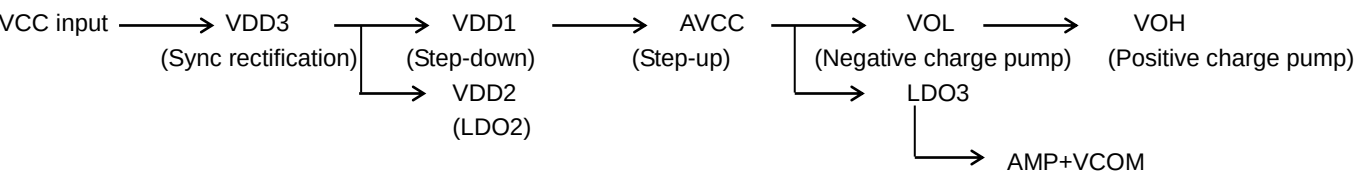


Figure 1. Typical 12V Input Application Diagram

2. Startup Sequence



3. Sequence Image Chart

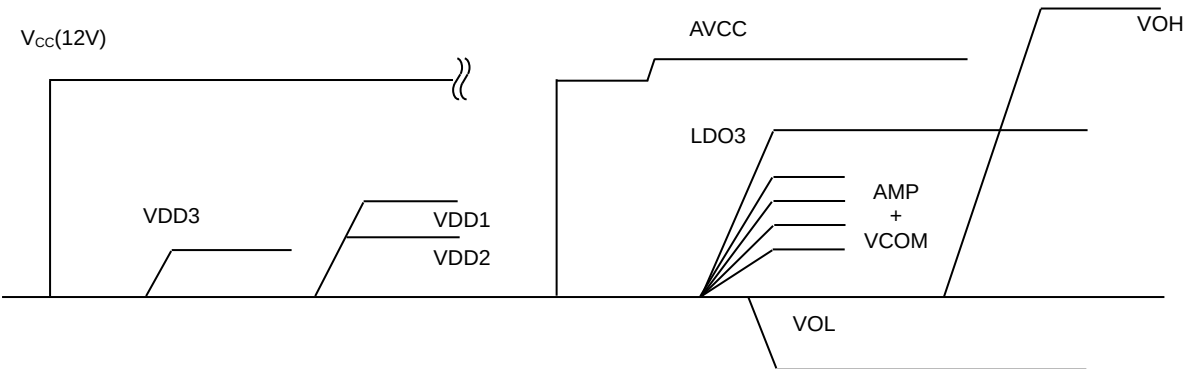


Figure 2. Sequence Chart

Typical Application Circuit (2)

1. Application used to input 5V:

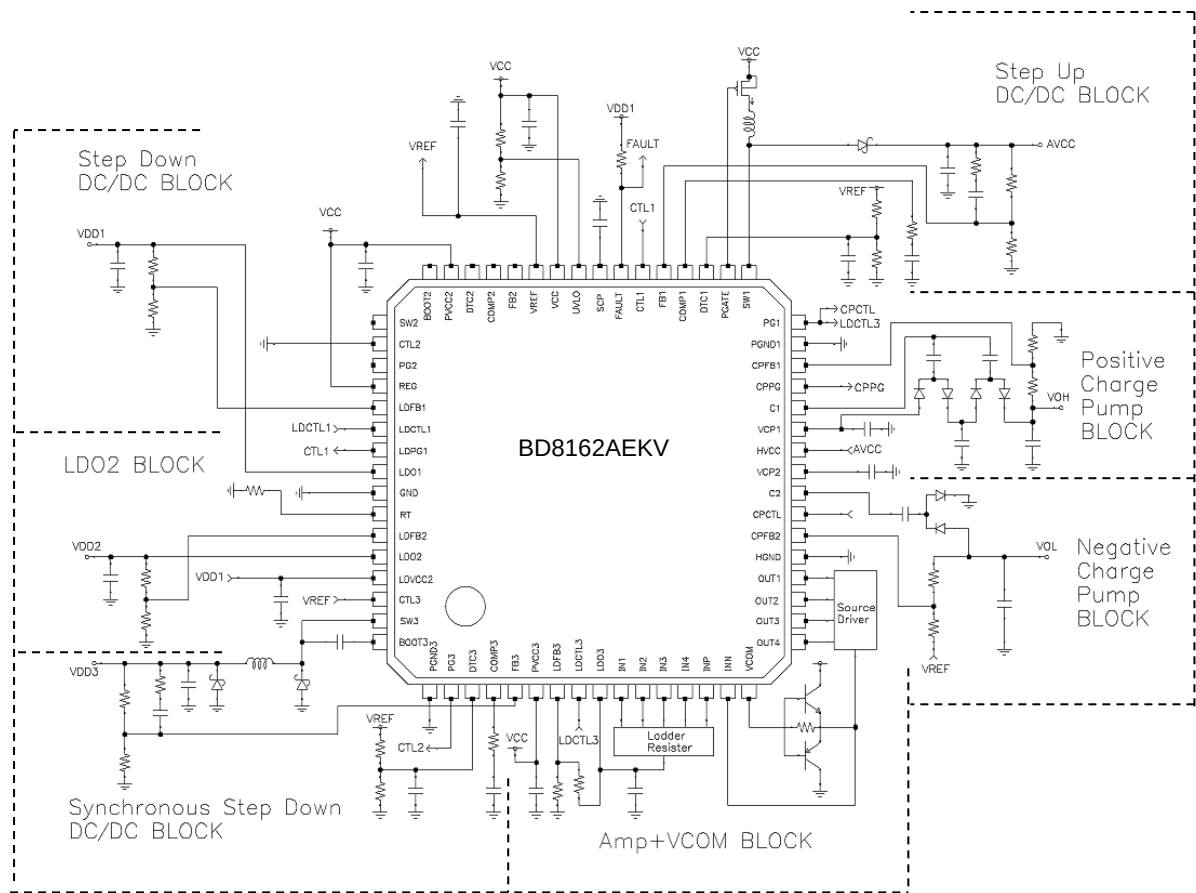
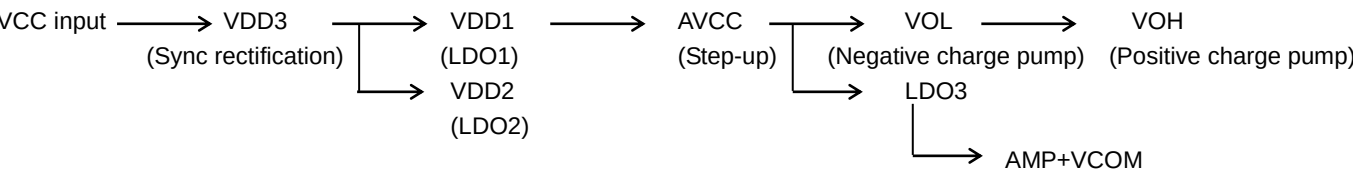


Figure 3. Typical 5V Input Application Circuit Diagram

2. Startup Sequence



3. Sequence Image Chart

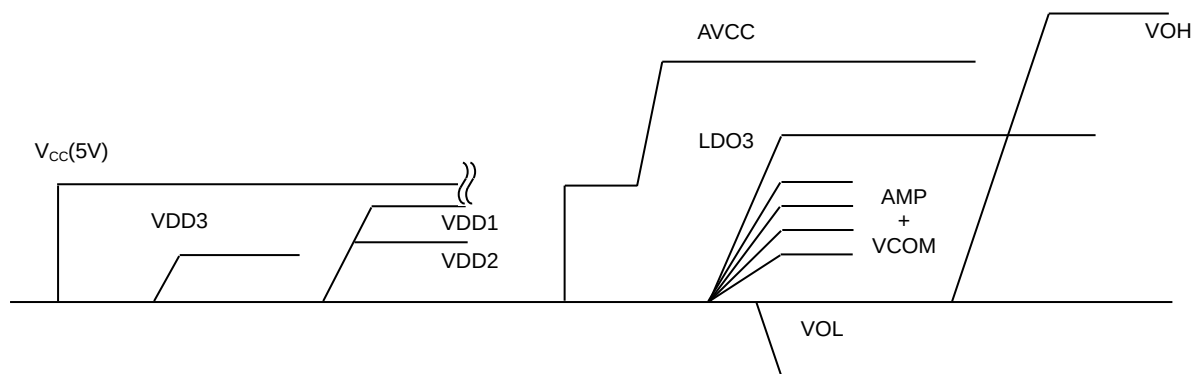
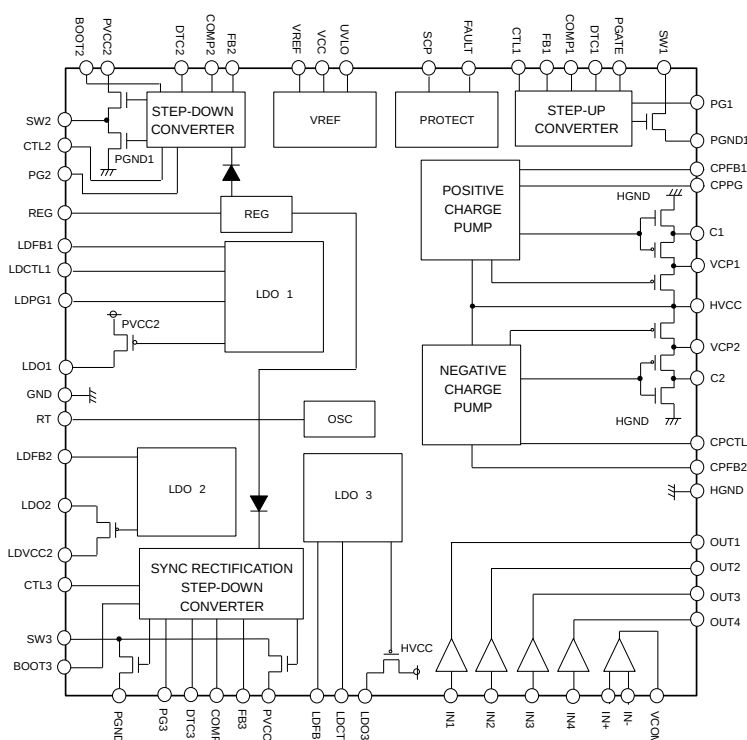


Figure 4. Sequence Chart

Block Diagram



Pin Description

PIN NO.	Pin name	Function	PIN NO.	Pin name	Function	PIN NO.	Pin name	Function
1	PGND3	Ground pin	23	CPCTL	CP control input pin	45	COMP2	Error amp output 2 pin
2	PG3	Power Good output 3 pin	24	C2	Charge pump output 2 pin	46	DTC2	Duty limit pin 2 pin
3	DTC3	Duty limit pin 3 pin	25	VCP2	Charge pump LDO output 2 pin	47	PVCC2	Power supply input pin
4	COMP3	Error amp output 3 pin	26	HVCC	Power supply input pin	48	BOOT2	Switch boot pin 2 pin
5	FB3	Feedback input 3 pin	27	VCP1	Charge pump LDO output 1 pin	49	SW2	Switching output 2 pin
6	PVCC3	Power supply input pin	28	C1	Charge pump output 1 pin	50	CTL2	Control input 2 pin
7	LDFB3	LDO feedback input 3 pin	29	CPPG	CP Power Good output pin	51	PG2	Power Good output 2 pin
8	LDCTL3	LDO3 control input pin	30	CPFB1	Charge pump feedback 1 pin	52	REG	Boot LDO output pin
9	LDO3	LDO output 3 pin	31	PGND1	Ground pin	53	LDFB1	LDO feedback 1 pin
10	IN1	AMP input 1 pin	32	PG1	Power Good output 1 pin	54	LDCTL1	LDO1 control input pin
11	IN2	AMP input 2 pin	33	SW1	Switching output 1 pin	55	LDPG1	LDO1 Power Good output pin
12	IN3	AMP input 3 pin	34	PGATE	Pch gate drive output pin	56	LDO1	LDO output 1 pin
13	IN4	AMP input 4 pin	35	DTC1	Duty limit pin 1 pin	57	GND	Ground pin
14	IN+	COM input + pin	36	COMP1	Error amp output 1 pin	58	RT	Frequency setting pin
15	IN-	COM input – pin	37	FB1	Feedback input 1 pin	59	LDFB2	LDO feedback 2 pin
16	VCOM	COM output pin	38	CTL1	Control input 1 pin	60	LDO2	LDO output 2 pin
17	OUT4	AMP output 4 pin	39	FAULT	Protection detection output pin	61	LDVCC2	Power supply input pin
18	OUT3	AMP output 3 pin	40	SCP	Short-circuit protection delay pin	62	CTL3	Control input 3 pin
19	OUT2	AMP output 2 pin	41	UVLO	Under-voltage lockout protection setting pin	63	SW3	Switching output 3 pin
20	OUT1	AMP output 1 pin	42	VCC	Power supply input pin	64	BOOT3	Switch boot pin 3 pin
21	HGND	Ground pin	43	VREF	Reference voltage output pin			
22	CPFB2	Charge pump feedback 2 pin	44	FB2	Feedback input 2 pin			

Absolute Maximum Ratings (Ta=25°C)

Parameter	Symbol	Rating	Unit
Power Supply Voltage 1	V _{CC} , V _{PVCC2, 3}	15	V
Power Supply Voltage 2	V _{LDVCC2}	7	V
Power Supply Voltage 3	V _{HVCC}	20	V
SW1 Pin Voltage	V _{SW1}	20	V
Maximum Junction Temperature	T _{jmax}	150	°C
Power Dissipation	P _d	5.20 (Note 1)	W
Operating Temperature Range	T _{opr}	-40 to +105	°C
Storage Temperature Range	T _{stg}	-55 to +150	°C

(Note 1) To use the IC at temperatures over Ta=25°C, derate power rating by 41.6mW/°C.

When mounted on a four-layer glass epoxy board measuring 70 mm x 70 mm x 1.6 mm (with reverse side of copper foil measuring 70mm x 70 mm).

Caution: Operating the IC over the absolute maximum ratings may damage the IC. The damage can either be a short circuit between pins or an open circuit between pins and the internal circuitry. Therefore, it is important to consider circuit protection measures, such as adding a fuse, in case the IC is operated over the absolute maximum ratings.

Recommended Operating Conditions (Ta=-40°C to +105°C)

Parameter	Symbol	Min	Max	Unit
Power Supply Voltage 1	V _{CC} , V _{PVCC2, 3}	4.2	14	V
Power Supply Voltage 2	V _{LDVCC2}	-	5.5	V
Power Supply Voltage 3	V _{HVCC}	6	18	V
SW1 Pin Voltage	V _{SW1}	-	18	V
SW1 Pin Current	I _{SW1}	-	3	A
SW2, 3 Pin Current	I _{SW2,3}	-	2	A
Fault Detection Pull-up Voltage	V _{FAULT}	-	5.5	V
Power Good Pull-up Voltage	V _{PG}	-	5.5	V
Switching Frequency	f _{sw}	200	800	kHz

Electrical Characteristics (Unless otherwise noted, Ta=25°C, V_{CC}=12V, V_{HVCC}=15V)

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
[DC/DC Converter Controller Block]						
Step-up Feedback Voltage	V _{FB1}	1.230	1.250	1.270	V	
Step-down Feedback Voltage	V _{FB2}	1.225	1.250	1.275	V	
Sync Rectification Feedback Voltage	V _{FB3}	0.882	0.900	0.918	V	
Input Bias Current	I _{FB}	-1.2	-0.1	+1.2	μA	V _{FB} =1.5V
COMP Source Current	I _{CSO}	15	40	65	μA	
COMP Sink Current	I _{CSI}	-65	-40	-15	μA	
SW1, 2, 3 Max Duty Ratio	MDT	85	92	99	%	
DTC at 0% Duty	V _{DTCMIN}	-	0.1	-	V	V _{FB} =0V
DTC at Max Duty	V _{DTCMAX}	-	0.9	-	V	V _{FB} =0V
DTC Bias Current	I _{DTC}	-1.2	-0.1	+1.2	μA	V _{DTC} =0V
DTC Sink Current	I _{DTC}	1	2	4	mA	
SW1 On Resistance	R _{ON1}	-	0.2	-	Ω	I _{SW} =1A
SW1 Current Limit	I _{SW1OCP}	3	-	-	A	
SW2 High Level On Resistance	R _{ON2H}	-	0.2	-	Ω	I _{SW} =1A
SW2 Low Level On Resistance	R _{ON2L}	-	2	-	Ω	I _{SW} =20mA
SW3 High Level On Resistance	R _{ON3H}	-	0.2	-	Ω	I _{SW} =1A
SW3 Low Level On Resistance	R _{ON3L}	-	0.2	-	Ω	I _{SW} =1A
SW1, 2, 3 Leak Current	I _{SWLEAK}	-5	0	+5	μA	
PGATE Sink Current	I _{PGTSI}	4	9	14	μA	V _{PG} =5V
PGATE Source Current	I _{PGTSO}	4	8	15	mA	V _{PG} =5V
PG On Resistance	R _{ONPG}	0.5	1.0	1.5	kΩ	
PG Leak Current	I _{PGLEAK}	-5	0	+5	μA	
PG1, 2, 3 On Voltage	PGH	-	90	-	%	
PG1, 2, 3 Off Voltage	PGL	-	60	-	%	
[LDO1, 2, 3 Block]						
Feedback Voltage 1, 2 ,3	V _{LDFB123}	1.231	1.250	1.269	V	
Input Bias Current	I _{LDFB123}	-1.2	-0.1	+1.2	μA	
LDO1 Output Voltage Range 1	V _{LDO1}	0	-	V _{PVCC2}	V	
LDO1 Output Voltage Range 2	V _{LDO2}	0	-	V _{LDVCC2}	V	
LDO1 Output Voltage Range 3	V _{LDO3}	0	-	V _{HVCC}	V	
I/O Voltage Difference 1	V _{DPLD1}	0.3	0.75	1.6	V	V _{LDFB1} =0V, I _o =500mA
I/O Voltage Difference 2	V _{DPLD2}	0.1	0.33	0.75	V	V _{LDFB2} =0V, I _o =200mA
I/O Voltage Difference 3	V _{DPLD3}	0.14	0.3	0.65	V	V _{LDFB3} =0V, I _o =20mA
LDPG1 ON Voltage	LDPG1H	-	90	-	%	
LDPG1 OFF Voltage	LDPG1L	-	60	-	%	

Electrical Characteristics – continued (Unless otherwise noted, Ta=25°C, V_{CC}=12V, V_{HVCC}=15V)

Parameter	Symbol	Limit			Unit	Conditions
		Min	Typ	Max		
[Charge Pump Block]						
Positive/Negative Feedback Voltage	V _{CPFB12}	1.225	1.250	1.275	V	
Input Bias Current	I _{CPFB12}	-1.2	-0.1	+1.2	μA	
VCP I/O Voltage Difference	V _{DP12}	0.28	0.7	1.55	V	I _O =100mA
C1, 2 High Level On Resistance	R _{ONCH}	-	3	-	Ω	
C1, 2 Low Level On Resistance	R _{ONCL}	-	3	-	Ω	
CPPG1 On Voltage	CPPGH	-	80	-	%	
CPPG1 Off Voltage	CPPGL	-	60	-	%	
[Operation Amplifier Block]						
Input Offset Voltage	V _{OFF}	-15	0	+15	mV	
Input Bias Current	I _{BAMP}	-1.2	0	+1.2	μA	
AMP Output Current Capability	I _{AMP}	30	50	200	mA	
VCOMP Output Current Capability	I _{COM}	60	150	400	mA	
AMP Slew Rate	SRAMP	-	4	-	V/μs	
VCOM Slew Rate	SRCOM	-	4	-	V/μs	
Load Stability	ΔV _O	-15	0	+15	mV	I _O =+1mA to -1mA
Max Output Voltage	V _{OH}	V _{HVCC} -1.0	V _{HVCC} -0.8	-	V	I _O =-1mA, V _{IN} =V _{HVCC} -0.8V
Min Output Voltage	V _{OL}	-	0.1	0.16	V	I _O =1mA, V _{IN} =0V
[Overall]						
Reference Output Voltage	V _{VREF}	2.44	2.50	2.66	V	
REG Output Voltage	V _{REG}	4.7	5.0	5.3	V	
Oscillating Frequency	f _{SW}	450	550	650	kHz	R _{RT} =51kΩ
UVLO Pin ON Voltage	V _{UVLOON}	0.88	1.00	1.12	V	
UVLO Pin OFF Voltage	V _{UVLOOFF}	0.93	1.05	1.17	V	
VCC Under-Voltage Lockout Protection ON/OFF Voltage	V _{CCUV}	3.5	-	4.2	V	
HVCC Under-Voltage Lockout Protection ON/OFF Voltage	V _{HVUV}	4.3	-	5.3	V	
CTL ON Voltage	V _{CTLON}	2	-	-	V	
CTL OFF Voltage	V _{CTLOF}	-	-	0.2	V	
CTL Bias Current	I _{CTL}	-20	-12.5	-5	μA	V _{CTLX} =0V
SCP Source Current	I _{SCPSO}	2	5	8	μA	
SCP Sink Current	I _{SCPSI}	2	5	10	mA	
SCP Threshold Voltage	V _{SCP}	-	1.25	-	V	
Fault Detection ON Resistance	R _{ONFLT}	0.5	1	1.5	kΩ	
Average Consumption Current 1 (VCC, PVCC2, 3)	I _{CC}	-	5	11	mA	No Switching
Average Consumption Current 2 (HVCC)	I _{HICC}	-	2.8	6	mA	No Switching

Description of Operation of Each Block and Procedure for Selecting Application Components

(1) Step-up DC/DC Converter Block

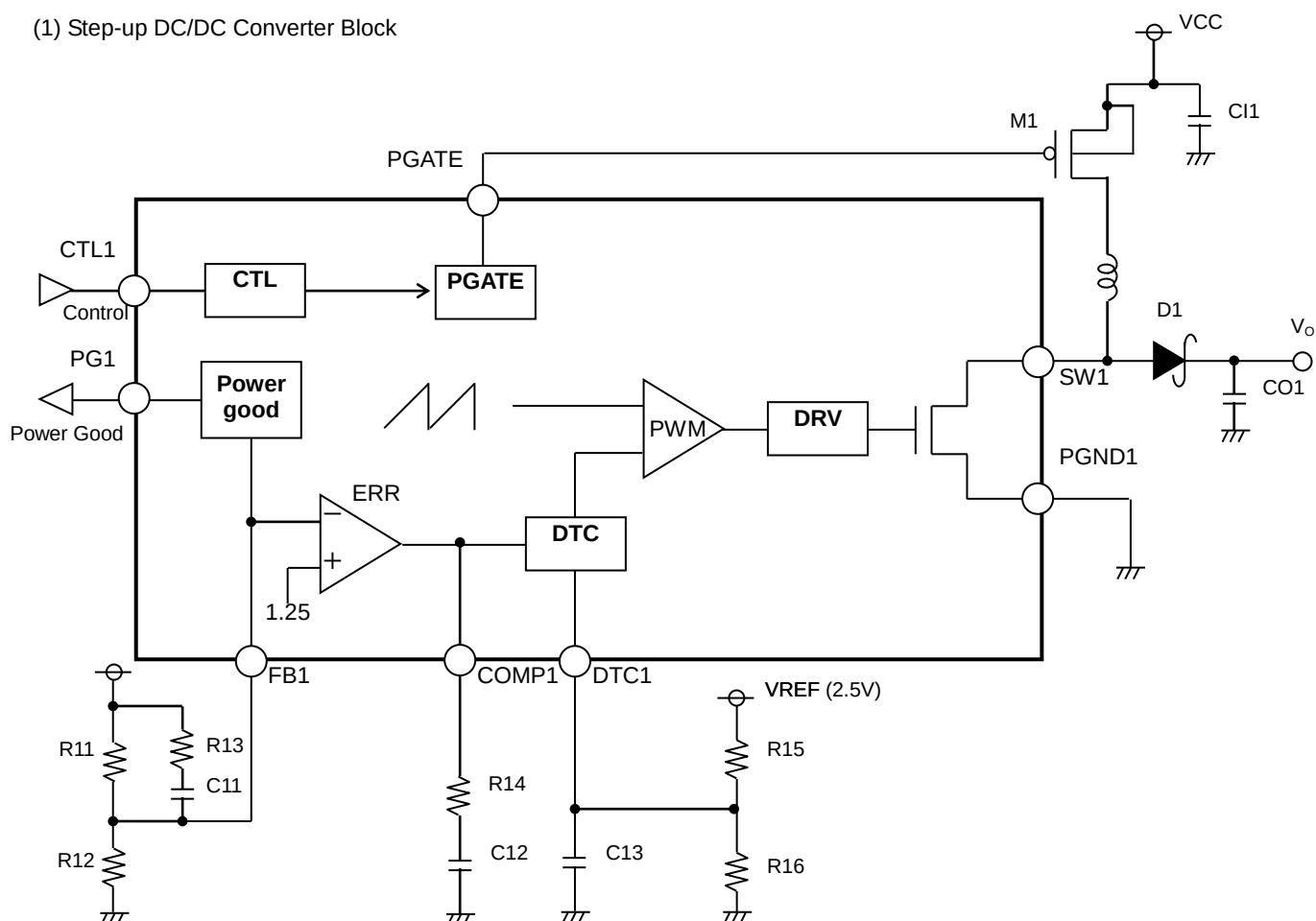


Figure 5. Step-up DC/DC Converter Block

This is a step-up DC/DC converter block that outputs a step-up voltage upon receipt of a signal from CTL1.

When the high-level signal is input to CTL1, a current will be pulled up from PGATE to turn ON input switch M1. At the time of startup, since the switching duty is limited by the DTC1 pin voltage, a soft start is operated. When output reaches 90% of the set voltage, the Power Good signal will be output from PG1.

(1.1) Selecting input switch M1

Input switch M1 will serve as a switch to block the path from VCC to output when a low-level control signal is input to CTL1. Select the input switch with careful attention paid to the following conditions.

Recommended ICs: RSQ and RTQ Series

Maximum inductor current:	$I_{INMAX} + \frac{\Delta I_L}{2}$	< Rated current of FET
Power supply voltage:	VCC	< Rated voltage of FET
Power supply voltage:	VCC	< ON voltage of FET gate

When the CTL1 control input is switched to the high level, a 9 μ A (Typ) sink current will be pulled from the PGATE pin to turn ON the input switch.

(1.2) Selecting the output L constant

The coil L to be used for output is determined by the rated current I_{LR} and the maximum input current value I_{INMAX} of the coil.

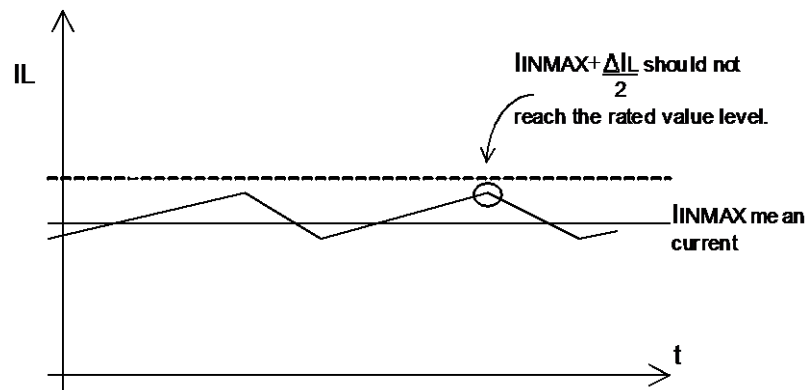


Figure 6. Coil Current Waveform (Step-up DC/DC Converter)

Make adjustments so that $I_{INMAX} + \Delta I_L / 2$ will not reach the rated current I_{LR} . At this time, ΔI_L is obtained by the following equation.

$$\Delta I_L = \frac{1}{L} V_{CC} \times \frac{V_O - V_{CC}}{V_O} \times \frac{1}{f} \quad [A]$$

Where:

f = Switching frequency

In addition, since the coil L value may have variations in the range of approximately $\pm 30\%$, set this value with sufficient margin.

If the coil current exceeds the rated current I_{LR} , the internal IC element may be damaged.

(1.3) Output capacitor setting

For capacitor C to be used for output, set it to the permissible value of the ripple voltage V_{PP} or that of the drop voltage at the time of a sudden load change, whichever is larger.

The output ripple voltage is obtained by the following equation.

$$\Delta V_{PP} = I_{LMAX} \times R_{ESR} + \frac{1}{fC_O} \times \frac{V_{CC}}{V_O} \times (I_{LMAX} - \frac{\Delta I_L}{2})$$

Make this setting so that the voltage will fall within the permissible ripple voltage range.

For the drop voltage V_{DR} during a sudden load change, estimate the V_{DR} with the following equation.

$$V_{DR} = \frac{\Delta I}{C_O} \times 10 \mu\text{sec} \quad [V]$$

Wherein, 10 μsec is the estimate of DC/DC response speed.

Set C_O so that these two values will fall within the limit values.

Since the DC/DC converter causes a peak current to flow between input and output, capacitors must also be mounted on the input side. For this reason, it is recommended to use low-ESR capacitors above 10 μF and below 100m Ω as the input capacitors. Using input capacitors outside of this range may superimpose excess ripple voltage upon the input voltage, causing the IC to malfunction.

However, since the aforementioned conditions vary with load current, input voltage, output voltage, inductor value, and switching frequency, be sure to verify the margin using the actual product.

(1.4) Output rectifier diode setting

For the rectifier diodes to be used as the output stage of the DC/DC converter, it is recommended to use Schottky diodes. Select diodes with careful attention paid to the maximum inductance current, maximum output voltage, and power supply voltage.

$$\text{Maximum inductance current: } I_{\text{INMAX}} + \frac{\Delta I_L}{2} < \text{Rated current of diode}$$

$$\text{Maximum output voltage: } V_{\text{OMAX}} < \text{Rated voltage of diode}$$

In addition, since each parameter has variations in current and voltage of 30% to 40%, design systems with sufficient margin.

(1.5) Output voltage setting

Set output voltage using the following equation with feedback resistance composed of R11 and R12.

$$V_o = \frac{R11 + R12}{R12} \times 1.25 \text{ [V]}$$

Set the maximum output voltage to not more than 18V so that it will not exceed the rating of the SW1 pin.

It is recommended to apply a setting range of 10kΩ to 330kΩ. Setting the feedback resistance to not more than 10kΩ will result in degraded voltage efficiency, while setting it to not less than 330kΩ will result in higher offset voltage due to an input bias current of 0.1μA (Typ) of the internal error amplifier.

(1.6) Phase compensation setting

Phase setting procedure:

The following conditions are required to ensure the stability of the negative feedback system.

- When the gain is set to “1” (0 dB), the phase lag should not be more than 150° (i.e., phase margin should not be less than 30°).

In addition, since DC/DC converter applications are sampled according to the switching frequency, the overall system GBW should be set to not more than 1/10 of the switching frequency. The targeted characteristics of the applications can be summarized as follows.

- When the gain is set to “1” (0 dB), the phase lag should not be more than 150° (i.e., phase margin should not be less than 30°).
- The GBW at that time (i.e., frequency when the gain is set to “0 dB”) should not be more than 1/10 of the switching frequency.

The responsiveness is determined by the GBW limitation. Consequently, to raise the responsiveness, higher switching frequencies are required.

To ensure the stability through the phase compensation, it is necessary to cancel the secondary phase delay (-180°) caused by LC resonance with the secondary phase lead (in other words, by adding two phase leads).

The GBW (i.e., frequency when the gain is set to “0 dB”) is determined by phase compensation capacitance connected to the error amplifier. If GBW needs to be reduced, increase the capacitance of the capacitor.

(i) Ordinary integrator (Low-pass filter)

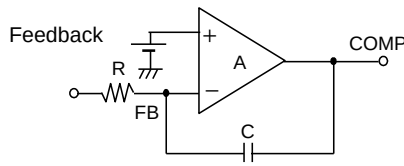


Figure 7

$$f_a = \frac{1}{2\pi R C A} \text{ [Hz]}$$

(ii) Open loop characteristics of integrator

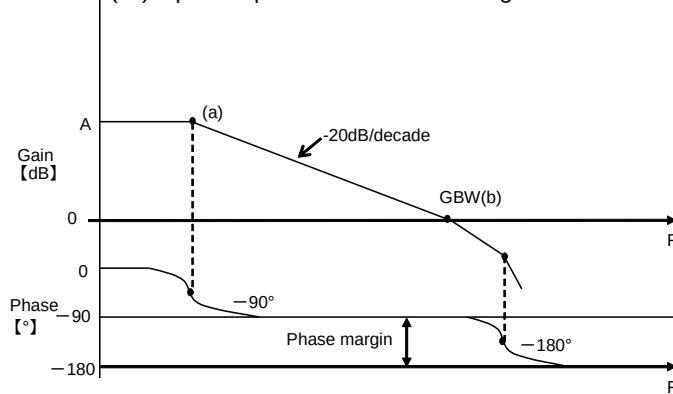


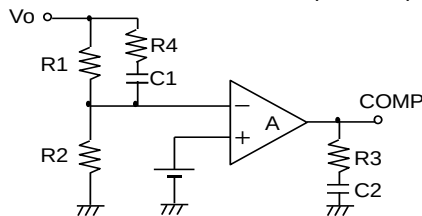
Figure 8

$$f_b = \text{GBW} = \frac{1}{2\pi R C} \text{ [Hz]}$$

Since the phase compensation like that shown in (a) and (b) applies to the error amplifier, it will act as a low-pass filter.

For DC/DC converter applications, R represents feedback resistors connected in parallel.

According to the LC resonance of the output, two phase leads should be added.



$$\text{LC resonant frequency } f_p = \frac{1}{2\pi\sqrt{LC}} \text{ [Hz]}$$

$$\text{Phase lead } f_{z1} = \frac{1}{2\pi C1 R1} \text{ [Hz]}$$

$$\text{Phase lead } f_{z2} = \frac{1}{2\pi C2 R3} \text{ [Hz]}$$

Figure 9

Set the lead frequency of one of the phases close to the LC resonant frequency for the purpose of canceling the LC resonance.

Note: If high-frequency noise occurs in output, it will pass through capacitor C1 and affect the feedback. To avoid this problem, add resistor R4 of approximately 1kΩ in series with capacitor C1.

(1.7) Duty cycle limit setting

Applying a voltage to the DTC pin makes it possible to fix the maximum duty cycle. Furthermore, since the upper limit value of the maximum duty cycle is fixed within the IC, it will not increase beyond the upper limit value. Figure 10 shows the relationship between the DTC voltage and the maximum duty cycle. Refer to this figure to make the DTC voltage setting. Subsequently, set R15 and R16 so that the DTC voltage will reach the level shown in the figure.

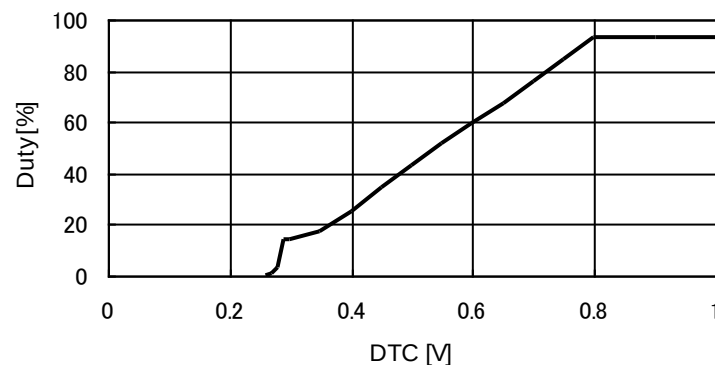


Figure 10. Characteristics of DTC Voltage vs Duty Cycle

Set the maximum duty cycle with sufficient margin so that it will not reach the maximum duty cycle for normal use. For step-up converters, the range normally used is as follows.

$$\text{Max On duty cycle} = \frac{V_{OMAX} - V_{CCMIN}}{V_{OMAX}} < \text{Max set duty cycle}$$

(1.8) Soft-start time setting

Adding capacitor C13 to DTC resistive dividers R15 and R16 makes it possible to use the soft-start function. The soft-start function is needed to prevent an excessive increase in coil current and output voltage overshoot at startup. The capacitance and soft-start time are obtained by the following equation.

$$t_{ss} = -C13 \times \frac{R15 \times R16}{R15 + R16} \times \ln \left(1 - \frac{\frac{V_o - V_{CC}}{V_o} \times 0.62 + 0.28}{2.5 \times \frac{R16}{R15 + R16}} \right) \text{ [sec]}$$

(1.9) Control and Power Good functions

When the control pin (CTL) is set to low-level input, the relevant block will stop operation. The control pin voltage is internally pulled up to the reference voltage VREF, whereby operating the relevant block in the open state.

The Power Good terminal (PG) is designed in an open-drain pattern to use as the control pin of a different block or an external power-good signal. The PG pin outputs a low-level signal while in the rising mode and, when the output voltage reaches 90% of the set voltage, will enter a high impedance state. At this time, the CTL pin at the destination will be switched to high-level input by the use of a pull-up resistor. In contrast, when the output voltage falls below 60% of the set voltage, the CTL pin will switch to low-level output.

To use the PG pin output as an external signal, connect a pull-up resistor.

A pull-up resistance ranging from 51kΩ to 200kΩ is recommended.

Typical application:

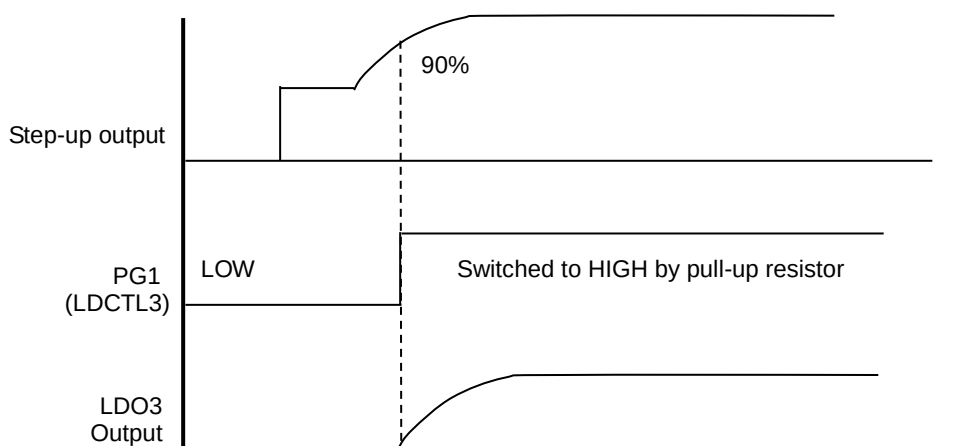
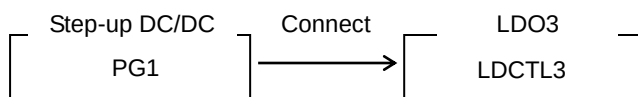


Figure 11. Typical Application of Control / Power Good Functions

(2) Step-down DC/DC Converter Block and
Sync Rectification Step-down DC/DC Converter Block

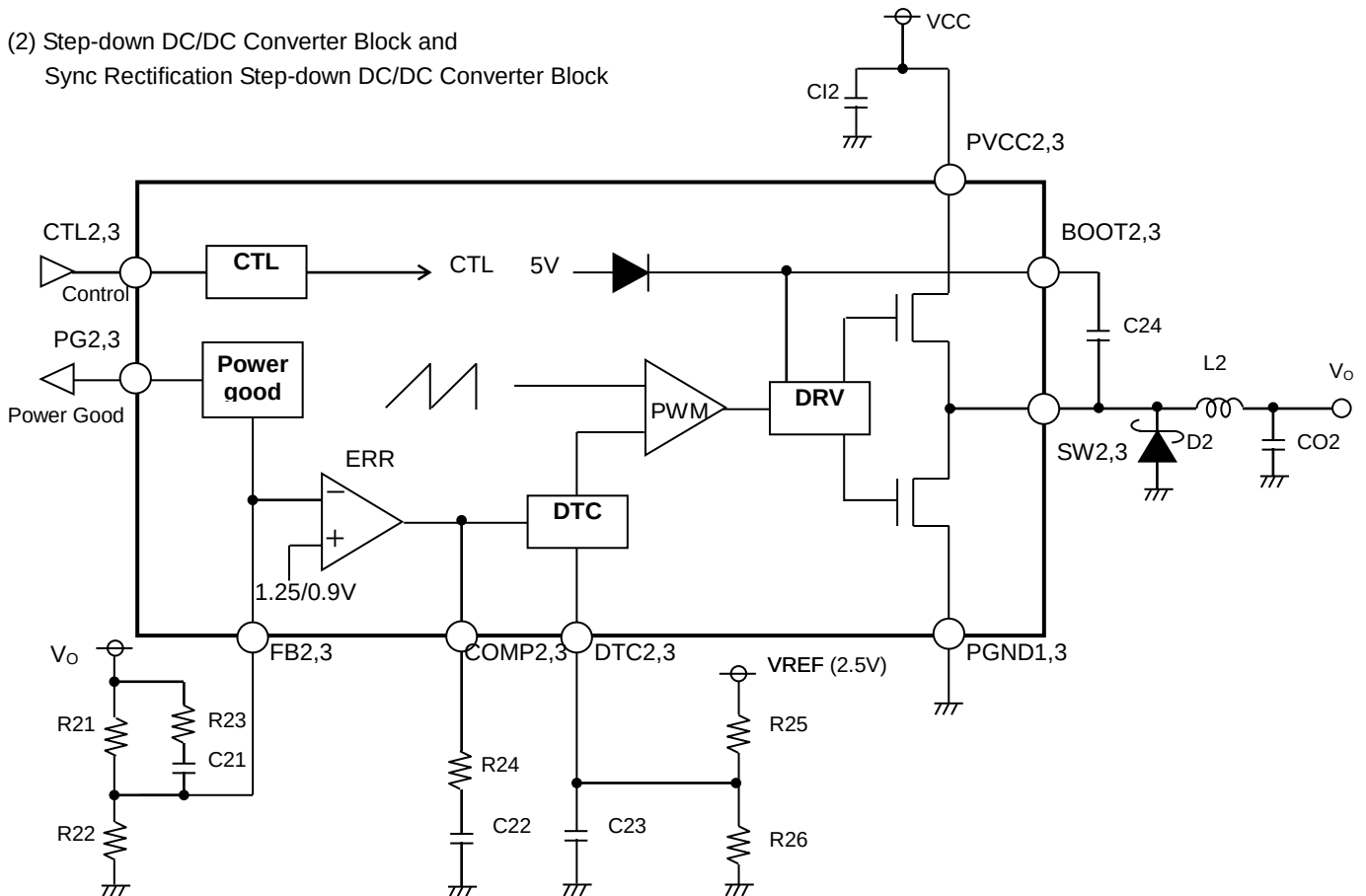


Figure 12. Step-down DC/DC Converter Block and Sync Rectification Step-down DC/DC Converter Block

The step-down DC/DC converter block and the sync rectification step-down DC/DC converter block differ in the feedback voltage and SW low-level On resistance, but have about the same configuration.

While the control signal remains at low level, the low-level SW turns ON to output a low voltage. When the control signal is switched to a high level, output voltage will start rising with the soft start function in operation.

When the output voltage reaches 90% of the set voltage, the Power Good signal will be output.

(2.1) Selecting the output L constant

The inductance L to be used for output is determined by the rated current I_{LR} and the maximum output current value I_{OMAX} of the inductor.

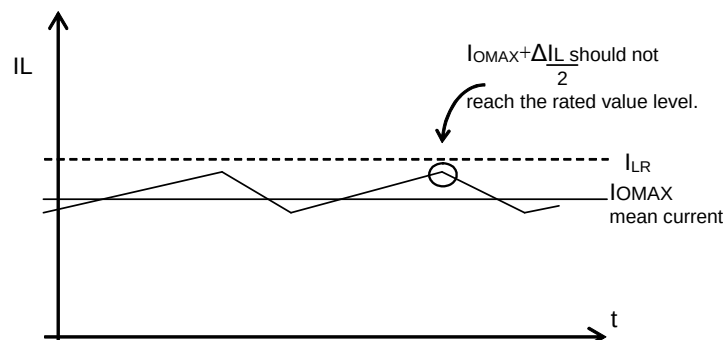


Figure 13. Coil Current Waveform (Step-down DC/DC Converter)

Make adjustments so that $I_{OMAX} + \Delta I_L / 2$ will not reach the rated current I_{LR} . At this time, ΔI_L is obtained by the following equation.

$$\Delta I_L = \frac{1}{L} \times (V_{CC} - V_o) \times \frac{V_o}{V_{CC}} \times \frac{1}{f} \quad [A]$$

In addition, since the inductance L value may have variations in the range of approximately $\pm 30\%$, set this value with sufficient margin.

If the coil current exceeds the rated current I_{LR} , the internal IC element may be damaged.

(2.2) Selecting I/O capacitors

To select I/O capacitors, refer to information in Section (1.3).

However, the output ripple voltage of the step-down DC/DC converter is obtained by the following equation.

$$\Delta V_{PP} = \Delta I_L \times R_{ESR} + \frac{\Delta I_L}{2C_O} \times \frac{V_O}{V_{CC}} \times \frac{1}{f} \quad [V]$$

(2.3) Output rectifier diode setting

For the rectifier diodes to be used as the output stage of the DC/DC converter, it is recommended to use Schottky diodes. Select diodes with careful attention paid to the maximum inductance current, maximum output voltage, and power supply voltage.

$$\text{Maximum inductance current: } I_{OMAX} + \frac{\Delta I_L}{2} < \text{Rated current of diode}$$

$$\text{Power supply voltage: } V_{CC} < \text{Rated voltage of diode}$$

In addition, since each parameter has variations in current and voltage of 30% to 40%, design systems with sufficient margin.

(2.4) Output voltage setting

Set output voltage using the following equation with feedback resistance composed of R21 and R22.

$$V_O = \frac{R21 + R22}{R22} \times V_{FB} \quad [V]$$

Where:

V_{FB} : Set to 1.25 for the step-down DC/DC converter (FB2) and 0.9 for the sync rectification step-down DC/DC converter (FB3).

It is recommended to apply a setting range of 10kΩ to 330kΩ. Setting the feedback resistance to not more than 10kΩ will result in degraded voltage efficiency, while setting it to not less than 330kΩ will result in higher offset voltage due to an input bias current of 0.1μA (Typ) of the internal error amplifier.

(2.5) Phase compensation setting

For details of phase compensation setting, refer to information in Section (1.6).

(2.6) Duty cycle limit setting

For details of duty cycle limit setting, refer to information in Section (1.7).

For step-down converters, however, the range normally used comes to the following:

$$\text{Max On duty cycle} = \frac{V_{OMAX}}{V_{CCMIN}} < \text{Max set duty cycle}$$

(2.7) Soft start time setting

Adding the capacitor C23 to the DTC resistive dividers R25 and R26 makes it possible to use the soft start function. The soft start function is needed to prevent an excessive increase in coil current at startup and output voltage overshoot at startup. The capacitance and soft start time are obtained by the following equation:

$$t_{ss} = -C23 \times \frac{R25 \times R26}{R25 + R26} \times \ln \left(1 - \frac{\frac{V_o}{V_{CC}} \times 0.62 + 0.28}{2.5 \times \frac{R26}{R25 + R26}} \right) \quad [s]$$

(2.8) Control and Power Good functions

For details of the control and Power Good functions, refer to information in Section (1.9).

(3) LDO1 to LDO3 blocks

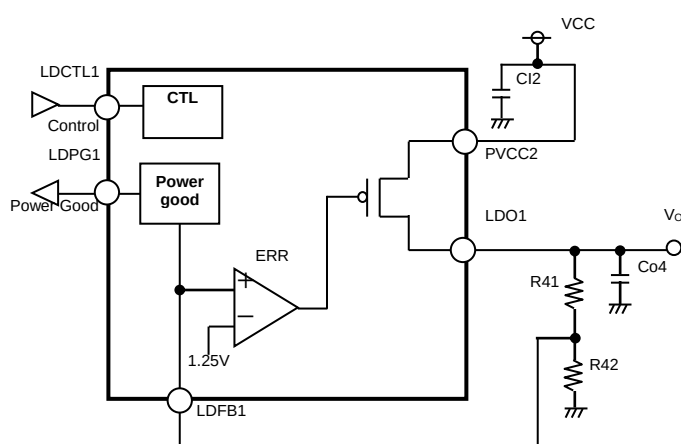


Figure 14. LDO1 Block

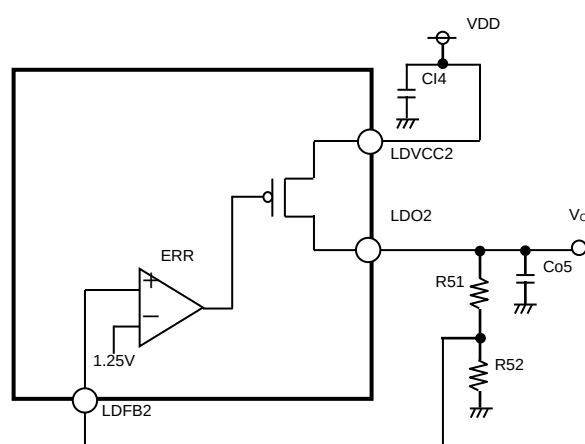


Figure 15. LDO2 Block

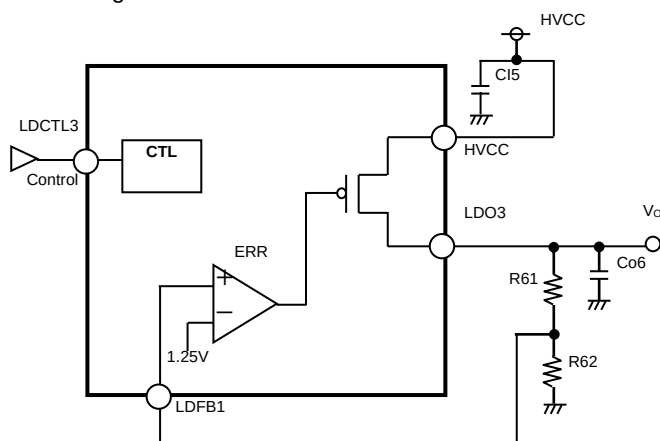


Figure 16. LDO3 Block

(3.1) Selecting I/O capacitors

The LDO1 to LDO3 blocks are ceramic capacitor compatible. Capacitance in the range of 1μF to 100μF is recommended.

(3.2) Output voltage setting

Set output voltage using the following equation with feedback resistance composed of R*1 and R*2.

$$V_o = \frac{R^*1 + R^*2}{R^*2} \times 1.25 \text{ [V]}$$

*:4 to 6

It is recommended to apply a setting range of 10kΩ to 330kΩ. Setting the feedback resistance to not more than 10kΩ will result in degraded voltage efficiency, while setting it to not less than 330kΩ will result in higher offset voltage due to an input bias current of 0.1μA (Typ) of the internal error amplifier.

The following table shows the output voltage setting ranges and current capabilities.

	Minimum setting	Maximum setting (with maximum output current)	Output current capability
LDO1	1.5V	VCC (Max 14V) -1.6V	Up to 500mA
LDO2	1.5V	LDVCC2 (Max 5.5V) -0.75V	Up to 200mA
LDO3	1.5V	HVCC (Max 18V) -0.65V	Up to 20mA

(3.3) Control and Power Good functions

For details of the control and Power Good functions, refer to information in Section (1.9).

For the LDO3 block, however, set output voltage so that the signal will be input into the control pin after HVCC, which serves as a power source, starts up.

(4) Charge Pump Block

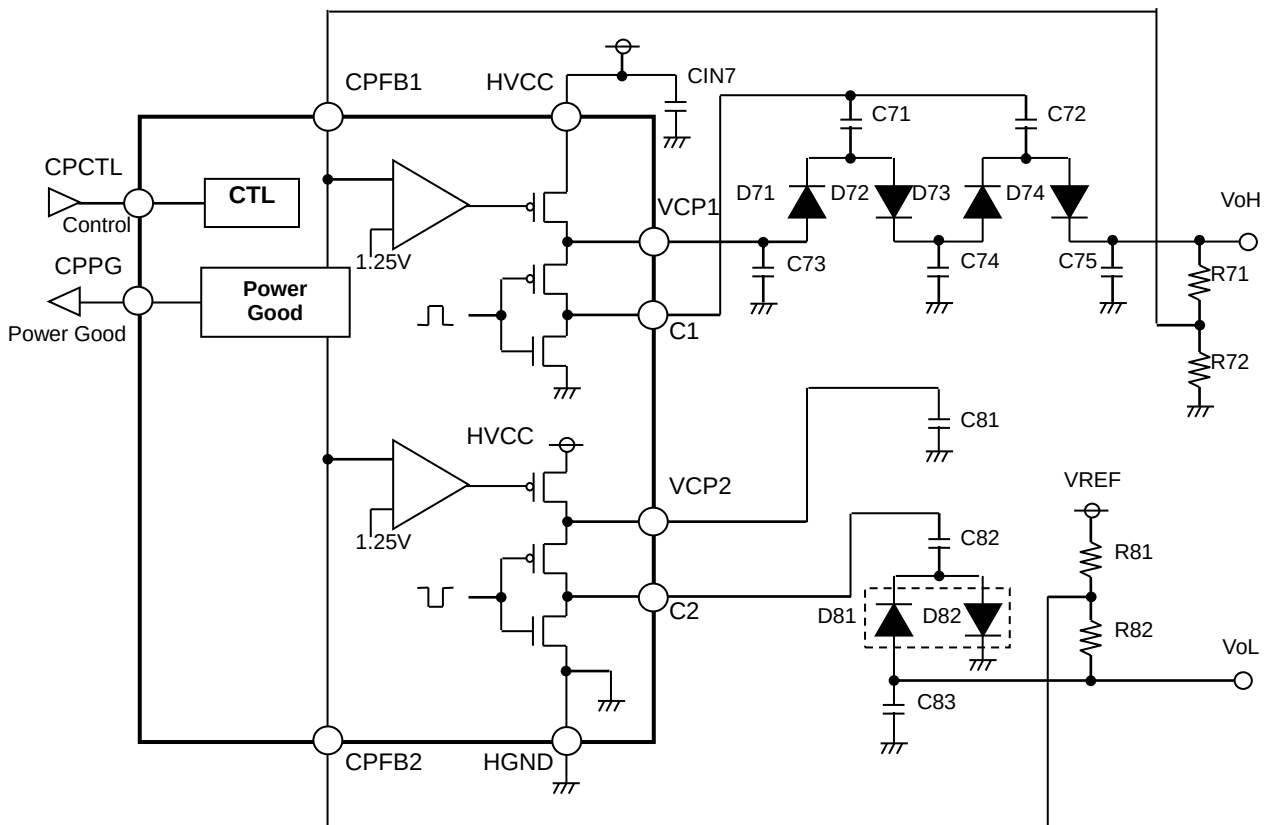


Figure 17. Charge Pump Block

When the charge pump block receives the control input signal, the negative-side charge pump will start operation. The startup sequences are internally fixed. Consequently, when the negative-side charge pump reaches 80% of the set voltage, the positive-side charge pump will start operation. When both negative- and positive-side charge pumps reach 80% of the set voltage, the power-good signal will be output from the CPPG pin.

(4.1) Selecting output diodes

For diodes D71 to D74, and D81 and D82, select Schottky diodes having a current capability three times (positive side) or two times (negative side) as high as the maximum output current and a withstand voltage higher than the output voltage.

Due to the aforementioned requirements, it is recommended to use the RB550EA dual Schottky barrier diode.

(4.2) Selecting output capacitors

Capacitors C73 and C81 serve as output capacitors for the charge pump regulators; a capacitance in the range of 1μF to 10μF is recommended. Capacitors C71, C72, and C82 serve as flying capacitors; a capacitance in the range of 0.1μF to 1μF is recommended. Capacitors C74, C75, and C83 serve as charge pump output capacitors; a capacitance in the range of 0.1μF to 10μF is recommended.

(4.3) Output voltage setting

Set output voltage using the following equation with feedback resistance.

$$V_{oH} = \frac{R71 + R72}{R72} \times 1.25 \text{ [V]}$$

$$V_{oL} = V_{REF} - \frac{R81 + R82}{R81} (V_{REF} - 1.25)$$

$$= 2.5 - \frac{R81 + R82}{R81} \times 1.25 \text{ [V]}$$

It is recommended to apply a setting range of 10kΩ to 330kΩ. Setting the feedback resistance to not more than 10kΩ will result in degraded voltage efficiency, while setting it to not less than 330kΩ will result in higher offset voltage due to an input bias current of 0.1μA (Typ) of the internal error amplifier.

(4.4) Control and Power Good functions

Make the sequence setting by inputting the power-good signal output from a different block. However, make this setting so that the signal will be input into the CPCTL control pin after HVCC, which serves as a power source, starts up. For example, to generate HVCC in the step-up DC/DC converter block, do not set the sequence so that the same Power Good pin is connected to CTL1 and CPCTL. Since the sequences of the negative- and positive-side charge pumps are internally fixed, the sequence of the negative-side pump starts up first and is followed by that of the positive-side pump. When both negative- and positive-side charge pumps reach 80% of the set voltage, the power-good signal will be output from the CPPG pin. The power-good signal output pattern is the same as that of different blocks. For details, refer to information in Section (1.9).

(5) Common Amplifier + 4ch Buffer Amplifier

The AMP and VCOM amplifiers operate in the range of 0.1V to HVCC–0.8V (Typ). Normally, use the VCOM amplifier as a buffer type amplifier as shown in (a). Use the output voltage of the LDO3 block for power supply on the reference side.

To increase the current drive capability, use the PNP and NPN transistors as shown in (b).

When the VCOM amplifier is not used, set the block to the buffer type as shown in (a) and ground the V+ pin. In this case, it is recommended to set the R3 and R4 resistors in the range of 10kΩ to 100kΩ. Setting them to not more than 10kΩ may increase current consumption, thus resulting in degraded power efficiency. Setting them to not less than 100kΩ may result in higher offset voltage due to the input bias current of 0.1μA (Typ).

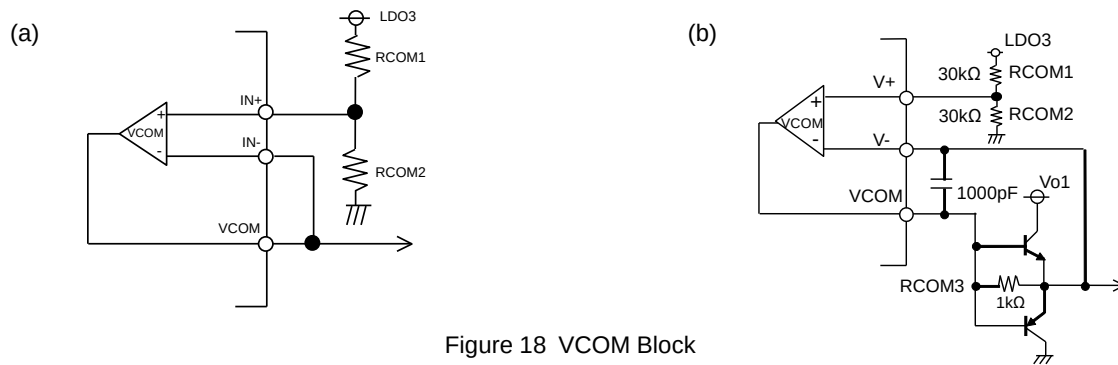


Figure 18 VCOM Block

$$V_{COM} = \frac{R_{COM2}}{R_{COM1} + R_{COM2}} \times LDO3 \text{ [V]}$$

Resistance of approximately 1kΩ is recommended for RCOM3.

(6) Common Block

(6.1) UVLO function

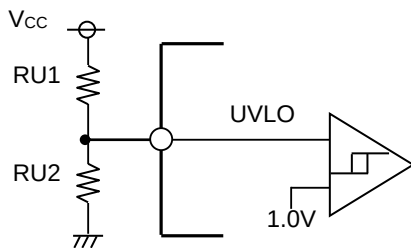


Figure 19. UVLO Block

Set the UVLO voltage with RU1 and RU2. The UVLO protection function will be implemented when the UVLO pin voltage falls below 1.0V (Typ) and canceled when it exceeds 1.05V (Typ). The VCC voltage at which the UVLO function is activated is expressed by the following equation.

$$V_{UVLO} = \frac{RU1 + RU2}{RU2} \text{ [V]}$$

It is recommended to set resistance in the range of 10kΩ to 200kΩ.

In addition, the VCC pin incorporates a fixed UVLO function. Consequently, when the UVLO pin voltage falls below 3.8V (Typ), the UVLO protection function will be operated even if the external UVLO voltage is set below 3.8V (Typ).

(6.2) SCP function

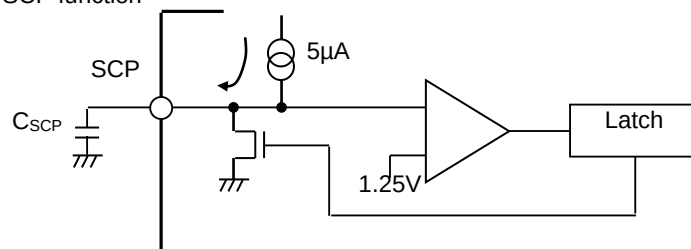


Figure 20. SCP Block

The SCP function protects against short-circuits in the outputs of the step-up DC/DC converter, step-down DC/DC converter, sync rectification step-down DC/DC converter, LDO1, and charge pump blocks. When any one of these outputs falls below 60% of the set voltage, it will be regarded as a short-circuit in output, thus activating the short-circuit protection function.

If a short-circuit is detected, source current of 5μA (Typ) will be output from the SCP pin. Then, delay time will be set with external capacitance. When the SCP pin voltage exceeds 1.25V (Typ), the state will be latched to shut down all outputs. Once the state has been latched, it will not be canceled unless VCC restarts. The delay time setting is obtained by using the following equation.

$$TL [s] = (C_{SCP} \times 1.25) / (5 \times 10^{-6})$$

Even if none of the output startup sequences is complete at startup of the IC, short-circuits will be detected and the SCP function activated. For this reason, set the delay time substantially longer than the startup time.

(6.3) Fault detection function

This IC has the built-in fault detection function that alerts the operating status of protection circuits.

If any of the protection circuits turns ON, the FAULT pin will be pulled up to output low voltage.

In stable operating status, the pin outputs high voltage. In this case, resistance ranging from 10kΩ to 220kΩ is recommended. Setting resistance to not more than 10kΩ may generate offset voltage due to the internal On resistance, thus disabling proper output of low voltage. In contrast, setting it to not less than 220kΩ may not output proper high-level voltage due to leak current.

The FAULT pin is switched to low voltage output under any of the following conditions.

- When the UVLO (under-voltage protection) function is activated;
- When the TSD (thermal shutdown circuit) function is activated;
- When the OCP (overcurrent protection circuit) function is activated, or;
- When the SCP (short-circuit protection) function is activated.

(6.4) Variable oscillator

Changing the timing resistance RT enables switching frequency adjustment. Set resistance referring to Figure 21. Set frequency in the range of 200kHz to 800kHz.

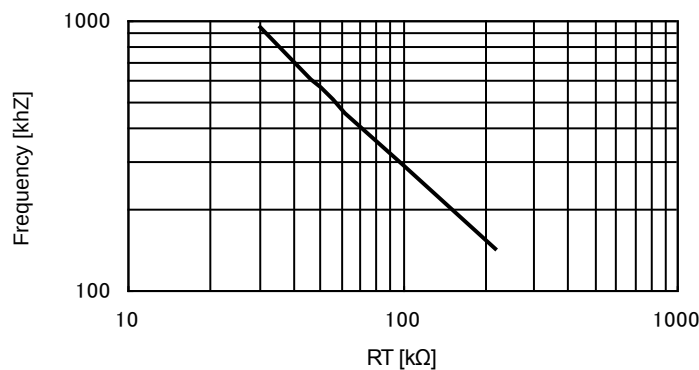


Figure 21. RT vs Switching Frequency

I/O Equivalent Circuits

2.PG3	3.DTC3	4.COMP3 36.COMP1 40.SCP 45.COMP2
5.FB3 22.CPFB2 44.FB2 53.LDFB1 59.LDFB2	7.LDFB3	8.LDCTL3 23.CPCTL
9.LDO3	10.IN1 11.IN2 12.IN3 13.IN4 14.IN+ 15.IN-	16.VCOM 17.OUT4 18.OUT3 19.OUT2 20.OUT1 24.C2 28.C1
25.VCP2 27.VCP1	29.CPPG	30.CPFB1
32.PG1 51.PG2 55.LDPG1	33.SW1	34.PGATE

I/O Equivalent Circuits - Continued

35.DTC1 46.DTC2	37.FB1	38.CTL1 50.CTL2 62.CTL3 54.LDCTL1
39.FAULT	41.UVLO	43.VREF
48.BOOT2 64.BOOT3	49.SW2 63.SW3	52.REG
56.LDO1	58.RT	60.LDO2
61.LDVCC2		

Operational Notes

1. Reverse Connection of Power Supply

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC's power supply pins.

2. Power Supply Lines

Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. Ground Voltage

Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition.

4. Ground Wiring Pattern

When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. Thermal Consideration

Should by any chance the maximum junction temperature rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the maximum junction temperature rating.

6. Recommended Operating Conditions

These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. Inrush Current

When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. Operation Under Strong Electromagnetic Field

Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. Testing on Application Boards

When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC's power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

10. Inter-pin Short and Mounting Errors

Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.

11. Unused Input Pins

Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

Operational Notes – continued

12. Regarding the Input Pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When $GND > Pin\ A$ and $GND > Pin\ B$, the P-N junction operates as a parasitic diode.

When $GND > Pin\ B$, the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

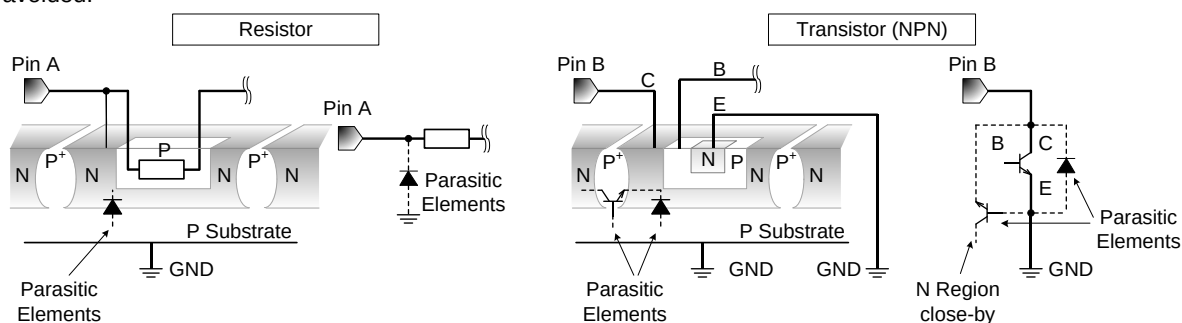


Figure 22. Example of monolithic IC structure

13. Thermal Shutdown Circuit(TSD)

This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature (T_j) will rise which will activate the TSD circuit that will turn OFF all output pins. When the T_j falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

14. Over Current Protection Circuit (OCP)

This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.

15. Discontinuous mode

The step-up and step-down DC/DC converters used in this IC have been designed on the assumption that the converters are used in continuous mode. Using the IC constantly while in discontinuous mode may result in malfunctions. To avoid this problem, make coil adjustments or insert a resistor between output and GND to prevent the IC from entering discontinuous mode while in use.

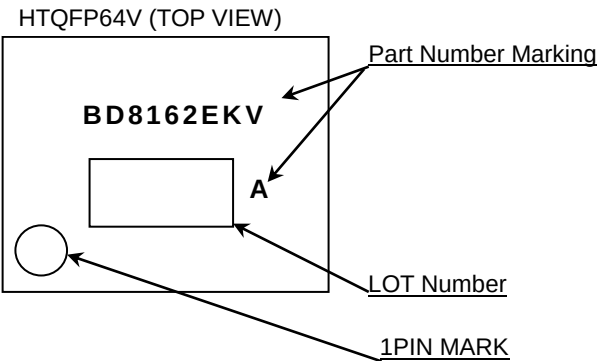
16. PCB layout for open-drain pin (SW1) of step-up DC/DC converter

Connect the open-drain pin of the FET built in the step-up DC/DC converter to the coil / diode with as thick and short of a line as possible. Particularly, making the line distance between the open-drain pin and the external diode longer or routing it with the use of a through-hole may form parasitic impedance due to patterns and cause the open-drain pin to generate a high surge voltage, thus leading to IC destruction. For this reason, ensure that the open-drain pin voltage (direct mounting to IC pin) will never exceed the absolute maximum ratings in practical applications of this IC.

Ordering Information

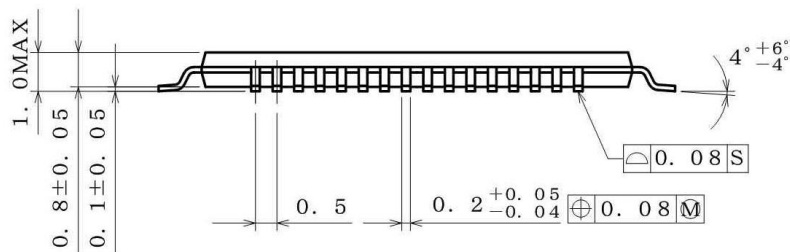
B D 8 1 6 2 A E K V										-	
Part number					Package EKV : HTQFP64V					Packaging and forming specification None: Tray	

Marking Diagram



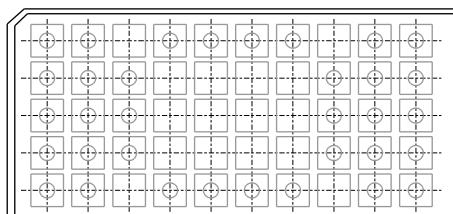
Package Name

HTQFP64V



(UNIT:mm)
PKG:HTQFP64V
Drawing No. EX282-5001

Container	Tray (with dry pack)
Quantity	1000pcs
Direction of feed	Direction of product is fixed in a tray



***Order quantity needs to be multiple of the minimum quantity.**

Revision History

Date	Revision	Changes
15.Feb.2016	001	New Release

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CLASS IV		CLASS III	

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 - Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - Sealing or coating our Products with resin or other coating materials
 - Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - Use of the Products in places subject to dew condensation
- The Products are not subject to radiation-proof design.
- Please verify and confirm characteristics of the final or mounted products in using the Products.
- In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
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- In principle, the reflow soldering method must be used on a surface-mount products, the flow soldering method must be used on a through hole mount products. If the flow soldering method is preferred on a surface-mount products, please consult with the ROHM representative in advance.

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 - [d] the Products are exposed to high Electrostatic
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Компания «Life Electronics» занимается поставками электронных компонентов импортного и отечественного производства от производителей и со складов крупных дистрибьюторов Европы, Америки и Азии.

С конца 2013 года компания активно расширяет линейку поставок компонентов по направлению коаксиальный кабель, кварцевые генераторы и конденсаторы (керамические, пленочные, электролитические), за счёт заключения дистрибьюторских договоров

Мы предлагаем:

- Конкурентоспособные цены и скидки постоянным клиентам.
- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
- Поставку компонентов, требующих военную и космическую приемку.
- Входной контроль качества.
- Наличие сертификата ISO.

В составе нашей компании организован Конструкторский отдел, призванный помогать разработчикам, и инженерам.

Конструкторский отдел помогает осуществить:

- Регистрацию проекта у производителя компонентов.
- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



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