



# BIPOLAR ANALOG INTEGRATED CIRCUIT

## $\mu$ PC2745TB, $\mu$ PC2746TB

### 3 V, SUPER MINIMOLD SILICON MMIC WIDEBAND AMPLIFIER FOR MOBILE COMMUNICATIONS

#### DESCRIPTION

The  $\mu$ PC2745TB and  $\mu$ PC2746TB are silicon monolithic integrated circuits designed as buffer amplifier for mobile communications. These low current amplifiers operate on 3.0 V (1.8 V MIN.).

These ICs are manufactured using our 20 GHz fr NESATIII silicon bipolar process. This process uses silicon nitride passivation film and gold electrodes. These materials can protect chip surface from external pollution and prevent corrosion/migration. Thus, these IC have excellent performance, uniformity and reliability.

#### FEATURES

- Supply voltage : Recommended  $V_{CC} = 2.7$  to  $3.3$  V  
Circuit operation  $V_{CC} = 1.8$  to  $3.3$  V
- Upper limit operating frequency :  $\mu$ PC2745TB;  $f_u = 2.7$  GHz TYP. @3 dB bandwidth  
 $\mu$ PC2746TB;  $f_u = 1.5$  GHz TYP. @3 dB bandwidth
- High isolation :  $\mu$ PC2745TB; ISL = 38 dB TYP. @f = 500 MHz  
 $\mu$ PC2746TB; ISL = 45 dB TYP. @f = 500 MHz
- Power gain :  $\mu$ PC2745TB;  $G_P = 12$  dB TYP. @f = 500 MHz  
 $\mu$ PC2746TB;  $G_P = 19$  dB TYP. @f = 500 MHz
- Saturated output power :  $\mu$ PC2745TB;  $P_{O(sat)} = -1$  dBm TYP. @f = 500 MHz  
 $\mu$ PC2746TB;  $P_{O(sat)} = 0$  dBm TYP. @f = 500 MHz
- High-density surface mounting : 6-pin super minimold package ( $2.0 \times 1.25 \times 0.9$  mm)

#### APPLICATIONS

- 1.5 GHz to 2.5 GHz communication system :  $\mu$ PC2745TB
- 800 MHz to 900 MHz communication system :  $\mu$ PC2746TB

#### ORDERING INFORMATION

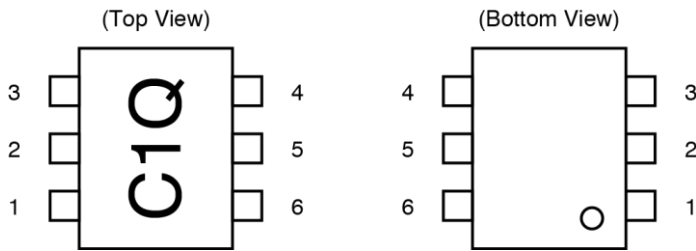
Part Number	Package	Marking	Supplying Form
$\mu$ PC2745TB-E3-A	6-pin super minimold	C1Q	• Embossed tape 8 mm wide • 1, 2, 3 pins face the perforation side of the tape • Qty 3 kpcs/reel
$\mu$ PC2746TB-E3-A		C1R	

**Remark** To order evaluation samples, contact your nearby sales office.  
Part number for sample order:  $\mu$ PC2745TB-A,  $\mu$ PC2746TB-A

**Caution: Observe precautions when handling because these devices are sensitive to electrostatic discharge.**

The information in this document is subject to change without notice. Before using this document, please confirm that this is the latest version.

**PIN CONNECTION**



Marking is an example of  $\mu$ PC2745TB

Pin No.	Pin Name
1	INPUT
2	GND
3	GND
4	OUTPUT
5	GND
6	V <sub>CC</sub>

**PRODUCT LINE-UP (T<sub>A</sub> = +25°C, V<sub>CC</sub> = 3.0 V, Z<sub>S</sub> = Z<sub>L</sub> = 50 Ω)**

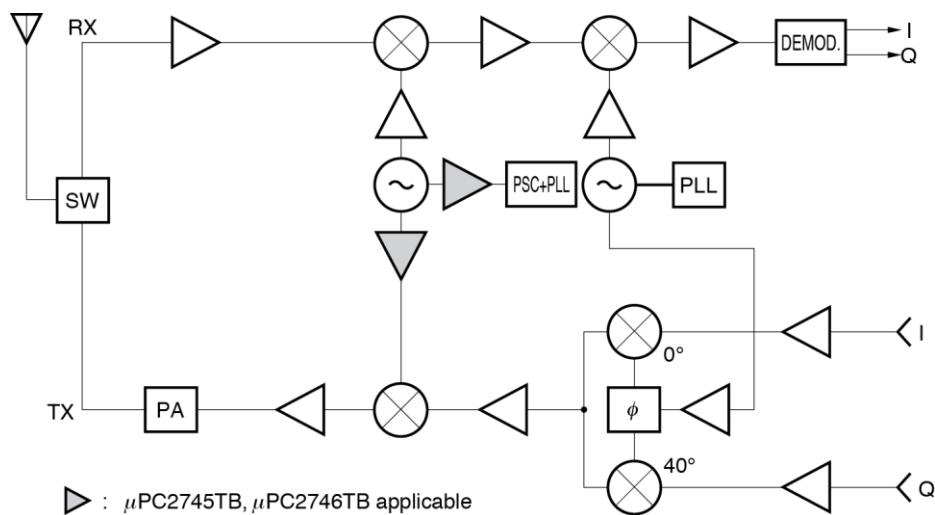
Part No.	f <sub>u</sub> (GHz)	P <sub>O(sat)</sub> (dBm)	G <sub>P</sub> (dB)	NF (dB)	I <sub>CC</sub> (mA)	Package	Making
$\mu$ PC2745T	2.7	-1.0	12	6.0	7.5	6-pin minimold	C1Q
$\mu$ PC2745TB						6-pin super minimold	
$\mu$ PC2746T	1.5	0	19	4.0	7.5	6-pin minimold	C1R
$\mu$ PC2746TB						6-pin super minimold	
$\mu$ PC2747T	1.8	-7.0	12	3.3	5.0	6-pin minimold	C1S
$\mu$ PC2747TB						6-pin super minimold	
$\mu$ PC2748T	0.2 to 1.5	-3.5	19	2.8	6.0	6-pin minimold	C1T
$\mu$ PC2748TB						6-pin super minimold	
$\mu$ PC2749T	2.9	-6.0	16	4.0	6.0	6-pin minimold	C1U
$\mu$ PC2749TB						6-pin super minimold	

**Remark** Typical performance. Please refer to **ELECTRICAL CHARACTERISTICS** in detail.

**Caution** The package size distinguish between minimold and super minimold.

**SYSTEM APPLICATION EXAMPLE**

**DIGITAL CELLULAR SYSTEM BLOCK DIAGRAM**



**PIN EXPLANATION**

Pin No.	Pin Name	Applied Voltage (V)	Pin Voltage (V) <small>Note</small>	Function and Applications	Internal Equivalent Circuit
1	INPUT	—	0.87 ----- 0.82	Signal input pin. A internal matching circuit, configured with resistors, enables 50 $\Omega$ connection over a wide band. this pin must be coupled to signal source with capacitor for DC cut.	
2 3 5	GND	0	—	Ground pin. This pin should be connected to system ground with minimum inductance. Ground pattern on the board should be formed as wide as possible. All the ground pins must be connected together with wide ground pattern to decrease impedance difference.	
4	OUTPUT	—	1.95 ----- 2.54	Signal output pin. A internal matching circuit, configured with resistors, enables 50 $\Omega$ connection over a wide band. This pin must be coupled to next stage with capacitor for DC cut.	
6	V <sub>CC</sub>	2.7 to 3.3	—	Power supply pin. This pin should be externally equipped with bypass capacity to minimize ground impedance.	

**Note** Pin voltage is measured at V<sub>CC</sub> = 3.0 V. Above:  $\mu$ PC2745TB, Below:  $\mu$ PC2746TB

**ABSOLUTE MAXIMUM RATINGS**

Parameter	Symbol	Conditions	Ratings	Unit
Supply Voltage	V <sub>CC</sub>	T <sub>A</sub> = +25°C	4.0	V
Circuit Current	I <sub>CC</sub>	T <sub>A</sub> = +25°C	16	mA
Power Dissipation	P <sub>D</sub>	T <sub>A</sub> = +85°C <b>Note</b>	270	mW
Operating Ambient Temperature	T <sub>A</sub>		-40 to +85	°C
Storage Temperature	T <sub>stg</sub>		-55 to +150	°C
Input Power	P <sub>in</sub>	T <sub>A</sub> = +25°C	0	dBm

**Note** Mounted on double-sided copper-clad 50 × 50 × 1.6 mm epoxy glass PWB

**RECOMMENDED OPERATING RANGE**

Parameter	Symbol	MIN.	TYP.	MAX.	Unit
Supply Voltage	V <sub>CC</sub>	2.7	3.0	3.3	V

**ELECTRICAL CHARACTERISTICS**

(T<sub>A</sub> = +25°C, V<sub>CC</sub> = 3.0 V, Z<sub>S</sub> = Z<sub>L</sub> = 50 Ω, unless otherwise specified)

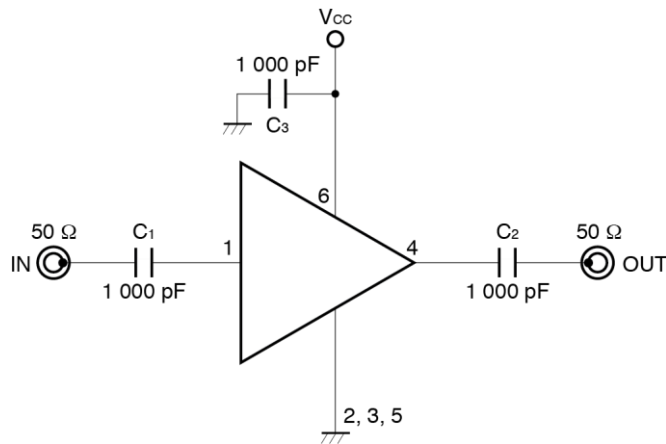
Parameter	Symbol	Test Conditions	$\mu$ PC2745TB			$\mu$ PC2746TB			Unit
			MIN.	TYP.	MAX.	MIN.	TYP.	MAX.	
Circuit Current	I <sub>CC</sub>	No signal	5.0	7.5	10.0	5.0	7.5	10.0	mA
Power Gain	G <sub>P</sub>	f = 500 MHz	9	12	14	16	19	21	dB
Noise Figure	NF	f = 500 MHz	—	6.0	7.5	—	4.0	5.5	dB
Upper Limit Operating Frequency	f <sub>u</sub>	3 dB down below from gain at f = 0.1 GHz	2.3	2.7	—	1.1	1.5	—	GHz
Isolation	ISL	f = 500 MHz	33	38	—	40	45	—	dB
Input Return Loss	RL <sub>in</sub>	f = 500 MHz	8	11	—	10	13	—	dB
Output Return Loss	RL <sub>out</sub>	f = 500 MHz	2.5	5.5	—	5.5	8.5	—	dB
Saturated Output Power	P <sub>O(sat)</sub>	f = 500 MHz, P <sub>in</sub> = -6 dBm	-4.0	-1.0	—	-3.0	0	—	dBm

**STANDARD CHARACTERISTICS FOR REFERENCE (T<sub>A</sub> = +25°C, V<sub>CC</sub> = 3.0 V, Z<sub>S</sub> = Z<sub>L</sub> = 50 Ω)**

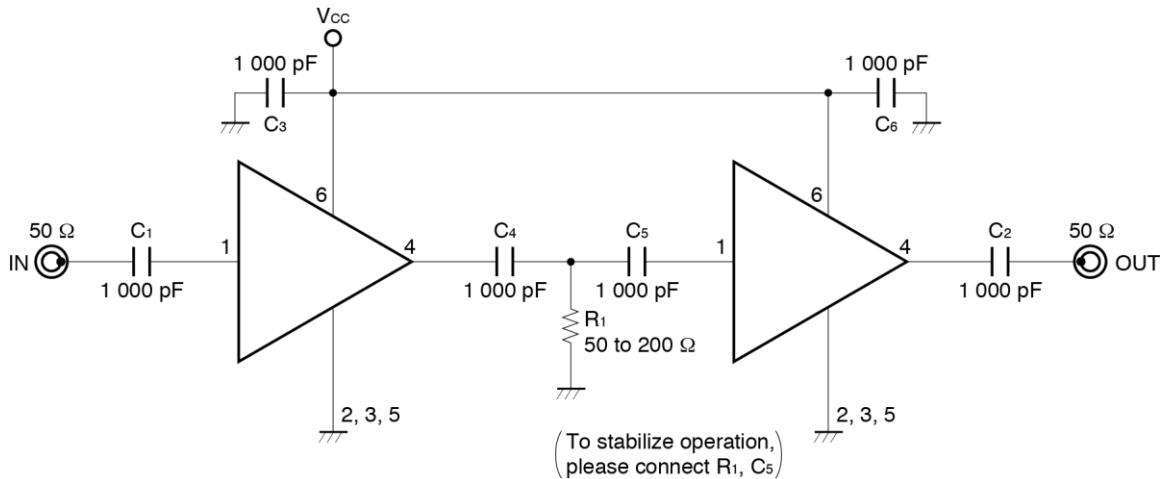
Parameter	Symbol	Test Conditions	Reference Value		Unit
			$\mu$ PC2745TB	$\mu$ PC2746TB	
Circuit Current	I <sub>CC</sub>	V <sub>CC</sub> = 1.8 V, No signal	4.5	4.5	mA
Power Gain	G <sub>P</sub>	V <sub>CC</sub> = 3.0 V, f = 1.0 GHz	12.0	18.5	dB
		V <sub>CC</sub> = 3.0 V, f = 2.0 GHz	11.0	—	
		V <sub>CC</sub> = 1.8 V, f = 0.5 GHz	7.0	14.0	
Noise Figure	NF	V <sub>CC</sub> = 3.0 V, f = 1.0 GHz	5.5	4.2	dB
		V <sub>CC</sub> = 3.0 V, f = 2.0 GHz	5.7	—	
		V <sub>CC</sub> = 1.8 V, f = 0.5 GHz	8.0	5.0	
Upper Limit Operating Frequency	f <sub>u</sub>	V <sub>CC</sub> = 1.8 V, 3 dB down below from gain at f = 0.1 GHz	1.8	1.1	GHz
Isolation	ISL	V <sub>CC</sub> = 3.0 V, f = 1.0 GHz	33	38	dB
		V <sub>CC</sub> = 3.0 V, f = 2.0 GHz	30	—	
		V <sub>CC</sub> = 1.8 V, f = 0.5 GHz	35	37	
Input Return Loss	RL <sub>in</sub>	V <sub>CC</sub> = 3.0 V, f = 1.0 GHz	13.0	10.0	dB
		V <sub>CC</sub> = 3.0 V, f = 2.0 GHz	14.0	—	
		V <sub>CC</sub> = 1.8 V, f = 0.5 GHz	6.5	10.0	
Output Return Loss	RL <sub>out</sub>	V <sub>CC</sub> = 3.0 V, f = 1.0 GHz	6.5	8.5	dB
		V <sub>CC</sub> = 3.0 V, f = 2.0 GHz	8.5	—	
		V <sub>CC</sub> = 1.8 V, f = 0.5 GHz	6.0	9.5	
Saturated Output Power	P <sub>O(sat)</sub>	V <sub>CC</sub> = 3.0 V, f = 1.0 GHz, P <sub>in</sub> = -6 dBm	-2.5	-1.0	dBm
		V <sub>CC</sub> = 3.0 V, f = 2.0 GHz, P <sub>in</sub> = -6 dBm	-3.5	—	
		V <sub>CC</sub> = 1.8 V, f = 0.5 GHz, P <sub>in</sub> = -10 dBm	-11.0	-8.0	
3rd Order Intermodulation Distortion	IM <sub>3</sub>	V <sub>CC</sub> = 3.0 V, P <sub>out</sub> = -10 dBm, f <sub>1</sub> = 500 MHz, f <sub>2</sub> = 502 MHz	-30.0	-26.0	dBc
		V <sub>CC</sub> = 1.8 V, P <sub>out</sub> = -20 dBm, f <sub>1</sub> = 500 MHz, f <sub>2</sub> = 502 MHz	-31.0	-37.0	
		V <sub>CC</sub> = 3.0 V, P <sub>out</sub> = -10 dBm, f <sub>1</sub> = 1 000 MHz, f <sub>2</sub> = 1 002 MHz	-26.0	—	



**TEST CIRCUIT**



**EXAMPLE OF APPLICATION CIRCUIT**



The application circuits and their parameters are for references only and are not intended for use in actual design-ins.

**CAPACITORS FOR THE V<sub>CC</sub>, INPUT, AND OUTPUT PINS**

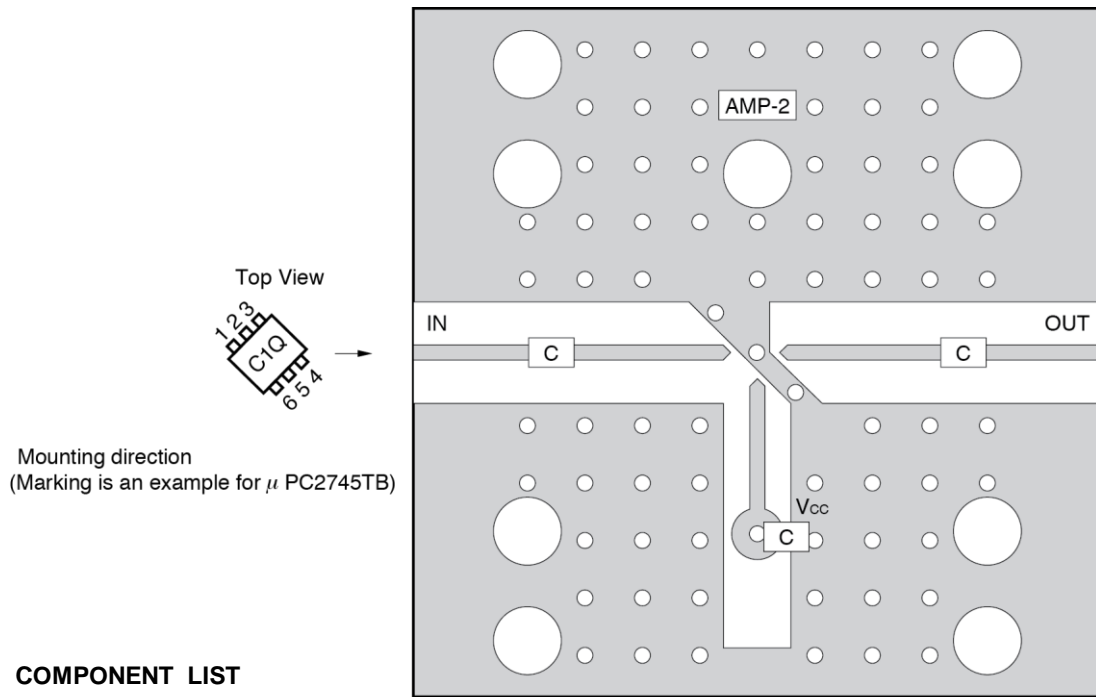
Capacitors of 1 000 pF are recommendable as the bypass capacitor for the V<sub>CC</sub> pin and the coupling capacitors for the input and output pins.

The bypass capacitor connected to the V<sub>CC</sub> pin is used to minimize ground impedance of V<sub>CC</sub> pin. So, stable bias can be supplied against V<sub>CC</sub> fluctuation.

The coupling capacitors, connected to the input and output pins, are used to cut the DC and minimize RF serial impedance. Their capacitance are therefore selected as lower impedance against a 50 Ω load. The capacitors thus perform as high pass filters, suppressing low frequencies to DC.

To obtain a flat gain from 100 MHz upwards, 1 000 pF capacitors are used in the test circuit. In the case of under 10 MHz operation, increase the value of coupling capacitor such as 10 000 pF. Because the coupling capacitors are determined by equation,  $f_c = 1/(2\pi RC)$ .

ILLUSTRATION OF THE TEST CIRCUIT ASSEMBLED ON EVALUATION BOARD



COMPONENT LIST

	Value
C	1 000 pF

Notes

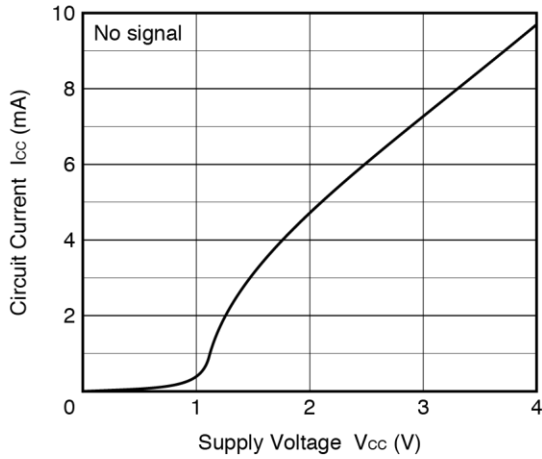
1. 30 × 30 × 0.4 mm double sided copper clad polyimide board.
2. Back side: GND pattern
3. Solder plated on pattern
4.  $\oplus \oplus \oplus$ : Through holes

For more information on the use of this IC, refer to the following application note: **USAGE AND APPLICATIONS OF 6-PIN MINI-MOLD, 6-PIN SUPER MINI-MOLD SILICON HIGH-FREQUENCY WIDEBAND AMPLIFIER MMIC (P11976E).**

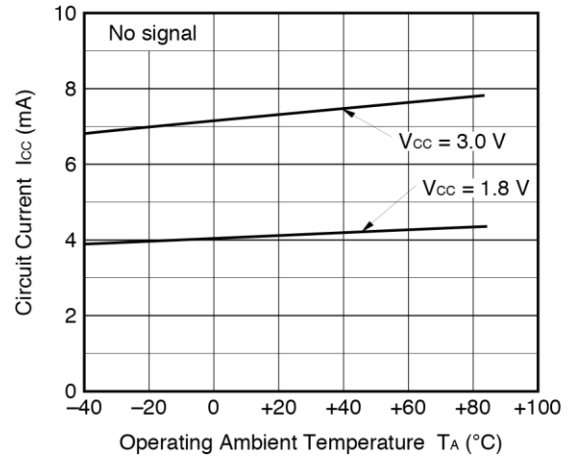
**TYPICAL CHARACTERISTICS ( $T_A = +25^\circ\text{C}$ , unless otherwise specified)**

—  $\mu$ PC2745TB —

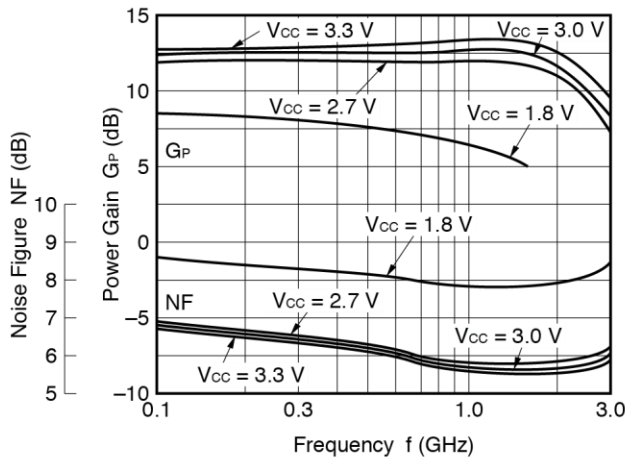
CIRCUIT CURRENT vs. SUPPLY VOLTAGE



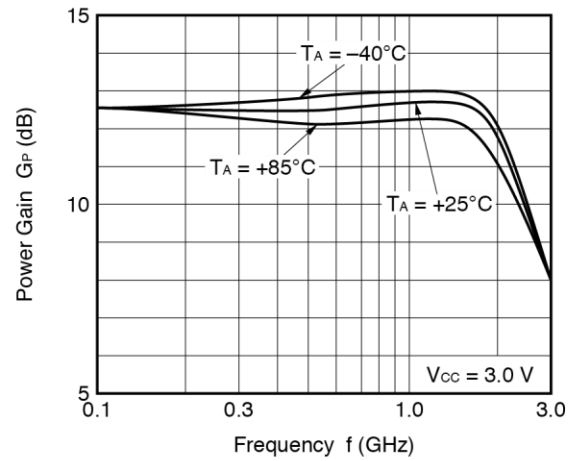
CIRCUIT CURRENT vs. OPERATING AMBIENT TEMPERATURE



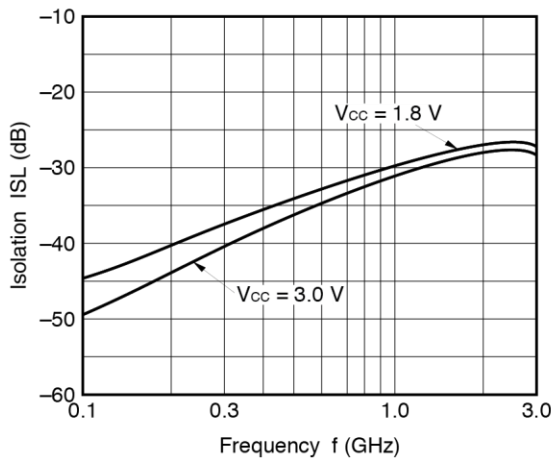
NOISE FIGURE, POWER GAIN vs. FREQUENCY



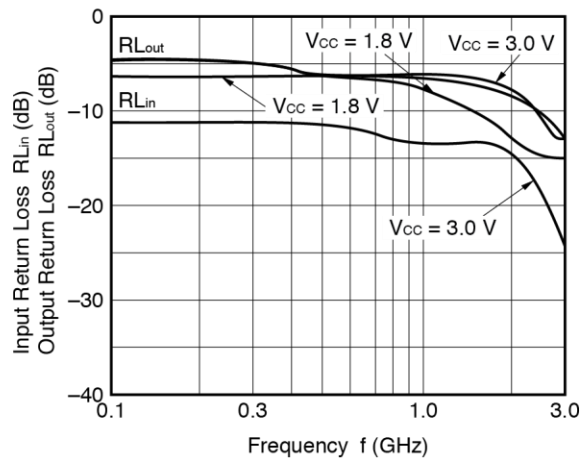
POWER GAIN vs. FREQUENCY



ISOLATION vs. FREQUENCY



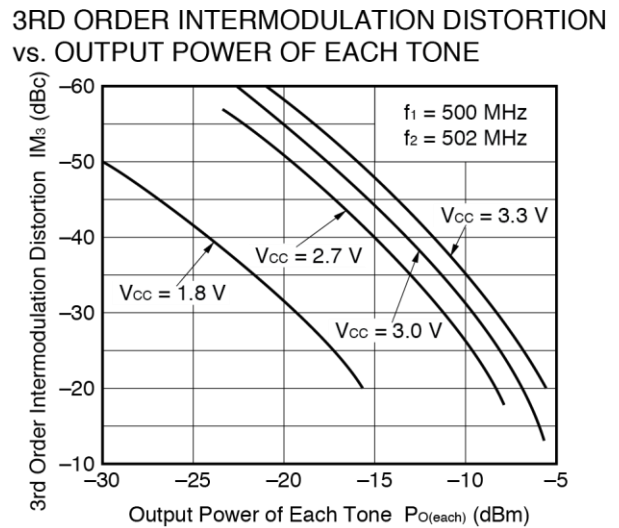
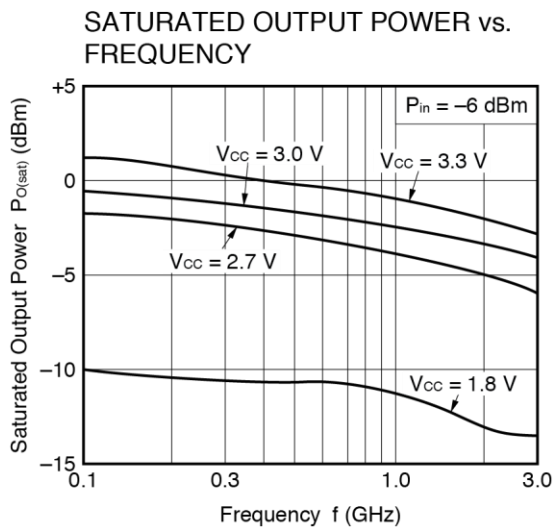
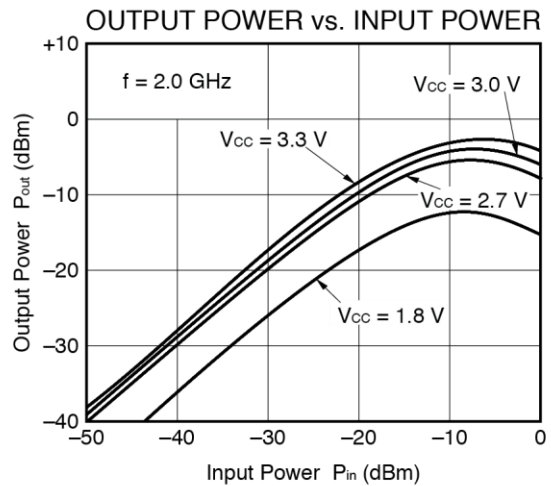
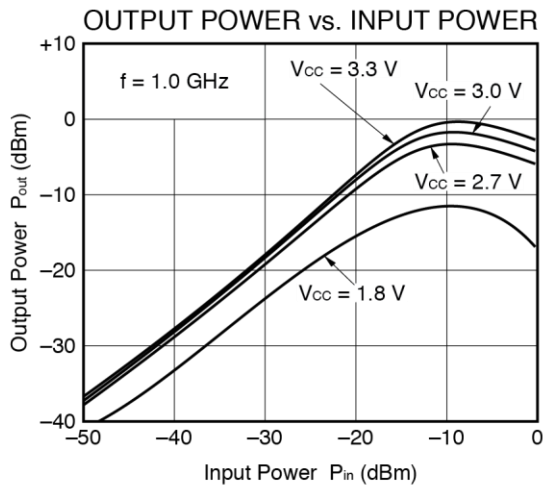
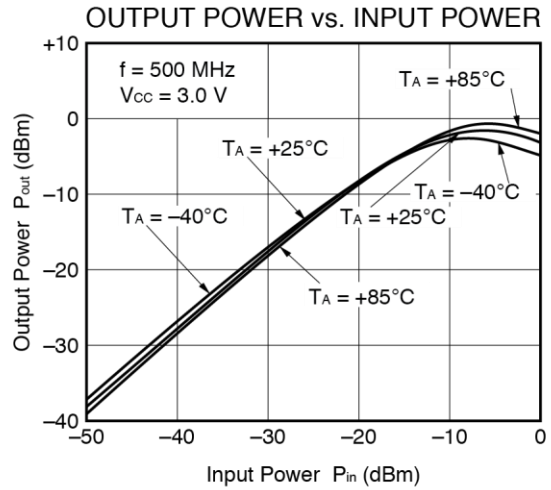
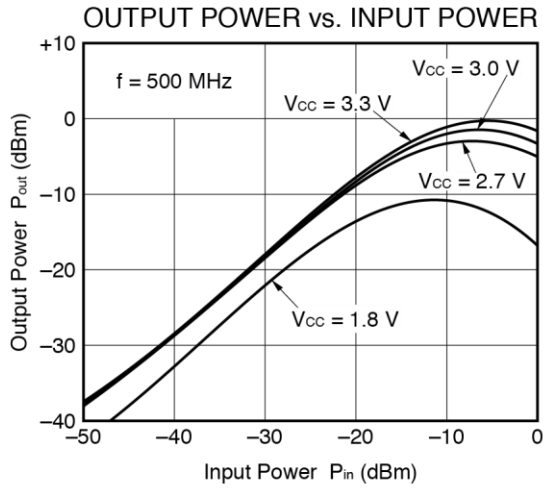
INPUT RETURN LOSS, OUTPUT RETURN LOSS vs. FREQUENCY



**Remark** The graphs indicate nominal characteristics.



—  $\mu$ PC2745TB —

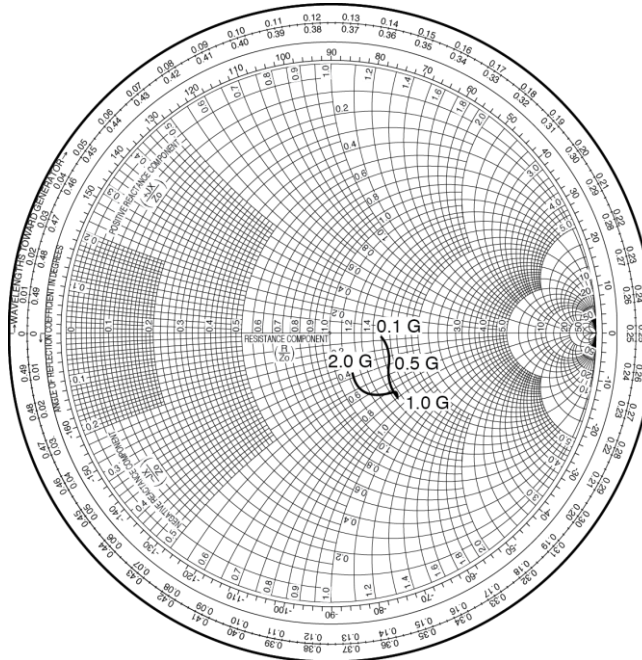


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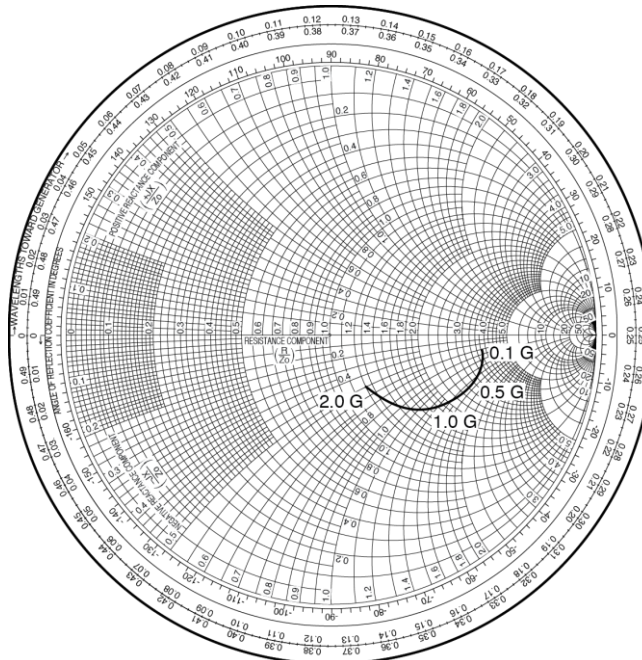
**SMITH CHART ( $T_A = +25^\circ\text{C}$ ,  $V_{CC} = 3.0\text{ V}$ )**

—  $\mu$ PC2745TB —

S<sub>11</sub>-FREQUENCY



S<sub>22</sub>-FREQUENCY



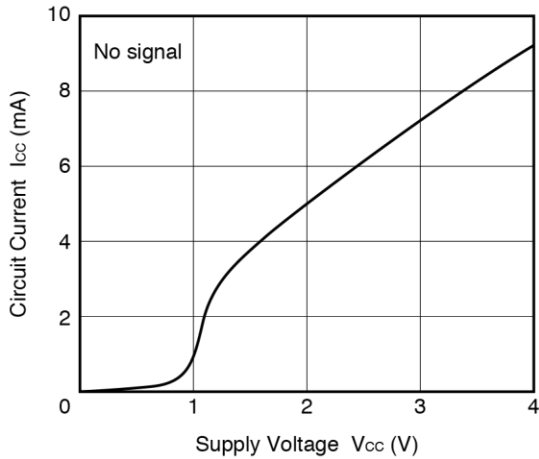
**S-PARAMETERS**

- S-parameters and noise parameters are provided on our Web site in a format (S2P) that enables the direct import of the parameters to microwave circuit simulators without the need for keyboard inputs.
- [Click here to download S-parameters.](#)
- [RF and Microwave] ® [Device Parameters]
- URL <http://www.necel.com/microwave/en/>

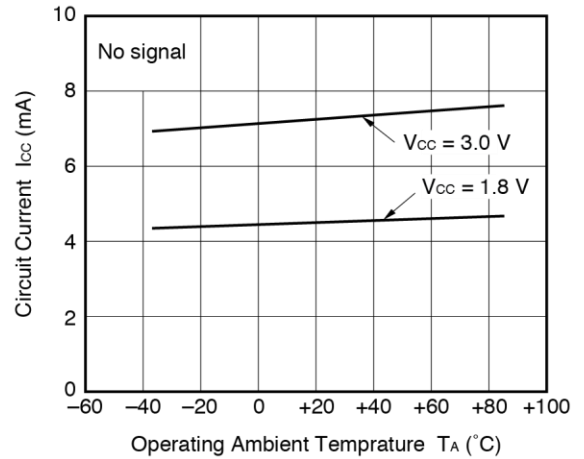
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—  $\mu$ PC2746TB —

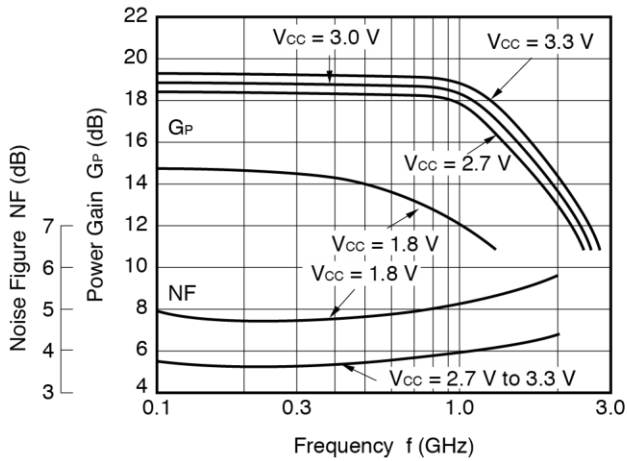
CIRCUIT CURRENT vs. SUPPLY VOLTAGE



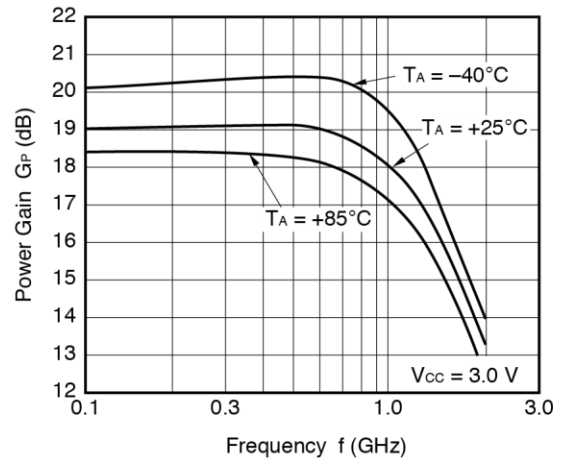
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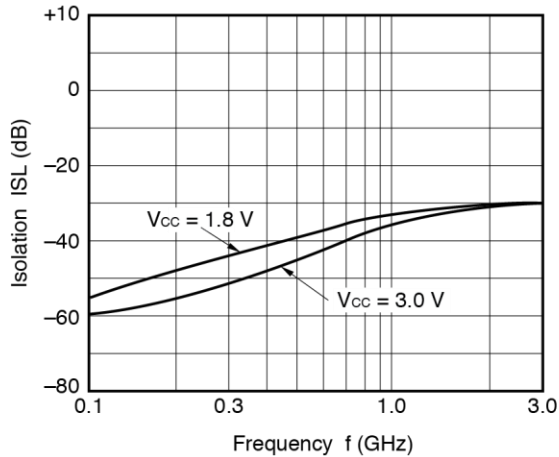
NOISE FIGURE, POWER GAIN vs. FREQUENCY



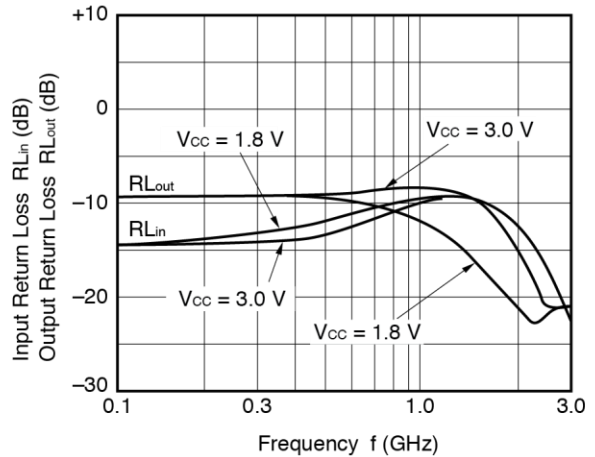
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ISOLATION vs. FREQUENCY



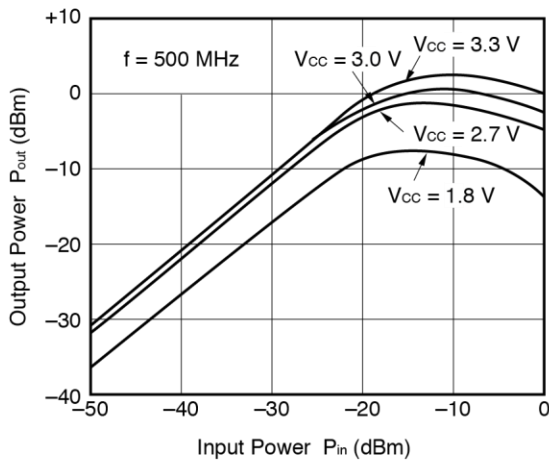
INPUT RETURN LOSS, OUTPUT RETURN LOSS vs. FREQUENCY



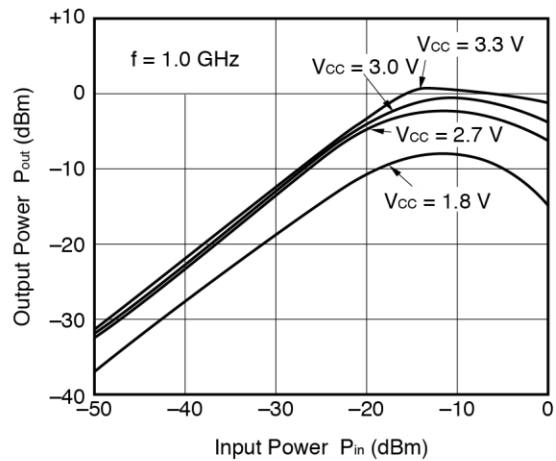
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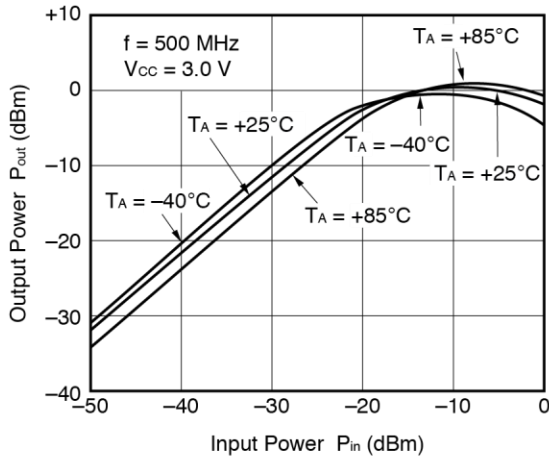
OUTPUT POWER vs. INPUT POWER



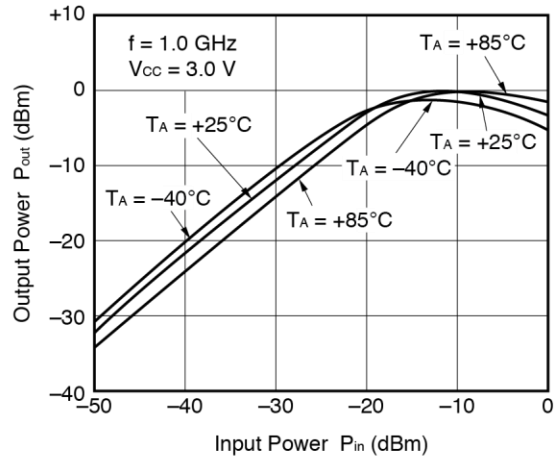
OUTPUT POWER vs. INPUT POWER



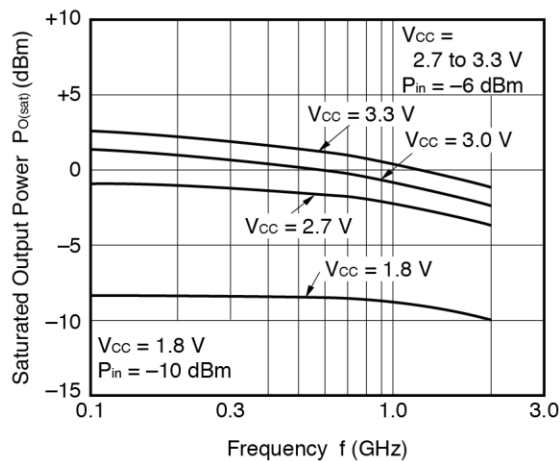
OUTPUT POWER vs. INPUT POWER



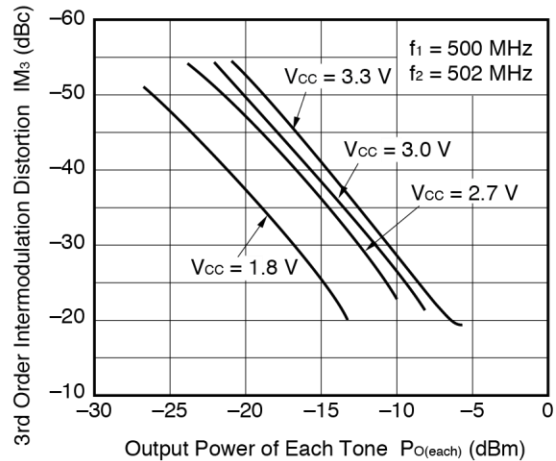
OUTPUT POWER vs. INPUT POWER



SATURATED OUTPUT POWER vs. FREQUENCY



3RD ORDER INTERMODULATION DISTORTION vs. OUTPUT POWER OF EACH TONE

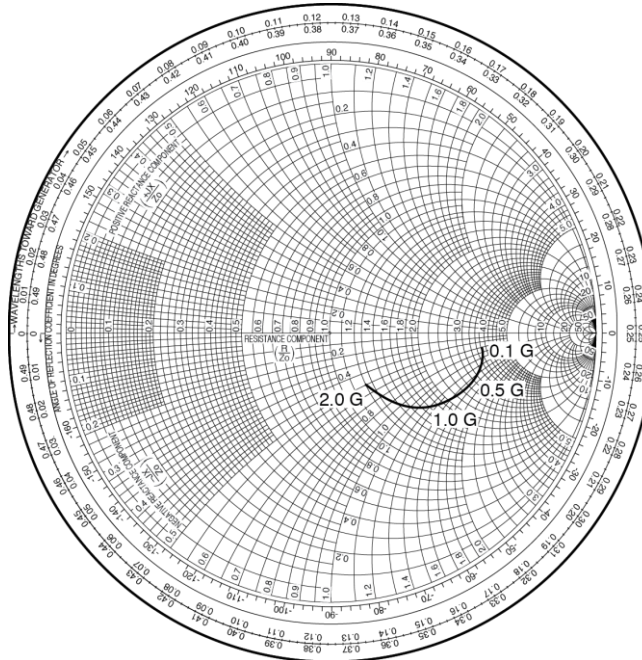


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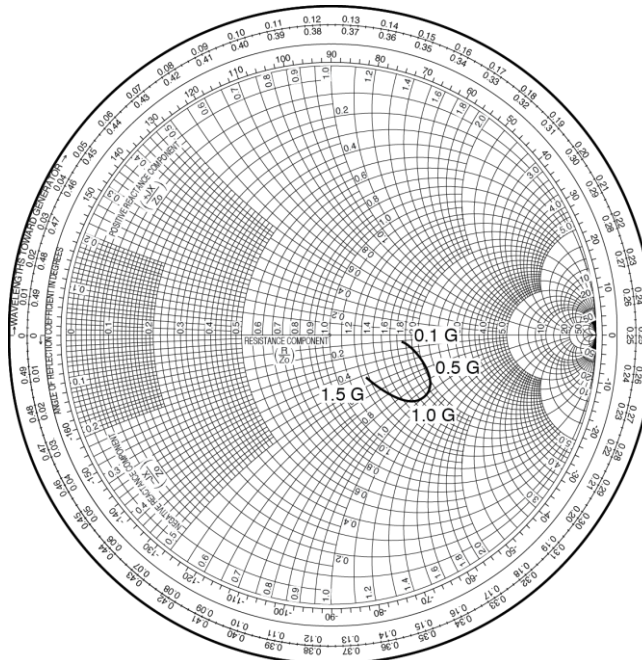
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S<sub>22</sub>-FREQUENCY

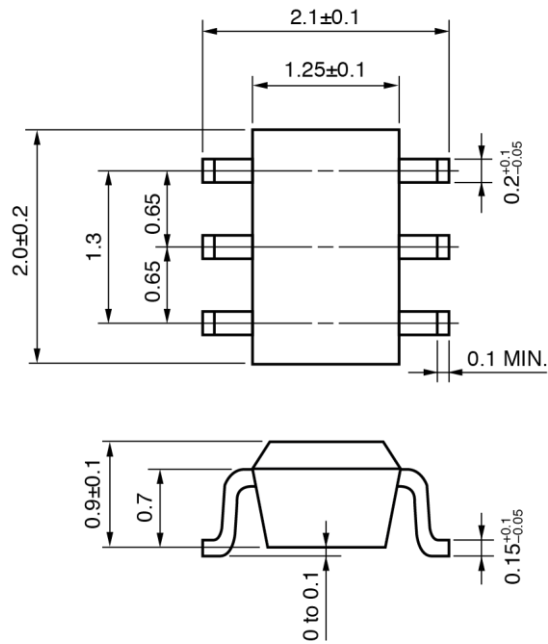


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PACKAGE DIMENSIONS

6-PIN SUPER MINIMOLD (UNIT: mm)





**NOTES ON CORRECT USE**

- (1) Observe precautions for handling because of electro-static sensitive devices.
- (2) Form a ground pattern as widely as possible to minimize ground impedance (to prevent undesired oscillation).  
All the ground pins must be connected together with wide ground pattern to decrease impedance difference.
- (3) The bypass capacitor should be attached to the Vcc pin.
- (4) The DC cut capacitor must be attached to input pin and output pin.

**RECOMMENDED SOLDERING CONDITIONS**

This product should be soldered and mounted under the following recommended conditions. For soldering methods and conditions other than those recommended below, contact your nearby sales office.

Soldering Method	Soldering Conditions	Condition Symbol
Infrared Reflow	Peak temperature (package surface temperature) : 260°C or below Time at peak temperature : 10 seconds or less Time at temperature of 220°C or higher : 60 seconds or less Preheating time at 120 to 180°C : 120±30 seconds Maximum number of reflow processes : 3 times Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	IR260
VPS	Peak temperature (package surface temperature) : 215°C or below Time at temperature of 200°C or higher : 25 to 40 seconds Preheating time at 120 to 150°C : 30 to 60 seconds Maximum number of reflow processes : 3 times Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	VP215
Wave Soldering	Peak temperature (molten solder temperature) : 260°C or below Time at peak temperature : 10 seconds or less Preheating temperature (package surface temperature) : 120°C or below Maximum number of flow processes : 1 time Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	WS260
Partial Heating	Peak temperature (pin temperature) : 350°C or below Soldering time (per side of device) : 3 seconds or less Maximum chlorine content of rosin flux (% mass) : 0.2%(Wt.) or below	HS350

**Caution Do not use different soldering methods together (except for partial heating).**

Компания «Life Electronics» занимается поставками электронных компонентов импортного и отечественного производства от производителей и со складов крупных дистрибьюторов Европы, Америки и Азии.

С конца 2013 года компания активно расширяет линейку поставок компонентов по направлению коаксиальный кабель, кварцевые генераторы и конденсаторы (керамические, пленочные, электролитические), за счёт заключения дистрибьюторских договоров

Мы предлагаем:

- Конкурентоспособные цены и скидки постоянным клиентам.
- Специальные условия для постоянных клиентов.
- Подбор аналогов.
- Поставку компонентов в любых объемах, удовлетворяющих вашим потребностям.
- Приемлемые сроки поставки, возможна ускоренная поставка.
- Доставку товара в любую точку России и стран СНГ.
- Комплексную поставку.
- Работу по проектам и поставку образцов.
- Формирование склада под заказчика.
- Сертификаты соответствия на поставляемую продукцию (по желанию клиента).
- Тестирование поставляемой продукции.
- Поставку компонентов, требующих военную и космическую приемку.
- Входной контроль качества.
- Наличие сертификата ISO.

В составе нашей компании организован Конструкторский отдел, призванный помогать разработчикам, и инженерам.

Конструкторский отдел помогает осуществить:

- Регистрацию проекта у производителя компонентов.
- Техническую поддержку проекта.
- Защиту от снятия компонента с производства.
- Оценку стоимости проекта по компонентам.
- Изготовление тестовой платы монтаж и пусконаладочные работы.



Тел: +7 (812) 336 43 04 (многоканальный)

Email: [org@lifeelectronics.ru](mailto:org@lifeelectronics.ru)